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# 16

# H8/3039 Group, H8/3039F-ZTAT™

# Hardware Manual

Renesas 16-Bit Single-Chip Microcomputer H8 Family / H8/300H Series

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H8/3039 H8/3038	HD64F3039F HD64F3039TE HD64F3039VF HD64F3039VTE HD6433039F HD6433039VF HD6433039VF HD6433039VTE HD6433038F HD6433038VF HD6433038VF HD6433038VTE	H8/3037 H8/3036	HD HD HD HD HD	6433037F 6433037VF 6433037VF 6433036F 6433036F 6433036VF 6433036VF 6433036VTE
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#### General Precautions in the Handling of MPU/MCU Products

The following usage notes are applicable to all MPU/MCU products from Renesas. For detailed usage notes on the products covered by this manual, refer to the relevant sections of the manual. If the descriptions under General Precautions in the Handling of MPU/MCU Products and in the body of the manual differ from each other, the description in the body of the manual takes precedence.

#### 1. Handling of Unused Pins

Handle unused pins in accord with the directions given under Handling of Unused Pins in the manual.

- The input pins of CMOS products are generally in the high-impedance state. In operation with an unused pin in the open-circuit state, extra electromagnetic noise is induced in the vicinity of LSI, an associated shoot-through current flows internally, and malfunctions may occur due to the false recognition of the pin state as an input signal. Unused pins should be handled as described under Handling of Unused Pins in the manual.
- 2. Processing at Power-on

The state of the product is undefined at the moment when power is supplied.

 The states of internal circuits in the LSI are indeterminate and the states of register settings and pins are undefined at the moment when power is supplied.

In a finished product where the reset signal is applied to the external reset pin, the states of pins are not guaranteed from the moment when power is supplied until the reset process is completed.

In a similar way, the states of pins in a product that is reset by an on-chip power-on reset function are not guaranteed from the moment when power is supplied until the power reaches the level at which resetting has been specified.

3. Prohibition of Access to Reserved Addresses

Access to reserved addresses is prohibited.

- The reserved addresses are provided for the possible future expansion of functions. Do
  not access these addresses; the correct operation of LSI is not guaranteed if they are
  accessed.
- 4. Clock Signals

After applying a reset, only release the reset line after the operating clock signal has become stable. When switching the clock signal during program execution, wait until the target clock signal has stabilized.

- When the clock signal is generated with an external resonator (or from an external oscillator) during a reset, ensure that the reset line is only released after full stabilization of the clock signal. Moreover, when switching to a clock signal produced with an external resonator (or by an external oscillator) while program execution is in progress, wait until the target clock signal is stable.
- 5. Differences between Products

Before changing from one product to another, i.e. to one with a different type number, confirm that the change will not lead to problems.

 The characteristics of MPU/MCU in the same group but having different type numbers may differ because of the differences in internal memory capacity and layout pattern. When changing to products of different type numbers, implement a system-evaluation test for each of the products.

#### Renesas

# Preface

The H8/3039 Group comprises high-performance single-chip microcomputers (MCUs) that integrate system supporting functions together with an H8/300H CPU core.

The H8/300H CPU has a 32-bit internal architecture with sixteen 16-bit general registers, and a concise, optimized instruction set designed for speed. It can address a 16-Mbyte linear address space.

The on-chip system supporting functions include ROM, RAM, a 16-bit integrated timer unit (ITU), a programmable timing pattern controller (TPC), a watchdog timer (WDT), a serial communication interface (SCI), an A/D converter, I/O ports, and other facilities. Of the two SCI channels, one has been expanded to support the ISO/IEC 7816-3 smart card interface. Functions have also been added to reduce power consumption in battery-powered applications: individual modules can be placed in standby, and the frequency of the system clock supplied to the chip can be divided down under software control.

The five MCU operating modes offer a choice of expanded mode, single-chip mode, and address space size, enabling the H8/3039 Group to adapt quickly and flexibly to a variety of conditions.

In addition to its mask-ROM versions, the H8/3039 Group has an F-ZTAT<sup>TM</sup> version with user programmable on-chip flash memory that can be programmed on-board. These versions enable users to respond quickly and flexibly to changing application specifications.

This manual describes the H8/3039 Group hardware. For details of the instruction set, refer to the H8/300H Series Software Manual.

Note: F-ZTAT is a trademark of Renesas Technology Corp.

#### Renesas

# Main Revisions for This Edition

Item	Page	Revision (See Manual for Details)
All	_	<ul> <li>Notification of change in company name amended (Before) Hitachi, Ltd. → (After) Renesas Technology Corp.</li> <li>Product naming convention amended (Before) H8/3039 Series → (After) H8/3039 Group</li> </ul>
2.3 Address Space Figure 2.2 Memory Map	20	Figure amended H'0000 H'FFFF

		1. N	ormal r	node (6	64-Kby	te mod	e)			
5.2.2 Interrupt Priority	91	Description amended								
Registers A and B		Bit	7	6	5	4	3	2	1	0
(IPRA, IPRB)		Initial value	IPRB7 0	IPRB6 0	0	0	IPRB3 0	IPRB2 0	IPRB1 0	0
Interrupt Priority Register B (IPRB)		Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
5.2.3 IRQ Status	93	Descriptio	on ame	nded						
Register (ISR)	Bits 5, 4, 1 and 0—IRQ <sub>5</sub> , IRQ <sub>4</sub> , IRQ <sub>1</sub> and IRQ <sub>0</sub> Fi IRQ4F, IRQ1F, and IRQ0F): These bits indicate IRQ <sub>5</sub> , IRQ <sub>4</sub> , IRQ <sub>1</sub> , and IRQ <sub>0</sub> interrupt requests.							cate th		
5.2.4 IRQ Enable	94	Description amended								
Register (IER)		Bit		7	6					
				—						
		Initial valu	е	0	0					
		Read/Writ	e _	R/W	R/V	/				
	Reserved bi									
				10001		•				
Bits 5, 4, 1, and 0—IRQ <sub>5</sub> , IRQ <sub>4</sub> , IRQ <sub>1</sub> , and IRQ <sub>0</sub> Enable (IRQ IRQ4E, IRQ1E, IRQ0E): These bits enable or disable IRQ5 IRQ4, IRQ1, IRQ0, interrupts.										

Item	Page	Revision (See Manual for Details)
5.3.3 Interrupt Vector Table	98	Table amended
Table 5.3 Interrupt		WOVI (interval timer)
Sources, Vector Addresses, and Priority		
5.5.4 Usage Notes	109	Figure amended
Figure 5.9 IRQnF Flag when Interrupt Exception Handling is not Executed		1 read 0 written 1 read 0 written 1 read 0 written
		(Inadvertent clearing)
		Generation condition (2)
6.4.2 Precautions on	131	Description amended
Setting ASTCR and ABWCR*		Modes 5 and 7
ADWOR		ASTCR0 = 0 ABWCR = H'FC
11.2.8 Bit Rate	349	Description added
Register (BRR)		The baud rate generator is controlled separately for the individual channels, so different values may be set for each.
Table 11.3 Examples	351	Table amended
of Bit Rates and BRR Settings in		φ (MHz)
Asynchronous Mode		
		Bit Rate         Error           (bits/s)         n         N         (%)           300         2         77         0.16
11.3.4 Synchronous	376	Description amended
Operation Clock		An internal clock generated by the on-chip baud rate generator or an external clock input from the SCK pin can be selected by clearing or setting the CKE1 and CKE0 bits in SCR and the C/A bit in SMR. See table 11.9.

Item	Page	Revision (See Manual for Details)									
16.2.1 Connecting a Crystal Resonator	500	Preliminary	delete	d							
Table 16.2 Crystal Resonator Parameters											
18.1.3 AC Characteristics	532	Table amen	ded	Con	dition A	Con	dition B	Con	dition C		
Table 18.5 Control				8	MHz	10 MHz			8 MHz	_	
Signal Timing		Item	Symbol		Max	Min	Max	Min	Max	Unit	Test Conditions
Signal Timing		RES setup time RES pulse width	t <sub>RESS</sub>	200 10	_	200 10	_	200 10	-	ns tcyc	Figure 18.10
		Mode programming setup time (MD <sub>o</sub> , MD, MD <sub>2</sub> )	t <sub>MDS</sub>	200	_	200	-	200	-	ns	
18.1.4 A/D Conversion Characteristics	535	Newly added	b								_
18.2.2 DC Characteristics	541	Table amended									
Table 18.10		Item									
Permissible Output Currents		Permissible output Total of 27 pins low current (total) including ports 1, 2, 5 and B									
A.1 Instruction List	576	Table amen	ded								
8. Block transfer instructions		Mnemonic EEPMOV. W		≐0 then eat @R5	→ @R6 → R5						
			unti else r	R4-1 R4=0	$\rightarrow$ R6 $\rightarrow$ R4						
A.3 Number of States Required for Execution	584	Table amended									
Table A.4 Number of						Acce		Operati			
Cycles per Instruction		Instruction Mner				м		N			
		BSR BSR	d:16	Norm				2 2			
				Advar	icea	.i					

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# Contents

Secti	on 1	Overview	1
1.1	Overv	iew	1
1.2	Block	Diagram	6
1.3	Pin De	escription	7
	1.3.1	Pin Arrangement	7
	1.3.2	Pin Functions	8
1.4	Pin Fu	nctions	12
Secti	on 2	CPU	17
2.1	Overv	iew	17
	2.1.1	Features	17
	2.1.2	Differences from H8/300 CPU	18
2.2	CPU (	Dperating Modes	19
2.3		ss Space	20
2.4	Regist	er Configuration	21
	2.4.1	Overview	21
	2.4.2	General Registers	22
	2.4.3	Control Registers	23
	2.4.4	Initial CPU Register Values	24
2.5	Data F	Pormats	25
	2.5.1	General Register Data Formats	25
	2.5.2	Memory Data Formats	27
2.6	Instruc	ction Set	28
	2.6.1	Instruction Set Overview	28
	2.6.2	Instructions and Addressing Modes	29
	2.6.3	Tables of Instructions Classified by Function	31
	2.6.4	Basic Instruction Formats	40
	2.6.5	Notes on Use of Bit Manipulation Instructions	41
2.7	Addre	ssing Modes and Effective Address Calculation	41
	2.7.1	Addressing Modes	41
	2.7.2	Effective Address Calculation	45
2.8	Proces	sing States	49
	2.8.1	Overview	49
	2.8.2	Program Execution State	49
	2.8.3	Exception-Handling State	50
	2.8.4	Exception-Handling Sequences	51
	2.8.5	Reset State	53

	2.8.6 Power-Down State	53
2.9	Basic Operational Timing	54
	2.9.1 Overview	54
	2.9.2 On-Chip Memory Access Timing	54
	2.9.3 On-Chip Supporting Module Access Timing	55
	2.9.4 Access to External Address Space	56
Secti	ion 3 MCU Operating Modes	57
3.1	Overview	57
	3.1.1 Operating Mode Selection	57
	3.1.2 Register Configuration	58
3.2	Mode Control Register (MDCR)	59
3.3	System Control Register (SYSCR)	60
3.4	Operating Mode Descriptions	62
	3.4.1 Mode 1	62
	3.4.2 Mode 3	62
	3.4.3 Mode 5	62
	3.4.4 Mode 6	62
	3.4.5 Mode 7	62
3.5	Pin Functions in Each Operating Mode	63
3.6	Memory Map in Each Operating Mode	63
3.7	Restrictions on Use of Mode 6	72
Secti	ion 4 Exception Handling	75
4.1	Overview	75
	4.1.1 Exception Handling Types and Priority	75
	4.1.2 Exception Handling Operation	75
	4.1.3 Exception Vector Table	76
4.2	Reset	78
	4.2.1 Overview	78
	4.2.2 Reset Sequence	78
	4.2.3 Interrupts after Reset	80
4.3	Interrupts	80
4.4	Trap Instruction	81
4.5	Stack Status after Exception Handling	81
4.6	Notes on Stack Usage	82
Secti	ion 5 Interrupt Controller	83
5.1	Overview	83
	5.1.1 Features	83
	5.1.2 Block Diagram	84

	5.1.3	Pin Configuration	85
	5.1.4	Register Configuration	85
5.2	Registe	Pr Descriptions	86
	5.2.1	System Control Register (SYSCR)	86
	5.2.2	Interrupt Priority Registers A and B (IPRA, IPRB)	88
	5.2.3	IRQ Status Register (ISR)	
	5.2.4	IRQ Enable Register (IER)	94
	5.2.5	IRQ Sense Control Register (ISCR)	95
5.3	Interru	pt Sources	96
	5.3.1	External Interrupts	
	5.3.2	Internal Interrupts	97
	5.3.3	Interrupt Vector Table	97
5.4	Interru	pt Operation	100
	5.4.1	Interrupt Handling Process	100
	5.4.2	Interrupt Sequence	
	5.4.3	Interrupt Response Time	106
5.5	Usage	Notes	
	5.5.1	Contention between Interrupt and Interrupt-Disabling Instruction	107
	5.5.2	Instructions that Inhibit Interrupts	108
	5.5.3	Interrupts during EEPMOV Instruction Execution	108
	5.5.4	Usage Notes	108
~ .			
Secti		Bus Controller	
6.1		ew	
	6.1.1	Factures	111
		Features	
	6.1.2	Block Diagram	112
	6.1.2 6.1.3	Block Diagram Input/Output Pins	112 113
	6.1.3 6.1.4	Block Diagram Input/Output Pins Register Configuration	112 113 113
6.2	6.1.3 6.1.4 Registe	Block Diagram Input/Output Pins Register Configuration r Descriptions	112 113 113 114
6.2	6.1.3 6.1.4 Registe 6.2.1	Block Diagram Input/Output Pins Register Configuration er Descriptions Access State Control Register (ASTCR)	112 113 113 114 114
6.2	<ul> <li>6.1.3</li> <li>6.1.4</li> <li>Register</li> <li>6.2.1</li> <li>6.2.2</li> </ul>	Block Diagram Input/Output Pins Register Configuration er Descriptions Access State Control Register (ASTCR) Wait Control Register (WCR)	112 113 113 114 114 115
6.2	6.1.3 6.1.4 Registe 6.2.1	Block Diagram Input/Output Pins Register Configuration r Descriptions Access State Control Register (ASTCR) Wait Control Register (WCR) Wait State Controller Enable Register (WCER)	112 113 113 114 114 115 116
6.2	6.1.3 6.1.4 Register 6.2.1 6.2.2 6.2.3 6.2.4	Block Diagram Input/Output Pins Register Configuration er Descriptions Access State Control Register (ASTCR) Wait Control Register (WCR) Wait State Controller Enable Register (WCER) Address Control Register (ADRCR)	<ol> <li>112</li> <li>113</li> <li>113</li> <li>114</li> <li>114</li> <li>115</li> <li>116</li> <li>117</li> </ol>
<ul><li>6.2</li><li>6.3</li></ul>	6.1.3 6.1.4 Register 6.2.1 6.2.2 6.2.3 6.2.4	Block Diagram Input/Output Pins Register Configuration r Descriptions Access State Control Register (ASTCR) Wait Control Register (WCR) Wait State Controller Enable Register (WCER)	<ol> <li>112</li> <li>113</li> <li>113</li> <li>114</li> <li>114</li> <li>115</li> <li>116</li> <li>117</li> </ol>
	6.1.3 6.1.4 Registe 6.2.1 6.2.2 6.2.3 6.2.4 Operat	Block Diagram Input/Output Pins Register Configuration er Descriptions Access State Control Register (ASTCR) Wait Control Register (WCR) Wait State Controller Enable Register (WCER) Address Control Register (ADRCR) ion	<ol> <li>112</li> <li>113</li> <li>113</li> <li>114</li> <li>114</li> <li>115</li> <li>116</li> <li>117</li> </ol>
	6.1.3 6.1.4 Registe 6.2.1 6.2.2 6.2.3 6.2.4 Operat	Block Diagram Input/Output Pins Register Configuration er Descriptions Access State Control Register (ASTCR) Wait Control Register (WCR) Wait State Controller Enable Register (WCER) Address Control Register (ADRCR) ion	<ol> <li>112</li> <li>113</li> <li>114</li> <li>114</li> <li>115</li> <li>116</li> <li>117</li> <li>119</li> <li>119</li> </ol>
	6.1.3 6.1.4 Register 6.2.1 6.2.2 6.2.3 6.2.4 Operatt 6.3.1 6.3.2 6.3.3	Block Diagram Input/Output Pins Register Configuration er Descriptions Access State Control Register (ASTCR) Wait Control Register (WCR) Wait State Controller Enable Register (WCER) Address Control Register (ADRCR) Area Division Bus Control Signal Timing Wait Modes	<ol> <li>112</li> <li>113</li> <li>114</li> <li>114</li> <li>115</li> <li>116</li> <li>117</li> <li>119</li> <li>119</li> <li>121</li> <li>123</li> </ol>
	6.1.3 6.1.4 Register 6.2.1 6.2.2 6.2.3 6.2.4 Operatt 6.3.1 6.3.2 6.3.3 6.3.4	Block Diagram Input/Output Pins	<ol> <li>112</li> <li>113</li> <li>114</li> <li>114</li> <li>115</li> <li>116</li> <li>117</li> <li>119</li> <li>121</li> <li>123</li> <li>129</li> </ol>
	6.1.3 6.1.4 Register 6.2.1 6.2.2 6.2.3 6.2.4 Operate 6.3.1 6.3.2 6.3.3 6.3.4 Usage	Block Diagram Input/Output Pins	<ul> <li>112</li> <li>113</li> <li>113</li> <li>114</li> <li>114</li> <li>115</li> <li>116</li> <li>117</li> <li>119</li> <li>121</li> <li>123</li> <li>129</li> <li>131</li> </ul>
6.3	6.1.3 6.1.4 Register 6.2.1 6.2.2 6.2.3 6.2.4 Operatt 6.3.1 6.3.2 6.3.3 6.3.4	Block Diagram Input/Output Pins	<ol> <li>112</li> <li>113</li> <li>114</li> <li>114</li> <li>115</li> <li>116</li> <li>117</li> <li>119</li> <li>121</li> <li>123</li> <li>129</li> <li>131</li> <li>131</li> </ol>

Secti	on 7 l	I/O Ports	133
7.1	Overvie	ew	133
7.2	Port 1		137
	7.2.1	Overview	137
	7.2.2	Register Descriptions	138
	7.2.3	Pin Functions in Each Mode	140
7.3	Port 2		142
	7.3.1	Overview	142
	7.3.2	Register Descriptions	143
	7.3.3	Pin Functions in Each Mode	145
	7.3.4	Input Pull-Up Transistors	147
7.4	Port 3.		148
	7.4.1	Overview	148
	7.4.2	Register Descriptions	148
	7.4.3	Pin Functions in Each Mode	150
7.5	Port 5		152
	7.5.1	Overview	152
	7.5.2	Register Descriptions	153
	7.5.3	Pin Functions in Each Mode	155
	7.5.4	Input Pull-Up Transistors	156
7.6	Port 6		157
	7.6.1	Overview	157
	7.6.2	Register Descriptions	158
	7.6.3	Pin Functions in Each Mode	160
7.7	Port 7		163
	7.7.1	Overview	163
	7.7.2	Register Description	163
7.8	Port 8		164
	7.8.1	Overview	164
	7.8.2	Register Descriptions	165
	7.8.3	Pin Functions	167
7.9	Port 9		168
	7.9.1	Overview	168
	7.9.2	Register Descriptions	168
	7.9.3	Pin Functions	170
7.10	Port A.		172
	7.10.1	Overview	172
		Register Descriptions	
		Pin Functions	
7.11	Port B.		182
	7.11.1	Overview	182

	7.11.2	Register Descriptions	182
	7.11.3	Pin Functions	184
Secti	on 8 1	6-Bit Integrated Timer Unit (ITU)	191
8.1	Overvie		191
	8.1.1	Features	191
	8.1.2	Block Diagrams	194
	8.1.3	Input/Output Pins	199
	8.1.4	Register Configuration	201
8.2	Registe	r Descriptions	204
	8.2.1	Timer Start Register (TSTR)	204
	8.2.2	Timer Synchro Register (TSNC)	206
	8.2.3	Timer Mode Register (TMDR)	208
	8.2.4	Timer Function Control Register (TFCR)	211
	8.2.5	Timer Output Master Enable Register (TOER)	214
	8.2.6	Timer Output Control Register (TOCR)	216
	8.2.7	Timer Counters (TCNT)	218
	8.2.8	General Registers (GRA, GRB)	219
	8.2.9	Buffer Registers (BRA, BRB)	220
	8.2.10	Timer Control Registers (TCR)	221
	8.2.11	Timer I/O Control Register (TIOR)	223
	8.2.12	Timer Status Register (TSR)	225
	8.2.13	Timer Interrupt Enable Register (TIER)	
8.3	CPU In	terface	228
	8.3.1	16-Bit Accessible Registers	228
	8.3.2	8-Bit Accessible Registers	231
8.4	Operati	on	232
	8.4.1	Overview	232
	8.4.2	Basic Functions	234
	8.4.3	Synchronization	243
	8.4.4	PWM Mode	245
	8.4.5	Reset-Synchronized PWM Mode	249
	8.4.6	Complementary PWM Mode	252
	8.4.7	Phase Counting Mode	261
	8.4.8	Buffering	263
	8.4.9	ITU Output Timing	269
8.5	Interrup	ots	272
	8.5.1	Setting of Status Flags	272
	8.5.2	Clearing of Status Flags	274
	8.5.3	Interrupt Sources	275
8.6	Usage 1	Notes	276

Secti	on 9	Programmable Timing Pattern Controller	291
9.1	Overvi	ew	291
	9.1.1	Features	291
	9.1.2	Block Diagram	292
	9.1.3	TPC Pins	293
	9.1.4	Registers	294
9.2	Registe	r Descriptions	295
	9.2.1	Port A Data Direction Register (PADDR)	295
	9.2.2	Port A Data Register (PADR)	295
	9.2.3	Port B Data Direction Register (PBDDR)	296
	9.2.4	Port B Data Register (PBDR)	296
	9.2.5	Next Data Register A (NDRA)	297
	9.2.6	Next Data Register B (NDRB)	299
	9.2.7	Next Data Enable Register A (NDERA)	301
	9.2.8	Next Data Enable Register B (NDERB)	302
	9.2.9	TPC Output Control Register (TPCR)	303
	9.2.10	TPC Output Mode Register (TPMR)	306
9.3	Operati	on	308
	9.3.1	Overview	308
	9.3.2	Output Timing	309
	9.3.3	Normal TPC Output	310
	9.3.4	Non-Overlapping TPC Output	312
	9.3.5	TPC Output Triggering by Input Capture	314
9.4	Usage I	Notes	
	9.4.1	Operation of TPC Output Pins	315
	9.4.2	Note on Non-Overlapping Output	315
Secti		Watchdog Timer	
10.1	Overvi	ew	317
	10.1.1	Features	317
	10.1.2	Block Diagram	318
	10.1.3	Pin Configuration	318
		Register Configuration	
10.2	Registe	r Descriptions	319
	10.2.1	Timer Counter (TCNT)	319
	10.2.2	Timer Control/Status Register (TCSR)	320
	10.2.3	Reset Control/Status Register (RSTCSR)	322
	10.2.4	Notes on Register Access	324
10.3	Operati	on	326
	10.3.1	Watchdog Timer Operation	
	10.3.2	Interval Timer Operation	327

	10.3.3	Timing of Setting of Overflow Flag (OVF)	328
	10.3.4	Timing of Setting of Watchdog Timer Reset Bit (WRST)	329
10.4	Interru	Dts	329
10.5	Usage 1	Notes	330
	-		
Secti	on 11	Serial Communication Interface	331
11.1	Overvie	ew	331
	11.1.1	Features	331
	11.1.2	Block Diagram	333
	11.1.3	Input/Output Pins	334
	11.1.4	Register Configuration	335
11.2	Registe	r Descriptions	336
	11.2.1	Receive Shift Register (RSR)	336
	11.2.2	Receive Data Register (RDR)	336
	11.2.3	Transmit Shift Register (TSR)	337
	11.2.4	Transmit Data Register (TDR)	337
	11.2.5	Serial Mode Register (SMR)	338
	11.2.6	Serial Control Register (SCR)	341
	11.2.7	Serial Status Register (SSR)	345
	11.2.8	Bit Rate Register (BRR)	349
11.3	Operati	on	358
	11.3.1	Overview	358
	11.3.2	Operation in Asynchronous Mode	360
	11.3.3	Multiprocessor Communication	369
	11.3.4	Synchronous Operation	376
11.4	SCI Int	errupts	384
11.5	Usage l	Notes	385
	-		
Secti	on 12	Smart Card Interface	391
12.1	Overvie	ew	391
	12.1.1	Features	391
	12.1.2	Block Diagram	392
	12.1.3	Pin Configuration	393
	12.1.4	Register Configuration	393
12.2	Registe	r Descriptions	394
	12.2.1	Smart Card Mode Register (SCMR)	394
	12.2.2	Serial Status Register (SSR)	396
12.3	Operati	on	
	12.3.1	Overview	397
	12.3.2	Pin Connections	398
	12.3.3	Data Format	399

	12.3.4	Register Settings	401
	12.3.5	Clock	403
	12.3.6	Data Transfer Operations	405
12.4	Usage 1	Note	411
Secti	on 13	A/D Converter	415
13.1	Overvie	ew	415
	13.1.1	Features	415
	13.1.2	Block Diagram	416
	13.1.3	Input Pins	417
		Register Configuration	
13.2	Registe	r Descriptions	419
	13.2.1	A/D Data Registers A to D (ADDRA to ADDRD)	419
	13.2.2	A/D Control/Status Register (ADCSR)	420
	13.2.3	A/D Control Register (ADCR)	422
13.3	CPU In	terface	423
13.4	Operati	on	424
	13.4.1	Single Mode (SCAN = 0)	424
	13.4.2	Scan Mode (SCAN = 1)	426
	13.4.3	Input Sampling and A/D Conversion Time	428
	13.4.4	External Trigger Input Timing	429
13.5	Interrup	pts	430
13.6	Usage I	Notes	430
Secti	on 14	RAM	435
14.1		ew	
	14.1.1	Block Diagram	436
		Register Configuration	
14.2	System	Control Register (SYSCR)	437
14.3	Operati	on	438
		ROM	
15.1		ew	
15.2		ew of Flash Memory	
		Features	
		Block Diagram	
	15.2.3	6	
		Register Configuration	
15.3	-	r Descriptions	
		Flash Memory Control Register (FLMCR)	
	15.3.2	Erase Block Register (EBR)	447

	15.3.3	RAM Control Register (RAMCR)	449
	15.3.4	Flash Memory Status Register (FLMSR)	451
15.4	On-Boa	ard Programming Modes	452
	15.4.1	Boot Mode	455
	15.4.2	User Program Mode	460
15.5	Program	nming/Erasing Flash Memory	462
	15.5.1	Program Mode	463
	15.5.2	Program-Verify Mode	464
	15.5.3	Erase Mode	466
	15.5.4	Erase-Verify Mode	466
15.6	Flash N	Memory Protection	468
	15.6.1	Hardware Protection	468
	15.6.2	Software Protection	470
		Error Protection	
	15.6.4	NMI Input Disable Conditions	473
15.7	Flash N	Iemory Emulation by RAM	474
15.8	Flash N	Iemory PROM Mode	475
	15.8.1	PROM Mode Setting	475
		Memory Map	
	15.8.3	PROM Mode Operation	476
	15.8.4	Memory Read Mode	479
	15.8.5	Auto-Program Mode	482
	15.8.6	Auto-Erase Mode	
	15.8.7	Status Read Mode	485
	15.8.8	PROM Mode Transition Time	487
	15.8.9	Notes on Memory Programming	488
15.9	Notes of	on Flash Memory Programming/Erasing	488
15.10	Mask F	ROM Overview	494
	15.10.1	Block Diagram	494
15.11	Notes of	on Ordering Mask ROM Version Chip	495
Secti		Clock Pulse Generator	
16.1		ew	
		Block Diagram	
16.2		tor Circuit	
	16.2.1	Connecting a Crystal Resonator	499
		External Clock Input	
16.3	-	djustment Circuit	
16.4		ers	
16.5	-	ncy Divider	
	16.5.1	Register Configuration	504

	16.5.2	Division Control Register (DIVCR)	505
	16.5.3	Usage Notes	506
Secti	on 17	Power-Down State	507
17.1	Overvie	ew	507
17.2	Registe	r Configuration	509
	17.2.1	System Control Register (SYSCR)	509
	17.2.2	Module Standby Control Register (MSTCR)	511
17.3	Sleep N	10de	513
	17.3.1	Transition to Sleep Mode	513
	17.3.2	Exit from Sleep Mode	513
17.4	Softwar	re Standby Mode	514
	17.4.1	Transition to Software Standby Mode	514
	17.4.2	Exit from Software Standby Mode	514
	17.4.3	Selection of Oscillator Waiting Time after Exit from Software Standby Mode	515
	17.4.4	Sample Application of Software Standby Mode	516
	17.4.5	Usage Note	
17.5	Hardwa	are Standby Mode	517
	17.5.1	Transition to Hardware Standby Mode	517
	17.5.2	Exit from Hardware Standby Mode	517
	17.5.3	Timing for Hardware Standby Mode	518
17.6	Module	e Standby Function	519
	17.6.1	Module Standby Timing	519
	17.6.2	Read/Write in Module Standby	519
	17.6.3	Usage Notes	519
17.7		Clock Output Disabling Function	
Secti	on 18	Electrical Characteristics	521
18.1	Electric	cal Characteristics of Mask ROM Version	521
	18.1.1	Absolute Maximum Ratings	521
	18.1.2	DC Characteristics	522
	18.1.3	AC Characteristics	530
	18.1.4	A/D Conversion Characteristics	535
18.2	Electric	cal Characteristics of Flash Memory Version	536
	18.2.1	Absolute Maximum Ratings	536
	18.2.2	DC Characteristics	537
	18.2.3	AC Characteristics	543
	18.2.4	A/D Conversion Characteristics	548
	18.2.5	Flash Memory Characteristics	549
18.3	Operati	onal Timing	552
	18.3.1	Bus Timing	552

	18.3.2 C	ontrol Signal Timing	556
	18.3.3 C	lock Timing	558
	18.3.4 T	PC and I/O Port Timing	558
		۲U Timing	
		CI Input/Output Timing	
Anne	endix A	Instruction Set	561
A.1		n List	
A.2		Code Maps	
A.3		of States Required for Execution	
11.0	i (unioti o	Total States Required for Encourton and an anti-	200
Appe	endix B	Internal I/O Register Field	590
B.1	Addresses	-	590
B.2	Function.		597
Anne	ndix C	I/O Block Diagrams	655
C.1		bek Diagram	
C.1 C.2		ock Diagram	
C.2 C.3		ock Diagram	
C.4		ock Diagram	
C.4		ock Diagram	
C.6		ock Diagram	
C.7		ock Diagram	
C.8		ock Diagram	
C.9		ock Diagram	
C.10		ock Diagram	
Appe	endix D	Pin States	674
D.1	Port States	s in Each Mode	674
D.2	Pin States	at Reset	676
Anne	endix E	Timing of Transition to and Recovery	
· · ppc		from Hardware Standby Mode	679
			5.7
Appe	endix F	Product Lineup	680
Anne	endix G	Package Dimensions	681
1 Phy			001

# Section 1 Overview

#### 1.1 Overview

The H8/3039 Group comprises microcomputers (MCUs) that integrate system supporting functions together with an H8/300H CPU core featuring an original Renesas Technology architecture.

The H8/300H CPU has a 32-bit internal architecture with sixteen 16-bit general registers, and a concise, optimized instruction set designed for speed. It can address a 16-Mbyte linear address space. Its instruction set is upward-compatible at the object-code level with the H8/300 CPU, enabling easy porting of software from the H8/300 Series.

The on-chip system supporting functions include ROM, RAM, a 16-bit integrated timer unit (ITU), a programmable timing pattern controller (TPC), a watchdog timer (WDT), a serial communication interface (SCI), an A/D converter, I/O ports, and other facilities.

The H8/3039 Group consists of four models: the H8/3039 with 128 kbytes of ROM and 4 kbytes of RAM, the H8/3038 with 64 kbytes of ROM and 2 kbytes of RAM, the H8/3037 with 32 kbytes of ROM and 1 kbytes of RAM, and the H8/3036 with 16 kbytes of ROM and 512 bytes of RAM.

The five MCU operating modes offer a choice of expanded mode, single-chip mode and address space size.

In addition to the mask-ROM version of the H8/3039 Group, an F-ZTAT<sup>TM</sup> version with an onchip flash memory that can be freely programmed and reprogrammed by the user after the board is installed is also available. This version enables users to respond quickly and flexibly to changing application specifications, growing production volumes, and other conditions.

Table 1.1 summarizes the features of the H8/3039 Group.

Note: F-ZTAT is a trademark of Renesas Technology Corp.

#### Renesas

#### Table 1.1 Features

Feature	Description				
CPU	Upward-compatible with the H8/300 CPU at the object-code level				
	General-register machine				
	Sixteen 16-bit general registers				
	(also useable as sixteen 8-bit registers or eight 32-bit registers)				
	High-speed operation				
	Maximum clock rate: 18 MHz				
	Add/subtract: 111 ns				
	Multiply/divide: 778 ns				
	Two CPU operating modes				
	Normal mode (64-kbyte address space)				
	Advanced mode (16-Mbyte address space)				
	Instruction features				
	8/16/32-bit data transfer, arithmetic, and logic instructions				
	• Signed and unsigned multiply instructions (8 bits × 8 bits, 16 bits × 16 bits)				
	• Signed and unsigned divide instructions (16 bits ÷ 8 bits, 32 bits ÷ 16 bits)				
	Bit accumulator function				
	Bit manipulation instructions with register-indirect specification of bit				
	positions				

Feature	Description					
Memory	H8/3039					
	ROM: 128 kbytes					
	RAM: 4 kbytes					
	H8/3038					
	ROM: 64 kbytes					
	<ul> <li>RAM: 2 kbytes</li> </ul>					
	H8/3037					
	ROM: 32 kbytes					
	RAM: 1 kbyte					
	H8/3036					
	ROM: 16 kbytes					
	RAM: 512 bytes					
Interrupt controller	• Five external interrupt pins: NMI, $\overline{IRQ}_{0}$ , $\overline{IRQ}_{1}$ , $\overline{IRQ}_{4}$ , $\overline{IRQ}_{5}$					
	25 internal interrupts					
	Three selectable interrupt priority levels					
Bus controller	Address space can be partitioned into eight areas, with independent bus specifications in each area					
	Two-state or three-state access selectable for each area					
	Selection of four wait modes					
16-bit integrated timer unit (ITU)	• Five 16-bit timer channels, capable of processing up to 12 pulse outputs or 10 pulse inputs					
	16-bit timer counter (channels 0 to 4)					
	• Two multiplexed output compare/input capture pins (channels 0 to 4)					
	Operation can be synchronized (channels 0 to 4)					
	PWM mode available (channels 0 to 4)					
	Phase counting mode available (channel 2)					
	Buffering available (channels 3 and 4)					
	Reset-synchronized PWM mode available (channels 3 and 4)					
	Complementary PWM mode available (channels 3 and 4)					

#### Section 1 Overview

Feature	Descripti	on				
Programmable	<ul> <li>Maximum 15-bit pulse output, using ITU as time base</li> </ul>					
timing pattern controller (TPC)	• Up to three 4-bit pulse output groups and one 3-bit pulse output group (or one 15-bit group, one 8-bit group, or one 7-bit group)					
	Non-o	verlap mode availa	ble			
Watchdog timer	Reset	signal can be gene	rated by overflow			
(WDT), 1 channel		signal can be outpu T version.)	ut externally (Howe	ever, not available wi	th the	
	Usable	e as an interval time	er			
Serial	Select	ion of asynchronou	s or synchronous	mode		
communication interface (SCI),	• Full du	uplex: can transmit	and receive simult	aneously		
2 channels	On-ch	ip baud-rate genera	ator			
	Smart	card interface func	tions added (SCI0	only)		
A/D converter	Resolution	ution: 10 bits				
	Eight channels, with selection of single or scan mode					
		le analog conversio				
	Sample-and-hold function					
		e externally triggere	ed			
I/O ports	55 input/output pins					
		e where each have				
Operating modes		operating modes				
	Mode	Address Space	Address Pins	Bus Width		
	Mode 1	1 Mbyte	$A_0$ to $A_{19}$	8 bits		
	Mode 3	16 Mbytes	$A_{23}$ to $A_0$	8 bits		
	Mode 5	1 Mbyte	$A_0$ to $A_{19}$	8 bits		
	Mode 6	64 kbytes	—			
	Mode 7	1 Mbyte	—			
On-chip ROM is disabled in modes 1 and 3						
Power-down state	Sleep mode					
	Software standby mode					
	Hardware standby mode					
	Module standby function					
Programmable System clock frequency division						

Feature	Description						
Other features	On-chip clock oscillator						
Product lineup	Model (5 V)	Model (3 V)*	Package	ROM			
	HD64F3039F	HD64F3039VF	80-pin QFP (FP-80A)	Flash memory			
	HD64F3039TE	HD64F3039VTE	80-pin TQFP (TFP-80C)				
	HD6433039F	HD6433039VF	80-pin QFP (FP-80A)	Mask ROM			
	HD6433039TE	HD6433039VTE	80-pin TQFP (TFP-80C)				
	HD6433038F	HD6433038VF	80-pin QFP (FP-80A)	Mask ROM			
	HD6433038TE	HD6433038VTE	80-pin TQFP (TFP-80C)				
	HD6433037F	HD6433037VF	80-pin QFP (FP-80A)	Mask ROM			
	HD6433037TE	HD6433037VTE	80-pin TQFP (TFP-80C)				
	HD6433036F	HD6433036VF	80-pin QFP (FP-80A)	Mask ROM			
	HD6433036TE	HD6433036VTE	80-pin TQFP (TFP-80C)				
	and or	ne with $V_{cc}$ = 3.0 V to	: one with V <sub>cc</sub> = 2.7 V to 5.5 V 5.5 V and $\phi$ = 2 to 10 MHz. H with V <sub>cc</sub> = 3.0 to 5.5 V and $\phi$	lowever, there is only			

#### 1.2 Block Diagram

Figure 1.1 shows an internal block diagram of the H8/3039 Group.

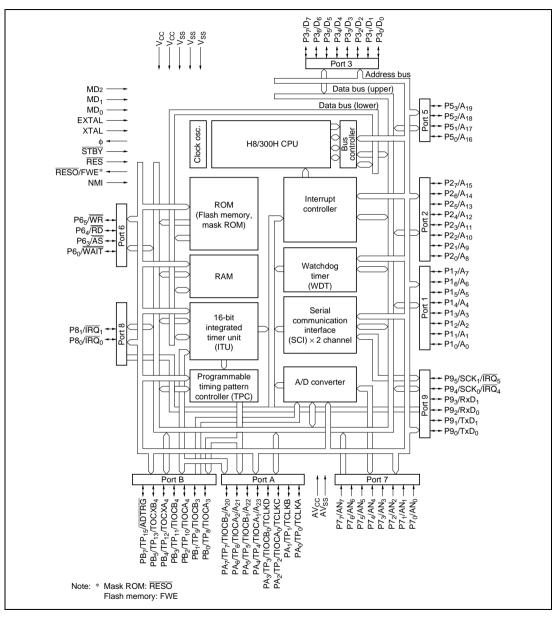


Figure 1.1 Block Diagram

Rev.3.00 Mar. 26, 2007 Page 6 of 682 REJ09B0353-0300

#### 1.3 Pin Description

#### 1.3.1 Pin Arrangement

Figure 1.2 shows the pin arrangement of the H8/3039 Group.

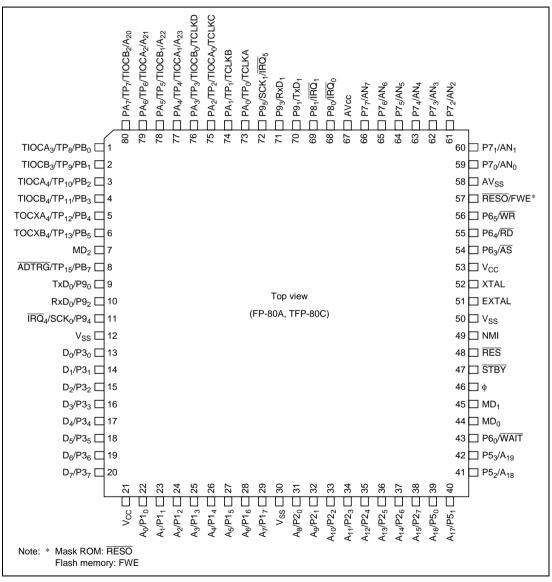


Figure 1.2 Pin Arrangement (FP-80A, TFP-80C Top View)

#### 1.3.2 Pin Functions

#### Pin Assignments in Each Mode

Table 1.2 lists the FP-80A and TFP-80C pin assignments in each mode.

#### Table 1.2 FP-80A and TFP-80C Pin Assignments in Each Mode

	Pin Name						
Pin No.	Mode 1	Mode 3	Mode 5	Mode 6	Mode 7	PROM Mode Flash memory	
1	PB <sub>0</sub> /TP <sub>8</sub> / TIOCA <sub>3</sub>	NC					
2	PB₁/TP₅/ TIOCB₃	PB₁/TP₅/ TIOCB₃	PB₁/TP₅/ TIOCB₃	PB₁/TP₅/ TIOCB₃	PB₁/TP₅/ TIOCB₃	NC	
3	PB <sub>2</sub> /TP <sub>10</sub> / TIOCA <sub>4</sub>	NC					
4	PB <sub>3</sub> /TP <sub>11</sub> / TIOCB <sub>4</sub>	NC					
5	PB <sub>4</sub> /TP <sub>12</sub> / TOCXA <sub>4</sub>	NC					
6	PB <sub>5</sub> /TP <sub>13</sub> / TOCXB <sub>4</sub>	PB₅/TP₁₃/ TOCXB₄	NC				
7	MD <sub>2</sub>	V <sub>ss</sub>					
8	PB <sub>7</sub> /TP <sub>15</sub> / ADTRG	PB,/TP <sub>15</sub> / ADTRG	NC				
9	P9 <sub>0</sub> /TxD <sub>0</sub>	NC					
10	P9 <sub>2</sub> /RxD <sub>0</sub>	$P9_2/RxD_0$	P9 <sub>2</sub> /RxD <sub>0</sub>	P9 <sub>2</sub> /RxD <sub>0</sub>	P9 <sub>2</sub> /RxD <sub>0</sub>	V <sub>ss</sub>	
11	P9₄/SCK₀/ IRQ₄	P9₄/SCK₀/ IRQ₄	P9₄/SCK₀/ IRQ₄	P9₄/SCK₀/ IRQ₄	P9₄/SCK₀/ IRQ₄	NC	
12	V <sub>ss</sub>	V <sub>ss</sub>					
13	D <sub>0</sub>	D <sub>o</sub>	D <sub>o</sub>	P3 <sub>0</sub>	P3 <sub>0</sub>	I/O <sub>0</sub>	
14	D <sub>1</sub>	D <sub>1</sub>	D <sub>1</sub>	P3 <sub>1</sub>	P3 <sub>1</sub>	I/O <sub>1</sub>	
15	D <sub>2</sub>	D <sub>2</sub>	D <sub>2</sub>	P3 <sub>2</sub>	P3 <sub>2</sub>	I/O <sub>2</sub>	
16	D <sub>3</sub>	D <sub>3</sub>	D <sub>3</sub>	P3 <sub>3</sub>	P3 <sub>3</sub>	I/O <sub>3</sub>	
17	$D_4$	$D_4$	$D_4$	P3 <sub>4</sub>	P3 <sub>4</sub>	I/O <sub>4</sub>	

Pin Name

	Pin Name							
Pin No.	Mode 1	Mode 3	Mode 5	Mode 6	Mode 7	PROM Mode Flash memory		
18	$D_{_5}$	$D_{s}$	$D_{s}$	P3₅	P3₅	I/O <sub>5</sub>		
19	D <sub>6</sub>	$D_6$	$D_6$	P3 <sub>6</sub>	P3 <sub>6</sub>	I/O <sub>6</sub>		
20	D <sub>7</sub>	D <sub>7</sub>	D <sub>7</sub>	P3,	P3,	I/O <sub>7</sub>		
21	V <sub>cc</sub>	V <sub>cc</sub>	V <sub>cc</sub>	V <sub>cc</sub>	V <sub>cc</sub>	V <sub>cc</sub>		
22	A <sub>o</sub>	A <sub>o</sub>	P1 <sub>0</sub> /A <sub>0</sub>	P1 <sub>0</sub>	P1 <sub>0</sub>	A <sub>o</sub>		
23	A <sub>1</sub>	A <sub>1</sub>	P1,/A,	P1,	P1,	A <sub>1</sub>		
24	A <sub>2</sub>	A <sub>2</sub>	P1 <sub>2</sub> /A <sub>2</sub>	P1 <sub>2</sub>	P1 <sub>2</sub>	A <sub>2</sub>		
25	A <sub>3</sub>	A <sub>3</sub>	P1 <sub>3</sub> /A <sub>3</sub>	P1 <sub>3</sub>	P1 <sub>3</sub>	A <sub>3</sub>		
26	A <sub>4</sub>	A <sub>4</sub>	P1 <sub>4</sub> /A <sub>4</sub>	P1 <sub>4</sub>	P1 <sub>4</sub>	A <sub>4</sub>		
27	A <sub>5</sub>	A <sub>5</sub>	P1 <sub>5</sub> /A <sub>5</sub>	P1₅	P1₅	A <sub>5</sub>		
28	A <sub>6</sub>	A <sub>6</sub>	P1 <sub>6</sub> /A <sub>6</sub>	P1 <sub>6</sub>	P1 <sub>6</sub>	A <sub>6</sub>		
29	A <sub>7</sub>	A <sub>7</sub>	P1 <sub>7</sub> /A <sub>7</sub>	P1,	P1,	A <sub>7</sub>		
30	V <sub>ss</sub>	V <sub>ss</sub>	V <sub>ss</sub>	V <sub>ss</sub>	V <sub>ss</sub>	V <sub>ss</sub>		
31	A <sub>8</sub>	A <sub>8</sub>	P2 <sub>0</sub> /A <sub>8</sub>	P2 <sub>0</sub>	P2 <sub>0</sub>	A <sub>8</sub>		
32	A <sub>9</sub>	A <sub>9</sub>	P2 <sub>1</sub> /A <sub>9</sub>	P2,	P2,	A <sub>9</sub>		
33	A <sub>10</sub>	A <sub>10</sub>	P2 <sub>2</sub> /A <sub>10</sub>	P2 <sub>2</sub>	P2 <sub>2</sub>	A <sub>10</sub>		
34	A <sub>11</sub>	A <sub>11</sub>	P2 <sub>3</sub> /A <sub>11</sub>	P2 <sub>3</sub>	P2 <sub>3</sub>	A <sub>11</sub>		
35	A <sub>12</sub>	A <sub>12</sub>	P2 <sub>4</sub> /A <sub>12</sub>	P2 <sub>4</sub>	P2 <sub>4</sub>	A <sub>12</sub>		
36	A <sub>13</sub>	A <sub>13</sub>	P2 <sub>5</sub> /A <sub>13</sub>	P2₅	P2₅	A <sub>13</sub>		
37	A <sub>14</sub>	A <sub>14</sub>	P2 <sub>6</sub> /A <sub>14</sub>	P2 <sub>6</sub>	P2 <sub>6</sub>	A <sub>14</sub>		
38	A <sub>15</sub>	A <sub>15</sub>	P2,/A <sub>15</sub>	P2,	P2,	A <sub>15</sub>		
39	A <sub>16</sub>	A <sub>16</sub>	P5 <sub>0</sub> /A <sub>16</sub>	P5 <sub>0</sub>	P5 <sub>0</sub>	A <sub>16</sub>		
40	A <sub>17</sub>	A <sub>17</sub>	P5 <sub>1</sub> /A <sub>17</sub>	P5 <sub>1</sub>	P5 <sub>1</sub>	V <sub>ss</sub>		
41	A <sub>18</sub>	A <sub>18</sub>	P5 <sub>2</sub> /A <sub>18</sub>	P5 <sub>2</sub>	P5 <sub>2</sub>	V <sub>ss</sub>		
42	A <sub>19</sub>	A <sub>19</sub>	P5 <sub>3</sub> /A <sub>19</sub>	P5₃	P5₃	V <sub>ss</sub>		
43	P6 <sub>0</sub> /WAIT	P6₀/WAIT	P6₀/WAIT	P6 <sub>0</sub>	P6 <sub>0</sub>	NC		
44	MD <sub>0</sub>	MD <sub>o</sub>	MD <sub>o</sub>	MD <sub>o</sub>	MD	V <sub>ss</sub>		
45	MD <sub>1</sub>	MD <sub>1</sub>	MD <sub>1</sub>	MD <sub>1</sub>	MD <sub>1</sub>	V <sub>ss</sub>		
46	¢	φ	φ	φ	φ	NC		

Pin Name						
Pin No.	Mode 1	Mode 3	Mode 5	Mode 6	Mode 7	PROM Mode Flash memory
47	STBY	STBY	STBY	STBY	STBY	V <sub>cc</sub>
48	RES	RES	RES	RES	RES	RES
49	NMI	NMI	NMI	NMI	NMI	V <sub>cc</sub>
50	V <sub>ss</sub>	V <sub>ss</sub>				
51	EXTAL	EXTAL	EXTAL	EXTAL	EXTAL	EXTAL
52	XTAL	XTAL	XTAL	XTAL	XTAL	XTAL
53	V <sub>cc</sub>	V <sub>cc</sub>				
54	AS	AS	AS	P6 <sub>3</sub>	P6 <sub>3</sub>	NC
55	RD	RD	RD	P6 <sub>4</sub>	P6 <sub>4</sub>	NC
56	WR	WR	WR	P6₅	P6 <sub>5</sub>	V <sub>cc</sub>
57	RESO/ FWE*	RESO/ FWE*	RESO/ FWE*	RESO/ FWE*	RESO/ FWE*	FWE
58	AV <sub>ss</sub>	V <sub>ss</sub>				
59	P7 <sub>0</sub> /AN <sub>0</sub>	NC				
60	P7 <sub>1</sub> /AN <sub>1</sub>	NC				
61	P7 <sub>2</sub> /AN <sub>2</sub>	NC				
62	P7 <sub>3</sub> /AN <sub>3</sub>	NC				
63	P7 <sub>4</sub> /AN <sub>4</sub>	P7₄/AN₄	P7₄/AN₄	P7₄/AN₄	P7 <sub>4</sub> /AN <sub>4</sub>	NC
64	P7₅/AN₅	P7₅/AN₅	P7₅/AN₅	P7₅/AN₅	P7 <sub>5</sub> /AN <sub>5</sub>	NC
65	P7 <sub>6</sub> /AN <sub>6</sub>	NC				
66	P7 <sub>7</sub> /AN <sub>7</sub>	NC				
67	AV <sub>cc</sub>	V <sub>cc</sub>				
68	P8 <sub>0</sub> /IRQ <sub>0</sub>	V <sub>ss</sub>				
69	P8,/IRQ	P8,/IRQ	P8 <sub>1</sub> /IRQ <sub>1</sub>	P8,/IRQ	P8 <sub>1</sub> /IRQ <sub>1</sub>	V <sub>ss</sub>
70	P9 <sub>1</sub> /TxD <sub>1</sub>	NC				
71	P9 <sub>3</sub> /RxD <sub>1</sub>	NC				
72	P9₅/SCK₁/ IRQ₅	$\frac{P9_{5}}{IRQ_{5}}$	P9₅/SCK₁/ IRQ₅	P9₅/SCK₁/ IRQ₅	P9₅/SCK₁/ IRQ₅	V <sub>cc</sub>
73	PA,/TP,/ TCLKA	PA,/TP,/ TCLKA	PA /TP / TCLKA	PA /TP / TCLKA	PA,/TP,/ TCLKA	CE

	Pin Name								
Pin No.	Mode 1	Mode 3	Mode 5	Mode 6	Mode 7	PROM Mode Flash memory			
74	PA₁/TP₁/ TCLKB	PA <sub>1</sub> /TP <sub>1</sub> / TCLKB	PA,/TP,/ TCLKB	PA <sub>1</sub> /TP <sub>1</sub> / TCLKB	PA₁/TP₁/ TCLKB	ŌĒ			
75	PA_/TP_/ TIOCA_/ TCLKC	PA2/TP2/ TIOCA0/ TCLKC	PA <sub>2</sub> /TP <sub>2</sub> / TIOCA <sub>0</sub> / TCLKC	PA2/TP2/ TIOCA7/ TCLKC	PA2/TP2/ TIOCA0/ TCLKC	WE			
76	PA <sub>3</sub> /TP <sub>3</sub> / TIOCB <sub>0</sub> / TCLKD	PA₃/TP₃/ TIOCB₀/ TCLKD	PA <sub>3</sub> /TP <sub>3</sub> / TIOCB <sub>6</sub> / TCLKD	PA,/TP,/ TIOCB,/ TCLKD	PA <sub>3</sub> /TP <sub>3</sub> / TIOCB <sub>0</sub> / TCLKD	NC			
77	PA₄/TP₄/ TIOCA₁	PA₄/TP₄/ TIOCA₁/A₂₃	PA₄/TP₄/ TIOCA₁	PA₄/TP₄/ TIOCA₁	PA₄/TP₄/ TIOCA₁	NC			
78	PA <sub>5</sub> /TP <sub>5</sub> / TIOCB <sub>1</sub>	PA <sub>5</sub> /TP <sub>5</sub> / TIOCB <sub>1</sub> /A <sub>22</sub>	PA <sub>5</sub> /TP <sub>5</sub> / TIOCB <sub>1</sub>	PA₅/TP₅/ TIOCB₁	PA <sub>5</sub> /TP <sub>5</sub> / TIOCB <sub>1</sub>	NC			
79	PA <sub>6</sub> /TP <sub>6</sub> / TIOCA <sub>2</sub>	PA <sub>6</sub> /TP <sub>6</sub> / TIOCA <sub>2</sub> /A <sub>21</sub>	PA <sub>6</sub> /TP <sub>6</sub> / TIOCA <sub>2</sub>	PA <sub>6</sub> /TP <sub>6</sub> / TIOCA <sub>2</sub>	PA <sub>6</sub> /TP <sub>6</sub> / TIOCA <sub>2</sub>	NC			
80	PA <sub>7</sub> /TP <sub>7</sub> / TIOCB <sub>2</sub>	A <sub>20</sub>	PA <sub>7</sub> /TP <sub>7</sub> / TIOCB <sub>2</sub>	PA <sub>7</sub> /TP <sub>7</sub> / TIOCB <sub>2</sub>	PA <sub>7</sub> /TP <sub>7</sub> / TIOCB <sub>2</sub>	NC			

Notes: Pins marked NC should be left unconnected. For details about PROM mode see section 15, ROM.

> \* Mask ROM: RESO Flash Memory: FWE

#### **1.4 Pin Functions**

Table 1.3 summarizes the pin functions.

#### Table 1.3Pin Functions

Туре	Symbol	Pin No.	I/O	Name	and Fu	nction		
Power	V <sub>cc</sub>	21, 53	Input	<b>Power:</b> For connection to the power supply. Connect all $V_{cc}$ pins to the system power supply.				
	V <sub>ss</sub>	12, 30, 50	Input	<b>Ground:</b> For connection to ground (0 V). Connect all $V_{ss}$ pins to the 0-V system power supply.				
Clock	XTAL	52	Input	For connection to a crystal resonator For examples of crystal resonator and external clock input, see section 16, Clock Pulse Generator.				
	EXTAL	51	Input	For connection to a crystal resonator or input of an external clock signal. For examples of crystal resonator and external clock input, see section 16, Clock Pulse Generator. System clock: Supplies the system clock to external devices				
	φ	46	Output					
Operating mode control	MD <sub>2</sub> , MD <sub>1</sub> , MD <sub>0</sub>	7, 45, 44	Input	<b>Mode 2 to mode 0:</b> For setting the operating mode, as follows. These pins should not be changed during operation.				
				$MD_{2}$	MD₁	MD₀	Operating Mode	
				0	0	0	_	
				0	0	1	Mode 1	
				0	1	0	_	
				0	1	1	Mode 3	
				1	0	0	_	
				1	0	1	Mode 5	
				1	1	0	Mode 6	
				1	1	1	Mode 7	

Туре	Symbol	Pin No.	I/O	Name and Function
System control	RES	48	Input	<b>Reset input:</b> When driven low, this pin resets the chip
	RESO/ FWE	57	Output/ Input	<b>Reset output</b> (Mask ROM version): Outputs WDT-generated reset signal to an external device.
				Write enable signal (F-ZTAT version): Flash memory write control signal.
	STBY	47	Input	<b>Standby:</b> When driven low, this pin forces a transition to hardware standby mode
Interrupts	NMI	49	Input	Nonmaskable interrupt: Requests a nonmaskable interrupt
	$\frac{\overline{IRQ}_{F}}{\overline{IRQ}_{I}}, \frac{\overline{IRQ}_{F}}{\overline{IRQ}_{I}}$	72, 11, 69, 68	Input	Interrupt request 5, 4, 1, 0: Maskable interrupt request pins
Address bus	$A_{23}$ to $A_{20}$ , $A_{19}$ to $A_{8}$ , $A_{7}$ to $A_{0}$	77 to 80, 42 to 31, 29 to 22	Output	Address bus: Outputs address signals
Data bus	$D_7$ to $D_0$	20 to 13	Input/ output	Data bus: Bidirectional data bus
Bus control	ĀS	54	Output	Address strobe: Goes low to indicate valid address output on the address bus
	RD	55	Output	<b>Read:</b> Goes low to indicate reading from the external address space.
	WR	56	Output	Write: Goes low to indicate writing to the external address space indicates valid data on the data bus.
	WAIT	43	Input	<b>Wait:</b> Requests insertion of wait states in bus cycles during access to the external address space.
16-bit integrated	TCLKD to TCLKA	76 to 73	Input	Clock input A to D: External clock inputs
timer unit (ITU)	$TIOCA_4$ to $TIOCA_0$	3, 1, 79, 77, 75	Input/ Output	Input capture/output compare A4 to A0: GRA4 to GRA0 output compare or input capture, or PWM output
	$TIOCB_4$ to $TIOCB_0$	4, 2, 80, 78, 76	Input/ output	Input capture/output compare B4 to B0 GRB4 to GRB0 output compare or input capture, or PWM output
	TOCXA <sub>4</sub>	5	Output	Output compare XA4: PWM output
	TOCXB <sub>4</sub>	6	Output	Output compare XB4: PWM output

Туре	Symbol	Pin No.	I/O	Name and Function
Programmable timing pattern controller (TPC)	$\begin{array}{c} TP_{15},\\ TP_{13} \text{ to } TP_{0} \end{array}$	8, 6 to 1 80 to 73	Output	TPC output 15, 13 to 0 : Pulse output
Serial com- munication	TxD₁, TxD₀	70, 9	Output	Transmit data:(channels 0 and 1): SCI data output
interface (SCI)	RxD <sub>1</sub> , RxD <sub>0</sub>	71, 10	Input	Receive data: (channels 0 and 1): SCI data input
	SCK <sub>1</sub> , SCK <sub>0</sub>	72, 11	Input/ output	Serial clock:(channels 0 and 1): SCI clock input/output
A/D	$AN_7$ to $AN_0$	66 to 59	Input	Analog 7 to 0: Analog input pins
converter	ADTRG	8	Input	<b>A/D trigger:</b> External trigger input for starting A/D conversion
	AV <sub>cc</sub>	67	Input	Power supply pin and reference voltage input pin for the A/D converter. Connect to the system power supply when not using the A/D converter.
	AV <sub>ss</sub>	58	Input	Ground pin for the A/D converter. Connect to system power-supply (0 V).
I/O ports	$P1_7$ to $P1_0$	29 to 22	Input/ output	<b>Port 1:</b> Eight input/output pins. The direction of each pin can be selected in the port 1 data direction register (P1DDR).
	$P2_7$ to $P2_0$	38 to 31	Input/ output	<b>Port 2:</b> Eight input/output pins. The direction of each pin can be selected in the port 2 data direction register (P2DDR).
	$P3_7$ to $P3_0$	20 to 13	Input/ output	<b>Port 3:</b> Eight input/output pins. The direction of each pin can be selected in the port 3 data direction register (P3DDR).
	$P5_{3}$ to $P5_{0}$	42 to 39	Input/ output	<b>Port 5:</b> Four input/output pins. The direction of each pin can be selected in the port 5 data direction register (P5DDR).
	$P6_{5}$ to $P6_{3}$ , $P6_{0}$	56 to 54, 43	Input/ output	<b>Port 6:</b> Four input/output pins. The direction of each pin can be selected in the port 6 data direction register (P6DDR).
	$P7_7$ to $P7_0$	66 to 59	Input	Port 7: Eight input pins
	P8 <sub>1</sub> , P8 <sub>0</sub>	69, 68	Input/ output	<b>Port 8:</b> Two input/output pins. The direction of each pin can be selected in the port 8 data direction register (P8DDR).

Section 1 Overview

Туре	Symbol	Pin No.	I/O	Name and Function
I/O ports	P9₅ to P9₀	72, 11 71, 10 70, 9	Input/ output	<b>Port 9:</b> Six input/output pins. The direction of each pin can be selected in the port 9 data direction register (P9DDR).
	PA <sub>7</sub> to PA <sub>0</sub>	80 to 73	Input/ output	<b>Port A:</b> Eight input/output pins. The direction of each pin can be selected in the port A data direction register (PADDR).
	$PB_7, PB_5$ to $PB_0$	8, 6to1	Input/ output	<b>Port B:</b> Seven input/output pins. The direction of each pin can be selected in the port B data direction register (PBDDR).

# Section 2 CPU

# 2.1 Overview

The H8/300H CPU is a high-speed central processing unit with an internal 32-bit architecture that is upward-compatible with the H8/300 CPU. The H8/300H CPU has sixteen 16-bit general registers, can address a 16-Mbyte linear address space, and is ideal for realtime control.

#### 2.1.1 Features

The H8/300H CPU has the following features.

- Upward compatibility with H8/300 CPU Can execute H8/300 series object programs without alteration
- General-register architecture Sixteen 16-bit general registers (also usable as sixteen 8-bit registers or eight 32-bit registers)
- Sixty-two basic instructions
  - 8/16/32-bit arithmetic and logic instructions
  - Multiply and divide instructions
  - Powerful bit-manipulation instructions
- Eight addressing modes
  - Register direct [Rn]
  - Register indirect [@ERn]
  - Register indirect with displacement [@(d:16, ERn) or @(d:24, ERn)]
  - Register indirect with post-increment or pre-decrement [@ERn+ or @-ERn]
  - Absolute address [@aa:8, @aa:16, or @aa:24]
  - Immediate [#xx:8, #xx:16, or #xx:32]
  - Program-counter relative [@(d:8, PC) or @(d:16, PC)]
  - Memory indirect [@@aa:8]
- 16-Mbyte linear address space
- High-speed operation
  - All frequently-used instructions execute in two to four states
  - Maximum clock frequency: 18 MHz
  - 8/16/32-bit register-register add/subtract: 111 ns
  - 8 × 8-bit register-register multiply: 778 ns
  - $16 \div 8$ -bit register-register divide: 778 ns

# Renesas

- 16  $\times$  16-bit register-register multiply: 1222 ns
- $-32 \div 16$ -bit register-register divide: 1222 ns
- Two CPU operating modes
  - Normal mode
  - Advanced mode
- Low-power mode

Transition to power-down state by SLEEP instruction

# 2.1.2 Differences from H8/300 CPU

In comparison to the H8/300 CPU, the H8/300H has the following enhancements.

• More general registers

Eight 16-bit registers have been added.

- Expanded address space
  - Advanced mode supports a maximum 16-Mbyte address space.
  - Normal mode supports the same 64-kbyte address space as the H8/300 CPU.
- Enhanced addressing

The addressing modes have been enhanced to make effective use of the 16-Mbyte address space.

- Enhanced instructions
  - Data transfer, arithmetic, and logic instructions can operate on 32-bit data.
  - Signed multiply/divide instructions and other instructions have been added.

# 2.2 CPU Operating Modes

The H8/300H CPU has two operating modes: normal and advanced. Normal mode supports a maximum 64-kbyte address space. Advanced mode supports up to 16 Mbytes. See figure 2.1.

Unless specified otherwise, all descriptions in this manual refer to advanced mode.

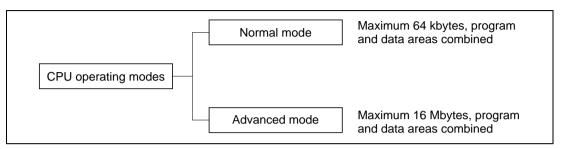


Figure 2.1 CPU Operating Modes

# 2.3 Address Space

The maximum address space of the H8/300H CPU is 16 Mbytes. This LSI allows selection of a normal mode and advanced mode 1-Mbyte mode or 16-Mbyte mode for the address space depending on the MCU operation mode. Figure 2.2 shows the address ranges of the H8/3039 Group. For further details see section 3.6, Memory Map in Each Operating Mode.

The 1-Mbyte operating mode uses 20-bit addressing. The upper 4 bits of effective addresses are ignored.

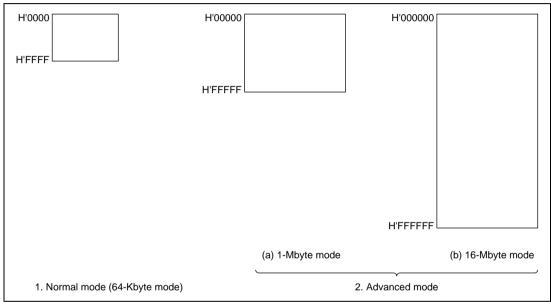


Figure 2.2 Memory Map

# 2.4 Register Configuration

#### 2.4.1 Overview

The H8/300H CPU has the internal registers shown in figure 2.3. There are two types of registers: general registers and control registers.

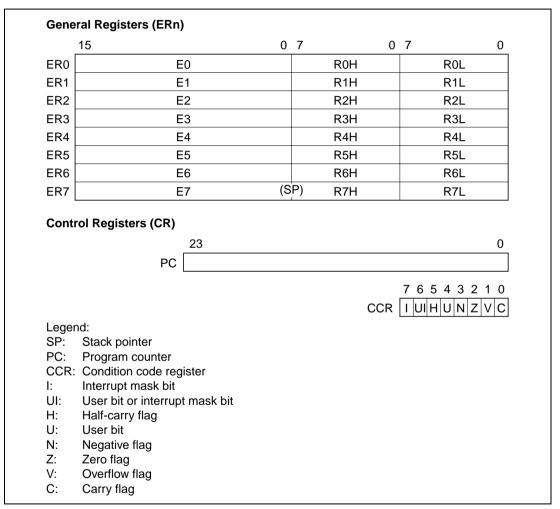


Figure 2.3 CPU Registers

### 2.4.2 General Registers

The H8/300H CPU has eight 32-bit general registers. These general registers are all functionally alike and can be used without distinction between data registers and address registers. When a general register is used as a data register, it can be accessed as a 32-bit, 16-bit, or 8-bit register. When the general registers are used as 32-bit registers or as address registers, they are designated by the letters ER (ER0 to ER7).

The ER registers divide into 16-bit general registers designated by the letters E (E0 to E7) and R (R0 to R7). These registers are functionally equivalent, providing a maximum sixteen 16-bit registers. The E registers (E0 to E7) are also referred to as extended registers.

The R registers divide into 8-bit general registers designated by the letters RH (R0H to R7H) and RL (R0L to R7L). These registers are functionally equivalent, providing a maximum sixteen 8-bit registers.

Figure 2.4 illustrates the usage of the general registers. The usage of each register can be selected independently.

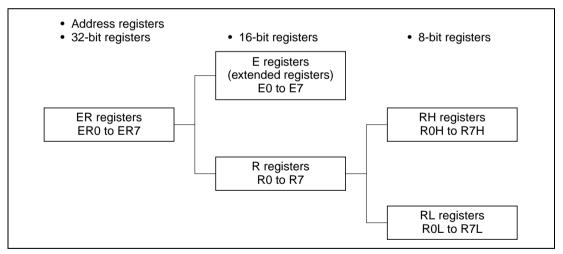


Figure 2.4 Usage of General Registers

General register ER7 has the function of stack pointer (SP) in addition to its general-register function, and is used implicitly in exception handling and subroutine calls. Figure 2.5 shows the stack.

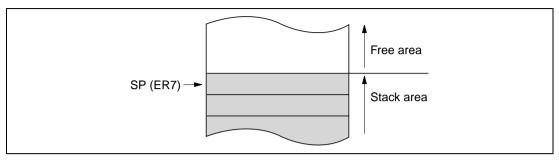


Figure 2.5 Stack

#### 2.4.3 Control Registers

The control registers are the 24-bit program counter (PC) and the 8-bit condition code register (CCR).

#### **Program Counter (PC)**

This 24-bit counter indicates the address of the next instruction the CPU will execute. The length of all CPU instructions is 2 bytes (one word) or a multiple of 2 bytes, so the least significant PC bit is ignored. When an instruction is fetched, the least significant PC bit is regarded as 0.

#### Condition Code Register (CCR)

This 8-bit register contains internal CPU status information, including the interrupt mask bit (I) and half-carry (H), negative (N), zero (Z), overflow (V), and carry (C) flags.

**Bit 7—Interrupt Mask Bit (I):** Masks interrupts other than NMI when set to 1. NMI is accepted regardless of the I bit setting. The I bit is set to 1 at the start of an exception-handling sequence.

**Bit 6—User Bit or Interrupt Mask Bit (UI):** Can be written and read by software using the LDC, STC, ANDC, ORC, and XORC instructions. This bit can also be used as an interrupt mask bit. For details see section 5, Interrupt Controller.

**Bit 5—Half-Carry Flag (H):** When the ADD.B, ADDX.B, SUB.B, SUBX.B, CMP.B, or NEG.B instruction is executed, this flag is set to 1 if there is a carry or borrow at bit 3, and cleared to 0 otherwise. When the ADD.W, SUB.W, CMP.W, or NEG.W instruction is executed, the H flag is set to 1 if there is a carry or borrow at bit 11, and cleared to 0 otherwise. When the ADD.L, SUB.L, CMP.L, or NEG.L instruction is executed, the H flag is set to 1 if there is a carry or borrow at bit 27, and cleared to 0 otherwise.

**Bit 4—User Bit (U):** Can be written and read by software using the LDC, STC, ANDC, ORC, and XORC instructions.

Bit 3—Negative Flag (N): Indicates the most significant bit (sign bit) of data.

Bit 2—Zero Flag (Z): Set to 1 to indicate zero data, and cleared to 0 to indicate non-zero data.

**Bit 1—Overflow Flag (V):** Set to 1 when an arithmetic overflow occurs, and cleared to 0 at other times.

Bit 0—Carry Flag (C): Set to 1 when a carry occurs, and cleared to 0 otherwise. Used by:

- Add instructions, to indicate a carry
- Subtract instructions, to indicate a borrow
- Shift and rotate instructions, to store the value shifted out of the end bit

The carry flag is also used as a bit accumulator by bit manipulation instructions.

Some instructions leave flag bits unchanged. Operations can be performed on CCR by the LDC, STC, ANDC, ORC, and XORC instructions. The N, Z, V, and C flags are used by conditional branch (Bcc) instructions.

For the action of each instruction on the flag bits, see appendix A.1, Instruction List. For the I and UI bits, see section 5, Interrupt Controller.

# 2.4.4 Initial CPU Register Values

In reset exception handling, PC is initialized to a value loaded from the vector table, and the I bit in CCR is set to 1. The other CCR bits and the general registers are not initialized. In particular, the stack pointer (ER7) is not initialized. The stack pointer must therefore be initialized by an MOV.L instruction executed immediately after a reset.

# 2.5 Data Formats

The H8/300H CPU can process 1-bit, 4-bit (BCD), 8-bit (byte), 16-bit (word), and 32-bit (longword) data. Bit-manipulation instructions operate on 1-bit data by accessing bit n (n = 0, 1, 2, ..., 7) of byte operand data. The DAA and DAS decimal-adjust instructions treat byte data as two digits of 4-bit BCD data.

#### 2.5.1 General Register Data Formats

Figures 2.6 and 2.7 show the data formats in general registers.

Data Type	General Register	Data Format
1-bit data	RnH	7 0 7 6 5 4 3 2 1 0 Don't care
1-bit data	RnL	7         0           Don't care         7         6         5         4         3         2         1         0
4-bit BCD data	RnH	7 4 3 0 Upper digit Lower digit Don't care
4-bit BCD data	RnL	7     4     3     0       Don't care     Upper digit     Lower digit
Byte data	RnH	7 0 Don't care
Byte data	RnL	7 0 Don't care MSB LSB
Legend: RnH: General regist RnL: General regist		

Figure 2.6 General Register Data Formats

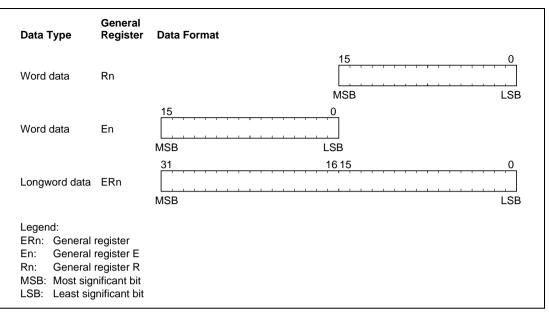


Figure 2.7 General Register Data Formats

#### 2.5.2 Memory Data Formats

Figure 2.8 shows the data formats on memory. The H8/300H CPU can access word data and longword data on memory, but word or longword data must begin at an even address. If an attempt is made to access word or longword data at an odd address, no address error occurs but the least significant bit of the address is regarded as 0, so the access starts at the preceding address. This also applies to instruction fetches.

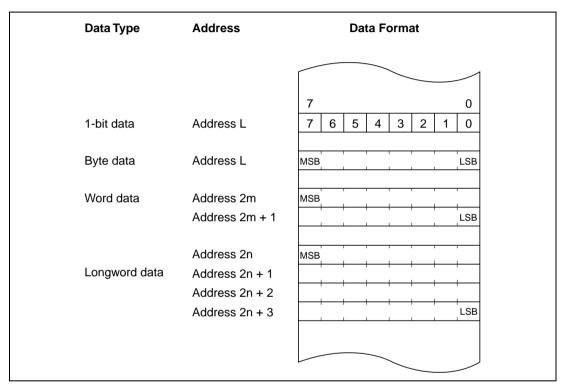


Figure 2.8 Memory Data Formats

When ER7 (SP) is used as an address register to access the stack, the operand size should be word size or longword size.

#### Renesas

# 2.6 Instruction Set

# 2.6.1 Instruction Set Overview

The H8/300H CPU has 62 types of instructions, which are classified as shown in table 2.1.

Function	Instruction	Types
Data transfer	MOV, PUSH* <sup>1</sup> , POP* <sup>1</sup> , MOVTPE* <sup>2</sup> , MOVFPE* <sup>2</sup>	3
Arithmetic operations	ADD, SUB, ADDX, SUBX, INC, DEC, ADDS, SUBS, DAA, DAS, MULXU, MULXS, DIVXU, DIVXS, CMP, NEG, EXTS, EXTU	18
Logic operations	AND, OR, XOR, NOT	4
Shift operations	SHAL, SHAR, SHLL, SHLR, ROTL, ROTR, ROTXL, ROTXR	8
Bit manipulation	BSET, BCLR, BNOT, BTST, BAND, BIAND, BOR, BIOR, BXOR, BIXOR, BLD, BILD, BST, BIST	14
Branch	Bcc* <sup>3</sup> , JMP, BSR, JSR, RTS	5
System control	TRAPA, RTE, SLEEP, LDC, STC, ANDC, ORC, XORC, NOP	9
Block data transfer	EEPMOV	1

 Table 2.1
 Instruction Classification

Total 62 types

- Notes: 1. POP.W Rn is identical to MOV.W @SP+, Rn. PUSH.W Rn is identical to MOV.W Rn, @-SP. POP.L ERn is identical to MOV.L @SP+, Rn. PUSH.L ERn is identical to MOV.L Rn, @-SP.
  - 2. These instructions are not available on the H8/3039 Group.
  - 3. Bcc is a generic branching instruction.

#### 2.6.2 Instructions and Addressing Modes

Table 2.2 indicates the instructions available in the H8/300H CPU.

 Table 2.2
 Instructions and Addressing Modes

						1	Addres	ssing	Mode	S				
Function	Instruction	XX#	Rn	@ERn	@(d:16,ERn)	@(d:24, ERn)	@ERn+/@-ERn	@aa:8	@aa:16	@aa:24	@(d:8, PC)	@(d:16,PC)	@ @aa:8	Implied
Data	MOV	BWL	BWL	BWL	BWL	BWL	BWL	В	BWL	BWL	_			—
transfer	POP, PUSH	—	_	_	-	_	—	_	_	_	_	_	_	WL
	MOVFPE, MOVTPE	-	—	—	—			_	В		—	—	—	—
Arithmetic	ADD, CMP	BWL	BWL	_	-	_	—	_	_	_	_	_	_	—
operations	SUB	WL	BWL	—	—	—	—	_	—	—	—	—	_	—
	ADDX, SUBX	В	В			—	—	—		_	—			—
	ADDS, SUBS	-	L			_	—	_		_	_			—
	INC, DEC	_	BWL				_			_	_			—
	DAA, DAS	_	В		—			I	_	_	_		_	—
	MULXU, MULXS, DIVXU, DIVXS	_	BW	_	_				_	Ι	—	_	_	—
	NEG	—	BWL	—	—		—		—		—	—	_	—
	EXTU, EXTS	—	WL	—	—		_		—		—	—	—	
Logic operations	AND, OR, XOR	BWL	BWL	—	—	_	_	_	—	—	_	—	—	—
	NOT	—	BWL	—	—		_		—	_	_	—	—	
Shift instruc	Shift instructions		BWL	—	—	—	—	_	—	—	_	—	—	—
Bit manipula	Bit manipulation		В	В	—	—	—	В	—	—	_	—	—	—
Branch	Bcc, BSR	—		_	—	—	_	_	—		0	0		—
	JMP, JSR	—	_	0	—	—	—		—	0	—	—	0	—
	RTS	_	—	—	—	_	—	_	_		_	—	—	0

		Addressing Modes												
Function	Instruction	xx#	Rn	@ERn	@(d:16,ERn)	@(d:24, ERn)	@ERn+/@-ERn	@aa:8	@aa:16	@aa:24	@(d:8, PC)	@(d:16,PC)	@ @aa:8	Implied
System	TRAPA		_		_		—		_	—	_		_	0
control	RTE	—	—	_	—	_	_	_	—	_	—	_	_	0
	SLEEP	—	—	—	—	—		—	—		—	—	—	0
	LDC	В	В	W	W	W	W	—	W	W	—	—	—	—
	STC	—	В	W	W	W	W	—	W	W	—	—	—	—
	ANDC, ORC, XORC	В												
	NOP	_	_	_	—	_	_	_	—	_	—	_	_	0
Block data t	ransfer	—	_	—	_	—		—	_		_	—	—	BW

Legend:

B: Byte

W: Word

L: Longword

#### 2.6.3 Tables of Instructions Classified by Function

Tables 2.3 to 2.10 summarize the instructions in each functional category. The operation notation used in these tables is defined as follows.

#### **Operation Notation**

Rd	General register (destination)*
Rs	General register (source)*
Rn	General register*
ERn	General register (32-bit register or address register)
(EAd)	Destination operand
(EAs)	Source operand
CCR	Condition code register
N	N (negative) flag of CCR
Z	Z (zero) flag of CCR
V	V (overflow) flag of CCR
С	C (carry) flag of CCR
PC	Program counter
SP	Stack pointer
#IMM	Immediate data
disp	Displacement
+	Addition
_	Subtraction
×	Multiplication
÷	Division
^	Logical AND
v	Logical OR
$\oplus$	Exclusive logical OR
$\rightarrow$	Move
7	NOT (logical complement)
:3/:8/:16/:24	3-, 8-, 16-, or 24-bit length

to R7, E0 to E7), and 32-bit data or address registers (ER0 to ER7).

Instructio	n Size*	Function					
MOV B/W/L		$(EAs) \rightarrow Rd,  Rs \rightarrow (EAd)$					
		Moves data between two general registers or between a general register and memory, or moves immediate data to a general register.					
MOVFPE	В	$(EAs) \to Rd$					
		Cannot be used in the H8/3039 Group.					
MOVTPE	В	Rs  ightarrow (EAs)					
		Cannot be used in the H8/3039 Group.					
POP	W/L	$@SP+ \rightarrow Rn$					
		Pops a general register from the stack. POP.W Rn is identical to MOV.W @SP+, Rn. Similarly, POP.L ERn is identical to MOV.L @SP+, ERn.					
PUSH	W/L	$Rn \rightarrow @-SP$					
		Pushes a general register onto the stack. PUSH.W Rn is identical to MOV.W Rn, @–SP. Similarly, PUSH.L ERn is identical to MOV.L ERn, @–SP.					
Note: *	Size refers to the	ne operand size.					
	B: Byte						
	W: Word						
	L: Longword						

Instructio	n Size*	Function
ADD,	B/W/L	$Rd \pm Rs \to Rd, Rd \pm \#IMM \to Rd$
SUB		Performs addition or subtraction on data in two general registers, or on immediate data and data in a general register. (Immediate byte data cannot be subtracted from data in a general register. Use the SUBX or ADD instruction.)
ADDX,	В	$Rd \pm Rs \pm C \to Rd,  Rd \pm \#IMM \pm C \to Rd$
SUBX		Performs addition or subtraction with carry on data in two general registers, or on immediate data and data in a general register.
INC,	B/W/L	$Rd \pm 1 \rightarrow Rd, Rd \pm 2 \rightarrow Rd$
DEC		Increments or decrements a general register by 1 or 2. (Byte operands can be incremented or decremented by 1 only.)
ADDS,	L	$Rd \pm 1 \to Rd,  Rd \pm 2 \to Rd,  Rd \pm 4 \to Rd$
SUBS		Adds or subtracts the value 1, 2, or 4 to or from data in a 32-bit register.
DAA,	В	Rd decimal adjust $\rightarrow$ Rd
DAS		Decimal-adjusts an addition or subtraction result in a general register by referring to CCR to produce 4-bit BCD data.
MULXU	B/W	Rd  imes Rs  o Rd
		Performs unsigned multiplication on data in two general registers: either 8 bits $\times$ 8 bits $\rightarrow$ 16 bits or 16 bits $\times$ 16 bits $\rightarrow$ 32 bits.
MULXS	B/W	Rd  imes Rs  o Rd
		Performs signed multiplication on data in two general registers: either 8 bits $\times$ 8 bits $\rightarrow$ 16 bits or 16 bits $\times$ 16 bits $\rightarrow$ 32 bits.
Note: *	Size refers to	o the operand size.
	B: Byte	
	W: Word	
	I. Longwo	rd

L: Longword

#### Section 2 CPU

Instructio	on Size*	Function
DIVXU	B/W	$Rd \div Rs \to Rd$
		Performs unsigned division on data in two general registers: either 16 bits $\div$ 8 bits $\rightarrow$ 8-bit quotient and 8-bit remainder or 32 bits $\div$ 16 bits $\rightarrow$ 16-bit quotient and 16-bit remainder.
DIVXS	B/W	$Rd \div Rs \to Rd$
		Performs signed division on data in two general registers: either 16 bits $\div$ 8 bits $\rightarrow$ 8-bit quotient and 8-bit remainder, or 32 bits $\div$ 16 bits $\rightarrow$ 16-bit quotient and 16-bit remainder.
CMP	B/W/L	Rd – Rs, Rd – #IMM
		Compares data in a general register with data in another general register or with immediate data, and sets CCR according to the result.
NEG	B/W/L	$0 - Rd \rightarrow Rd$
		Takes the two's complement (arithmetic complement) of data in a general register.
EXTS	W/L	Rd (sign extension) $\rightarrow$ Rd
		Extends byte data in the lower 8 bits of a 16-bit register to word data, or extends word data in the lower 16 bits of a 32-bit register to longword data, by extending the sign bit.
EXTU	W/L	Rd (zero extension) $\rightarrow$ Rd
		Extends byte data in the lower 8 bits of a 16-bit register to word data, or extends word data in the lower 16 bits of a 32-bit register to longword data, by padding with zeros.
Note: *	Size refers to the	ne operand size.
	B: Byte	
	W: Word	
	I. I. and an end of the second	

L: Longword

Instruction	n Size*	Function		
AND	B/W/L	$Rd \land Rs \to Rd,  Rd \land \#IMM \to Rd$		
		Performs a logical AND operation on a general register and another general register or immediate data.		
OR	B/W/L	$Rd \lor Rs \to Rd,  Rd \lor \#IMM \to Rd$		
		Performs a logical OR operation on a general register and another general register or immediate data.		
XOR	B/W/L	$Rd \oplus Rs \to Rd,  Rd \oplus \#IMM \to Rd$		
		Performs a logical exclusive OR operation on a general register and another general register or immediate data.		
NOT	B/W/L	$\neg \operatorname{Rd} \rightarrow \operatorname{Rd}$		
		Takes the one's complement of general register contents.		
Note: *	Size refers to	the operand size.		
	B: Byte			
	W: Word			

#### Table 2.5 Logic Operation Instructions

L: Longword

#### Table 2.6Shift Instructions

Instructio	n Size*	Function
SHAL,	B/W/L	$Rd (shift) \rightarrow Rd$
SHAR		Performs an arithmetic shift on general register contents.
SHLL,	B/W/L	$Rd (shift) \rightarrow Rd$
SHLR		Performs a logical shift on general register contents.
ROTL,	B/W/L	$Rd \text{ (rotate)} \rightarrow Rd$
ROTR		Rotates general register contents.
ROTXL,	B/W/L	$Rd \text{ (rotate)} \rightarrow Rd$
ROTXR		Rotates general register contents through the carry bit.
Note: * Size refers to the operand size.		the operand size.
	B: Byte	
	W: Word	

L: Longword

1 abic 2.7	Dit Manipu			
Instruction	Size*	Function		
BSET	В	$1 \rightarrow (\text{ of })$		
		Sets a specified bit in a general register or memory operand to 1. The bit number is specified by 3-bit immediate data or the lower 3 bits of a general register.		
BCLR	В	$0 \rightarrow (\text{ of })$		
		Clears a specified bit in a general register or memory operand to 0. The bit number is specified by 3-bit immediate data or the lower 3 bits of a general register.		
BNOT	В	¬ ( <bit-no.> of <ead>) → (<bit-no.> of <ead>)</ead></bit-no.></ead></bit-no.>		
		Inverts a specified bit in a general register or memory operand. The bit number is specified by 3-bit immediate data or the lower 3 bits of a general register.		
BTST	В	$\neg$ ( <bit-no.> of <ead>) <math>\rightarrow</math> Z</ead></bit-no.>		
		Tests a specified bit in a general register or memory operand and sets or clears the Z flag accordingly. The bit number is specified by 3-bit immediate data or the lower 3 bits of a general register.		
BAND	В	$C \land (<\!bit\text{-}No.\!> of <\!\mathsf{EAd\!\!>}) \to C$		
		ANDs the carry flag with a specified bit in a general register or memory operand and stores the result in the carry flag.		
BIAND	В	$C \land [\neg (<\!bit\!-\!No.\!> of <\!\mathsf{E\!Ad\!\!>)] \!\rightarrow \!C}$		
		ANDs the carry flag with the inverse of a specified bit in a general register or memory operand and stores the result in the carry flag.		
		The bit number is specified by 3-bit immediate data.		
BOR	В	$C \lor (<\!bit\!-\!No.\!> of <\!\mathsf{EAd\!\!>}) \to C$		
		ORs the carry flag with a specified bit in a general register or memory operand and stores the result in the carry flag.		
BIOR	В	$C \lor [\neg (<\!bit-No.\!> of <\!\mathsf{EAd\!\!>)}] \to C$		
		ORs the carry flag with the inverse of a specified bit in a general register or memory operand and stores the result in the carry flag.		
		The bit number is specified by 3-bit immediate data.		

#### Table 2.7 Bit Manipulation Instructions

Instruction	Size*	Function
BXOR	В	$C \oplus (<\!bit\!-\!No.\!> of <\!EAd\!>) \to C$
		Exclusive-ORs the carry flag with a specified bit in a general register or memory operand and stores the result in the carry flag.
BIXOR	В	$C \oplus [\neg ( of )] \to C$
		Exclusive-ORs the carry flag with the inverse of a specified bit in a general register or memory operand and stores the result in the carry flag.
		The bit number is specified by 3-bit immediate data.
BLD	В	( <bit-no.> of <ead>) <math>\rightarrow</math> C</ead></bit-no.>
		Transfers a specified bit in a general register or memory operand to the carry flag.
BILD	В	$\neg \text{ ( of )} \rightarrow C$
		Transfers the inverse of a specified bit in a general register or memory operand to the carry flag.
		The bit number is specified by 3-bit immediate data.
BST	В	$C \rightarrow (\text{ of })$
		Transfers the carry flag value to a specified bit in a general register or memory operand.
BIST	В	$C \rightarrow \neg$ ( <bit-no.> of <ead>)</ead></bit-no.>
		Transfers the inverse of the carry flag value to a specified bit in a general register or memory operand.
		The bit number is specified by 3-bit immediate data.

B: Byte

#### Table 2.8Branching Instructions

Instruction	Size	Function		
Bcc	—		pecified address if a specified cor tions are listed below.	ndition is true. The
		Mnemonic	Description	Condition
		BRA (BT)	Always (true)	Always
		BRN (BF)	Never (false)	Never
		BHI	High	$C \lor Z = 0$
		BLS	Low or same	C ∨ Z = 1
		Bcc (BHS)	Carry clear (high or same)	C = 0
		BCS (BLO)	Carry set (low)	C = 1
		BNE	Not equal	Z = 0
		BEQ	Equal	Z = 1
		BVC	Overflow clear	V = 0
		BVS	Overflow set	V = 1
		BPL	Plus	N = 0
		BMI	Minus	N = 1
		BGE	Greater or equal	$N \oplus V = 0$
		BLT	Less than	N ⊕ V = 1
		BGT	Greater than	$Z \lor (N \oplus V) = 0$
		BLE	Less or equal	$Z \lor (N \oplus V) = 1$
JMP	_	Branches uncor	ditionally to a specified address	
BSR	—	Branches to a se	ubroutine at a specified address	
JSR	_	Branches to a subroutine at a specified address		
RTS	_	Returns from a subroutine		

Instruction	n Size*	Function
TRAPA		Starts trap-instruction exception handling
RTE	_	Returns from an exception-handling routine
SLEEP	_	Causes a transition to the power-down state
LDC	B/W	$(EAs) \to CCR$
		Moves the source operand contents to the condition code register. The condition code register size is one byte, but in transfer from memory, data is read by word access.
STC	B/W	$CCR \rightarrow (EAd)$
		Transfers the CCR contents to a destination location. The condition code register size is one byte, but in transfer to memory, data is written by word access.
ANDC	В	$CCR \land \#IMM \rightarrow CCR$
		Logically ANDs the condition code register with immediate data.
ORC	В	$CCR \lor \#IMM \rightarrow CCR$
		Logically ORs the condition code register with immediate data.
XORC	В	$CCR \oplus \#IMM \to CCR$
		Logically exclusive-ORs the condition code register with immediate data.
NOP	_	$PC + 2 \rightarrow PC$
		Only increments the program counter.
Note: *	Size refers B: Byte	to the operand size.

#### Table 2.9 System Control Instructions

W: Word

Instruction	Size	Function	
EEPMOV.B	_	if R4L ≠ 0 then	
		repeat @ER5+ $\rightarrow$ @ER6+, R4L – 1 $\rightarrow$ R4L	
		until $R4L = 0$	
		else next;	
EEPMOV.W		if R4 ≠ 0 then	
		repeat @ER5+ $\rightarrow$ @ER6+, R4 – 1 $\rightarrow$ R4	
		until $R4 = 0$	
		else next;	
		Transfers a data block according to parameters set in general registers R4L or R4, ER5, and ER6.	
		R4L or R4:Size of block (bytes)ER5:Starting source addressER6:Starting destination address	
		ecution of the next instruction begins as soon as the transfer is mpleted.	

#### Table 2.10Block Transfer Instruction

#### 2.6.4 Basic Instruction Formats

The H8/300H instructions consist of 2-byte (1-word) units. An instruction consists of an operation field (OP field), a register field (r field), an effective address extension (EA field), and a condition field (cc).

**Operation Field:** Indicates the function of the instruction, the addressing mode, and the operation to be carried out on the operand. The operation field always includes the first 4 bits of the instruction. Some instructions have two operation fields.

**Register Field:** Specifies a general register. Address registers are specified by 3 bits, data registers by 3 bits or 4 bits. Some instructions have two register fields. Some have no register field.

**Effective Address Extension:** Eight, 16, or 32 bits specifying immediate data, an absolute address, or a displacement. A 24-bit address or displacement is treated as 32-bit data in which the first 8 bits are 0 (H'00).

Condition Field: Specifies the branching condition of Bcc instructions.

Figure 2.9 shows examples of instruction formats.

	NOP, RTS, etc.			
Operation field a	nd register fields			
	ор	rn	rm	ADD.B Rn, Rm, etc.
Dperation field, r	egister fields, and op	d effective address	s extension rm	
Operation field, r	ор			MOV.B @(d:16, Rn), Rm
	op EA (	rn	rm	MOV.B @(d:16, Rn), Rm

Figure 2.9 Instruction Formats

#### 2.6.5 Notes on Use of Bit Manipulation Instructions

The BSET, BCLR, BNOT, BST, and BIST instructions read a byte of data, modify a bit in the byte, then write the byte back. Care is required when these instructions are used to access registers with write-only bits, or to access ports.

The BCLR instruction can be used to clear flags in the on-chip registers. In an interrupt-handling routine, for example, if it is known that the flag is set to 1, it is not necessary to read the flag ahead of time.

# 2.7 Addressing Modes and Effective Address Calculation

#### 2.7.1 Addressing Modes

The H8/300H CPU supports the eight addressing modes listed in table 2.11. Each instruction uses a subset of these addressing modes. Arithmetic and logic instructions can use the register direct and immediate modes. Data transfer instructions can use all addressing modes except programcounter relative and memory indirect. Bit manipulation instructions use register direct, register indirect, or absolute (@aa:8) addressing mode to specify an operand, and register direct (BSET,

BCLR, BNOT, and BTST instructions) or immediate (3-bit) addressing mode to specify a bit number in the operand.

# Table 2.11 Addressing Modes

No.	Addressing Mode	Symbol
1	Register direct	Rn
2	Register indirect	@ERn
3	Register indirect with displacement	@(d:16, ERn)/@d:24, ERn)
4	Register indirect with post-increment	@Ern+
	Register indirect with pre-decrement	@-ERn
5	Absolute address	@aa:8/@aa:16/@aa:24
6	Immediate	#xx:8/#xx:16/#xx:32
7	Program-counter relative	@(d:8, PC)/@(d:16, PC)
8	Memory indirect	@@aa:8

#### 1. Register Direct—Rn

The register field of the instruction code specifies an 8-, 16-, or 32-bit register containing the operand. R0H to R7H and R0L to R7L can be specified as 8-bit registers. R0 to R7 and E0 to E7 can be specified as 16-bit registers. ER0 to ER7 can be specified as 32-bit registers.

#### 2. Register Indirect—@ERn

The register field of the instruction code specifies an address register (ERn), the lower 24 bits of which contain the address of the operand.

# 3. Register Indirect with Displacement—@(d:16, ERn) or @(d:24, ERn)

A 16-bit or 24-bit displacement contained in the instruction code is added to the contents of an address register (ERn) specified by the register field of the instruction, and the lower 24 bits of the sum specify the address of a memory operand. A 16-bit displacement is sign-extended when added.

#### 4. Register Indirect with Post-Increment or Pre-Decrement—@ERn+ or @-ERn

• Register indirect with post-increment—@ERn+

The register field of the instruction code specifies an address register (ERn) the lower 24 bits of which contain the address of a memory operand. After the operand is accessed, 1, 2, or 4 is added to the address register contents (32 bits) and the sum is stored in the address register. The value added is 1 for byte access, 2 for word access, or 4 for longword access. For word or longword access, the register value should be even.

• Register indirect with pre-decrement—@-ERn The value 1, 2, or 4 is subtracted from an address register (ERn) specified by the register field in the instruction code, and the lower 24 bits of the result become the address of a memory operand. The result is also stored in the address register. The value subtracted is 1 for byte access, 2 for word access, or 4 for longword access. For word or longword access, the resulting register value should be even.

#### 5. Absolute Address—@aa:8, @aa:16, or @aa:24

The instruction code contains the absolute address of a memory operand. The absolute address may be 8 bits long (@aa:8), 16 bits long (@aa:16), or 24 bits long (@aa:24). For an 8-bit absolute address, the upper 16 bits are all assumed to be 1 (H'FFFF). For a 16-bit absolute address the upper 8 bits are a sign extension. A 24-bit absolute address can access the entire address space. Table 2.12 indicates the accessible address ranges.

Absolute Address	1-Mbyte Modes	16-Mbyte Modes
8 bits (@aa:8)	H'FFF00 to H'FFFFF (1,048,320 to 1,048,575)	H'FFFF00 to H'FFFFFF (16,776,960 to 16,777,215)
16 bits (@aa:16)	H'00000 to H'07FFF, H'F8000 to H'FFFFF (0 to 32,767, 1,015,808 to 1,048,575)	H'000000 to H'007FFF, H'FF8000 to H'FFFFFF (0 to 32,767, 16,744,448 to 16,777,215)
24 bits (@aa:24)	H'00000 to H'FFFFF (0 to 1,048,575)	H'000000 to H'FFFFF (0 to 16,777,215)

#### Table 2.12 Absolute Address Access Ranges

#### Renesas

### 6. Immediate—#xx:8, #xx:16, or #xx:32

The instruction code contains 8-bit (#xx:8), 16-bit (#xx:16), or 32-bit (#xx:32) immediate data as an operand.

The instruction codes of the ADDS, SUBS, INC, and DEC instructions contain immediate data implicitly. The instruction codes of some bit manipulation instructions contain 3-bit immediate data specifying a bit number. The TRAPA instruction code contains 2-bit immediate data specifying a vector address.

# 7. Program-Counter Relative—@(d:8, PC) or @(d:16, PC)

This mode is used in the Bcc and BSR instructions. An 8-bit or 16-bit displacement contained in the instruction code is sign-extended to 24 bits and added to the 24-bit PC contents to generate a 24-bit branch address. The PC value to which the displacement is added is the address of the first byte of the next instruction, so the possible branching range is -126 to +128 bytes (-63 to +64 words) or -32766 to +32768 bytes (-16383 to +16384 words) from the branch instruction. The resulting value should be an even number.

# 8. Memory Indirect—@@aa:8

This mode can be used by the JMP and JSR instructions. The instruction code contains an 8-bit absolute address specifying a memory operand. This memory operand contains a branch address. The memory operand is accessed by longword access. The first byte of the memory operand is ignored, generating a 24-bit branch address. See figure 2.10. The upper bits of the 8-bit absolute address are assumed to be 0 (H'0000), so the address range is 0 to 255 (H'000000 to H'0000FF). Note that the first part of this range is also the exception vector area. For further details see section 5, Interrupt Controller.

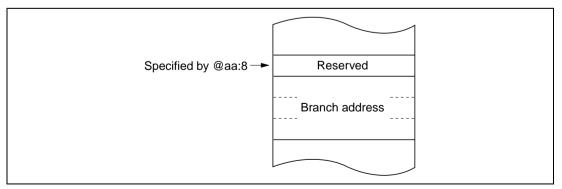
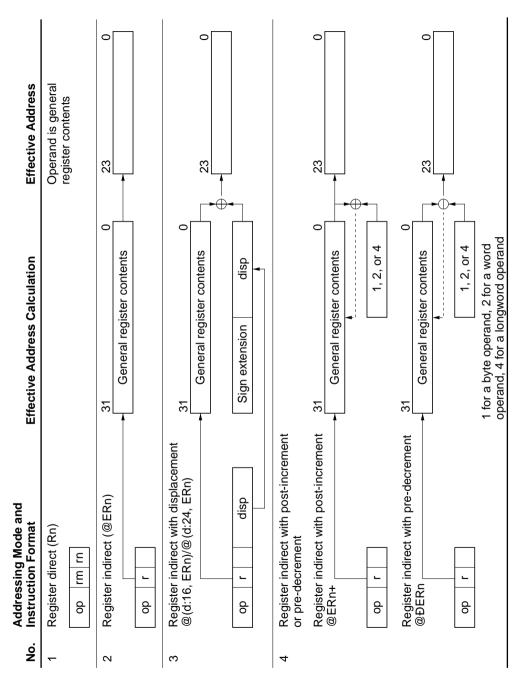


Figure 2.10 Memory-Indirect Branch Address Specification

When a word-size or longword-size memory operand is specified, or when a branch address is specified, if the specified memory address is odd, the least significant bit is regarded as 0. The accessed data or instruction code therefore begins at the preceding address. See section 2.5.2, Memory Data Formats.

#### 2.7.2 Effective Address Calculation

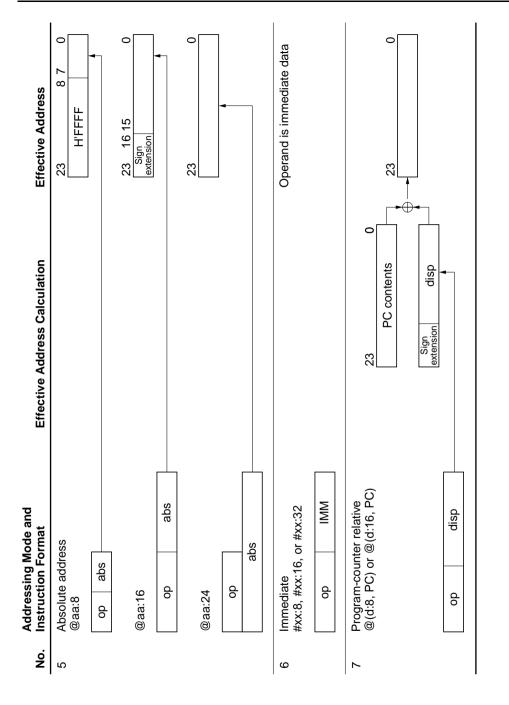
Table 2.13 explains how an effective address is calculated in each addressing mode. In the 1-Mbyte operating modes the upper 4 bits of the calculated address are ignored in order to generate a 20-bit effective address.



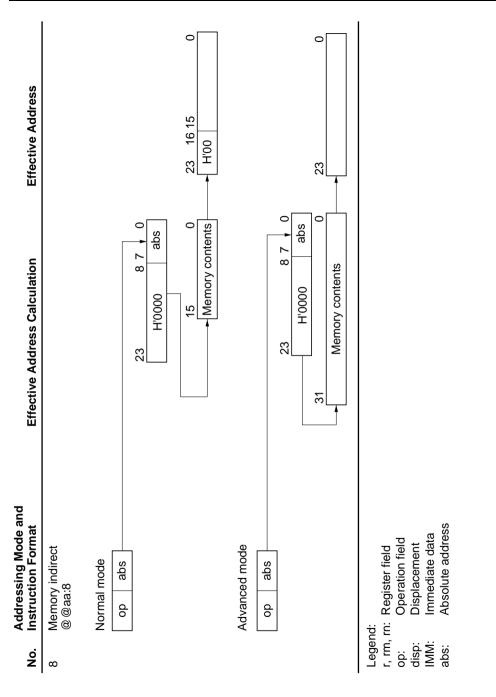
Section 2

CPU

Rev.3.00 Mar. 26, 2007 Page 46 of 682 REJ09B0353-0300



Renesas



## 2.8 Processing States

#### 2.8.1 Overview

The H8/300H CPU has four processing states: the program execution state, exception-handling state, power-down state, and reset state. The power-down state includes sleep mode, software standby mode, and hardware standby mode. Figure 2.11 classifies the processing states. Figure 2.13 indicates the state transitions.

Processing states	<ul> <li>Program execution state</li> </ul>	
	The CPU executes program instructions in	sequence
-	Exception-handling state	
	A transient state in which the CPU execute (saving PC and CCR, fetching a vector, et interrupt, or other exception	
_	Reset state	
	The CPU and all on-chip supporting modu	les are initialized and halted
	Power-down state	Sleep mode
	The CPU is halted to conserve power	
		Software standby mode
		Hardware standby mode

Figure 2.11 Processing States

#### 2.8.2 Program Execution State

In this state the CPU executes program instructions in normal sequence.

## 2.8.3 Exception-Handling State

The exception-handling state is a transient state that occurs when the CPU alters the normal program flow due to a reset, interrupt, or trap instruction. The CPU fetches a starting address from the exception vector table and branches to that address. In interrupt and trap exception handling the CPU references the stack pointer (ER7) and saves the program counter and condition code register.

**Types of Exception Handling and Their Priority:** Exception handling is performed for resets, interrupts, and trap instructions. Table 2.14 indicates the types of exception handling and their priority. Trap instruction exceptions are accepted at all times in the program execution state.

Priority	Type of Exception	Detection Timing	Start of Exception Handling
High	Reset	Synchronized with clock	Exception handling starts immediately when RES changes from low to high
	Interrupt	End of instruction execution or end of exception handling*	When an interrupt is requested, exception handling starts at the end of the current instruction or current exception-handling sequence
Low	Trap instruction	When TRAPA instruction is executed	Exception handling starts when a trap (TRAPA) instruction is executed
Low Note: *	Interrupts are not detecte	ed at the end of the ANDC, O	executed RC, XORC, and LDC instructio

 Table 2.14
 Exception Handling Types and Priority

Note: \* Interrupts are not detected at the end of the ANDC, ORC, XORC, and LDC instructions, or immediately after reset exception handling.

Figure 2.12 classifies the exception sources. For further details about exception sources, vector numbers, and vector addresses, see section 4, Exception Handling, and section 5, Interrupt Controller.

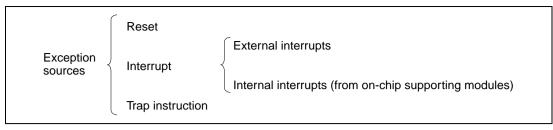


Figure 2.12 Classification of Exception Sources

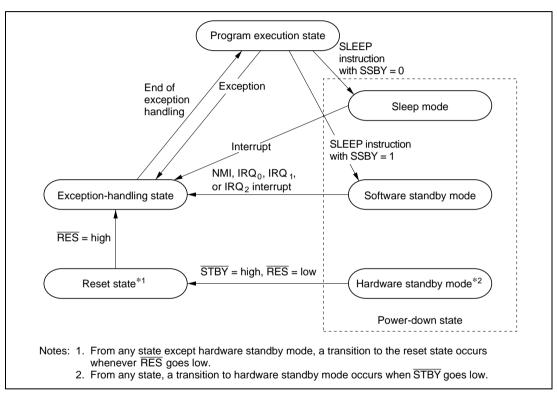


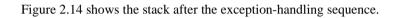
Figure 2.13 State Transitions

#### 2.8.4 Exception-Handling Sequences

**Reset Exception Handling:** Reset exception handling has the highest priority. The reset state is entered when the  $\overline{\text{RES}}$  signal goes low. Reset exception handling starts after that, when  $\overline{\text{RES}}$  changes from low to high. When reset exception handling starts the CPU fetches a start address from the exception vector table and starts program execution from that address. All interrupts,

including NMI, are disabled during the reset exception-handling sequence and immediately after it ends.

**Interrupt Exception Handling and Trap Instruction Exception Handling:** When these exception-handling sequences begin, the CPU references the stack pointer (ER7) and pushes the program counter and condition code register on the stack. Next, if the UE bit in the system control register (SYSCR) is set to 1, the CPU sets this set to 1, the CPU sets the I bit in the condition code register to 1. If the UE bit is cleared to 0, the CPU sets both the I bit and the UI bit in the condition code register to 1. Then the CPU fetches a start address from the exception vector table and execution branches to that address.



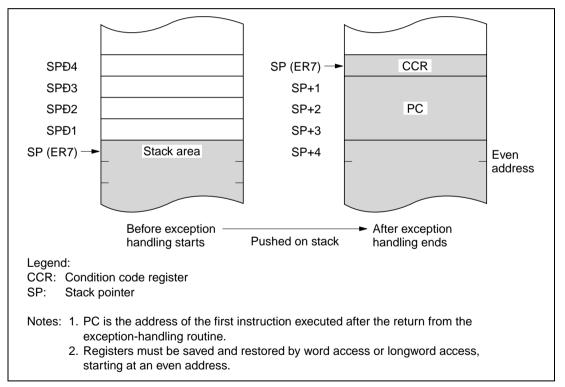


Figure 2.14 Stack Structure after Exception Handling

#### 2.8.5 Reset State

When the  $\overline{\text{RES}}$  input goes low all current processing stops and the CPU enters the reset state. The I bit in the condition code register is set to 1 by a reset. All interrupts are masked in the reset state. Reset exception handling starts when the  $\overline{\text{RES}}$  signal changes from low to high.

The reset state can also be entered by a watchdog timer overflow. For details see section 10, Watchdog Timer.

#### 2.8.6 Power-Down State

In the power-down state the CPU stops operating to conserve power. There are three modes: sleep mode, software standby mode, and hardware standby mode.

**Sleep Mode:** A transition to sleep mode is made if the SLEEP instruction is executed while the SSBY bit is cleared to 0 in the system control register (SYSCR). CPU operations stop immediately after execution of the SLEEP instruction, but the contents of CPU registers are retained.

**Software Standby Mode:** A transition to software standby mode is made if the SLEEP instruction is executed while the SSBY bit is set to 1 in SYSCR. The CPU and clock halt and all on-chip supporting modules stop operating. The on-chip supporting modules are reset, but as long as a specified voltage is supplied the contents of CPU registers and on-chip RAM are retained. The I/O ports also remain in their existing states.

**Hardware Standby Mode:** A transition to hardware standby mode is made when the **STBY** input goes low. As in software standby mode, the CPU and clock halt and the on-chip supporting modules are reset, but as long as a specified voltage is supplied, on-chip RAM contents are retained.

For further information see section 17, Power-Down State.

## 2.9 Basic Operational Timing

## 2.9.1 Overview

The H8/300H CPU operates according to the system clock ( $\phi$ ). The interval from one rise of the system clock to the next rise is referred to as a "state." A memory cycle or bus cycle consists of two or three states. The CPU uses different methods to access on-chip memory, the on-chip supporting modules, and the external address space. Access to the external address space can be controlled by the bus controller.

## 2.9.2 On-Chip Memory Access Timing

On-chip memory is accessed in two states. The data bus is 16 bits wide, permitting both byte and word access. Figure 2.15 shows the on-chip memory access cycle. Figure 2.16 indicates the pin states.

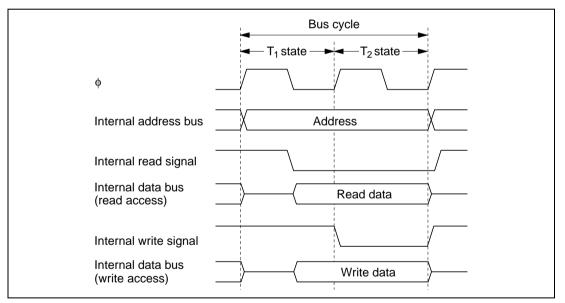


Figure 2.15 On-Chip Memory Access Cycle

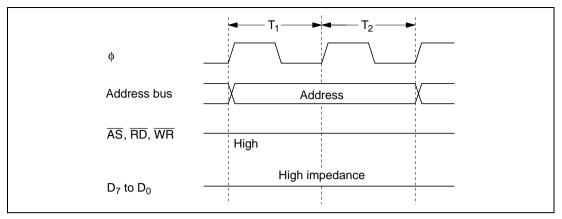


Figure 2.16 Pin States during On-Chip Memory Access

## 2.9.3 On-Chip Supporting Module Access Timing

The on-chip supporting modules are accessed in three states. The data bus is 8 or 16 bits wide, depending on the register being accessed. Figure 2.17 shows the on-chip supporting module access timing. Figure 2.18 indicates the pin states.

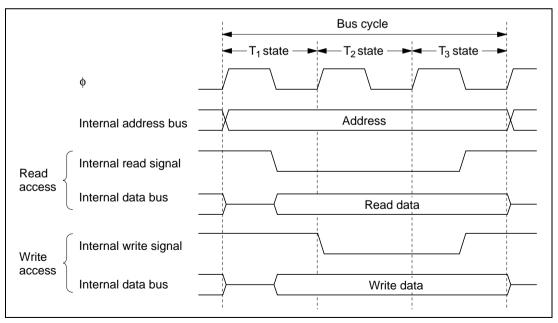
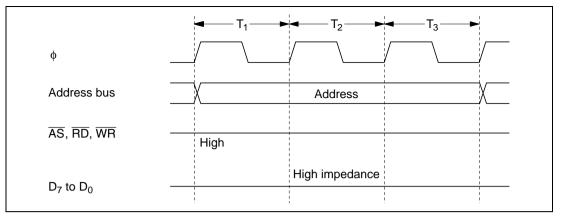
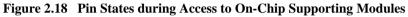


Figure 2.17 Access Cycle for On-Chip Supporting Modules





## 2.9.4 Access to External Address Space

The external address space is divided into eight areas (areas 0 to 7). Bus-controller settings determine whether each area accessed in two or three states. For details see section 6, Bus Controller.

# Section 3 MCU Operating Modes

## 3.1 Overview

### 3.1.1 Operating Mode Selection

The H8/3039 Group has five operating modes (modes 1, 3, 5 to7) that are selected by the mode pins ( $MD_2$  to  $MD_0$ ) as indicated in table 3.1. The input at these pins determines expanded mode or single-chip mode.

Mode Pins			ins	Description					
Operating Mode	MD <sub>2</sub>	MD <sub>1</sub>	MD₀	Address Space	Initial Bus Mode <sup>*1</sup>	On-Chip ROM	On-Chip RAM		
	0	0	0	_		—	_		
Mode 1	0	0	1	Expanded mode	8 bits	Disabled	Enabled*1		
Mode 2	0	1	0	—	_	_	_		
Mode 3	0	1	1	Expanded mode	8 bits	Disabled	Enabled*1		
Mode 4	1	0	0	—	_	_	_		
Mode 5	1	0	1	Expanded mode	8 bits	Enabled	Enabled*1		
Mode 6	1	1	0	Single-chip normal mode	_	Enabled	Enabled*2		
Mode 7	1	1	1	Single-chip advanced mode	» —	Enabled	Enabled*2		

#### Table 3.1 Operating Mode Selection

Notes: 1. If the RAM enable bit (RAME) in the system control register (SYSCR) is cleared to 0, these addresses become external addresses.

 In mode 6 and 7, clearing bit RAME in SYSCR to 0 and reading the on-chip RAM always return H'FF, and write access is ignored. For details, see section 14.3, Operation.

For the address space size there are three choices: 64 kbytes, 1 Mbyte, or 16 Mbytes.

Modes 1 and 3 are on-chip ROM disable expanded modes capable of accessing external memory and peripheral devices.

Mode 1 supports a maximum address space of 1 Mbyte.

Mode 3 supports a maximum address space of 16 Mbytes.

Mode 5 is externally expanded mode that enables access to external memory and peripheral devices and also enables access to the on-chip ROM. Mode 5 supports a maximum address space of 1 Mbyte.

Modes 6 and 7 are single-chip modes that operate using the on-chip ROM, RAM, and registers. All I/O ports are available. Mode 6 is a normal mode with 64-kbyte address space. Mode 7 is an advanced mode with a maximum address space of 1 Mbyte.

The H8/3039 Group can be used only in modes 1, 3, or 5 to 7. The inputs at the mode pins must select one of these seven modes. The inputs at the mode pins must not be changed during operation.

## 3.1.2 Register Configuration

The H8/3039 Group has a mode control register (MDCR) that indicates the inputs at the mode pins ( $MD_2$  to  $MD_0$ ), and a system control register (SYSCR). Table 3.2 summarizes these registers.

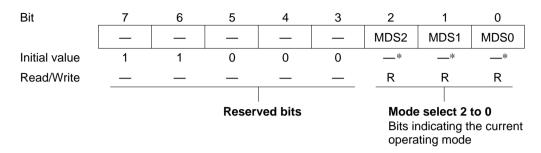
Table 3.2	Registers
-----------	-----------

Address*	Name	Abbreviation	R/W	Initial Value
H'FFF1	Mode control register	MDCR	R	Undetermined
H'FFF2	System control register	SYSCR	R/W	H'0B

Note: \* The lower 16 bits of the address are indicated.

## 3.2 Mode Control Register (MDCR)

MDCR is an 8-bit read-only register that indicates the current operating mode of the H8/3039 Group.



Note: Determined by pins  $MD_2$  to  $MD_0$ .

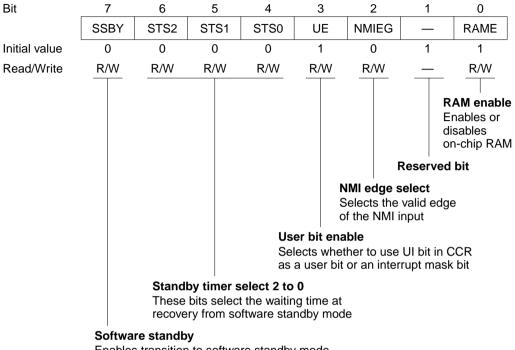
Bits 7 and 6—Reserved: These bits cannot be modified and are always read as 1.

Bits 5 to 3—Reserved: These bits cannot be modified and are always read as 0.

**Bits 2 to 0—Mode Select 2 to 0** (**MDS**<sub>2</sub> **to MDS**<sub>0</sub>): These bits indicate the logic levels at pins  $MD_2$  to  $MD_0$  (the current operating mode).  $MDS_2$  to  $MDS_0$  correspond to  $MD_2$  to  $MD_0$ .  $MDS_1$  and  $MDS_0$  are read-only bits. The mode pin ( $MD_2$  to  $MD_0$ ) levels are latched when MDCR is read.

## **3.3** System Control Register (SYSCR)

SYSCR is an 8-bit register that controls the operation of the H8/3039 Group.



Enables transition to software standby mode

**Bit 7—Software Standby (SSBY):** Enables transition to software standby mode. (For further information about software standby mode see section 17, Power-Down State.)

When software standby mode is exited by an external interrupt, this bit remains set to 1. To clear this bit, write 0.

Bit 7 SSBY	Description	
0	SLEEP instruction causes transition to sleep mode	(Initial value)
1	SLEEP instruction causes transition to software standby mode	

**Bits 6 to 4—Standby Timer Select (STS2 to STS0):** These bits select the length of time the CPU and on-chip supporting modules wait for the internal clock oscillator to settle when software standby mode is exited by an external interrupt. Set these bits so that the waiting time will be at

Bit 6 STS2	Bit 5 STS1	Bit 4 STS0	Description	
0	0	0	Waiting time = 8,192 states	(Initial value)
0	0	1	Waiting time = 16,384 states	
0	1	0	Waiting time = 32,768 states	
0	1	1	Waiting time = 65,536 states	
1	0	0	Waiting time = 131,072 states	
1	0	1	Waiting time = 1,024 states	
1	1		Illegal setting	

least 7 ms at the system clock rate. For further information about waiting time selection, see section 17.4.3, Selection of Oscillator Waiting Time after Exit from Software Standby Mode.

Bit 3—User Bit Enable (UE): Selects whether to use the UI bit in the condition code register as a user bit or an interrupt mask bit.

Bit 3		
UE	Description	
0	UI bit in CCR is used as an interrupt mask bit	
1	UI bit in CCR is used as a user bit	(Initial value)

Bit 2-NMI Edge Select (NMIEG): Selects the valid edge of the NMI input.

Bit 2 NMIEG	Description	
0	An interrupt is requested at the falling edge of NMI	(Initial value)
1	An interrupt is requested at the rising edge of NMI	

Bit 1—Reserved: This bit cannot be modified and is always read as 1.

**Bit 0—RAM Enable (RAME):** Enables or disables the on-chip RAM. The RAME bit is initialized by the rising edge of the  $\overline{\text{RES}}$  signal. It is not initialized in software standby mode.

Bit 0	Description	
RAME	Description	
0	On-chip RAM is disabled	
1	On-chip RAM is enabled	(Initial value)

## **3.4 Operating Mode Descriptions**

## 3.4.1 Mode 1

Ports 1, 2, and 5 function as address pins  $A_{19}$  to  $A_0$ , permitting access to a maximum 1-Mbyte address space. The initial bus mode after a reset is 8 bits, with 8-bit access to all areas.

### 3.4.2 Mode 3

Ports 1, 2, and 5 and part of port A function as address pins  $A_{23}$  to  $A_0$ , permitting access to a maximum 16-Mbyte address space. The initial bus mode after a reset is 8 bits, with 8-bit access to all areas.  $A_{23}$  to  $A_{21}$  are valid when 0 is written in bits 7 to 5 of the bus release control register (BRCR). (In this mode  $A_{20}$  is always used for address output.)

### 3.4.3 Mode 5

Ports 1, 2, and 5 can function as address pins  $A_{19}$  to  $A_{07}$  permitting access to a maximum 1-Mbyte address space, but following a reset they are input ports. To use ports 1, 2, and 5 as an address bus, the corresponding bits in their data direction registers (P1DDR, P2DDR, and P5DDR) must be set to 1. The address bus width can be selected freely by setting DDR of ports 1, 2, and 5. The initial bus mode after a reset is 8 bits, with 8-bit access to all areas.

### 3.4.4 Mode 6

This mode operates using the on-chip ROM, RAM, and registers. All I/O ports are available. Mode 6 is a normal mode with 64-kbyte address space.

### 3.4.5 Mode 7

This mode is an advanced mode with a 1-Mbyte address space which operates using the on-chip ROM, RAM, and registers. All I/O ports are available.

Note: The H8/3039 Group cannot be used in mode 2 and 4.

## 3.5 Pin Functions in Each Operating Mode

The pin functions of ports 1 to 3, port 5 and port A vary depending on the operating mode. Table 3.3 indicates their functions in each operating mode.

Port	Mode 1	Mode 2*1	Mode 3	Mode 4*1	Mode 5	Mode 6	Mode 7
Port 1	$A_7$ to $A_0$	_	$A_7$ to $A_0$	_	P1 <sub>7</sub> to P1 <sub>0</sub> * <sup>2</sup>	$P1_7$ to $P1_0$	P1 <sub>7</sub> to P1 <sub>0</sub>
Port 2	$A_{15}$ to $A_8$	_	$A_{15}$ to $A_8$	_	$P2_{7}$ to $P2_{0}^{*^{2}}$	$P2_7$ to $P2_0$	P2 <sub>7</sub> to P2 <sub>0</sub>
Port 3	D <sub>7</sub> to D <sub>0</sub>	_	D <sub>7</sub> to D <sub>0</sub>	_	D <sub>7</sub> to D <sub>0</sub>	P3 <sub>7</sub> to P3 <sub>0</sub>	P3 <sub>7</sub> to P3 <sub>0</sub>
Port 5	$\rm A_{_{19}}$ to $\rm A_{_{16}}$	_	$A_{19}$ to $A_{16}$	_	$P5_{_3}$ to $P5_{_0}^{*^2}$	$P5_{_3}$ to $P5_{_0}$	$P5_{_3}$ to $P5_{_0}$
Port A	PA <sub>7</sub> to PA <sub>4</sub>		PA <sub>6</sub> to PA <sub>4</sub> * <sup>3</sup> , A <sub>20</sub>	_	PA <sub>7</sub> to PA <sub>4</sub>	PA <sub>7</sub> to PA <sub>4</sub>	PA <sub>7</sub> to PA <sub>4</sub>

Table 3.3Pin Functions in Each Mode

Notes: 1. H8/3039 Group cannot be used in these modes.

2. Initial state. These pins become address output pins when the corresponding bits in the data direction registers (P1DDR, P2DDR, P5DDR) are set to 1.

3. Initial state  $A_{20}$  is always an address output pin.  $PA_6$  to  $PA_4$  are switched over to  $A_{23}$  to  $A_{21}$  output by writing 0 in bits 7 to 5 of ADRCR.

## **3.6 Memory Map in Each Operating Mode**

Figure 3.1 shows a memory map of the H8/3039. Figure 3.2 shows a memory map of the H8/3038. Figure 3.3 shows a memory map of the H8/3037. Figure 3.4 shows a memory map of the H8/3036. The address space is divided into eight areas.

Modes 1, 3 and 5 are the 8-bit bus mode.

The address locations of the on-chip RAM and on-chip registers differ between the 1-Mbyte modes (modes 1, 5, and 7) and 16-Mbyte mode (mode 3), and 64-kbyte mode (mode 6). The address range specifiable by the CPU in the 8- and 16-bit absolute addressing modes (@aa:8 and @aa:16) also differs.

	Mode 1				Mode 3		
	oyte expanded mo on-chip ROM disa		h		lbyte expanded modes with on-chip ROM disabled)		
H'00000	Vector area	s ite	half)	H'000000	Vector area	s ite	half)
H'000FF		8-bit memory-indirect branch addresses	16-bit absolute addresses (first half)	H'0000FF		8-bit memory-indirect branch addresses	16-bit absolute addresses (first half)
H'07FFF		8-bit m branch	16-b	H'007FFF		8-bit m branch	16-b
H'1FFFF		Area 0				Area 0	
H'20000		Area 1		H'1FFFFF			
H'3FFFF H'40000 H'5FFFF		Area 2		H'200000		Area 1	
H'60000 H'7FFF	External address	Area 3		H'3FFFFF H'400000			·
H'80000 H'9FFF	space	Area 4				Area 2	
H'A0000 H'BFFFF	:	Area 5		H'5FFFFF H'600000			
H'C0000 H'DFFFF		Area 6		H'7FFFFF	External address space	Area 3	
H'E0000		Area 7		H'800000		+	
						Area 4	
				H'9FFFFF			
H'F8000			1	H'A00000			
H'FEF0F			s			Area 5	
H'FEF10		1	sse	H'BFFFFF			
	On-chip RAM*	ses	dree	H'C00000			
H'FFF00		Ğ ∳	ado			Area 6	
H'FFF0F H'FFF10		llpp	f)	H'DFFFFF H'E00000			·
H'FFF1E	address	8-bit absolute addresses	16-bit absolute addresses (second half)			Area 7	
H'FFF1C		ab	6-bi				
	registers	3-bit	~ "	H'FF8000		+	<b>↓</b>
H'FFFF		Lt	<b>!</b>	H'FFEF0F			
				H'FFEF10	On-chip RAM*	es	absolute addresses nd half)
				H'FFFF00		ess	adc
				H'FFFF0F H'FFFF10	External	8-bit absolute addresses	16-bit absolute (second half)
					address space	olut	ab
				H'FFFF1B H'FFFF1C	-	abs(	5-bit
				H'FFFFFF	On-chip I/O registers	8-bit é	16 (St
							·
Note: * Ex	ternal addresses	can be	access	sed by disabl	ing on-chip RAN	1.	

Figure 3.1 H8/3039 Memory Map in Each Operating Mode (1)

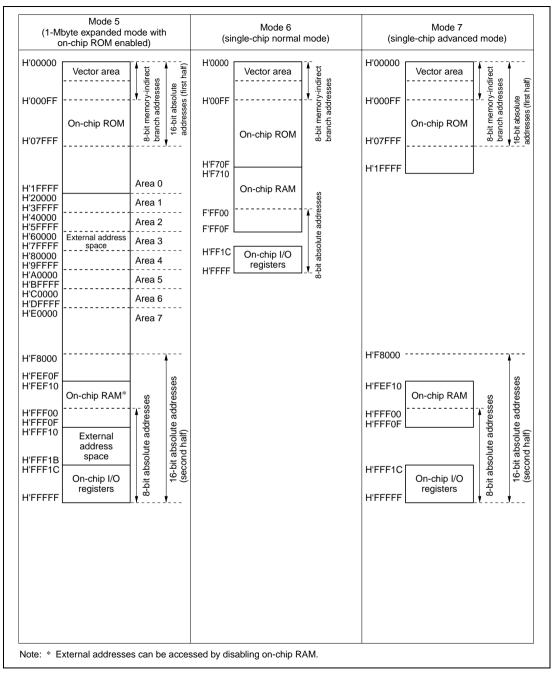
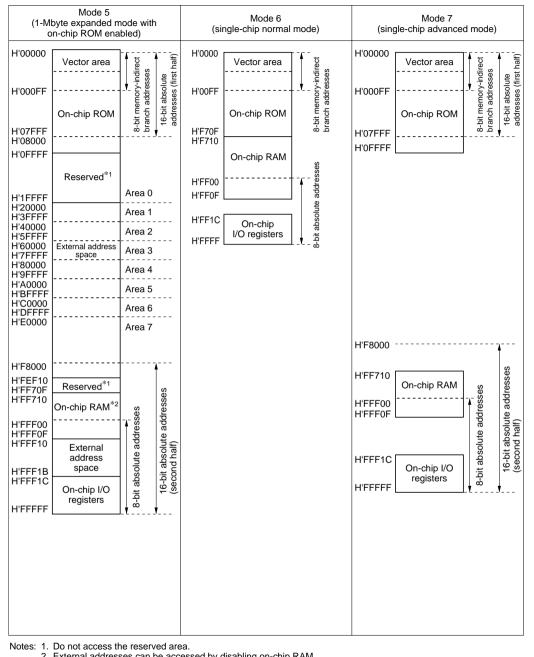


Figure 3.1 H8/3039 Memory Map in Each Operating Mode (2)

	Mode 1 yte expanded mo n-chip ROM disal		ı			(16-Mbyte expanded modes with on-chip ROM disabled)		
H'00000	Vector area	lirect s	half)	H'000000	Vector area	s s	half)	
H'000FF		8-bit memory-indirect branch addresses	16-bit absolute addresses (first	H'0000FF		8-bit memory-indirect branch addresses	16-bit absolute addresses (first	
H'07FFF		8-bit r branc	16-b addr	H'007FFF		8-bit r branc	16-b addr	
H'1FFFF		Area 0				Area 0		
H'20000 H'3FFFF		Area 1		H'1FFFFF H'200000				
H'40000 H'5FFFF		Area 2				Area 1		
H'60000 H'7FFFF	External address space	Area 3		H'3FFFFF H'400000				
H'80000 H'9FFFF		Area 4				Area 2		
H'A0000 H'BFFFF		Area 5		H'5FFFFF H'600000				
H'C0000 H'DFFFF		Area 6			External address	Area 3		
H'E0000		Area 7		H'7FFFFF	space			
				H'800000		Area 4		
H'F8000			<b>}</b>	H'9FFFFF H'A00000				
H'FEF10 H'FF70F	Reserved <sup>*1</sup>		S			Area 5		
H'FF710	On-chip RAM*2	St	esse.	H'BFFFFF H'C00000				
H'FFF00		esse	addr	H'DFFFFF		Area 6		
H'FFF0F H'FFF10	External	addr	olute alf)	H'E00000				
H'FFF1B	address space	8-bit absolute addresses	16-bit absolute addresses (second half)			Area 7		
H'FFF1C	On-chip I/O	t abs	16-bi secc					
H'FFFFF	registers	8-bi		H'FF8000			1	
	L		J	H'FFEF10 H'FFF70F	Reserved <sup>*1</sup>		es	
				H'FFF710	On-chip RAM*2	ses	dress	
				H'FFFF00 H'FFFF0F		lress	e adc	
				H'FFFF10	External	e adc	solut( alf)	
				H'FFFF1B	address space	solute	16-bit absolute addresses (second half)	
				H'FFFF1C	On-chip I/O	8-bit absolute addresses	16-bi (secc	
				H'FFFFFF	registers	, <del>.</del>	ł	

Figure 3.2 H8/3038 Memory Map in Each Operating Mode (1)

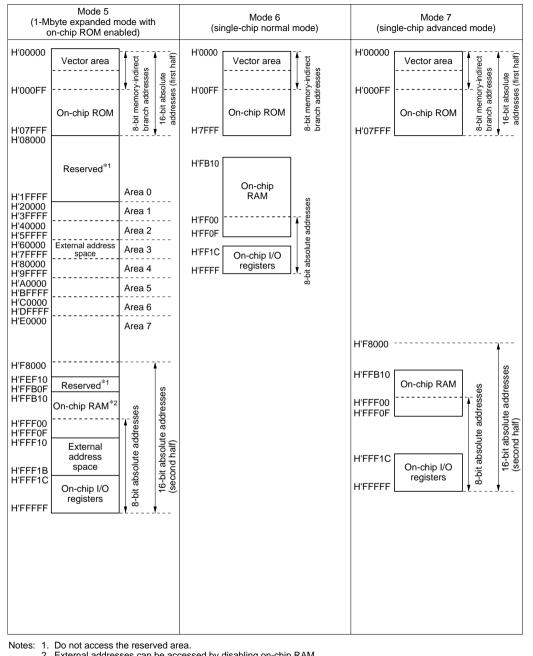


2. External addresses can be accessed by disabling on-chip RAM.

#### Figure 3.2 H8/3038 Memory Map in Each Operating Mode (2)

	Mode 1 oyte expanded mo n-chip ROM disal		h	Mode 3 (16-Mbyte expanded modes with on-chip ROM disabled)			n
H'00000	Vector area	direct	e st half)	H'000000	Vector area	direct es	e st half)
H'000FF		8-bit memory-indirect branch addresses	16-bit absolute addresses (first h	H'0000FF		8-bit memory-indirect branch addresses	16-bit absolute addresses (first half)
H'07FFF		8-bit bran	ad 6	H'007FFF		8-bit bran	ad 6
H'1FFFF		Area 0				Area 0	
H'20000 H'3FFFF		Area 1		H'1FFFFF H'200000			
H'40000 H'5FFFF		Area 2		11200000		Area 1	
H'60000 H'7FFFF	External address space	Area 3		H'3FFFFF H'400000			
H'80000 H'9FFFF		Area 4				Area 2	
H'A0000		Area 5		H'5FFFFF			
H'BFFFF H'C0000				H'600000	External address	Area 3	
H'DFFFF		Area 6		H'7FFFFF	space	Area 3	
H'E0000		Area 7		H'800000			
H'F8000			<b>-</b>	H'9FFFF H'A00000		Area 4	
H'FEF10	Reserved <sup>*1</sup>		Ī	11 A00000		Area 5	
H'FFB0F H'FFB10			sses	H'BFFFFF H'C00000			
H'FFF00	On-chip RAM <sup>*2</sup>	sses	ddre	11000000		Area 6	
H'FFF0F H'FFF10		Iddre	ute a f)	H'DFFFFF H'E00000			
	External address space	8-bit absolute addresses	16-bit absolute addresses (second half)			Area 7	
H'FFF1B H'FFF1C		abso	3-bit				
	On-chip I/O registers	3-bit	(s	H'FF8000			<b>h</b>
H'FFFFF	_	Lt	<b>!</b>	H'FFEF10 H'FFFB0F	Reserved <sup>*1</sup>		ŝ
				H'FFFB10	On-chip RAM*2	Se	16-bit absolute addresses (second half)
				H'FFFF00 H'FFFF0F		Iress	add
				H'FFFF10	External	e adc	solute nalf)
				H'FFFF1B	address space	8-bit absolute addresses	16-bit absolu second half)
				H'FFFF1C	On-chip I/O	oit ab	16-b (sec
				H'FFFFFF	registers	₩, 8,	ļ

Figure 3.3 H8/3037 Memory Map in Each Operating Mode (1)

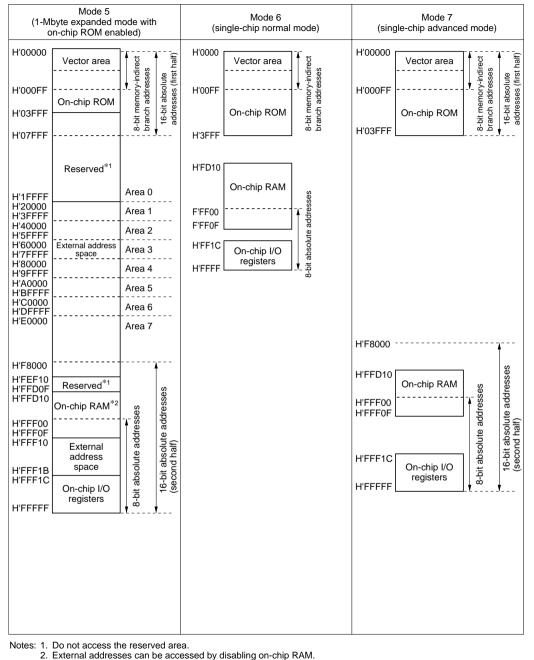


2. External addresses can be accessed by disabling on-chip RAM.

#### Figure 3.3 H8/3037 Memory Map in Each Operating Mode (2)

	Mode 1 yte expanded mo n-chip ROM disal		h		Mode 3 oyte expanded mo n-chip ROM disat		h
H'00000	Vector area	s ite	half)	H'000000	Vector area	s litect	half)
H'000FF		8-bit memory-indirect branch addresses	16-bit absolute addresses (first h	H'0000FF		a-bit memory-indirect branch addresses	16-bit absolute addresses (first half)
H'07FFF			<b>1</b>	H'007FFF		8-bi	a d
H'1FFFF		Area 0				Area 0	
H'20000 H'3FFFF		Area 1		H'1FFFFF H'200000			
H'40000 H'5FFFF H'60000	External address	Area 2		H'3FFFFF		Area 1	
H'7FFFF H'80000	space	Area 3 Area 4		H'400000		Area 2	
H'9FFFF H'A0000 H'BFFFF		Area 5		H'5FFFFF H'600000	External		
H'C0000 H'DFFFF H'E0000		Area 6		H'7FFFFF	address space	Area 3	
H E0000		Area 7		H'800000		Area 4	
H'F8000			<b>A</b>	H'9FFFFF H'A00000			
H'FEF10 H'FFD0F H'FFD10	Reserved <sup>*1</sup>		ses	H'BFFFFF		Area 5	
H'FFF00	On-chip RAM <sup>*2</sup>	sses	ddres	H'C00000		Area 6	
H'FFF0F H'FFF10	External	addre	olute a alf)	H'DFFFFF H'E00000			
H'FFF1B	address space	8-bit absolute addresses	16-bit absolute addresses (second half)			Area 7	
H'FFF1C	On-chip I/O registers	-bit ab	16-b (sec	H'FF8000			<b>A</b>
H'FFFFF		. <b>t</b>	<b>.</b>	H'FFEF10 H'FFFD0F	Reserved <sup>*1</sup>		es
				H'FFFD10	On-chip RAM <sup>*2</sup>	Ses	ldress
				H'FFFF00 H'FFFF0F H'FFFF10	External	addres	lute ac f)
				H'FFFF1B	External address space	8-bit absolute addresses	16-bit absolute addresses (second half)
				H'FFFF1C	On-chip I/O	bit abs	16-bi (secc
				H'FFFFFF	registers	, ¢	J

Figure 3.4 H8/3036 Memory Map in Each Operating Mode (1)



External addresses can be accessed by disabling on-chip RAM.

#### Figure 3.4 H8/3036 Memory Map in Each Operating Mode (2)

## **3.7** Restrictions on Use of Mode 6

In mode 6 (single-chip normal mode), on-chip ROM area data is undefined if address H'10000 or above (64 kbytes or above) is accessed, and therefore instruction code fetch and data read operations may not always be performed normally.

However, there is no problem with address H'10000 and above if the lower 16-bit address is an on-chip RAM (H'F710 to H'FF0F) or internal I/O register (H'FF1C to H'FFFF) address.

Table 3.4 shows the restrictions concerning each addressing mode.

	Со			
Addressing Mode	<b>Restricted Item</b>	Address Range	Operation	Restriction
Register direct (Rn)	_		No problem	_
Register indirect (@ERn)	Contents of ERn	H'00010000 or above, with lower 16 bits in range H'0000 to H'F710	Read data is undefined. Writes are invalid.	Set upper 16 bits of ERn to H'0000; or, write same data as in H'00000–H'0FFFF to H'10000–H'1FFFF in on-chip ROM.
Register indirect with displacement (@(d:16,ERn), @(d:16,ERn))	Value of ERn contents plus displacement	_		
Register indirect with post-increment (@ERn+)	Value of ERn contents incremented (or			
Register indirect with pre-decrement (@ERn-)	decremented) by 1, 2, or 4			
Absolute address (@aa:8)	_	_	No problem	_
Absolute address (@aa:16)	Value of @aa sign-extended to 24 bits	H'010000 or above, with lower 16 bits in range H'0000 to H'F710	Read data is undefined. Writes are invalid.	Do not specify H'8000 or above as absolute address; or, write same data as in H'00000– H'0FFFF to H'10000– H'1FFFF in on-chip ROM.

### Table 3.4 Access Restrictions in Mode 6 (Single-Chip Normal Mode)

	Со	nditions		
Addressing Mode	<b>Restricted Item</b>	Address Range	Operation	Restriction
Absolute address (@aa:24)	Value of @aa	H'010000 or above, with lower 16 bits in range H'0000 to H'F710	Read data is undefined. Writes are invalid.	Do not access addresses in range shown under conditions; or, write same data as in H'00000–H'0FFFF to H'10000–H'1FFFF in on-chip ROM.
Immediate	_	_	No problem	—
Program-counter relative (@(d:8,PC), @(d:16,PC))	Value of PC plus displacement	H'010000 or above, with lower 16 bits in range H'0000 to H'F710	Does not operate normally since instruction code is undefined.	Do not access addresses in range shown under conditions; or, write same data as in H'00000–H'0FFFF to H'10000–H'1FFFF in on-chip ROM.
Memory indirect (@@aa:8)	_	_	No problem	_

# Section 4 Exception Handling

## 4.1 Overview

## 4.1.1 Exception Handling Types and Priority

As table 4.1 indicates, exception handling may be caused by a reset, trap instruction, or interrupt. Exception handling is prioritized as shown in table 4.1. If two or more exceptions occur simultaneously, they are accepted and processed in priority order. Trap instruction exceptions are accepted at all times in the program execution state.

Table 4.1	Exception	Types and	Priority
-----------	-----------	-----------	----------

Priority	Exception Type	Start of Exception Handling
High	Reset	Starts immediately after a low-to-high transition at the RES pin
	Interrupt	Interrupt requests are handled when execution of the current instruction or handling of the current exception is completed
Low	Trap instruction (TRAPA)	Started by execution of a trap instruction (TRAPA)

### 4.1.2 Exception Handling Operation

Exceptions originate from various sources. Trap instructions and interrupts are handled as follows.

- 1. The program counter (PC) and condition code register (CCR) are pushed onto the stack.
- 2. The CCR interrupt mask bit is set to 1.
- 3. A vector address corresponding to the exception source is generated, and program execution starts from the address indicated in the vector address.

For a reset exception, steps 2 and 3 above are carried out.

## 4.1.3 Exception Vector Table

The exception sources are classified as shown in figure 4.1. Different vectors are assigned to different exception sources. Table 4.2 lists the exception sources and their vector addresses.

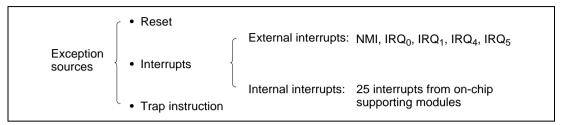


Figure 4.1 Exception Sources

			Vector Address* <sup>1</sup>	
Exception Source		Vector Number	Normal Mode	Advanced Mode
Reset	Reset		H'0000 to H'0001	H'0000 to H'0003
Reserved for syster	n use	1	H'0002 to H'0003	H'0004 to H'0007
		2	H'0004 to H'0005	H'0008 to H'000B
		3	H'0006 to H'0007	H'000C to H'000F
		4	H'0008 to H'0009	H'0010 to H'0013
		5	H'000A to H'000B	H'0014 to H'0017
		6	H'000C to H'000D	H'0018 to H'001B
External interrupt (N	NMI)	7	H'000E to H'000F	H'001C to H'001F
Trap instruction (4 sources)		8	H'0010 to H'0011	H'0020 to H'0023
		9	H'0012 to H'0013	H'0024 to H'0027
		10	H'0014 to H'0015	H'0028 to H'002B
		11	H'0016 to H'0017	H'002C to H'002F
External interrupt	$IRQ_{0}$	12	H'0018 to H'0019	H'0030 to H'0033
	IRQ <sub>1</sub>	13	H'001A to H'001B	H'0034 to H'0037
Reserved for system use		14	H'001C to H'001D	H'0038 to H'003B
		15	H'001E to H'001F	H'003C to H'003F
External interrupt	IRQ₄	16	H'0020 to H'0021	H'0040 to H'0043
	IRQ₅	17	H'0022 to H'0023	H'0044 to H'0047
Reserved for system	n use	18	H'0024 to H'0025	H'0048 to H'004B
		19	H'0026 to H'0027	H'004C to H'004F
Internal interrupts*2		20	H'0028 to H'0029	H'0050 to H'0053
		to 60	to H'0078 to H'0079	to H'00F0 to H'00F3
Natary 4 Lawred	0 1 1		11007010110079	

### Table 4.2Exception Vector Table

Notes: 1. Lower 16 bits of the address.

2. For the internal interrupt vectors, see section 5.3.3, Interrupt Vector Table.

## 4.2 Reset

## 4.2.1 Overview

A reset is the highest-priority exception. When the  $\overline{\text{RES}}$  pin goes low, all processing halts and the H8/3039 Group enters the reset state. A reset initializes the internal state of the CPU and the registers of the on-chip supporting modules. Reset exception handling begins when the  $\overline{\text{RES}}$  pin changes from low to high.

The chip can also be reset by overflow of the watchdog timer. For details see section 10, Watchdog Timer.

## 4.2.2 Reset Sequence

The H8/3039 Group enters the reset state when the  $\overline{\text{RES}}$  pin goes low.

To ensure that the chip is reset, hold the  $\overline{\text{RES}}$  pin low for at least 20 ms at power-up. To reset the chip during operation, hold the  $\overline{\text{RES}}$  pin low for at least 10 system clock ( $\phi$ ) cycles. When using the flash memory version, hold at "Low" level for a least 1usec. See appendix D.2, Pin States at Reset, for the states of the pins in the reset state.

When the  $\overline{\text{RES}}$  pin goes high after being held low for the necessary time, the H8/3039 Group chip starts reset exception handling as follows.

- The internal state of the CPU and the registers of the on-chip supporting modules are initialized, and the I bit is set to 1 in CCR.
- The contents of the reset vector address (H'0000 to H'0003 in advanced mode) are read, and program execution starts from the address indicated in the vector address.

Figure 4.2 shows the reset sequence in modes 5 and 7.

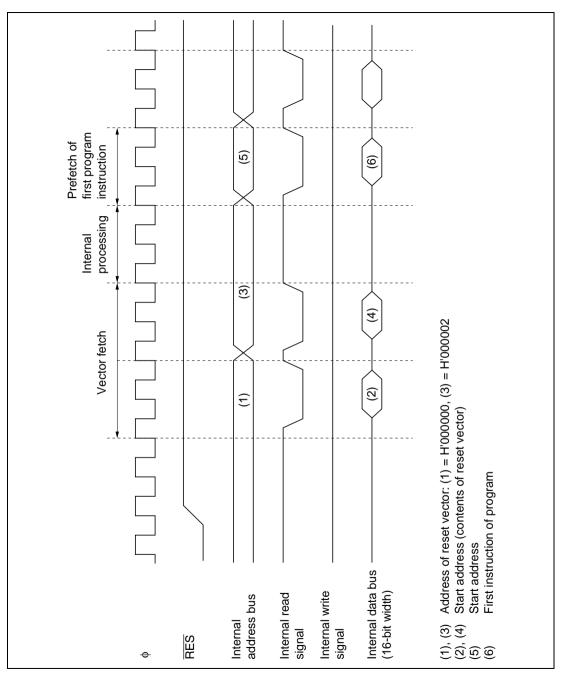


Figure 4.2 Reset Sequence (Modes 5 and 7)

## 4.2.3 Interrupts after Reset

If an interrupt is accepted after a reset but before the stack pointer (SP) is initialized, PC and CCR will not be saved correctly, leading to a program crash. To prevent this, all interrupt requests, including NMI, are disabled immediately after a reset. The first instruction of the program is always executed immediately after the reset state ends. This instruction should initialize the stack pointer (example: MOV.L #xx:32, SP).

## 4.3 Interrupts

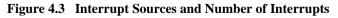
Interrupt exception handling can be requested by five external sources (NMI,  $IRQ_0$ ,  $IRQ_1$ ,  $IRQ_4$ ,  $IRQ_5$ ) and 25 internal sources in the on-chip supporting modules. Figure 4.3 classifies the interrupt sources and indicates the number of interrupts of each type.

The on-chip supporting modules that can request interrupts are the watchdog timer (WDT), 16-bit integrated timer unit (ITU), serial communication interface (SCI), and A/D converter. Each interrupt source has a separate vector address.

NMI is the highest-priority interrupt and is always accepted. Interrupts are controlled by the interrupt controller. The interrupt controller can assign interrupts other than NMI to two priority levels, and arbitrate between simultaneous interrupts. Interrupt priorities are assigned in interrupt priority registers A and B (IPRA and IPRB) in the interrupt controller.

For details on interrupts see section 5, Interrupt Controller.

External interrupts	$\left\{ \begin{array}{l} NMI (1) \\ IRQ_0, IRQ_1, IRQ_4, IRQ_5 (4) \end{array} \right.$			
Interrupts { Internal interrupts	WDT* (1) ITU (15) SCI (8) A/D converter (1)			
Notes: Numbers in parentheses are the number of interrupt sources. * When the watchdog timer is used as an interval timer, it generates an interrupt request at every counter overflow.				



## 4.4 Trap Instruction

Trap instruction exception handling starts when a TRAPA instruction is executed. If the UE bit is set to 1 in the system control register (SYSCR), the exception handling sequence sets the I bit to 1 in CCR. If the UE bit is 0, the I and UI bits are both set to 1. The TRAPA instruction fetches a start address from a vector table entry corresponding to a vector number from 0 to 3, which is specified in the instruction code.

## 4.5 Stack Status after Exception Handling

Figure 4.4 shows the stack after completion of trap instruction exception handling and interrupt exception handling.

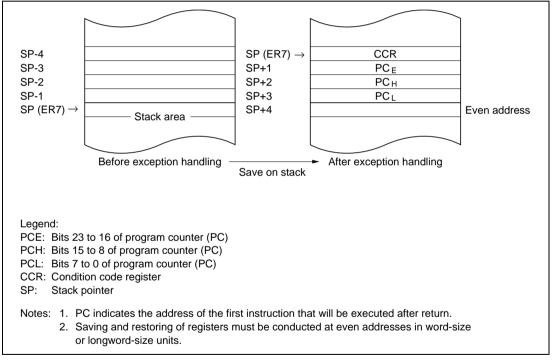


Figure 4.4 Stack after Completion of Exception Handling (Advanced Mode)

## 4.6 Notes on Stack Usage

When accessing word data or longword data, the H8/3039 Group regards the lowest address bit as 0. The stack should always be accessed by word access or longword access, and the value of the stack pointer (SP, ER7) should always be kept even. Use the following instructions to save registers:

PUSH.W Rn (or MOV.W Rn, @-SP) PUSH.L ERn (or MOV.L ERn, @-SP)

Use the following instructions to restore registers:

POP.W Rn (or MOV.W @SP+, Rn) POP.L ERn (or MOV.L @SP+, ERn)

Setting SP to an odd value may lead to a malfunction. Figure 4.5 shows an example of what happens when the SP value is odd.

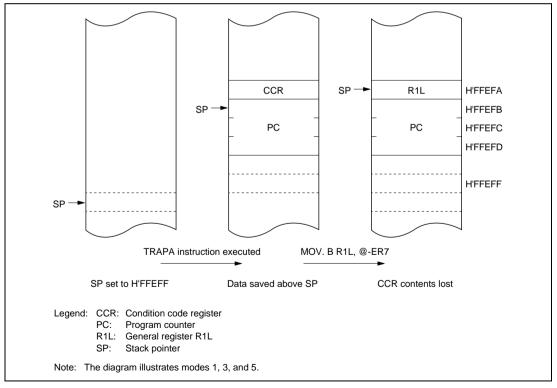


Figure 4.5 Operation when SP Value is Odd

Rev.3.00 Mar. 26, 2007 Page 82 of 682 REJ09B0353-0300

# Section 5 Interrupt Controller

## 5.1 Overview

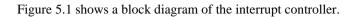
## 5.1.1 Features

The interrupt controller has the following features:

- Interrupt priority registers (IPRs) for setting interrupt priorities Interrupts other than NMI can be assigned to two priority levels on a module-by-module basis in interrupt priority registers A and B (IPRA and IPRB).
- Three-level masking by the I and UI bits in the CPU condition code register (CCR)
- Independent vector addresses All interrupts are independently vectored; the interrupt service routine does not have to identify the interrupt source.
- Five external interrupt pins

NMI has the highest priority and is always accepted; either the rising or falling edge can be selected. For each of  $IRQ_0$ ,  $IRQ_1$ ,  $IRQ_4$ , and  $IRQ_5$ , sensing of the falling edge or level sensing can be selected independently.

### 5.1.2 Block Diagram



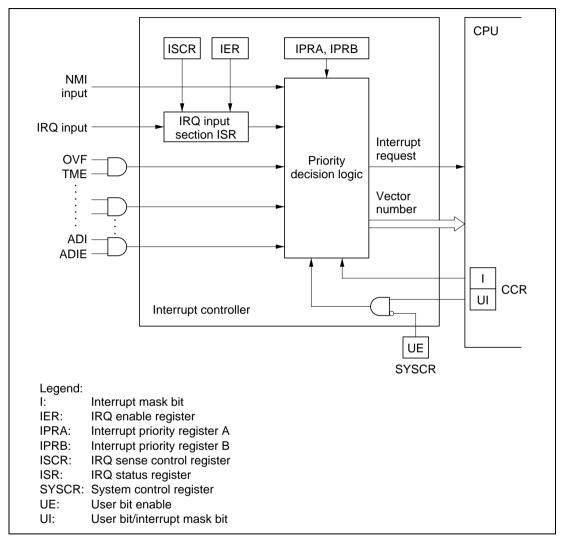


Figure 5.1 Interrupt Controller Block Diagram

#### 5.1.3 Pin Configuration

Table 5.1 lists the interrupt pins.

### Table 5.1Interrupt Pins

Name	Abbreviation	I/O	Function
Nonmaskable interrupt	NMI	Input	Nonmaskable interrupt, rising edge or falling edge selectable
External interrupt request 5, 4, 1, and 0	$\frac{\overline{IRQ}}{\overline{IRQ}}_{1}, \frac{\overline{IRQ}}{\overline{IRQ}}_{1}, \text{ and }$	Input	Maskable interrupts, falling edge or level sensing selectable

### 5.1.4 Register Configuration

Table 5.2 lists the registers of the interrupt controller.

### Table 5.2 Interrupt Controller Registers

Address*1	Name	Abbreviation	R/W	Initial Value
H'FFF2	System control register	SYSCR	R/W	H'0B
H'FFF4	IRQ sense control register	ISCR	R/W	H'00
H'FFF5	IRQ enable register	IER	R/W	H'00
H'FFF6	IRQ status register	ISR	R/(W)* <sup>2</sup>	H'00
H'FFF8	Interrupt priority register A	IPRA	R/W	H'00
H'FFF9	Interrupt priority register B	IPRB	R/W	H'00

Notes: 1. Lower 16 bits of the address.

2. Only 0 can be written, to clear flags.

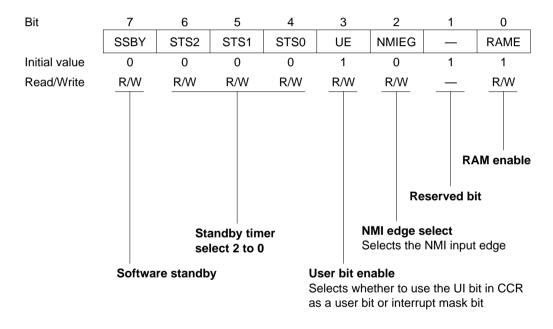
# 5.2 **Register Descriptions**

# 5.2.1 System Control Register (SYSCR)

SYSCR is an 8-bit readable/writable register that controls software standby mode, selects the action of the UI bit in CCR, selects the NMI edge, and enables or disables the on-chip RAM.

Only bits 3 and 2 are described here. For the other bits, see section 3.3, System Control Register (SYSCR).

SYSCR is initialized to H'0B by a reset and in hardware standby mode. It is not initialized in software standby mode.



Bit 3—User Bit Enable (UE): Selects whether to use the UI bit in CCR as a user bit or an interrupt mask bit.

Bit 3 UE	Description	
0	UI bit in CCR is used as interrupt mask bit	
1	UI bit in CCR is used as user bit	(Initial value)

### Bit 2—NMI Edge Select (NMIEG): Selects the NMI input edge.

Bit 2 NMIEG	Description	
0	Interrupt is requested at falling edge of NMI input	(Initial value)
1	Interrupt is requested at rising edge of NMI input	

# 5.2.2 Interrupt Priority Registers A and B (IPRA, IPRB)

IPRA and IPRB are 8-bit readable/writable registers that control interrupt priority.

### **Interrupt Priority Register A (IPRA)**

IPRA is an 8-bit readable/writable register in which interrupt priority levels can be set.

Bit	7	6	5	4	3	2	1	0
	IPRA7	IPRA6	_	IPRA4	IPRA3	IPRA2	IPRA1	IPRA0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
		Priority le	Reserved vel A6	interrupt r I <b>bit</b>	Priority Selects f WDT int evel A4 he priority l equests	Se of int riority lev elects the TU channe level A3 the priority errupt requ evel of IR0	Fiority level elects the ITU chan terrupt req Priority level 0 interrupt level of uests Q <sub>4</sub> and IR0	priority level nel 1 juests vel of pt requests
		Selects the	e priority l	evel of IR	Q <sub>1</sub> interrup	ot requests	;	
	Priority le	evel A7		) into muu				

Selects the priority level of IRQ<sub>0</sub> interrupt requests

# Renesas

IPRA is initialized to H'00 by a reset and in hardware standby mode.

Bit 7—Priority Level A7 (IPRA7): Selects the priority level of IRQ0 interrupt requests.

Bit7 IPRA7	Description	
0	IRQ0 interrupt requests have priority level 0 (low priority)	(Initial value)
1	IRQ0 interrupt requests have priority level 1 (high priority)	

Bit 6—Priority Level A6 (IPRA6): Selects the priority level of IRQ1 interrupt requests.

Bit6 IPRA6	Description	
0	IRQ1 interrupt requests have priority level 0 (low priority)	(Initial value)
1	IRQ1 interrupt requests have priority level 1 (high priority)	

Bit 5—Reserved bit: This bit can be written and read, but it does not affect interrupt priority.

Bit 4—Priority Level A4 (IPRA4): Selects the priority level of IRQ<sub>4</sub> and IRQ<sub>5</sub> interrupt requests.

Bit4 IPRA4	Description	
0	$IRQ_4$ , $IRQ_5$ interrupt requests have priority level 0 (low priority)	(Initial value)
1	$IRQ_4, IRQ_{\!\scriptscriptstyle 5}$ interrupt requests have priority level 1 (high priority)	

Bit 3—Priority Level A3 (IPRA3): Selects the priority level of WTD interrupt requests.

Bit3 IPRA3	Description	
0	WDT interrupt requests have priority level 0 (low priority)	(Initial value)
1	WDT interrupt requests have priority level 1 (high priority)	

### Renesas

Bit 2—Priority Level A2 (IPRA2): Selects the priority level of ITU channel 0 interrupt requests.

Bit2 IPRA2	Description	
0	ITU channel 0 interrupt requests have priority level 0 (low priority)	(Initial value)
1	ITU channel 0 interrupt requests have priority level 1 (high priority)	

Bit 1—Priority Level A1 (IPRA1): Selects the priority level of ITU channel 1 interrupt requests.

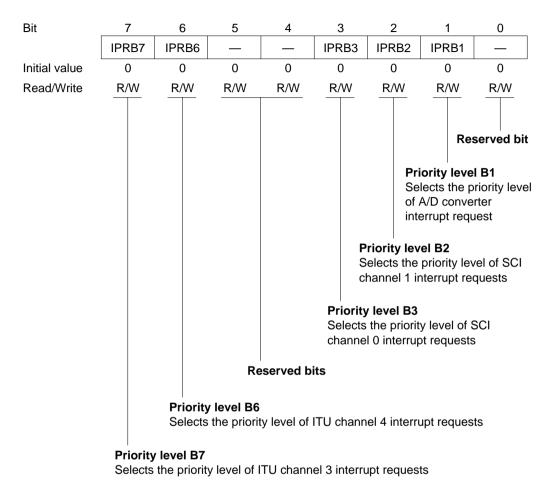
Bit1 IPRA1	Description	
0	ITU channel 1 interrupt requests have priority level 0 (low priority)	(Initial value)
1	ITU channel 1 interrupt requests have priority level 1 (high priority)	

Bit 0—Priority Level A0 (IPRA0): Selects the priority level of ITU channel 2 interrupt requests.

Bit0 IPRA0	Description	
0	ITU channel 2 interrupt requests have priority level 0 (low priority)	(Initial value)
1	ITU channel 2 interrupt requests have priority level 1 (high priority)	

### **Interrupt Priority Register B (IPRB)**

IPRB is an 8-bit readable/writable register in which interrupt priority levels can be set.



IPRB is initialized to H'00 by a reset and in hardware standby mode.

Bit 7—Priority Level B7 (IPRB7): Selects the priority level of ITU channel 3 interrupt requests.

Bit7 IPRB7	Description	
0	ITU channel 3 interrupt requests have priority level 0 (low priority)	(Initial value)
1	ITU channel 3 interrupt requests have priority level 1 (high priority)	

Bit 6—Priority Level B6 (IPRB6): Selects the priority level of ITU channel 4 interrupt requests.

Bit6 IPRB6	Description	
0	ITU channel 4 interrupt requests have priority level 0 (low priority)	(Initial value)
1	ITU channel 4 interrupt requests have priority level 1 (high priority)	

Bits 5 and 4—Reserved: These bits cannot be modified and are always read as 0.

Bit 3—Priority Level B3 (IPRB3): Selects the priority level of SCI channel 0 interrupt requests.

Bit3 IPRB3	Description	
0	SCI channel 0 interrupt requests have priority level 0 (low priority)	(Initial value)
1	SCI channel 0 interrupt requests have priority level 1 (high priority)	

Bit 2—Priority Level B2 (IPRB2): Selects the priority level of SCI channel 1 interrupt requests.

Bit2 IPRB2	Description	
0	SCI channel 1 interrupt requests have priority level 0 (low priority)	(Initial value)
1	SCI channel 1 interrupt requests have priority level 1 (high priority)	

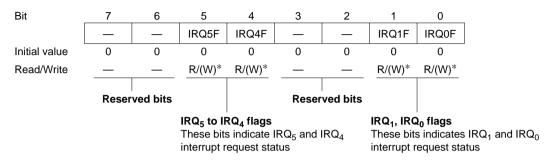
Bit 1—Priority Level B1 (IPRB1): Selects the priority level of A/D converter interrupt requests.

Bit1 IPRB1	Description	
0	A/D converter interrupt requests have priority level 0 (low priority)	(Initial value)
1	A/D converter interrupt requests have priority level 1 (high priority)	

Bit 0—Reserved: This bit cannot be modified and is always read as 0.

### 5.2.3 IRQ Status Register (ISR)

ISR is an 8-bit readable/writable register that indicates the status of  $IRQ_0$ ,  $IRQ_1$ ,  $IRQ_4$ , and  $IRQ_5$  interrupt requests.



Note: \* Only 0 can be written, to clear flags.

ISR is initialized to H'00 by a reset and in hardware standby mode.

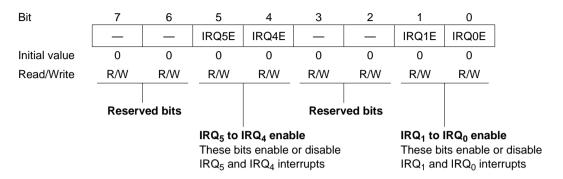
Bits 7, 6, 3 and 2—Reserved: These bits cannot be modified and are always read as 0.

**Bits 5, 4, 1 and 0—IRQ<sub>5</sub>, IRQ<sub>4</sub>, IRQ<sub>1</sub> and IRQ<sub>0</sub> Flags (IRQ5F, IRQ4F, IRQ1F, and IRQ0F):** These bits indicate the status of IRQ<sub>5</sub>, IRQ<sub>4</sub>, IRQ<sub>1</sub>, and IRQ<sub>0</sub> interrupt requests.

Bits 5, 4, 1, and 0 IRQ5F, IRQ4F,		
IRQ1F, and IRQ0F	Description	
0	[Clearing conditions]	(Initial value)
	0 is written in IRQnF after reading the IRQnF flag when	IRQnF = 1.
	• IRQnSC = 0, $\overline{IRQn}$ input is high, and interrupt exception	handling is
	carried out.	
	• IRQnSC = 1 and IRQn interrupt exception handling is ca	arried out.
1	[Setting conditions]	
	• IRQnSC = 0 and $\overline{IRQn}$ input is low.	
	• IRQnSC = 1 and IRQn input changes from high to low.	
Note: n = 5, 4, 1 and	10	

# 5.2.4 IRQ Enable Register (IER)

IER is an 8-bit readable/writable register that enables or disables  $IRQ_0$ ,  $IRQ_1$ ,  $IRQ_4$ , and  $IRQ_5$  interrupt requests.



IER is initialized to H'00 by a reset and in hardware standby mode.

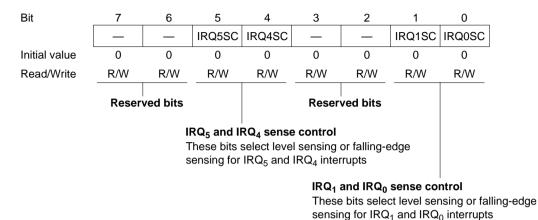
Bits 7, 6, 3, and 2—Reserved: These bits cannot be modified and are always read as 0.

**Bits 5, 4, 1, and 0—IRQ<sub>5</sub>, IRQ<sub>4</sub>, IRQ<sub>1</sub>, and IRQ<sub>0</sub> Enable (IRQ5E, IRQ4E, IRQ1E, IRQ0E):** These bits enable or disable IRQ5, IRQ4, IRQ1, IRQ0 interrupts.

Bits 5, 4, 1, and 0 IRQ5E, IRQ4E, IRQ1E, and IRQ0E	Description	
0	$IRQ_{5}$ , $IRQ_{4}$ , $IRQ_{1}$ , $IRQ_{0}$ interrupts are disabled	(Initial value)
1	$IRQ_5$ , $IRQ_4$ , $IRQ_1$ , $IRQ_0$ interrupts are enabled	

### 5.2.5 IRQ Sense Control Register (ISCR)

ISCR is an 8-bit readable/writable register that selects level sensing or falling-edge sensing of the inputs at pins  $\overline{IRQ}_5$ ,  $\overline{IRQ}_4$ ,  $\overline{IRQ}_1$ , and  $\overline{IRQ}_0$ 



ISCR is initialized to H'00 by a reset and in hardware standby mode.

Bits 7, 6, 3, and 2—Reserved: These bits are readable/writable and do not affect selection of level sensing or falling-edge sensing.

Bits 5, 4, 1, and 0—IRQ<sub>5</sub>, IRQ<sub>4</sub>, IRQ<sub>1</sub>, and IRQ<sub>0</sub> Sense Control (IRQ5SC, IRQ4SC, IRQ1SC, IRQ0SC): These bits selects whether interrupts  $IRQ_5$ ,  $IRQ_4$ ,  $IRQ_1$ ,  $IRQ_0$  are requested by level sensing of pins  $\overline{IRQ}_5$ ,  $\overline{IRQ}_4$ ,  $\overline{IRQ}_1$ ,  $\overline{IRQ}_0$  or by falling-edge sensing.

Bits 5, 4, 1, and 0 IRQ5SC, IRQ4SC, IRQ1SC, IRQ0SC	Description
0	Interrupts are requested when $\overline{IRQ}_{s}$ , $\overline{IRQ}_{4}$ , $\overline{IRQ}_{1}$ , $\overline{IRQ}_{0}$ inputs are low
	(Initial value)
1	Interrupts are requested by falling-edge input at $\overline{IRQ}_{5}$ , $\overline{IRQ}_{4}$ , $\overline{IRQ}_{1}$ , $\overline{IRQ}_{0}$

# 5.3 Interrupt Sources

The interrupt sources include external interrupts (NMI,  $IRQ_5$ ,  $IRQ_4$ ,  $IRQ_1$  and  $IRQ_0$ ) and 25 internal interrupts.

# 5.3.1 External Interrupts

There are five external interrupts: NMI, and  $IRQ_5$ ,  $IRQ_4$ ,  $IRQ_1$ , and  $IRQ_0$ . Of these, NMI, IRQ0, IRQ1, can be used to exit software standby mode.

**NMI:** NMI is the highest-priority interrupt and is always accepted, regardless of the states of the I and UI bits in CCR. The NMIEG bit in SYSCR selects whether an interrupt is requested by the rising or falling edge of the input at the NMI pin. NMI interrupt exception handling has vector number 7.

**IRQ**<sub>5</sub>, **IRQ**<sub>4</sub>, **IRQ**<sub>1</sub>, **IRQ**<sub>0</sub> **Interrupts:** These interrupts are requested by input signals at pins  $\overline{\text{IRQ}}_5$ ,  $\overline{\text{IRQ}}_4$ ,  $\overline{\text{IRQ}}_1$ ,  $\overline{\text{IRQ}}_0$ . The IRQ<sub>5</sub>, IRQ<sub>4</sub>, IRQ<sub>1</sub>, IRQ<sub>0</sub> interrupts have the following features.

- ISCR settings can select whether an interrupt is requested by the low level of the input at pins  $\overline{IRQ}_{5}$ ,  $\overline{IRQ}_{4}$ ,  $\overline{IRQ}_{1}$ ,  $\overline{IRQ}_{0}$ , or by the falling edge.
- IER settings can enable or disable the IRQ<sub>5</sub>, IRQ<sub>4</sub>, IRQ<sub>1</sub>, IRQ<sub>0</sub> interrupts. Interrupt priority levels can be assigned by four bits in IPRA (IPRA7, IPRA6, and IPRA4).
- The status of IRQ<sub>5</sub>, IRQ<sub>4</sub>, IRQ<sub>1</sub>, IRQ<sub>0</sub> interrupt requests is indicated in ISR. The ISR flags can be cleared to 0 by software.

Figure 5.2 shows a block diagram of interrupts IRQ<sub>5</sub>, IRQ<sub>4</sub>, IRQ<sub>1</sub>, IRQ<sub>0</sub>.

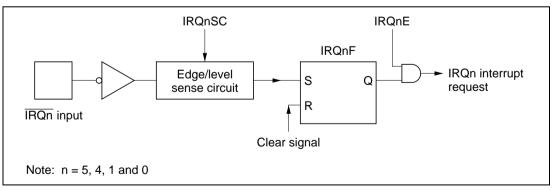


Figure 5.2 Block Diagram of Interrupts IRQ<sub>5</sub>, IRQ<sub>4</sub>, IRQ<sub>1</sub>, and IRQ<sub>0</sub>

Figure 5.3 shows the timing of the setting of the interrupt flags (IRQnF).

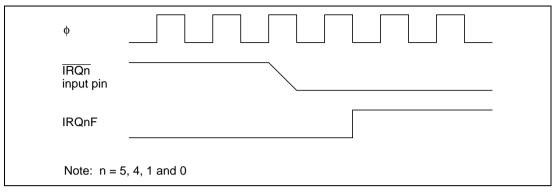


Figure 5.3 Timing of Setting of IRQnF

Interrupts  $IRQ_5$ ,  $IRQ_4$ ,  $IRQ_1$ ,  $IRQ_0$  have vector numbers 17, 16, 13, 12. These interrupts are detected regardless of whether the corresponding pin is set for input or output. When using a pin for external interrupt input, clear its DDR bit to 0 and do not use the pin for SCI input or output.

### 5.3.2 Internal Interrupts

Twenty-five internal interrupts are requested from the on-chip supporting modules.

- Each on-chip supporting module has status flags for indicating interrupt status, and enable bits for enabling or disabling interrupts.
- Interrupt priority levels can be assigned in IPRA and IPRB.

### 5.3.3 Interrupt Vector Table

Table 5.3 lists the interrupt sources, their vector addresses, and their default priority order. In the default priority order, smaller vector numbers have higher priority. The priority of interrupts other than NMI can be changed in IPRA and IPRB. The priority order after a reset is the default order shown in table 5.3.

		Vector	Vec	tor Address*		
Interrupt Source	Origin	Number	Normal Mode	Advanced Mode	IPR	- Priority
NMI	External pins	7	H'000E to H'000F	H'001C to H'001F	_	High
IRQ <sub>0</sub>		12	H'0018 to H'0019	H'0030 to H'0033	IPRA7	<sup>−</sup> ↑
IRQ,		13	H'001A to H'001B	H'0034 to H0037	IPRA6	-
Reserved	_	14	H'001C to H'001D	H'0038 to H'003B	_	-
		15	H'001E to H'001F	H'003C to H'003F		
IRQ <sub>4</sub>	External pins	16	H'0020 to H'0021	H'0040 to H'0043	IPRA4	-
IRQ₅		17	H'0022 to H'0023	H'0044 to H'0047	_	
Reserved	_	18	H'0024 to H'0025	H'0048 to H'004B	_	
		19	H'0026 to H'0027	H'004C to H'004F		
WOVI (interval timer)	Watchdog timer	20	H'0028 to H'0029	H'0050 to H'0053	IPRA3	_
Reserved	_	21	H'002A to H'002B	H'0054 to H'0057	_	
		22	H'002C to H'002D	H'0058 to H'005B		
		23	H'002E to H'002F	H'005C to H'005F		
IMIA0 (compare match/ input capture A0)	ITU channel 0	24	H'0030 to H'0031	H'0060 to H'0063	IPRA2	-
IMIB0 (compare match/ input capture B0)		25	H'0032 to H'0033	H'0064 to H'0067		
OVI0 (overflow 0)		26	H'0034 to H'0035	H'0068 to H'006B		
Reserved	_	27	H'0036 to H'0037	H'006C to H'006F	_	
IMIA1 (compare match/ input capture A1)	ITU channel 1	28	H'0038 to H'0039	H'0070 to H'0073	IPRA1	-
IMIB1 (compare match/ input capture B1)		29	H'003A to H'003B	H'0074 to H'0077		
OVI1 (overflow 1)		30	H'003C to H'003D	H'0078 to H'007B		
Reserved	_	31	H'003E to H'003F	H'007C to H'007F	_	
IMIA2 (compare match/ input capture A2)	ITU channel 2	32	H'0040 to H'0041	H'0080 to H'0083	IPRA0	-
IMIB2 (compare match/ input capture B2)		33	H'0042 to H'0043	H'0084 to H'0087		
OVI2 (overflow 2)		34	H'0044 to H'0045	H'0088 to H'008B		
Reserved	_	35	H'0046 to H'0047	H'008C to H'008F	_	

### Table 5.3 Interrupt Sources, Vector Addresses, and Priority

		Vector		tor Address*		
Interrupt Source	Origin	Number	Normal Mode	Advanced Mode	IPR	- Priority
IMIA3 (compare match/ input capture A3)	ITU channel 3	36	H'0048 to H'0049	H'0090 to H'0093	IPRB7	1
IMIB3 (compare match/ input capture B3)		37	H'004A to H'004B	H'0094 to H'0097		
OVI3 (overflow 3)		38	H'004C to H'004D	H'0098 to H'009B		
Reserved	_	39	H'004E to H'004F	H'009C to H'009F	-	
IMIA4 (compare match/ input capture A4)	ITU channel 4	40	H'0050 to H'0051	H'00A0 to H'00A3	IPRB6	-
IMIB4 (compare match/ input capture B4)		41	H'0052 to H'0053	H'00A4 to H'00A7		
OVI4 (overflow 4)		42	H'0054 to H'0055	H'00A8 to H'00AB		
Reserved	_	43	H'0056 to H'0057	H'00AC to H'00AF	_	-
		44	H'0058 to H'0059	H'00B0 to H'00B3		
		45	H'005A to H'005B	H'00B4 to H'00B7		
		46	H'005C to H'005D	H'00B8 to H'00BB		
		47	H'005E to H'005F	H'00BC to H'00BF		
		48	H'0060 to H'0061	H'00C0 to H'00C3		
		49	H'0062 to H'0063	H'00C4 to H'00C7		
		50	H'0064 to H'0065	H'00C8 to H'00CB		
		51	H'0066 to H'0067	H'00CC to H'00CF		
ERI0 (receive error 0)	SCI channel 0	52	H'0068 to H'0069	H'00D0 to H'00D3	IPRB3	-
RXI0 (receive data full 0)		53	H'006A to H'006B	H'00D4 to H'00D7		
TXI0 (transmit data empty 0)		54	H'006C to H'006D	H'00D8 to H'00DB		
TEI0 (transmit end 0)		55	H'006E to H'006F	H'00DC to H'00DF		
ERI1 (receive error 1)	SCI channel 1	56	H'0070 to H'0071	H'00E0 to H'00E3	IPRB2	-
RXI1 (receive data full 1)		57	H'0072 to H'0073	H'00E4 to H'00E7		
TXI1 (transmit data empty 1)		58	H'0074 to H'0075	H'00E8 to H'00EB		
TEI1 (transmit end 1)		59	H'0076 to H'0077	H'00EC to H'00EF		
ADI (A/D end)	A/D	60	H'0078 to H'0079	H'00F0 to H'00F3	IPRB1	Low

Note: \* Lower 16 bits of the address.

# 5.4 Interrupt Operation

# 5.4.1 Interrupt Handling Process

The H8/3039 Group handles interrupts differently depending on the setting of the UE bit. When UE = 1, interrupts are controlled by the I bit. When UE = 0, interrupts are controlled by the I and UI bits. Table 5.4 indicates how interrupts are handled for all setting combinations of the UE, I, and UI bits.

NMI interrupts are always accepted except in the reset and hardware standby states. IRQ interrupts and interrupts from the on-chip supporting modules have their own enable bits. Interrupt requests are ignored when the enable bits are cleared to 0.

SYSCR	CCR I UI		
UE			Description
1	0	_	All interrupts are accepted. Interrupts with priority level 1 have higher priority.
	1	_	No interrupts are accepted except NMI.
0	0	_	All interrupts are accepted. Interrupts with priority level 1 have higher priority.
	1	0	NMI and interrupts with priority level 1 are accepted.
		1	No interrupts are accepted except NMI.

Table 5.4 UE, I, and UI Bit Settings and Interrupt Handling

# **UE** = 1

Interrupts  $IRQ_0$ ,  $IRQ_1$ ,  $IRQ_4$ , and  $IRQ_5$  and interrupts from the on-chip supporting modules can all be masked by the I bit in the CPU's CCR. Interrupts are masked when the I bit is set to 1, and unmasked when the I bit is cleared to 0. Interrupts with priority level 1 have higher priority. Figure 5.4 is a flowchart showing how interrupts are accepted when UE = 1.

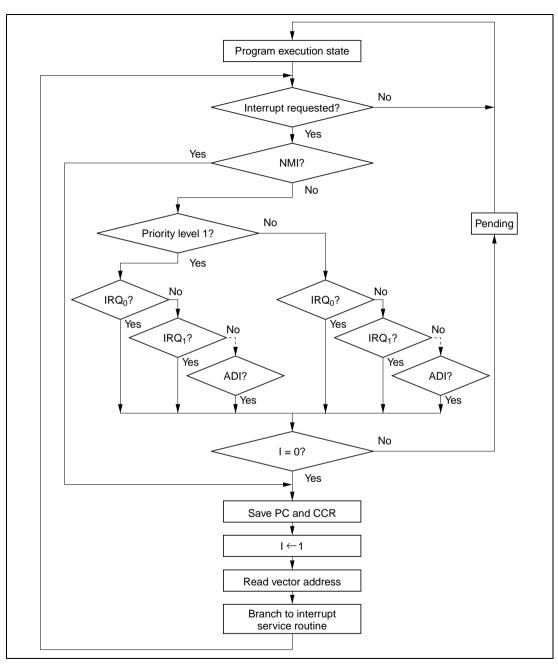


Figure 5.4 Process Up to Interrupt Acceptance when UE = 1

- If an interrupt condition occurs and the corresponding interrupt enable bit is set to 1, an interrupt request is sent to the interrupt controller.
- When the interrupt controller receives one or more interrupt requests, it selects the highestpriority request, following the IPR interrupt priority settings, and holds other requests pending. If two or more interrupts with the same IPR setting are requested simultaneously, the interrupt controller follows the priority order shown in table 5.3.
- The interrupt controller checks the I bit. If the I bit is cleared to 0, the selected interrupt request is accepted. If the I bit is set to 1, only NMI is accepted; other interrupt requests are held pending.
- When an interrupt request is accepted, interrupt exception handling starts after execution of the current instruction has been completed.
- In interrupt exception handling, PC and CCR are saved to the stack area. The PC value that is saved indicates the address of the first instruction that will be executed after the return from the interrupt service routine.
- Next the I bit is set to 1 in CCR, masking all interrupts except NMI.
- The vector address of the accepted interrupt is generated, and the interrupt service routine starts executing from the address indicated by the contents of the vector address.

# UE = 0

The I and UI bits in the CPU's CCR and the IPR bits enable three-level masking of  $IRQ_0$ ,  $IRQ_1$ ,  $IRQ_4$ , and  $IRQ_5$  interrupts and interrupts from the on-chip supporting modules.

- Interrupt requests with priority level 0 are masked when the I bit is set to 1, and are unmasked when the I bit is cleared to 0.
- Interrupt requests with priority level 1 are masked when the I and UI bits are both set to 1, and are unmasked when either the I bit or the UI bit is cleared to 0.

For example, if the interrupt enable bits of all interrupt requests are set to 1, IPRA is set to H'10, and IPRB is set to H'00 (giving  $IRQ_4$  and  $IRQ_5$  interrupt requests priority over other interrupts), interrupts are masked as follows:

- a. If I = 0, all interrupts are unmasked (priority order:  $NMI > IRQ_4 > IRQ_5 > IRQ0 ...$ ).
- b. If I = 1 and UI = 0, only NMI,  $IRQ_4$ , and  $IRQ_5$  are unmasked.
- c. If I = 1 and UI = 1, all interrupts are masked except NMI.

Figure 5.5 shows the transitions among the above states.

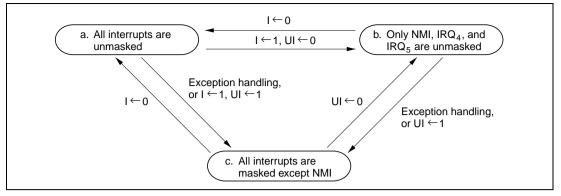


Figure 5.5 Interrupt Masking State Transitions (Example)

Figure 5.6 is a flowchart showing how interrupts are accepted when UE = 0.

- If an interrupt condition occurs and the corresponding interrupt enable bit is set to 1, an interrupt request is sent to the interrupt controller.
- When the interrupt controller receives one or more interrupt requests, it selects the highestpriority request, following the IPR interrupt priority settings, and holds other requests pending. If two or more interrupts with the same IPR setting are requested simultaneously, the interrupt controller follows the priority order shown in table 5.3.
- The interrupt controller checks the I bit. If the I bit is cleared to 0, the selected interrupt request is accepted regardless of its IPR setting, and regardless of the UI bit. If the I bit is set to 1 and the UI bit is cleared to 0, only NMI and interrupts with priority level 1 are accepted; interrupt requests with priority level 0 are held pending. If the I bit and UI bit are both set to 1, only NMI is accepted; all other interrupt requests are held pending.
- When an interrupt request is accepted, interrupt exception handling starts after execution of the current instruction has been completed.
- In interrupt exception handling, PC and CCR are saved to the stack area. The PC value that is saved indicates the address of the first instruction that will be executed after the return from the interrupt service routine.
- The I and UI bits are set to 1 in CCR, masking all interrupts except NMI.
- The vector address of the accepted interrupt is generated, and the interrupt service routine starts executing from the address indicated by the contents of the vector address.

### Renesas

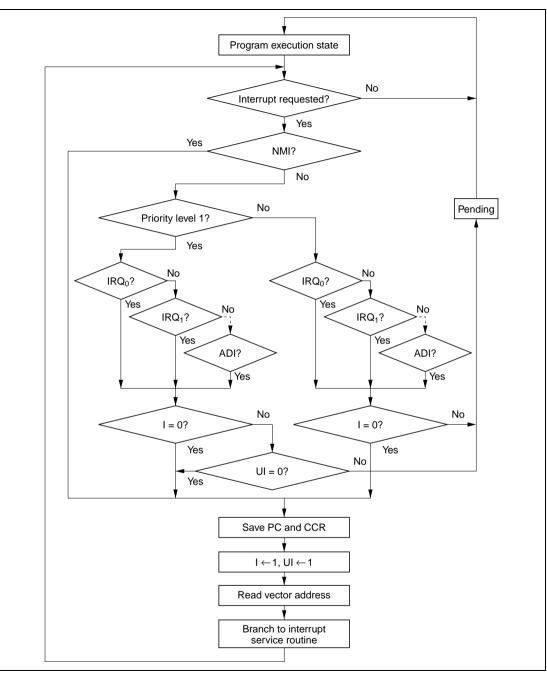
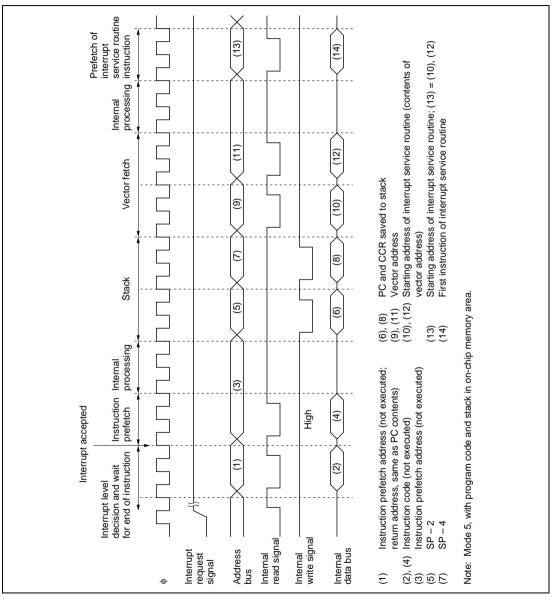


Figure 5.6 Process Up to Interrupt Acceptance when UE = 0

## 5.4.2 Interrupt Sequence

Figure 5.7 shows the interrupt sequence in mode 5 when the program code and stack are in an onchip memory area.





Renesas

# 5.4.3 Interrupt Response Time

Table 5.5 indicates the interrupt response time from the occurrence of an interrupt request until the first instruction of the interrupt service routine is executed.

### Table 5.5 Interrupt Response Time

			External Memory	
		On-Chip	8-Bit Bus	
No.	Item	Memory	2 States	3 States
1	Interrupt priority decision	<b>2</b> * <sup>1</sup>	<b>2</b> * <sup>1</sup>	<b>2</b> * <sup>1</sup>
2	Maximum number of states	1 to 23	1 to 27	1 to 31*4
	until end of current instruction			
3	Saving PC and CCR to stack	4	8	12* <sup>4</sup>
4	Vector fetch	4	8	12* <sup>4</sup>
5	Instruction prefetch*2	4	8	12* <sup>4</sup>
6	Internal processing*3	4	4	4
Total		19 to 41	31 to 57	43 to 73

Notes: 1. 1 state for internal interrupts.

2. Prefetch after the interrupt is accepted and prefetch of the first instruction in the interrupt service routine.

3. Internal processing after the interrupt is accepted and internal processing after prefetch.

4. The number of states increases if wait states are inserted in external memory access.

# 5.5 Usage Notes

# 5.5.1 Contention between Interrupt and Interrupt-Disabling Instruction

When an instruction clears an interrupt enable bit to 0 to disable the interrupt, the interrupt is not disabled until after execution of the instruction is completed. If an interrupt occurs while a BCLR, MOV, or other instruction is being executed to clear its interrupt enable bit to 0, at the instant when execution of the instruction ends the interrupt is still enabled, so its interrupt exception handling is carried out. If a higher-priority interrupt is also requested, however, interrupt exception handling for the higher-priority interrupt is carried out, and the lower-priority interrupt is ignored. This also applies to the clearing of an interrupt flag.

Figure 5.8 shows an example in which an IMIEA bit is cleared to 0 in the ITU's TIER.

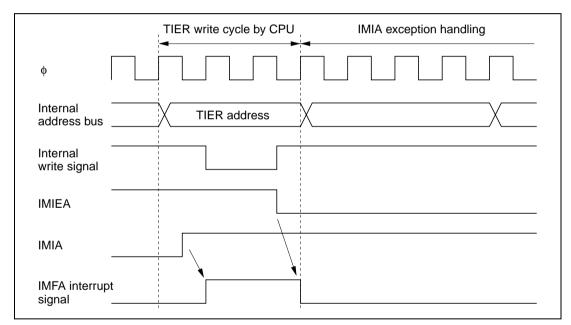


Figure 5.8 Contention between Interrupt and Interrupt-Disabling Instruction

This type of contention will not occur if the interrupt is masked when the interrupt enable bit or flag is cleared to 0.

# 5.5.2 Instructions that Inhibit Interrupts

The LDC, ANDC, ORC, and XORC instructions inhibit interrupts. When an interrupt occurs, after determining the interrupt priority, the interrupt controller requests a CPU interrupt. If the CPU is currently executing one of these interrupt-inhibiting instructions, however, when the instruction is completed the CPU always continues by executing the next instruction.

# 5.5.3 Interrupts during EEPMOV Instruction Execution

The EEPMOV.B and EEPMOV.W instructions differ in their reaction to interrupt requests.

When the EEPMOV.B instruction is executing a transfer, no interrupts are accepted until the transfer is completed, not even NMI.

When the EEPMOV.W instruction is executing a transfer, interrupt requests other than NMI are not accepted until the transfer is completed. If NMI is requested, NMI exception handling starts at a transfer cycle boundary. The PC value saved on the stack is the address of the next instruction. Programs should be coded as follows to allow for NMI interrupts during EEPMOV.W execution:

L1: EEPMO V.W MOV.W R4, R4 BNE L1

### 5.5.4 Usage Notes

The IRQnF flag specification calls for the flag to be cleared by writing 0 to it after it has been read while set to 1. However, it is possible for the IRQnF flag to be cleared by mistake simply by writing 0 to it, irrespective of whether it has been read while set to 1, with the result that interrupt exception handling is not executed. This will occur when the following conditions are met.

# 1. Setting Conditions

- (1) Multiple external interrupts (IRQa, IRQb) are being used.
- (2) Different clearing methods are being used: clearing by writing 0 for the IRQaF flag, and clearing by hardware for the IRQbF flag.
- (3) A bit-manipulation instruction is used on the IRQ status register for clearing the IRQaF flag, or else ISR is read as a byte unit, the IRQaF flag bit is cleared, and the values read in the other bits are written as a byte unit.

### 2. Generation Conditions

- (1) A read of the ISR register is executed to clear the IRQaF flag while it is set to 1, then the IRQbF flag is cleared by the execution of interrupt exception handling.
- (2) When the IRQaF flag is cleared, there is contention with IRQb generation (IRQaF flag setting). (IRQbF was 0 when ISR was read to clear the IRQaF flag, but IRQbF is set to 1 before ISR is written to.)

If the above setting conditions (1) to (3) and generation conditions (1) and (2) are all fulfilled, when the ISR write in generation condition (2) is performed the IRQbF flag will be cleared inadvertently, and interrupt exception handling will not be executed.

However, this inadvertent clearing of the IRQbF flag will not occur if 0 is written to this flag even once between generation conditions (1) and (2).

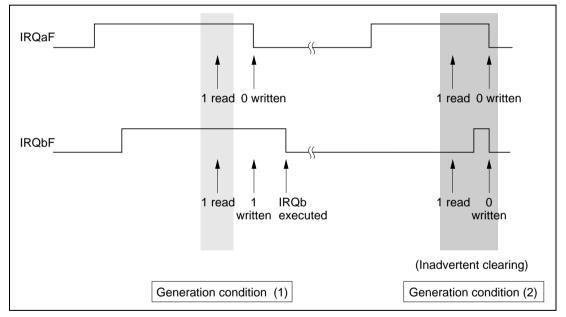


Figure 5.9 IRQnF Flag when Interrupt Exception Handling is not Executed

### Renesas

Either of the methods shown below should be used to prevent this problem.

**Method 1:** When clearing the IRQaF flag, read ISR as a byte unit instead of using a bitmanipulation instruction, and write a byte value that clears the IRQaF flag to 0 and sets the other bits to 1.

Example: When a = 0

MOV.B @ISR, R0L MOV.B #HFE, R0L MOV.B R0L, @ISR

**Method 2:** Perform dummy processing within the IRQb interrupt exception handling routine to clear the IRQbF flag.

Example: When b = 1

IRQB MOV.B #HFD, R0L MOV.B R0L, @ISR

# Section 6 Bus Controller

# 6.1 Overview

The H8/3039 Group has an on-chip bus controller that divides the external address space into eight areas and can assign different bus specifications to each. This enables different types of memory to be connected easily.

### 6.1.1 Features

Features of the bus controller are listed below.

- Independent settings for address areas 0 to 7
  - 128-kbyte areas in 1-Mbyte mode.
  - 2-Mbyte areas in 16-Mbyte mode.
  - Areas can be designated for two-state or three-state access.
- Four wait modes
  - Programmable wait mode, pin auto-wait mode, and pin wait modes 0 and 1 can be selected.
  - Zero to three wait states can be inserted automatically.

# Renesas

### 6.1.2 Block Diagram

Figure 6.1 shows a block diagram of the bus controller.

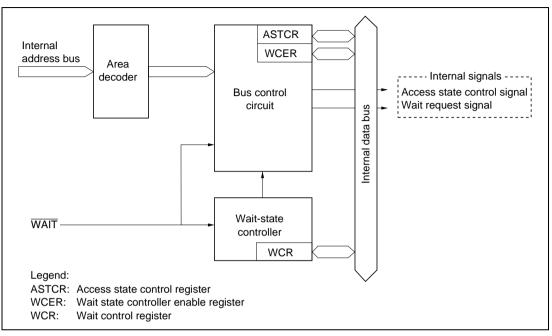


Figure 6.1 Block Diagram of Bus Controller

### 6.1.3 Input/Output Pins

Table 6.1 summarizes the bus controller's input/output pins.

Table 6.1	<b>Bus Controller</b>	Pins
-----------	-----------------------	------

Name	Abbreviation	I/O	Function
Address strobe	ĀS	Output	Strobe signal indicating valid address output on the address bus
Read	RD	Output	Strobe signal indicating reading from the external address space
Write	WR	Output	Strobe signal indicating writing to the external address space, with valid data on the data bus(D7 to D0)
Wait	WAIT	Input	Wait request signal for access to external three- state-access areas

### 6.1.4 Register Configuration

Table 6.2 summarizes the bus controller's registers.

### Table 6.2Bus Controller Registers

Address*	Name	Abbreviation	R/W	Initial Value
H'FFED	Access state control register	ASTCR	R/W	H'FF
H'FFEE	Wait control register	WCR	R/W	H'F3
H'FFEF	Wait state controller enable register	WCER	R/W	H'FF
H'FFF3	Address control register	ADRCR	R/W	H'FE

Note: \* Lower 16 bits of the address.

# 6.2 **Register Descriptions**

# 6.2.1 Access State Control Register (ASTCR)

ASTCR is an 8-bit readable/writable register that selects whether each area is accessed in two states or three states.

Bit	7	6	5	4	3	2	1	0
	AST7	AST6	AST5	AST4	AST3	AST2	AST1	AST0
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W							

Bits selecting number of states for access to each area

ASTCR is initialized to H'FF by a reset and in hardware standby mode. It is not initialized in software standby mode.

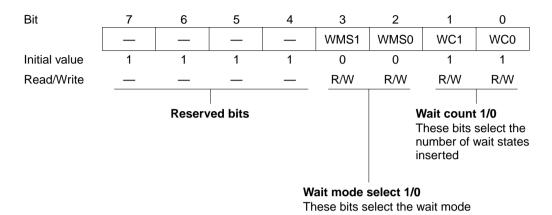
Bits 7 to 0—Area 7 to 0 Access State Control (AST7 to AST0): These bits select whether the corresponding area is accessed in two or three states.

Bits 7 to 0 AST7 to AST0	Description	
0	Areas 7 to 0 are accessed in two states	
1	Areas 7 to 0 are accessed in three states	(Initial value)

ASTCR specifies the number of states in which external areas are accessed. On-chip memory and registers are accessed in a fixed number of states that does not depend on ASTCR settings. Therefore, in the single-chip modes (modes 6 and 7), the set value is meaningless.

### 6.2.2 Wait Control Register (WCR)

WCR is an 8-bit readable/writable register that selects the wait mode for the wait-state controller (WSC) and specifies the number of wait states.



WCR is initialized to H'F3 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 to 4—Reserved: These bits cannot be modified and are always read as 1.

Bit3 WMS1	Bit2 WMS0	Description	
0	0	Programmable wait mode	(Initial value)
	1	No wait states inserted by wait-state controller	
1	0	Pin wait mode 1	
	1	Pin auto-wait mode	

Bits 1 and 0—Wait Count 1 and 0 (WC1/0): These bits select the number of wait states inserted in access to external three-state-access areas.

Bit1 WC1	Bit0 WC0	Description	
0	0	No wait states inserted by wait-state controller	
	1	1 state inserted	
1	0	2 states inserted	
	1	3 states inserted	(Initial value)

#### 6.2.3 Wait State Controller Enable Register (WCER)

WCER is an 8-bit readable/writable register that enables or disables wait-state control of external three-state-access areas by the wait-state controller.

Bit	7	6	5	4	3	2	1	0
	WCE7	WCE6	WCE5	WCE4	WCE3	WCE2	WCE1	WCE0
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W							

Wait state controller enable 7 to 0

These bits enable or disable wait-state control

WCER is initialized to H'FF by a reset and in hardware standby mode. It is not initialized in software standby mode.

**Bits 7 to 0—Wait-State Controller Enable 7 to 0 (WCE7 to WCE0):** These bits enable or disable wait-state control of external three-state-access areas.

Bits 7 to 0 WCE7 to WCE0	Description	
0	Wait-state control disabled (pin wait mode 0)	
1	Wait-state control enabled	(Initial value)

WCER enables or disables wait-state control of external three-state-access areas. Therefore, in the single-chip modes (modes 6 and 7), the set value is meaningless.

### 6.2.4 Address Control Register (ADRCR)

Bit	7	6	5	4	3	2	1	0
	A <sub>23</sub> E	A <sub>22</sub> E	A <sub>21</sub> E	—	_	_	_	—
Modes 1 (Initial value	1	1	1	1	1	1	1	0
and 5 to 7 (Read/Write	—	—	—	—	—	—	—	R/W
Mode 3	1	1	1	1	1	1	1	0
Read/Write	R/W	R/W	R/W	—	—	—	—	R/W
	Address 23 to 21 enable These bits enable $PA_6$ to $PA_4$ to be used for $A_{23}$ to $A_{21}$ address output				Re	served bi	ts	

ADRCR is an 8-bit readable/writable register that enables address output on bus lines A23 to A21.

ADRCR is initialized to H'FE by a reset and in hardware standby mode. It is not initialized in software standby mode.

**Bit 7—Address 23 Enable** ( $A_{23}$ E): Enables PA4 to be used as the  $A_{23}$  address output pin. Writing 0 in this bit enables  $A_{23}$  address output from PA4. In modes other than 3 this bit cannot be modified and PA4 has its ordinary input/output functions

Bit 7 A <sub>23</sub> E	Description	
0	$PA_4$ is the $A_{23}$ address output pin	
1	$PA_4$ is the $PA_4/TP_4/TIOCA_1$ input/output pin	(Initial value)

**Bit 6—Address 22 Enable** ( $A_{22}E$ ): Enables  $PA_5$  to be used as the  $A_{22}$  address output pin. Writing 0 in this bit enables  $A_{22}$  address output from  $PA_5$ . In modes other than 3 this bit cannot be modified and  $PA_5$  has its ordinary input/output functions.

Bit 6		
$A_{22}E$	Description	
0	$PA_{5}$ is the $A_{22}$ address output pin	
1	$PA_s$ is the $PA_s/TP_s/TIOCB_1$ input/output pin	(Initial value)

**Bit 5—Address 21 Enable**  $(A_{21}E)$ : Enables  $PA_6$  to be used as the  $A_{21}$  address output pin. Writing 0 in this bit enables  $A_{21}$  address output from  $PA_6$ . In modes other than 3 this bit cannot be modified and  $PA_6$  has its ordinary input/output functions.

Bit 5		
<b>A</b> <sub>21</sub> <b>E</b>	Description	
0	$PA_{_{\theta}}$ is the $A_{_{21}}$ address output pin	
1	PA <sub>6</sub> is the PA <sub>6</sub> /TP <sub>6</sub> /TIOCA <sub>2</sub> input/output pin	(Initial value)

# Bits 4 to 0—Reserved

# 6.3 Operation

# 6.3.1 Area Division

The external address space is divided into areas 0 to 7. Each area has a size of 128 kbytes in the

1-Mbyte mode and 2 Mbytes in the 16-Mbyte mode. Figure 6.2 shows a general view of the memory map.

H'00000		H'000000		H'00000	On-chip ROM*1
H'1FFFF	Area 0 (128 kbytes)	H'1FFFFF	Area 0 (2 Mbytes)	H'1FFFF	Area 0 (128 kbytes)
H'20000		H'200000		H'20000	
	Area 1 (128 kbytes)	11200000	Area 1 (2 Mbytes)	1120000	Area 1 (128 kbytes)
H'3FFFF	/ 100 · (120 · 10) (00)	H'3FFFFF	/ou / (oytoo)	H'3FFFF	
H'40000		H'400000		H'40000	
	Area 2 (128 kbytes)		Area 2 (2 Mbytes)		Area 2 (128 kbytes)
H'5FFFF		H'5FFFFF		H'5FFFF	
H'60000		H'600000		H'60000	
	Area 3 (128 kbytes)		Area 3 (2 Mbytes)		Area 3 (128 kbytes)
H'7FFFF		H'7FFFFF		H'7FFFF	
H'80000		H'800000		H'80000	
H'9FFFF	Area 4 (128 kbytes)	H'9FFFFF	Area 4 (2 Mbytes)	H'9FFFF	Area 4 (128 kbytes)
H'A0000		H'A00000		H'A0000	
	Area 5 (128 kbytes)		Area 5 (2 Mbytes)		Area 5 (128 kbytes)
H'BFFFF	/ 100 0 (120 http://00)	H'BFFFFF	/ 100 0 (2 110)(00)	H'BFFFF	
H'C0000		H'C00000		H'C0000	
	Area 6 (128 kbytes)		Area 6 (2 Mbytes)		Area 6 (128 kbytes)
H'DFFFF		H'DFFFFF		H'DFFFF	
H'E0000	Area 7 (128 kbytes)	H'E00000	Area 7 (2 Mbytes)	H'E0000	Area 7 (128 kbytes)
	On-chip RAM <sup>*1 *2</sup>		On-chip RAM <sup>*1 *2</sup>		On-chip RAM*1 *2
	External address space <sup>*3</sup>		External address space <sup>*3</sup>		External address space*3
H'FFFFF	On-chip I/O registers*1	H'FFFFFF	On-chip I/O registers*1	H'FFFFF	On-chip I/O registers*1
:	a. 1-Mbyte modes with on-chip ROM disabled (mode 1)		<ul> <li>b. 16-Mbyte modes with on-chip ROM disabled (mode 3)</li> </ul>		c. 1-Mbyte mode with on-chip ROM enabled (mode 5)

3. This area follows area 7 specifications.

### Figure 6.2 Access Area Map (Mode 1, 3, and 5)

The bus specifications for each area can be selected in ASTCR, WCER, and WCR as shown in table 6.3.

Table 6.3	<b>Bus Specifications</b>
-----------	---------------------------

ASTCR	WCER	WCR		Bus Specifications		
ASTn	WCEn	WMS1	WMS0	Bus Width	Access States	Wait Mode
0	—		—	8	2	Disabled
1	0		—	8	3	Pin wait mode 0
	1	0	0	8	3	Programmable wait mode
			1	8	3	Disabled
		1	0	8	3	Pin wait mode 1
			1	8	3	Pin auto-wait mode

Note: n = 0 to 7

### 6.3.2 Bus Control Signal Timing

# 8-Bit, Three-State-Access Areas

Figure 6.3 shows the timing of bus control signals for an 8-bit, three-state-access area. Wait states can be inserted.

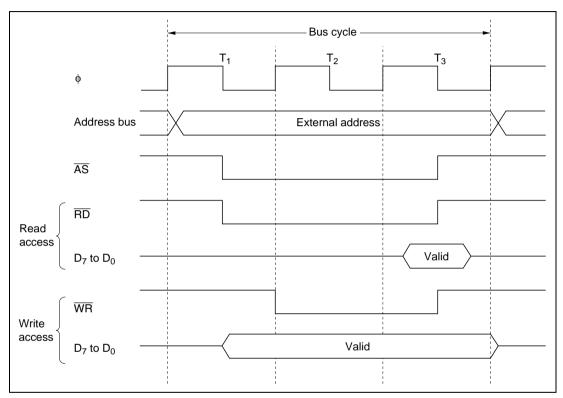


Figure 6.3 Bus Control Signal Timing for 8-Bit, Three-State-Access Area

## 8-Bit, Two-State-Access Areas

Figure 6.4 shows the timing of bus control signals for an 8-bit, two-state-access area. Wait states cannot be inserted.

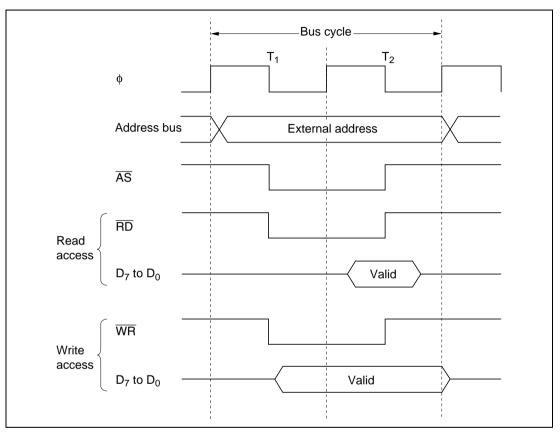


Figure 6.4 Bus Control Signal Timing for 8-Bit, Two-State-Access Area

### 6.3.3 Wait Modes

Four wait modes can be selected for each area as shown in table 6.4.

ASTCR	WCER	v			
ASTn Bit	WCEn Bit	WMS1 Bit	WMS0 Bit	WSC Control	Wait Mode
0	_	—	_	Disabled	No wait states
1	0	_	_	Disabled	Pin wait mode 0
	1	0	0	Enabled	Programmable wait mode
			1	Enabled	No wait states
		1	0	Enabled	Pin wait mode 1
			1	Enabled	Pin auto-wait mode

### Table 6.4 Wait Mode Selection

Note: n = 0 to 7

The ASTn and WCEn bits can be set independently for each area. Bits WMS1 and WMS0 apply to all areas. All areas for which WSC control is enabled operate in the same wait mode.

# Renesas

# Pin Wait Mode 0

The wait state controller is disabled. Wait states can only be inserted by  $\overline{WAIT}$  pin control. During access to an external three-state-access area, if the  $\overline{WAIT}$  pin is low at the fall of the system clock ( $\phi$ ) in the T<sub>2</sub> state, a wait state (T<sub>w</sub>) is inserted. If the  $\overline{WAIT}$  pin remains low, wait states continue to be inserted until the  $\overline{WAIT}$  signal goes high. Figure 6.5 shows the timing.

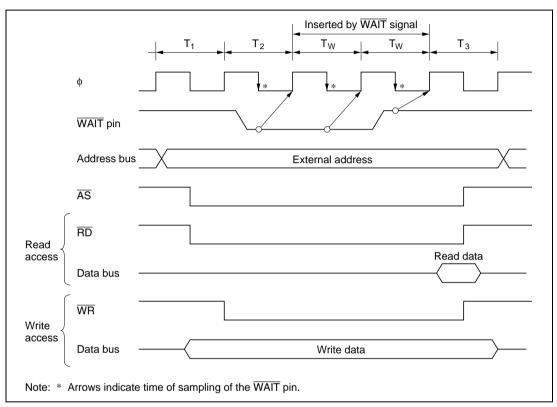


Figure 6.5 Pin Wait Mode 0

### Pin Wait Mode 1

In all accesses to external three-state-access areas, the number of wait states  $(T_w)$  selected by bits WC1 and WC0 are inserted. If the WAIT pin is low at the fall of the system clock ( $\phi$ ) in the last of these wait states, an additional wait state is inserted. If the WAIT pin remains low, wait states continue to be inserted until the WAIT signal goes high.

Pin wait mode 1 is useful for inserting four or more wait states, or for inserting different numbers of wait states for different external devices.

If the wait count is 0, this mode operates in the same way as pin wait mode 0.

Figure 6.6 shows the timing when the wait count is 1 (WC1 = 0, WC0 = 1) and one additional wait state is inserted by  $\overline{WAIT}$  input.

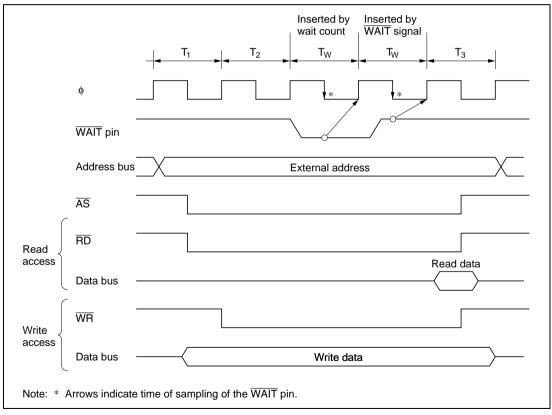


Figure 6.6 Pin Wait Mode 1

# Pin Auto-Wait Mode

If the  $\overline{WAIT}$  pin is low, the number of wait states (T<sub>w</sub>) selected by bits WC1 and WC0 are inserted.

In pin auto-wait mode, if the  $\overline{WAIT}$  pin is low at the fall of the system clock ( $\phi$ ) in the T<sub>2</sub> state, the number of wait states (T<sub>w</sub>) selected by bits WC1 and WC0 are inserted. No additional wait states are inserted even if the  $\overline{WAIT}$  pin remains low.

Figure 6.7 shows the timing when the wait count is 1.

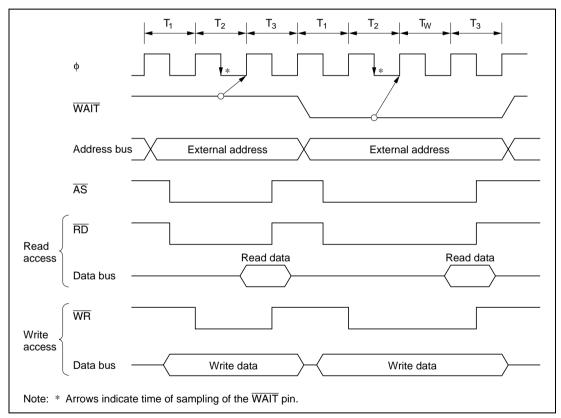


Figure 6.7 Pin Auto-Wait Mode

### **Programmable Wait Mode**

The number of wait states  $(T_w)$  selected by bits WC1 and WC0 are inserted in all accesses to external three-state-access areas. Figure 6.8 shows the timing when the wait count is 1 (WC1 = 0, WC0 = 1).

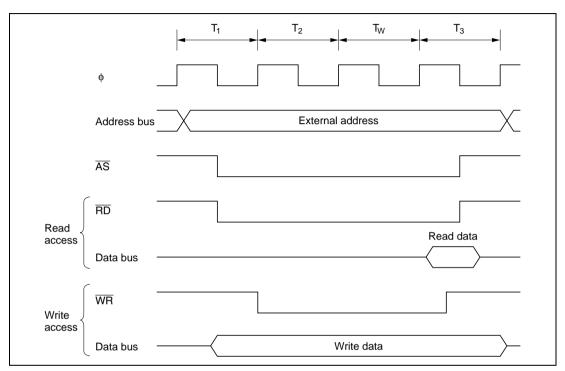


Figure 6.8 Programmable Wait Mode

# Renesas

# **Example of Wait State Control Settings**

A reset initializes ASTCR and WCER to H'FF and WCR to H'F3, selecting programmable wait mode and three wait states for all areas. Software can select other wait modes for individual areas by modifying the ASTCR, WCER, and WCR settings. Figure 6.9 shows an example of wait mode settings.

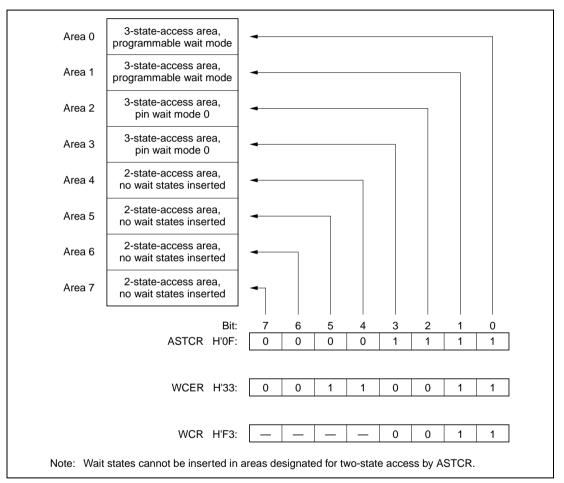


Figure 6.9 Wait Mode Settings (Example)

### 6.3.4 Interconnections with Memory (Example)

For each area, the bus controller can select two- or three-state access. In three-state-access areas, wait states can be inserted in a variety of modes, simplifying the connection of both high-speed and low-speed devices.

Figure 6.10 shows a memory map for this example.

A 32-kword  $\times$  8-bit EPROM is connected to area 2. This device is accessed in three states via an 8-bit bus.

Two 32-kword  $\times$  8-bit SRAM devices (SRAM1 and SRAM2) are connected to area 3. These devices are accessed in two states via an 8-bit bus.

One 32-kword  $\times$  8-bit SRAM (SRAM3) is connected to area 7. This device is accessed via an 8-bit bus, using three-state access with an additional wait state inserted in pin auto-wait mode.

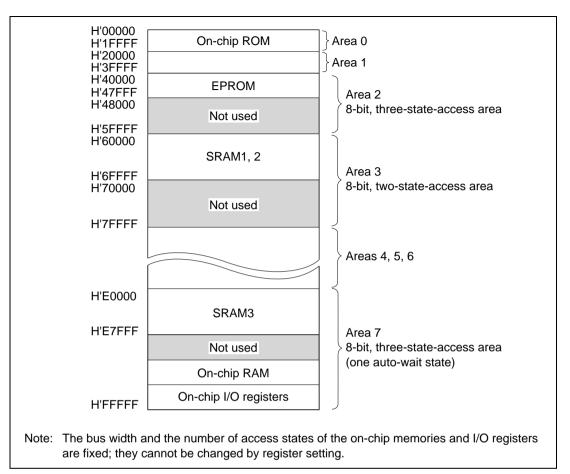


Figure 6.10 Memory Map (H8/3039 Mode 5)

# 6.4 Usage Notes

# 6.4.1 Register Write Timing

# ASTCR and WCER Write Timing

Data written to ASTCR or WCER takes effect starting from the next bus cycle. Figure 6.11 shows the timing when an instruction fetched from area 2 changes area 2 from three-state access to two-state access.

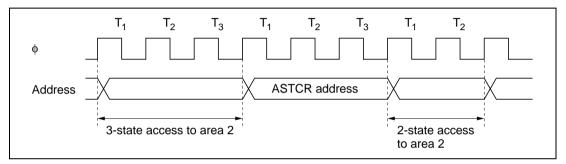


Figure 6.11 ASTCR Write Timing

# 6.4.2 Precautions on Setting ASTCR and ABWCR\*

Use the H8/3039 Group on-chip program to set ASTCR and ABWCR as shown below, so that the on-chip ROM access cycle for H8/3039 Group can be emulated using the evaluation chip for support tools.

Modes 5 and 7

ASTCR0 = 0ABWCR = H'FC

Note: \* The ABWCR (bus width control register; lower 16-bit address: H'FFEC) is not built onto this LSI. For detailed features of the ABWCR, see the H8/3048 Group, H8/3048F-ZTAT<sup>™</sup> Hardware Manual.

# Section 7 I/O Ports

# 7.1 Overview

The H8/3039 Group has nine input/output ports (ports 1, 2, 3, 5, 6, 8, 9, A, and B) and one input port (port 7). Table 7.1 summarizes the port functions. The pins in each port are multiplexed as shown in table 7.1.

Each port has a data direction register (DDR) for selecting input or output, and a data register (DR) for storing output data. In addition to these registers, ports 2, and 5 have an input pull-up control register (PCR) for switching input pull-up transistors on and off.

Ports 1 to 3 and ports 5, 6, and 8 can drive one TTL load and a 90-pF capacitive load. Ports 9, A, and B can drive one TTL load and a 30-pF capacitive load. Ports 1 to 3 and ports 5, 6, 8, 9, A, and B can drive a Darlington pair. Ports 1, 2, 5, and B can drive LEDs (with 10-mA current sink). Pins  $P8_1$ ,  $P8_0$ ,  $PA_7$  to  $PA_0$ , and  $PB_3$  to  $PB_0$  have Schmitt-trigger input circuits.

For block diagrams of the ports see appendix C, I/O Block Diagrams.

# Table 7.1Port Functions

Port	Description	Pins	Mode 1	Mode 3	Mode 5	Mode 6	Mode 7	
Port 1	<ul> <li>8-bit I/O port</li> <li>Can drive LEDs</li> </ul>	$P1_7$ to $P1_0/A_7$ to $A_0$	Address output pins ( $A_7$ to $A_0$ )		Address output $(A_7 \text{ to } A_0)$ and generic input	Generic inp	ut/ output	
			DDR = 0: generic input					
					DDR = 1: address output			
Port 2	<ul> <li>8-bit I/O port</li> </ul>	$P2_7 to P2_0/$	Address output	pins (A <sub>15</sub> to	Address output	Generic inp	ut/ output	
	<ul> <li>Input pull-up</li> </ul>	$A_{15}$ to $A_8$	A <sub>s</sub> )		(A <sub>15</sub> to A <sub>8</sub> ) and generic input			
	<ul> <li>Can drive LEDs</li> </ul>		ge		DDR = 0: generic input			
					DDR = 1: address output			
Port 3	8-bit I/O port	$P3_7$ to $P3_0$ / $D_7$ to $D_0$	Data input/outp	Generic inp	ut/ output			
		$P5_{3}$ to $P5_{0}$ /	Address output	(A <sub>19</sub> to A <sub>16</sub> )	Address output	Generic inp	ut/ output	
	<ul><li>Input pull-up</li><li>Can drive</li></ul>	$A_{19}$ to $A_{16}$			(A <sub>19</sub> to A <sub>16</sub> ) and 4-bit generic input			
	LEDs				DDR = 0: generic input			
					DDR = 1: address output			
Port 6	<ul> <li>4-bit I/O port</li> </ul>	P6₄/ <del>WR</del> , P6₄/ <del>RD</del> , P6₄/ <del>AS</del>	Bus control signa	al output (WR, Ì	RD, AS)	Generic inp	ut/ output	
		P6₀/WAIT	Bus control sign generic input/or		ut (WAIT) and 1-bit			
Port 7	<ul> <li>8-bit Input port</li> </ul>	$P7_7$ to $P7_0$ / AN <sub>7</sub> to AN <sub>0</sub>	Analog input (A	$N_7$ to $AN_0$ ) to $A$	A/D converter, and	generic inpu	t	
Port 8	-	P8,/ IRQ	$\overline{IRQ}_1$ input and	1-bit generic i	nput/output	IRQ, and IF		
	<ul> <li>P8, and P8, have Schmitt inputs</li> </ul>	P8 <sub>0</sub> /IRQ <sub>0</sub>	$\overline{IRQ}_0$ input and	1-bit generic i	nput/output	and generic input/ output		

Port	Description	Pins	Mode 1	Mode 3	Mode 5	Mode 6	Mode 7
Port 9	• 6-bit I/O port	P9 <sub>5</sub> /SCK <sub>1</sub> / <del>IRQ</del> <sub>5</sub> , P9 <sub>4</sub> /SCK <sub>2</sub> / <del>IRQ</del> <sub>4</sub> , P9 <sub>3</sub> /RxD <sub>1</sub> , P9 <sub>2</sub> /RxD <sub>0</sub> , P9 <sub>1</sub> /TxD <sub>1</sub> , P9 <sub>0</sub> /TxD <sub>0</sub>		interfaces 1 a	, RxD <sub>1</sub> , RxD <sub>0</sub> , TxD <sub>1</sub> nd 0 (SCl0, 1), IRC		
Port A	<ul> <li>8-bit I/O port</li> <li>Schmitt inputs</li> </ul>	PA,/TP,/ TIOCB,/A <sub>20</sub>	Output (TP <sub>7</sub> ) from pro- grammable timing pattern controller (TPC), input or output (TIOCB <sub>2</sub> ) for 16-bit integrated timer unit (ITU), and generic input/output	Address output (A <sub>20</sub> )	TPC output $(TP_7)$ , (TIOCB <sub>2</sub> ), and ger		
		PA <sub>6</sub> /TP <sub>6</sub> / TIOCA <sub>2</sub> /A <sub>21</sub> , PA <sub>5</sub> /TP <sub>6</sub> / TIOCB <sub>1</sub> /A <sub>22</sub> , PA <sub>4</sub> /TP <sub>4</sub> / TIOCA <sub>1</sub> /A <sub>23</sub>	TPC output (TP <sub>6</sub> to TP₄), ITU input and output (TIOCA₂, TIOCB₁, TIOCA,), and generic input/ output	TPC output (TP <sub>6</sub> to TP <sub>4</sub> ), ITU input and output (TIOCA <sub>2</sub> , TIOCB <sub>1</sub> , TIOCA <sub>1</sub> ), address output (A <sub>23</sub> to A <sub>21</sub> ), and generic input/output	TPC output (TP <sub>6</sub> tr output (TIOCA <sub>2</sub> , T generic input/outp	IOCB <sub>1</sub> , TIOC	
		PA,/TP,/ TIOCB,/ TCLKD, PA,/TP,/ TCLKC, PA,/TP,/ TCLKB, PA,/TP,/ TCLKA			input and output (T CA₀), and generic i		<b>⟨C</b> ,

### Section 7 I/O Ports

Port	Description	Pins	Mode 1	Mode 3	Mode 5	Mode 6	Mode 7		
Port B	<ul><li>7-bit I/Oport</li><li>Can drive</li></ul>	PB <sub>7</sub> /TP <sub>15</sub> / ADTRG	TPC output (TP $_{15}$ ), trigger input (ADTRG) to A/D converter, and generic input/output.						
	<ul> <li>PB<sub>3</sub> to PB<sub>0</sub> have Schmitt inputs</li> </ul>	PB <sub>s</sub> /TP <sub>13</sub> / TOCXB <sub>4</sub> , PB <sub>s</sub> /TP <sub>12</sub> / TOCXA <sub>4</sub> PB <sub>3</sub> /TP <sub>11</sub> / TIOCB <sub>4</sub> , PB <sub>2</sub> /TP <sub>10</sub> / TIOCA <sub>4</sub> , PB <sub>1</sub> /TP <sub>6</sub> / TIOCB <sub>3</sub> , PB <sub>0</sub> /TP <sub>8</sub> / TIOCA <sub>3</sub>		15 0,	input and output ( <sup>*</sup> ICA <sub>3</sub> ), and generic	4	-		

# 7.2 Port 1

### 7.2.1 Overview

Port 1 is an 8-bit input/output port with the pin configuration shown in figure 7.1. The pin functions differ between the expanded modes with on-chip ROM disabled, expanded modes with on-chip ROM enabled, and single-chip mode. In modes 1, 3 (expanded modes with on-chip ROM disabled), they are address bus output pins ( $A_7$  to  $A_0$ ).

In mode 5 (expanded modes with on-chip ROM enabled), settings in the port 1 data direction register (P1DDR) can designate pins for address bus output ( $A_7$  to  $A_0$ ) or generic input. In modes 6 and 7 (single-chip mode), port 1 is a generic input/output port.

Pins in port 1 can drive one TTL load and a 90-pF capacitive load. They can also drive a Darlington transistor pair.

	Port 1 pins	Modes 1 and 3	Mode 5	Modes 6 and 7
Port 1	$P1_7/A_7$ $P1_6/A_6$ $P1_5/A_5$ $P1_4/A_4$	$A_7$ (output) $A_6$ (output) $A_5$ (output) $A_4$ (output)	P1 <sub>7</sub> (input)/A <sub>7</sub> (output) P1 <sub>6</sub> (input)/A <sub>6</sub> (output) P1 <sub>5</sub> (input)/A <sub>5</sub> (output) P1 <sub>4</sub> (input)/A <sub>4</sub> (output)	P1 <sub>7</sub> (input/output) P1 <sub>6</sub> (input/output) P1 <sub>5</sub> (input/output) P1 <sub>4</sub> (input/output)
	$P1_{3}/A_{3}$ $P1_{2}/A_{2}$ $P1_{1}/A_{1}$ $P1_{0}/A_{0}$	$A_3$ (output) $A_2$ (output) $A_1$ (output) $A_0$ (output)	P1 <sub>3</sub> (input)/A <sub>3</sub> (output) P1 <sub>2</sub> (input)/A <sub>2</sub> (output) P1 <sub>1</sub> (input)/A <sub>1</sub> (output) P1 <sub>0</sub> (input)/A <sub>0</sub> (output)	P1 <sub>3</sub> (input/output) P1 <sub>2</sub> (input/output) P1 <sub>1</sub> (input/output) P1 <sub>0</sub> (input/output)

Figure 7.1 Port 1 Pin Configuration

# 7.2.2 Register Descriptions

Table 7.2 summarizes the registers of port 1.

# Table 7.2Port 1 Registers

				Initial Value			
Address*	Name	Abbreviation	R/W	Modes 1, 3	Modes 5 to 7		
H'FFC0	Port 1 data direction register	P1DDR	W	H'FF	H'00		
H'FFC2	Port 1 data register	P1DR	R/W	H'00	H'00		

Note: \* Lower 16 bits of the address.

# Port 1 Data Direction Register (P1DDR)

P1DDR is an 8-bit write-only register that can select input or output for each pin in port 1.

Bit		7	6	5	4	3	2	1	0
		P17DDR	P1 <sub>6</sub> DDR	P1₅DDR	P1 <sub>4</sub> DDR	P1 <sub>3</sub> DDR	P1 <sub>2</sub> DDR	P1 <sub>1</sub> DDR	P1 <sub>0</sub> DDR
	Initial valu	e 1	1	1	1	1	1	1	1
1, 3	Read/Writ	e —	—	—	—	—	—	—	—
	Initial valu	e 0	0	0	0	0	0	0	0
5 to 7	Read/Writ	e W	W	W	W	W	W	W	W
Port 1 data direction 7 to 0									

Port 1 data direction 7 to 0 These bits select input or output for port 1 pins

P1DDR is initialized to H'00 by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a P1DDR bit is set to 1, the corresponding pin maintains its output state in software standby mode.

### Port 1 Data Register (P1DR)

Bit	7	6	5	4	3	2	1	0			
	P17	P1 <sub>6</sub>	P1 <sub>5</sub>	P14	P1 <sub>3</sub>	P12	P1 <sub>1</sub>	P10			
Initial value	0	0	0	0	0	0	0	0			
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W			
Port 1 data 7 to 0											

P1DR is an 8-bit readable/writable register that stores data for pins P1<sub>7</sub> to P1<sub>0</sub>.

These bits store data for port 1 pins

When a bit in P1DDR is set to 1, if port 1 is read the value of the corresponding P1DR bit is returned directly, regardless of the actual state of the pin. When a bit in P1DDR is cleared to 0, if port 1 is read the corresponding pin level is read.

P1DR is initialized to H'00 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

# 7.2.3 Pin Functions in Each Mode

The pin functions of port 1 differ between mode 1, 3 (expanded mode with on-chip ROM disabled), mode 5 (expanded mode with on-chip ROM enabled), mode 6, and 7 (single-chip mode). The pin functions in each mode are described as follows.

### Modes 1 and 3

Address output can be selected for each pin in port 1. Figure 7.2 shows the pin functions in modes 1 and 3.

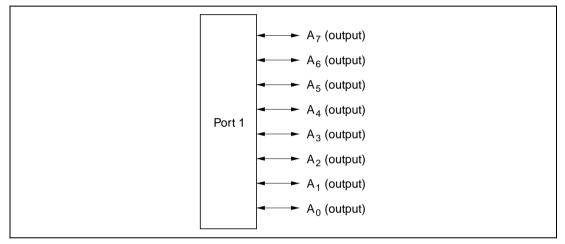


Figure 7.2 Pin Functions in Modes 1 and 3 (Port 1)

# Mode 5

Address output or generic input can be selected for each pin in port 1. A pin becomes an address output pin if the corresponding P1DDR bit is set to 1, and a generic input pin if this bit is cleared to 0. Following a reset, all pins are input pins. To use a pin for address output, its P1DDR bit must be set to 1. Figure 7.3 shows the pin functions in mode 5.

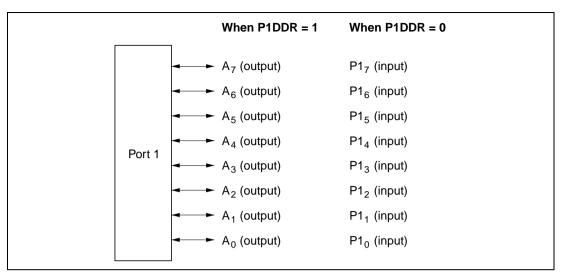


Figure 7.3 Pin Functions in Mode 5 (Port 1)

### Modes 6 and 7 (Single-Chip Mode)

Input or output can be selected separately for each pin in port 1. A pin becomes an output pin if the corresponding P1DDR bit is set to 1, and an input pin if this bit is cleared to 0. Figure 7.4 shows the pin functions in modes 6 and 7.

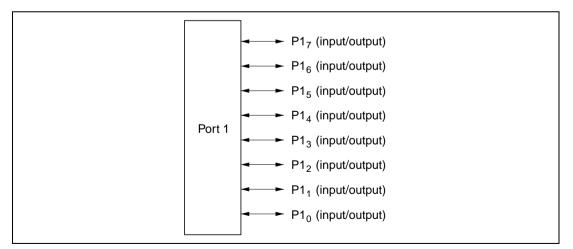


Figure 7.4 Pin Functions in Modes 6 and 7 (Port 1)

# Renesas

# 7.3 Port 2

# 7.3.1 Overview

Port 2 is an 8-bit input/output port with the pin configuration shown in figure 7.5. Pin functions differ according to operation mode.

In modes 1 and 3 (expanded mode with on-chip ROM disabled), port 2 consists of address bus output pins ( $A_{15}$  to  $A_8$ ). In mode 5 (expanded mode with on-chip ROM enabled), settings in the port 2 data direction register (P2DDR) can designate pins for address bus output ( $A_{15}$  to  $A_8$ ) or generic input. In modes 6 and 7 (single-chip mode), port 2 is a generic input/output port.

Port 2 has software-programmable built-in pull-up transistors. Pins in port 2 can drive one TTL load and a 90-pF capacitive load. They can also drive a Darlington transistor pair.

	Port 2 pins	Modes 1 and 3	Modes 5	Mode 6 and 7
		A <sub>15</sub> (output) A <sub>14</sub> (output)	$P2_7$ (input)/ $A_{15}$ (output) $P2_6$ (input)/ $A_{14}$ (output)	P2 <sub>7</sub> (input/output) P2 <sub>6</sub> (input/output)
	← P2 <sub>5</sub> /A <sub>13</sub>	A <sub>13</sub> (output)		-
Port 2	$  P2_4/A_{12}  P2_3/A_{11} $	A <sub>12</sub> (output) A <sub>11</sub> (output)	$P2_4$ (input)/ $A_{12}$ (output) $P2_3$ (input)/ $A_{11}$ (output)	P2 <sub>4</sub> (input/output) P2 <sub>3</sub> (input/output)
	► P2 <sub>2</sub> /A <sub>10</sub>	A <sub>10</sub> (output)	P2 <sub>2</sub> (input)/A <sub>10</sub> (output)	P2 <sub>2</sub> (input/output)
	← ► P2 <sub>1</sub> /A <sub>9</sub>	A <sub>9</sub> (output)	P2 <sub>1</sub> (input)/A <sub>9</sub> (output)	P2 <sub>1</sub> (input/output)
	► P2 <sub>0</sub> /A <sub>8</sub>	A <sub>8</sub> (output)	P2 <sub>0</sub> (input)/A <sub>8</sub> (output)	P2 <sub>0</sub> (input/output)

Figure 7.5 Port 2 Pin Configuration

#### 7.3.2 Register Descriptions

Table 7.3 summarizes the registers of port 2.

### Table 7.3Port 2 Registers

				Initial Value			
Address*	Name	Abbreviation	R/W	Modes 1 and 3	Modes 5 to 7		
H'FFC1	Port 2 data direction register	P2DDR	W	H'FF	H'00		
H'FFC3	Port 2 data register	P2DR	R/W	H'00	H'00		
H'FFD8	Port 2 input pull-up control register	P2PCR	R/W	H'00	H'00		

Note: \* Lower 16 bits of the address.

#### Port 2 Data Direction Register (P2DDR)

P2DDR is an 8-bit write-only register that can select input or output for each pin in port 2.

Bit		7	6	5	4	3	2	1	0
		P27DDR	P2 <sub>6</sub> DDR	$P2_5DDR$	P2 <sub>4</sub> DDR	P2 <sub>3</sub> DDR	P2 <sub>2</sub> DDR	P2 <sub>1</sub> DDR	P20DDR
	Initial valu		1	1	1	1	1	1	1
1 and 3	Read/Writ	e —	—	—	—	—	_	—	—
Modes 5 to 7	Initial valu	e 0	0	0	0	0	0	0	0
	Read/Writ	e W	W	W	W	W	W	W	W

Port 2 data direction 7 to 0 These bits select input or output for port 2 pins

P2DDR is initialized to H'00 by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a P2DDR bit is set to 1, the corresponding pin maintains its output state in software standby mode.

# Renesas

# Port 2 Data Register (P2DR)

Bit	7	6	5	4	3	2	1	0
	P27	P2 <sub>6</sub>	P25	P24	P23	P22	P21	P20
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
			-					

P2DR is an 8-bit readable/writable register that stores data for pins  $P2_7$  to  $P2_9$ .

Port 2 data 7 to 0

These bits store data for port 2 pins

When a bit in P2DDR is set to 1, if port 2 is read the value of the corresponding P2DR bit is returned directly, regardless of the actual state of the pin. When a bit in P2DDR is cleared to 0, if port 2 is read the corresponding pin level is read.

P2DR is initialized to H'00 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

# Port 2 Input Pull-Up Control Register (P2PCR)

P2PCR is an 8-bit readable/writable register that controls the MOS input pull-up transistors in port 2.

Bit	7	6	5	4	3	2	1	0
	P27PCR	P2 <sub>6</sub> PCR	$P2_5PCR$	P2 <sub>4</sub> PCR	P2 <sub>3</sub> PCR	P2 <sub>2</sub> PCR	P2 <sub>1</sub> PCR	P20PCR
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
				2 input pu	•			
			These	e bits conti	rol input pu	ull-up		
			transi	stors built	into port 2			

When a P2DDR bit is cleared to 0 (selecting generic input) in modes 7 to 5, if the corresponding bit from  $P2_{7}PCR$  to  $P2_{9}PCR$  is set to 1, the input pull-up transistor is turned on.

P2PCR is initialized to H'00 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

### 7.3.3 Pin Functions in Each Mode

The pin functions of port 2 differ between mode 1, 3 (expanded mode with on-chip ROM disabled), mode 5 (expanded mode with on-chip ROM enabled), mode 6, and 7 (single-chip mode). The pin functions in each mode are described followings.

#### Modes 1 and 3

Address output can be selected for each pin in port 2. Figure 7.6 shows the pin functions in modes 1 and 3.

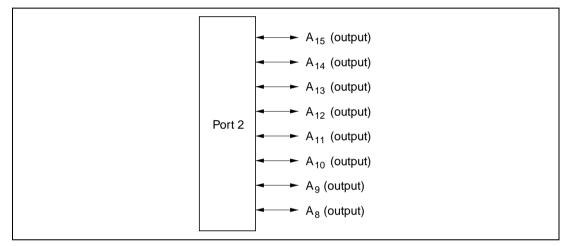


Figure 7.6 Pin Functions in Modes 1 and 3 (Port 2)

### Mode 5

Address output or generic input can be selected for each pin in port 2. A pin becomes an address output pin if the corresponding P2DDR bit is set to 1, and a generic input pin if this bit is cleared to 0. Following a reset, all pins are input pins. To use a pin for address output, its P2DDR bit must be set to 1. Figure 7.7 shows the pin functions in modes 5.

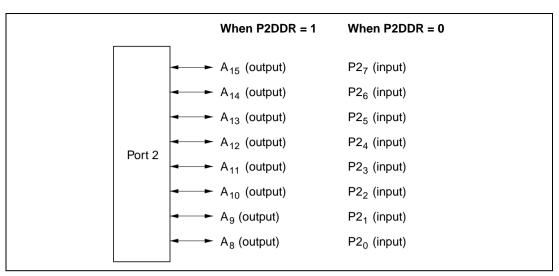


Figure 7.7 Pin Functions in Modes 1 and 3 (Port 2)

# Modes 6 and 7

Input or output can be selected separately for each pin in port 2. A pin becomes an output pin if the corresponding P2DDR bit is set to 1, and an input pin if this bit is cleared to 0. Figure 7.8 shows the pin functions in modes 6 and 7.

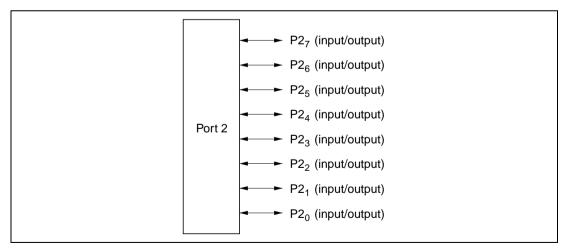


Figure 7.8 Pin Functions in Modes 6 and 7 (Port 2)

### 7.3.4 Input Pull-Up Transistors

Port 2 has built-in MOS input pull-up transistors that can be controlled by software. These input pull-up transistors can be turned on and off individually.

When a P2PCR bit is set to 1 and the corresponding P2DDR bit is cleared to 0, the input pull-up transistor is turned on.

The input pull-up transistors are turned off by a reset and in hardware standby mode. In software standby mode they retain their previous state.

Table 7.4 summarizes the states of the input pull-up transistors in each mode.

Mode	Reset	Hardware Standby Mode	Software Standby Mode	Other Modes
1 3	Off	Off	Off	Off
5 6 7	Off	Off	On/Off	On/Off

 Table 7.4
 Input Pull-Up Transistor States (Port 2)

Legend:

Off: The input pull-up transistor is always off.

On/Off: The input pull-up transistor is on if P2PCR = 1 and P2DDR = 0. Otherwise, it is off.

# Renesas

# 7.4 Port 3

# 7.4.1 Overview

Port 3 is an 8-bit input/output port with the pin configuration shown in figure 7.9. Port 3 is a data bus in modes 1, 3 and 5 (expanded modes) and a generic input/output port in mode 6 and 7 (single-chip mode).

Pins in port 3 can drive one TTL load and a 90-pF capacitive load. They can also drive a Darlington transistor pair.

	Port 3 pins	Modes 1, 3 and 5	Modes 6 and 7
Port 3	Port 3 pins $P3_7/D_7$ $P3_6/D_6$ $P3_5/D_5$ $P3_4/D_4$ $P3_3/D_3$ $P3_2/D_2$ $P3_1/D_1$	Modes 1, 3 and 5 $D_7$ (input/output) $D_6$ (input/output) $D_5$ (input/output) $D_4$ (input/output) $D_3$ (input/output) $D_2$ (input/output) $D_1$ (input/output)	Modes 6 and 7 $P3_7$ (input/output) $P3_6$ (input/output) $P3_5$ (input/output) $P3_4$ (input/output) $P3_2$ (input/output) $P3_1$ (input/output)
▲ →	P3 <sub>0</sub> /D <sub>0</sub>	D <sub>0</sub> (input/output)	P3 <sub>0</sub> (input/output)

Figure 7.9 Port 3 Pin Configuration

# 7.4.2 Register Descriptions

Table 7.5 summarizes the registers of port 3.

### Table 7.5Port 3 Registers

Address*	Name	Abbreviation	R/W	Initial Value
H'FFC4	Port 3 data direction register	P3DDR	W	H'00
H'FFC6	Port 3 data register	P3DR	R/W	H'00

Note: \* Lower 16 bits of the address.

### Port 3 Data Direction Register (P3DDR)

Bit	7	6	5	4	3	2	1	0
	P37DDR	P3 <sub>6</sub> DDR	$P3_5DDR$	P3 <sub>4</sub> DDR	P3 <sub>3</sub> DDR	P32DDR	P31DDR	P30DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W
			Port	3 data di	rection 7	to 0		

P3DDR is an 8-bit write-only register that can select input or output for each pin in port 3.

These bits select input or output for port 3 pins

Modes 1, 3, and 5: Port 3 functions as a data bus. P3DDR is ignored.

**Modes 6 and 7:** Port 3 functions as an input/output port. A pin in port 3 becomes an output pin if the corresponding P3DDR bit is set to 1, and an input pin if this bit is cleared to 0.

P3DDR is a write-only register. Its value cannot be read. All bits return 1 when read.

P3DDR is initialized to H'00 by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a P3DDR bit is set to 1, the corresponding pin maintains its output state in software standby mode.

# Port 3 Data Register (P3DR)

P3DR is an 8-bit readable/writable register that stores data for pins P3<sub>7</sub> to P3<sub>0</sub>.

Bit	7	6	5	4	3	2	1	0
	P37	P3 <sub>6</sub>	P3 <sub>5</sub>	P34	P3 <sub>3</sub>	P3 <sub>2</sub>	P3 <sub>1</sub>	P3 <sub>0</sub>
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
			Dent	2 data 7 i				

Port 3 data 7 to 0

These bits store data for port 3 pins

When a bit in P3DDR is set to 1, if port 3 is read the value of the corresponding P3DR bit is returned directly, regardless of the actual state of the pin. When a bit in P3DDR is cleared to 0, if port 3 is read the corresponding pin level is read.

P3DR is initialized to H'00 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

# Renesas

# 7.4.3 Pin Functions in Each Mode

The pin functions of port 3 differ between modes 1, 3 and 5 and modes 6 and 7. The pin functions in each mode are described below.

# Modes 1, 3 and 5

All pins of port 3 automatically become data input/output pins. Figure 7.10 shows the pin functions in modes 1, 3 and 5.

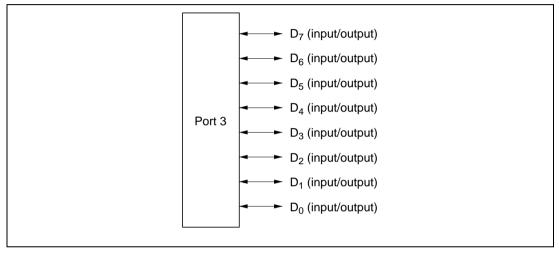


Figure 7.10 Pin Functions in Modes 1, 3 and 5 (Port 3)

### Modes 6 and 7

Input or output can be selected separately for each pin in port 3. A pin becomes an output pin if the corresponding P3DDR bit is set to 1, and an input pin if this bit is cleared to 0. Figure 7.11 shows the pin functions in modes 6 and 7.

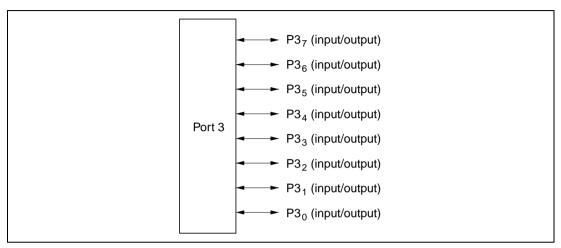


Figure 7.11 Pin Functions in Modes 6 and 7 (Port 3)

# 7.5 Port 5

# 7.5.1 Overview

Port 5 is a 4-bit input/output port with the pin configuration shown in figure 7.12. The pin functions differ depending on the operating mode.

In modes 1, 3 (expanded modes with on-chip ROM disabled), port 5 consists of address output pins (A19 to A16). In modes 5 (expanded modes with on-chip ROM enabled), settings in the port 5 data direction register (P5DDR) designate pins for address bus output ( $A_{19}$  to  $A_{16}$ ) or generic input. In mode 6 and 7 (single-chip mode), port 5 is a generic input/output port.

Port 5 has software-programmable built-in pull-up transistors. Port 5 can drive one TTL load and a 90-pF capacitive load. They can also drive an LED or a Darlington transistor pair.

	1	Port 5 pins	Modes 1, 3	Mode 5	Modes 6 and 7
	• •	P5 <sub>3</sub> /A <sub>19</sub>	A <sub>19</sub> (output)	P5 <sub>3</sub> (input)/A <sub>19</sub> (output)	P53 (input/output)
Dort 5	• •	P5 <sub>2</sub> /A <sub>18</sub>	A <sub>18</sub> (output)	P5 <sub>2</sub> (input)/A <sub>18</sub> (output)	P5 <sub>2</sub> (input/output)
Port 5	• •	P5 <sub>1</sub> /A <sub>17</sub>	A <sub>17</sub> (output)	P5 <sub>1</sub> (input)/A <sub>17</sub> (output)	P5 <sub>1</sub> (input/output)
	• •	P5 <sub>0</sub> /A <sub>16</sub>	A <sub>16</sub> (output)	P5 <sub>0</sub> (input)/A <sub>16</sub> (output)	P5 <sub>0</sub> (input/output)

Figure 7.12 Port 5 Pin Configuration

### 7.5.2 Register Descriptions

Table 7.6 summarizes the registers of port 5.

#### Table 7.6Port 5 Registers

				Initial	Value
Address*	Name	Abbreviation	R/W	Modes 1 and 3	Modes 5 to 7
H'FFC8	Port 5 data direction register	P5DDR	W	H'FF	H'F0
H'FFCA	Port 5 data register	P5DR	R/W	H'F0	H'F0
H'FFDB	Port 5 input pull-up control register	P5PCR	R/W	H'F0	H'F0

Note: \* Lower 16 bits of the address.

#### Port 5 Data Direction Register (P5DDR)

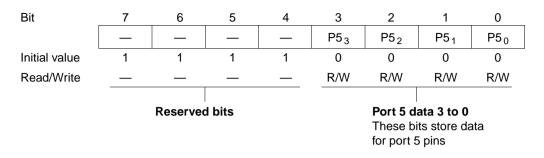
P5DDR is an 8-bit write-only register that can select input or output for each pin in port 5.

Bit		7	6	5	4	3	2	1	0
		_	_	—	—	P53DDR	P5 <sub>2</sub> DDR	P5 <sub>1</sub> DDR	P5 <sub>0</sub> DDR
Modes	Initial value	1	1	1	1	1	1	1	1
1 and 3	Read/Write	_	_	—	—	—	—	—	—
Modes	Initial value	1	1	1	1	0	0	0	0
5 to 7	Read/Write	_	—	—	_	W	W	W	W
Reserved bits						The	r <b>t 5 data c</b> ese bits se put for por	lect input	

P5DDR is initialized to H'F0 by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a P5DDR bit is set to 1, the corresponding pin maintains its output state in software standby mode.

# Port 5 Data Register (P5DR)

P5DR is an 8-bit readable/writable register that stores data for pins P5<sub>3</sub> to P5<sub>0</sub>.



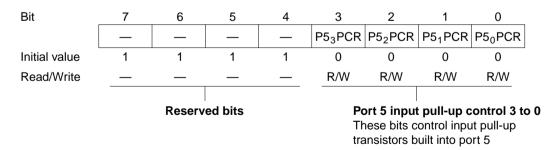
When a bit in P5DDR is set to 1, if port 5 is read the value of the corresponding P5DR bit is returned directly, regardless of the actual state of the pin. When a bit in P5DDR is cleared to 0, if port 5 is read the corresponding pin level is read.

Bits  $P5_7$  to  $P5_4$  are reserved. They cannot be modified and are always read as 1.

P5DR is initialized to H'F0 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

# Port 5 Input Pull-Up Control Register (P5PCR)

P5PCR is an 8-bit readable/writable register that controls the MOS input pull-up transistors in port 5.



When a P5DDR bit is cleared to 0 (selecting generic input) in modes 5 to 7, if the corresponding bit from  $P5_3PCR$  to  $P5_0PCR$  is set to 1, the input pull-up transistor is turned on.

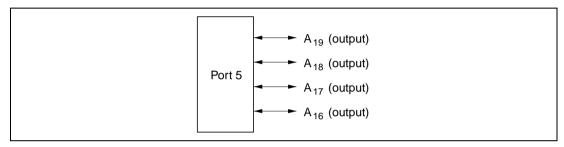
P5PCR is initialized to H'F0 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

### 7.5.3 Pin Functions in Each Mode

The pin functions differ between mode 1, 3 (expanded modes with on-chip ROM disabled), mode 5 (expanded modes with on-chip ROM enabled), mode 6, and 7 (single-chip mode). The pin functions in each mode are described below.

### Modes 1 and 3

Address output can be selected for each pin in port 5. Figure 7.13 shows the pin functions in modes 1 and 3.





### Mode 5

Address output or generic input can be selected for each pin in port 5. A pin becomes an address output pin if the corresponding P5DDR bit is set to 1, and a generic input pin if this bit is cleared to 0. Following a reset, all pins are input pins. To use a pin for address output, its P5DDR must be set to 1. Figure 7.14 shows the pin functions in mode 5.

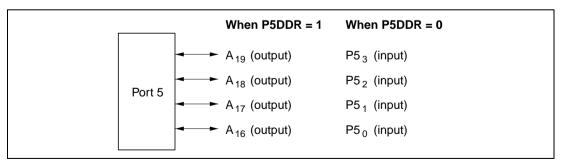


Figure 7.14 Pin Functions in Mode 5 (Port 5)

# Modes 6 and 7

Input or output can be selected separately for each pin in port 5. A pin becomes an output pin if the corresponding P5DDR bit is set to 1, and an input pin if this bit is cleared to 0. Figure 7.15 shows the pin functions in modes 6 and 7.

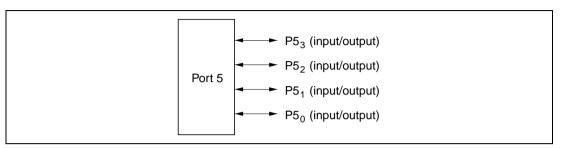


Figure 7.15 Pin Functions in Mode 6 and 7 (Port 5)

# 7.5.4 Input Pull-Up Transistors

Port 5 has built-in MOS input pull-up transistors that can be controlled by software. These input pull-up transistors can be turned on and off individually.

When a P5PCR bit is set to 1 and the corresponding P5DDR bit is cleared to 0, the input pull-up transistor is turned on.

The input pull-up transistors are turned off by a reset and in hardware standby mode. In software standby mode they retain their previous state.

Table 7.7 summarizes the states of the input pull-up transistors in each mode.

# Table 7.7 Input Pull-Up Transistor States (Port 5)

Mode	Reset	Hardware Standby Mode	Software Standby Mode	Other Modes
1 3	Off	Off	Off	Off
5 6 7	Off	Off	On/Off	On/Off

Legend:

Off: The input pull-up transistor is always off.

On/Off: The input pull-up transistor is on if P5PCR = 1 and P5DDR = 0. Otherwise, it is off.

#### 7.6 Port 6

#### 7.6.1 Overview

Port 6 is a 4-bit input/output port that is also used for input and output of bus control signals ( $\overline{WR}$ ,  $\overline{RD}$ ,  $\overline{AS}$ , and  $\overline{WAIT}$ ).

Figure 7.16 shows the pin configuration of port 6. In modes 1, 3 and 5, the pin functions are  $\overline{WR}$ ,  $\overline{RD}$ ,  $\overline{AS}$ , and  $P6_0/\overline{WAIT}$ . In modes 6 and 7, port 6 is a generic input/output port.

Pins in port 6 can drive one TTL load and a 90-pF capacitive load. They can also drive a Darlington transistor pair.

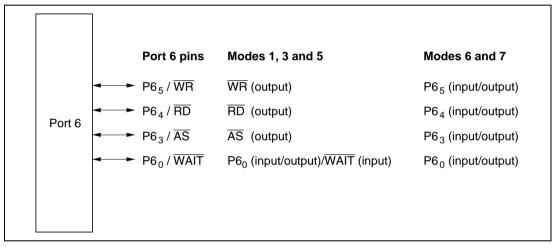


Figure 7.16 Port 6 Pin Configuration

#### Renesas

#### 7.6.2 Register Descriptions

Table 7.8 summarizes the registers of port 6.

#### Table 7.8Port 6 Registers

				Initial	/alue
Address*	Name	Abbreviation	R/W	Modes 1, 3, and 5	Modes 6 and 7
H'FFC9	Port 6 data direction register	P6DDR	W	H'F8	H'80
H'FFCB	Port 6 data register	P6DR	R/W	H'80	H'80

Note: \* Lower 16 bits of the address.

#### Port 6 Data Direction Register (P6DDR)

P6DDR is an 8-bit write-only register that can select input or output for each pin in port 6.

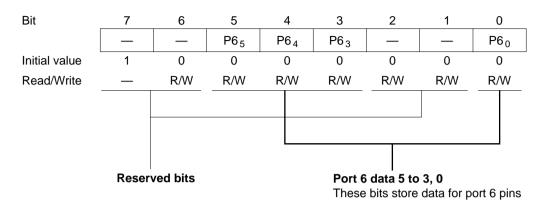
Bit	7	6	5	4	3	2	1	0
	_	—	P65DDR	P6 <sub>4</sub> DDR	P6 <sub>3</sub> DDR	—	_	P6 <sub>0</sub> DDR
Initial value	1	0	0	0	0	0	0	0
Read/Write	—	W	W	W	W	W	W	W
	Reser	ved bits			6 data dir			
				Thes	e bits sele	ct input or	output fo	or port 6 pins

Bits 7, 6, 2, and 1 are reserved.

P6DDR is a write-only register. Its value cannot be read. All bits return 1 when read.

P6DDR is initialized to H'80 by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a P6DDR bit is set to 1, the corresponding pin maintains its output state in software standby mode.

#### Port 6 Data Register (P6DR)



P6DR is an 8-bit readable/writable register that stores data for pins P6<sub>5</sub> to P6<sub>3</sub> and P6<sub>0</sub>.

When a bit in P6DDR is set to 1, if port 6 is read the value of the corresponding P6DR bit is returned directly. When a bit in P6DDR is cleared to 0, if port 6 is read the corresponding pin level is read. Bits 7, 6, 2, and 1 are reserved. Bit 7 cannot be modified and always reads 1. Bits 6, 2, and 1 can be written and read, but cannot be used as ports. If bit 6, 2, or 1 in P6DDR is read while its value is 1, the value of the corresponding bit in P6DR will be read. If bit 6, 2, or 1 in P6DDR is read while its read while its value is 0, it will always read 1.

P6DR is initialized to H'80 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

#### 7.6.3 Pin Functions in Each Mode

## Modes 1, 3, and 5

 $P6_5$  to  $P6_3$  function as bus control output pins.  $P6_0$  is either a bus control input pin or generic input/output pin, functioning as an output pin when bit  $P6_0DDR$  is set to 1 and an input pin when this bit is cleared to 0. Figure 7.17 and table 7.9 indicate the pin functions in modes 1, 3, and 5.

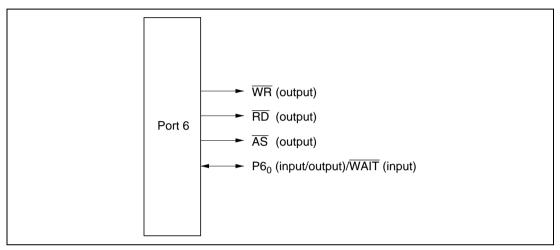


Figure 7.17 Pin Functions in Modes 1, 3, and 5 (Port 6)

Pin Functions a	nd Selection Me	ethod						
Functions as follows regardless of P6 <sub>5</sub> DDR								
P6₅DDR	(	0		1				
Pin function		WR	output					
Functions as folle	Functions as follows regardless of P6₄DDR							
P6₄DDR	(	C		1				
Pin function		RD c	output					
Functions as follows regardless of P6 <sub>3</sub> DDR								
P6 <sub>3</sub> DDR	(	0		1				
Pin function		AS c	output					
Bits WCE7 to WCE0 in WCER, bit WMS1 in WCR, and bit P6 <sub>0</sub> DDR select the pir function as follows								
WCER		All 1s		Not all 1s				
WMS1	(	0	1	—				
P6₀DDR	0	1	0*	0*				
Pin function	P6 <sub>0</sub> input	P6 <sub>0</sub> output		WAIT input				
Note: * Do no	ot set bit P6 <sub>0</sub> DDR	to 1.						
	Functions as follow         P6,DDR         Pin function         Functions as follow         P6,DDR         Pin function         Functions as follow         P6,DDR         Pin function         Bits WCE7 to We function as follow         WCER         WMS1         P6,DDR         Pin function	Functions as follows regardless of $P6_{5}DDR$ Pin function         Functions as follows regardless of $P6_{4}DDR$ Pin function         Functions as follows regardless of $P6_{4}DDR$ Pin function         Functions as follows regardless of $P6_{3}DDR$ Pin function         Bits WCE7 to WCE0 in WCER, b         function as follows         WCER         WMS1 $P6_{0}DDR$ $P6_{0}DDR$ $P6_{0}$ input	P6₅DDR       0         Pin function       WR of         Functions as follows regardless of P6₄DDR       P6₄DDR         P6₄DDR       0         Pin function       RD of         Functions as follows regardless of P6₃DDR       P6₃DDR         Functions as follows regardless of P6₃DDR       P6₃DDR         Bits WCE7 to WCE0 in WCER, bit WMS1 in WCR function as follows       All 1s         WCER       All 1s         WMS1       0         P6₀DDR       0         Pin function       P6₀ output	Functions as follows regardless of $P6_{9}DDR$ $P6_{9}DDR$ 0         Pin function       WR output         Functions as follows regardless of $P6_{4}DDR$ $P6_{4}DDR$ 0         Pin function       RD output         Functions as follows regardless of $P6_{4}DDR$ P6_{4}DDR       0         Pin function       RD output         Functions as follows regardless of $P6_{3}DDR$ P6_{3}DDR       0       RD         Pin function       AS output         Bits WCE7 to WCE0 in WCER, bit WMS1 in WCR, and bit I function as follows         WCER       All 1s         WMS1       0       1         P6_{0}DDR       0       1       0*         Pin function       P6_{0} input       P6_{0} output       1				

#### Table 7.9Port 6 Pin Functions in Modes 1, 3, and 5

#### Modes 6 and 7

Input or output can be selected separately for each pin in port 6. A pin becomes an output pin if the corresponding P6DDR bit is set to 1, and an input pin if this bit is cleared to 0. Figure 7.18 shows the pin functions in modes 6 and 7.

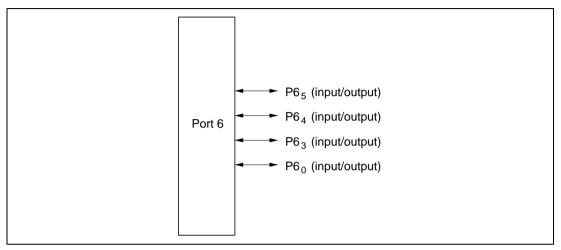


Figure 7.18 Pin Functions in Modes 6 and 7 (Port 6)

### 7.7 Port 7

#### 7.7.1 Overview

Port 7 is an 8-bit input port that is also used for analog input to the A/D converter. The pin functions are the same in all operating modes. Figure 7.19 shows the pin configuration of port 7.

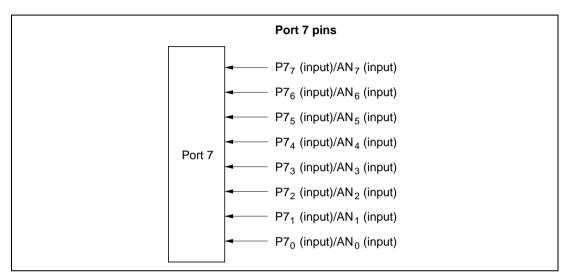


Figure 7.19 Port 7 Pin Configuration

#### 7.7.2 Register Description

Table 7.10 summarizes the port 7 register. Port 7 is an input-only port, so it has no data direction register.

#### Table 7.10 Port 7 Data Register

Address*	Name	Abbreviation	R/W	Initial Value
H'FFCE	Port 7 data register	P7DR	R	Undetermined
Mater W	Leven 40 bits of the coldinary			

Note: \* Lower 16 bits of the address.

#### Port 7 Data Register (P7DR)

Bit	7	6	5	4	3	2	1	0
	P7 <sub>7</sub>	P76	P75	P74	P73	P72	P7 <sub>1</sub>	P70
Initial value	*	*	*	*	*	*	*	*
Read/Write	R	R	R	R	R	R	R	R

Note: \* Determined by pins P<sub>7</sub> to P<sub>0</sub>.

When P7DR is read, the pin levels are always read.

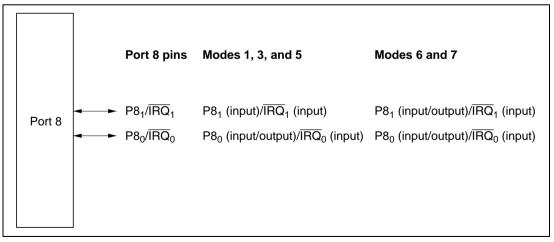
#### 7.8 Port 8

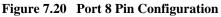
#### 7.8.1 Overview

Port 8 is a 2-bit input/output port that is also used for  $\overline{IRQ}_1$  and  $\overline{IRQ}_0$  input. Figure 7.20 shows the pin configuration of port 8.

Pin P8<sub>0</sub> functions as input/output pin or as an  $\overline{IRQ}_0$  input pin. Pins P8<sub>1</sub> function as either input pins or  $\overline{IRQ}_1$  input pins in modes 1, 3, and 5, and as input/output pins or  $\overline{IRQ}_1$  input pins in modes 6 and 7.

Pins in port 8 can drive one TTL load and a 90-pF capacitive load. They can also drive a Darlington transistor pair. Pins  $P8_1$  and  $P8_0$  have Schmitt-trigger inputs.





#### 7.8.2 Register Descriptions

Table 7.11 summarizes the registers of port 8.

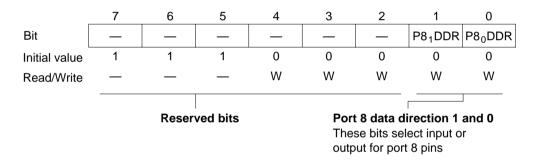
#### Table 7.11 Port 8 Registers

Address*	Name	Abbreviation	R/W	Initial Value
H'FFCD	Port 8 data direction register	P8DDR	W	H'E0
H'FFCF	Port 8 data register	P8DR	R/W	H'E0
Nata: *	Louver 10 bits of the address			

Note: \* Lower 16 bits of the address.

#### Port 8 Data Direction Register (P8DDR)

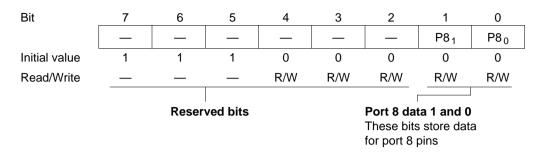
P8DDR is an 8-bit write-only register that can select input or output for each pin in port 8.



P8DDR is initialized to H'E0 by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a P8DDR bit is set to 1, the corresponding pin maintains its output state in software standby mode.

#### Port 8 Data Register (P8DR)

P8DR is an 8-bit readable/writable register that stores data for pins P8<sub>1</sub> to P8<sub>0</sub>.



When a bit in P8DDR is set to 1, if port 8 is read the value of the corresponding P8DR bit is returned directly. When a bit in P8DDR is cleared to 0, if port 8 is read the corresponding pin level is read.

Bits 7 to 2 are reserved. Bits 7 to 5 cannot be modified and always read 1. Bit 4, 3, and 2 can be written and read, but it cannot be used for port input or output. If bit 4, 3, and 2 of P8DDR is read while its value is 1, bit 4, 3 and 2 of P8DR is read directly. If bit 4, 3, and 2 of P8DDR is read while its value is 0, it always reads 1.

P8DR is initialized to H'E0 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

#### 7.8.3 Pin Functions

The port 8 pins are also used for  $\overline{IRQ}_1$  and  $\overline{IRQ}_0$ . Table 7.12 describes the selection of pin functions.

#### Table 7.12Port 8 Pin Functions

#### Pin Pin Functions and Selection Method

P8 <sub>1</sub> /IRQ <sub>1</sub>	Bit P8,DDR s	elects the pin function a	s follows	
	P8,DDR	0	1	
			Modes 6 and 7	
	Pin function	P8, input	Illegal setting	P8, output
			IRQ, input	

$P8_0/\overline{IRQ}_0$	Bit P8 <sub>0</sub> DDR s	elects the pin function as follows	
	P8 <sub>0</sub> DDR	0	1
	Pin function	P8 <sub>0</sub> input	P8, output
		IRQ	input
	L	1	

## 7.9 Port 9

#### 7.9.1 Overview

Port 9 is a 6-bit input/output port that is also used for input and output  $(TxD_0, TxD_1, RxD_0, RxD_1, SCK_0, SCK_1)$  by serial communication interface channels 0 and 1 (SCI0 and SCI1), and for  $\overline{IRQ}_5$  and  $\overline{IRQ}_4$  input.

Port 9 has the same set of pin functions in all operating modes. Figure 7.21 shows the pin configuration of port 9.

Pins in port 9 can drive one TTL load and a 30-pF capacitive load. They can also drive a Darlington transistor pair.

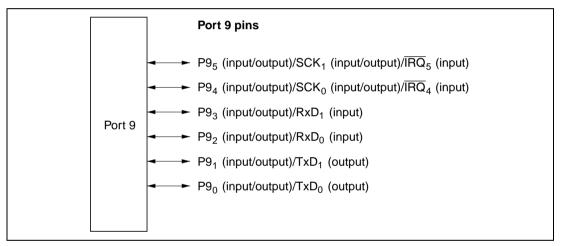


Figure 7.21 Port 9 Pin Configuration

#### 7.9.2 Register Descriptions

Table 7.13 summarizes the registers of port 9.

#### Table 7.13 Port 9 Registers

Address*	Name	Abbreviation	R/W	Initial Value
H'FFD0	Port 9 data direction register	P9DDR	W	H'C0
H'FFD2	Port 9 data register	P9DR	R/W	H'C0

Note: \* Lower 16 bits of the address.

#### Port 9 Data Direction Register (P9DDR)

Bit	7	6	5	4	3	2	1	0
	—		P95DDR	P94DDR	P9 <sub>3</sub> DDR	P9 <sub>2</sub> DDR	P91DDR	P9 <sub>0</sub> DDR
Initial value	1	1	0	0	0	0	0	0
Read/Write	—	—	W	W	W	W	W	W
	Reser	ved bits		The	<b>t 9 data d</b> se bits sel out for por	ect input o		

P9DDR is an 8-bit write-only register that can select input or output for each pin in port 9.

A pin in port 9 becomes an output pin if the corresponding P9DDR bit is set to 1, and an input pin if this bit is cleared to 0.

P9DDR is a write-only register. Its value cannot be read. All bits return 1 when read.

P9DDR is initialized to H'C0 by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a P9DDR bit is set to 1, the corresponding pin maintains its output state in software standby mode.

#### Port 9 Data Register (P9DR)

P9DR is an 8-bit readable/writable register that stores output data for pins P9, to P9,.

Bit	7	6	5	4	3	2	1	0
		_	P9 <sub>5</sub>	P94	P93	P92	P9 <sub>1</sub>	P9 <sub>0</sub>
Initial value	1	1	0	0	0	0	0	0
Read/Write	—	—	R/W	R/W	R/W	R/W	R/W	R/W
	Reserv	ved bits				a <b>ta 5 to 0</b> ts store da pins	ta	

When a bit in P9DDR is set to 1, if port 9 is read the value of the corresponding P9DR bit is returned. When a bit in P9DDR is cleared to 0, if port 9 is read the corresponding pin level is read.

Bits 7 and 6 are reserved. They cannot be modified and are always read as 1.

#### Renesas

P9DR is initialized to H'C0 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

#### 7.9.3 Pin Functions

The port 9 pins are also used for SCI input and output (TxD, RxD, SCK), and for  $\overline{IRQ}_5$  and  $\overline{IRQ}_4$  input. Table 7.14 describes the selection of pin functions.

#### Table 7.14Port 9 Pin Functions

#### Pin Pin Functions and Selection Method

P9₅/SCK₁/ ĪRQ₅	Bit C/A in SMF the pin functio		of SCI1, bits CKE0 and CKE1 in SCR of SCI1, and bit P9 <sub>5</sub> DDR select as follows						
	CKE1		0 1						
	C/Ā		0 1						
	CKE0	(	)	1	_	—			
	P9₅DDR	0	1	—	—	—			
	Pin function	P9₅ input	P9₅ output	SCK <sub>1</sub> input					
				$\overline{IRQ}_{\mathfrak{s}}$	input				

 $\begin{array}{l} P9_4/SCK_0/\\ \overline{IRQ}_4 \end{array} \quad \mbox{Bit C/A in SMR of SCI0, bits CKE0 and CKE1 in SCR of SCI, and bit P9_4DDR select} \\ the pin function as follows \end{array}$ 

CKE1			0		1			
C/A		(	)	1	—			
CKE0	(	C	1	_	—			
P9₄DDR	0	1	_	_	—			
Pin function	P9₄ input	P9₄ output	$SCK_0$ output	$SCK_0$ output	$SCK_0$ input			
	IRQ₄ input							

-----

#### P9,/RxD, Bit RE in SCR of SCI1 and bit P9, DDR select the pin function as follows

RE	(	)	1
P9 <sub>3</sub> DDR	0	1	—
Pin function	P9 <sub>3</sub> input	P9 <sub>3</sub> output	RxD <sub>1</sub> input

# $P9_2/RxD_0$ Bit RE in SCR of SCI<sub>0</sub>, bit SMIF in SCMR, and bit $P9_2DDR$ select the pin function as follows

SMIF		(	)	1
RE	(	)	1	—
P9 <sub>2</sub> DDR	0	1	_	_
Pin function	P9 <sub>2</sub> input	P9 <sub>2</sub> output	RxD₀ input	$RxD_0$ input

P9,/TxD, Bit TE in SCR of SCI1 and bit P91DDR select the pin function as follows

TE	(	)	1
P9,DDR	0	1	_
Pin function	P9₁ input	P9, output	TxD₁ output

P9<sub>0</sub>/TxD<sub>0</sub> Bit TE in SCR of SCI0, bit SMIF in SCMR, and bit P9<sub>0</sub>DDR select the pin function as follows

SMIF		(	)	1		
TE	(	)	1	—		
P9₀DDR	0	1	—	_		
Pin function	P9 <sub>0</sub> input	P9 <sub>0</sub> output	TxD <sub>0</sub> output	TxD <sub>0</sub> output*		

Note: \* Functions as the  $TxD_0$  output pin, but there are two states: one in which the pin is driven, and another in which the pin is at high impedance.

## 7.10 Port A

#### 7.10.1 Overview

Port A is an 8-bit input/output port that is also used for output ( $TP_7$  to  $TP_0$ ) from the programmable timing pattern controller (TPC), input and output ( $TIOCB_2$ ,  $TIOCA_2$ ,  $TIOCB_1$ ,  $TIOCA_1$ ,  $TIOCB_0$ ,  $TIOCA_0$ , TCLKD, TCLKC, TCLKB, TCLKA) by the 16-bit integrated timer unit (ITU), and address output ( $A_{23}$  to  $A_{20}$ ). Figure 7.22 shows the pin configuration of port A.

Pins in port A can drive one TTL load and a 30-pF capacitive load. They can also drive a Darlington transistor pair. Port A has Schmitt-trigger inputs.

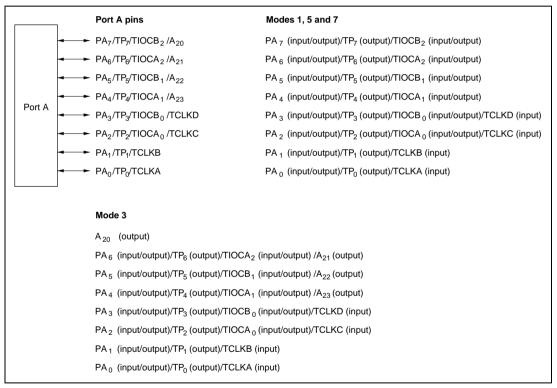


Figure 7.22 Port A Pin Configuration

#### 7.10.2 Register Descriptions

Table 7.15 summarizes the registers of port A.

#### Table 7.15 Port A Registers

				Initial Value			
Address*	Name	Abbreviation	R/W	Modes 1, 5, and 7	Mode 3		
H'FFD1	Port A data direction register	PADDR	W	H'00	H'80		
H'FFD3	Port A data register	PADR	R/W	H'00	H'00		

Note: \* Lower 16 bits of the address.

#### Port A Data Direction Register (PADDR)

PADDR is an 8-bit write-only register that can select input or output for each pin in port A. The corresponding PADDR bit should also be set when a pin is used as a TPC output.

	7	6	5	4	3	2	1	0
Bit	PA7DDR	PA <sub>6</sub> DDR	PA <sub>5</sub> DDR	PA <sub>4</sub> DDR	PA <sub>3</sub> DDR	PA <sub>2</sub> DDR	PA <sub>1</sub> DDR	PA <sub>0</sub> DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W
Initial value	1	0	0	0	0	0	0	0
Read/Write		W	W	W	W	W	W	W
	Initial value Read/Write Initial value	Initial value 0 Read/Write W Initial value 1	BitPA7DDRPA6DDRInitial value00Read/WriteWWInitial value10	BitPA7DDRPA6DDRPA5DDRInitial value000Read/WriteWWWInitial value100	BitPA7DDRPA6DDRPA5DDRPA4DDRInitial value0000Read/WriteWWWWInitial value1000	BitPA7DDRPA6DDRPA5DDRPA4DDRPA3DDRInitial value00000Read/WriteWWWWInitial value1000	BitPA7DDRPA6DDRPA5DDRPA4DDRPA3DDRPA2DDRInitial value000000Read/WriteWWWWWInitial value10000	Bit $PA_7DDR$ $PA_6DDR$ $PA_5DDR$ $PA_4DDR$ $PA_3DDR$ $PA_2DDR$ $PA_1DDR$ Initial value000000Read/WriteWWWWWWInitial value100000

Port A data direction 7 to 0

These bits select input or output for port A pins

A pin in port A becomes an output pin if the corresponding PADDR bit is set to 1, and an input pin if this bit is cleared to 0. However, in mode 3,  $PA_7$  DDR is fixed at 1, and PA7 functions as an address output pin.

PADDR is a write-only register. Its value cannot be read. All bits return 1 when read.

PADDR is initialized to H'00 in modes 1, 5 and 7 and to H'80 in mode 3 by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a PADDR bit is set to 1, the corresponding pin maintains its output state in software standby mode.

#### Port A Data Register (PADR)

Bit	7	6	5	4	3	2	1	0
	PA <sub>7</sub>	$PA_6$	$PA_5$	PA <sub>4</sub>	PA <sub>3</sub>	PA <sub>2</sub>	PA <sub>1</sub>	PA <sub>0</sub>
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
			F	Port A dat	a 7 to 0			

PADR is an 8-bit readable/writable register that stores data for pins PA<sub>7</sub> to PA<sub>0</sub>.

These bits store data for port A pins

When a bit in PADDR is set to 1, if port A is read the value of the corresponding PADR bit is returned directly. When a bit in PADDR is cleared to 0, if port A is read the corresponding pin level is read.

PADR is initialized to H'00 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

When port A pins are used for TPC output, PADR stores output data for TPC output groups 0 and 1. If a bit in the next data enable register (NDERA) is set to 1, the corresponding PADR bit cannot be written. In this case, PADR can be updated only when data is transferred from NDRA.

#### 7.10.3 **Pin Functions**

The port A pins are also used for TPC output ( $TP_7$  to  $TP_0$ ), ITU input/output ( $TIOCB_2$  to  $TIOCB_0$ ,  $TIOCA_2$  to  $TIOCA_0$ ) and input (TCLKD, TCLKC, TCLKB, TCLKA), and as address bus pins ( $A_{23}$  to  $A_{20}$ ). Table 7.16 describes the selection of pin functions.

#### Table 7.16 Port A Pin Functions

#### Pin Pin Functions and Selection Method

 $\begin{array}{ll} \mathsf{PA_{7}}/\mathsf{TP_{7}} & \text{The mode setting, ITU channel 2 settings (bit PWM2 in TMDR and bits IOB2 to IOB0 \\ \mathsf{TIOCB_{2}} & \text{in TIOR2}), \text{ bit NDER7 in NDERA, and bit PA7DDR in PADDR select the pin function} \\ \mathsf{A_{20}} & \text{as follows} \end{array}$ 

Mode			1, 5	to 7			3
ITU channel 2 settings	(1) in tab	le belov	W	(2)	low		
PA <sub>7</sub> DDR	-	_		0	1	1	—
NDER7	-	_		_	0	1	—
Pin function	TIOCB	output		PA <sub>7</sub> input	PA <sub>7</sub> output	TP <sub>7</sub> output	A <sub>20</sub> output
				TI			
Note: * TIOC	B2 input when	IOB2 =	1 and F	PWM2 = 0			
ITU channel 2 settings	(2)	(	1)				
IOB2	(	)					
IOB1	0	0	1	—			
IOB0	0	1	—				

IOA1

IOA0

0

0

0

1

Pin

PA/TP/

TIČCA<sub>2</sub>/ A<sub>21</sub>

#### Pin Functions and Selection Method

The mode setting, bit  $A_{21}E$  in BRCR, ITU channel 2 settings (bit PWM2 in TMDR and bits IOA2 to IOA0 in TIOR2), bit NDER6 in NDERA, and bit PA<sub>6</sub>DDR in PADDR select the pin function as follows

Mode		1, 5	to 7		3					
A <sub>21</sub> E		-	_		1					0
ITU channel 2 settings	(1) in table below	(2) i	(2) in table below				(2) in table below			_
PA <sub>6</sub> DDR	—	0	1	1	-	-	0	1	1	—
NDER6	—	_	0	1	-	_	_	0	1	—
Pin function	TIOCA <sub>2</sub> output	PA <sub>6</sub> input	PA <sub>6</sub> output	TP <sub>6</sub> output		CA <sub>2</sub>	PA <sub>6</sub> input	PA <sub>6</sub> output	TP <sub>6</sub> output	A <sub>21</sub> output
		TIC	DCA <sub>2</sub> inp	ut*			TIOCA <sub>2</sub> input*			
Note: *	TIOCA <sub>2</sub>	input wh	en IOA2	= 1.						
ITU channel 2 settings	(2)		(1)				(2)	(1)		
PWM2			(	)					1	
IOA2			0				1		_	

1

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PA<sub>5</sub>/TP<sub>5</sub>/ TIOCB<sub>1</sub>/ A<sub>22</sub> The mode setting, bit  $A_{22}E$  in BRCR, ITU channel 1 settings (bit PWM1 in TMDR and bits IOB2 to IOB0 in TIOR1), bit NDER5 in NDERA, and bit PA<sub>5</sub>DDR in PADDR select the pin function as follows

Mode		1, 5	to 7		3				
A <sub>22</sub> E		_	_		1				0
ITU channel 1 settings	(1) in table below	(2) ii	n table b	elow	(1) in table below	(2) in table below			—
PA₅DDR	—	0	1	1	—	0	1	1	—
NDER5	—	_	0 1		—	_	0	1	—
Pin function	TIOCB, output	PA₅ input	PA₅ output	TP₅ output	TIOCB, output	PA₅ input	PA₅ output	TP₅ output	A <sub>22</sub> output
		TIC	CB1 inp	ut*		TIOCB1 input*			
Note: *	TIOCB <sub>1</sub> in	nput whe	n IOB2 =	= 1 and F	PWM1 = 0	Э.			
ITU channel 1 settings	(2	2)	(*	1)	(2)				
IOB2		(	)		1				
IOB1	0	0 1			—				
IOB0	0	)	1	—	_				

A<sub>23</sub>

#### Pin **Pin Functions and Selection Method**

#### PA/TP/ The mode setting, bit A23 E in BRCR, ITU channel 1 settings (bit PWM1 in TMDR and bits IOA2 to IOA0 in TIOR1), bit NDER4 in NDERA, and bit PA\_DDR in PADDR select TIOCB,/ the pin function as follows

Mode			3							
A <sub>23</sub> E		_	_		1					0
ITU channel 1 settings	(1) in table below	(2) i	(2) in table below				(2) in table below			—
PA₄DDR	—	0	1	1		-	0	1	1	_
NDER4	—	_	0	1		-	_	0	1	_
Pin function	TIOCA₁ output	PA₄ input	PA₄ output	TP₅ output	TIOCA, output		PA₄ input	PA₄ output	TP₄ output	A <sub>23</sub> output
			TIOCA, input*				TIC	DCA <sub>1</sub> inp	ut*	
Note: *	TIOCA <sub>1</sub> in	nput whe	en IOA2 =	= 1.						
ITU channel 1 settings	(2)		(*	1)	(2) (1)					
PWM1			(	C					1	
IOA2			0			1			—	
IOA1	0		0 1						_	
IOA0	0		1	_				_		

0

IOB0

PA,/TP,/ TIỔCB, TCLKD

#### ITU channel 0 settings (bit PWM0 in TMDR and bits IOB2 to IOB0 in TIOR0), bits TPSC2 to TPSC0 in TCR4 to TCR0, bit NDER3 in NDERA, and bit PA3DDR in PADDR select the pin function as follows

ITU channel 0 settings	(1) in table below			(2) in table below			
PA₃DDR	-		0	1	1		
NDER3	-		—	0	1		
Pin	TIOCB		PA <sub>3</sub> input	PA <sub>3</sub> output	TP <sub>3</sub> output		
function			TIOCB₀ input*1				
	TCLKD input*2						
Notes: 1.	TIOCB0 input when	IOB2 = 1	and PWM	/10 = 0.			
	<ol> <li>TCLKD input when TPSC2 = TPSC1 = TPSC0 = 1 in any of TCR4 to TCR0.</li> </ol>						
ITU channel 0 settings	(2)	(1) (2)					
IOB2	(	C		1			
IOB1	0	0	1				

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PA\_/TP\_/ITU channel 0 settings (bit PWM0 in TMDR and bits IOA2 to IOA0 in TIOR0), bitsTIOCA\_/TPSC2 to TPSC0 in TCR4 to TCR0, bit NDER2 in NDERA, and bit PA\_DDR in PADDRTCLKCselect the pin function as follows

ITU channel 0 settings	(1) in table below	(2) in table below			
PA <sub>2</sub> DDR	_	0	1	1	
NDER2	_	_	0	1	
Pin	TIOCA <sub>0</sub> output	PA <sub>2</sub> input	PA <sub>2</sub> output	TP <sub>2</sub> output	
function		TIOCA <sub>0</sub> input* <sup>1</sup>			
	TCLKC i	nput*2			
Notes: 1. $TIOCA_{0}$ input when IOA2 = 1.					
	TCLKC input when TPSC2 = TPSC1 = TCR0.	1 and TPSC	C0 = 0 in any	of TCR4 to	

ITU channel 0 settings	(2)	(1) (2)			(1)
PWM0		(	1		
IOA2	0			1	—
IOA1	0	0	1		—
IOA0	0	1	—	_	—

PA,/TP,/ Bit NDER1 in NDERA and bit PA1DDR in PADDR select the pin function as follows TCLKB

PA <sub>1</sub> DDR	0	1	1		
NDER1	—	0	1		
Pin	PA₁ input	PA₁ output	TP <sub>1</sub> output		
function TCLKB input*					
Note: * TCLKB input when MDF = 1 in TMDR, or when TPSC2 = 1, TPSC1 = 0,					

and TPSC0 = 1 in any of TCR4 to TCR0.

TCLKA

PA<sub>0</sub>/TP<sub>0</sub>/ Bit NDER0 in NDERA and bit PA0DDR in PADDR select the pin function as follows

PA₀DDR	0	1	1	
NDER0	—	0	1	
Pin	PA <sub>0</sub> input	PA <sub>0</sub> output	TP₀ output	
function		TCLKA input*		
Note: * TCLKA input when MDF = 1 in TMDR, or when TPSC2 = 1 and TPSC1 = TPSC0 = 0 in any of TCR4 to TCR0.				

## 7.11 Port B

#### 7.11.1 Overview

Port B is a 7-bit input/output port that is also used for TPC output  $(TP_{15}, TP_{13} \text{ to } TP_{8})$ , ITU input/output  $(TIOCB_4, TIOCB_3, TIOCA_4, TIOCA_3)$  and ITU output  $(TOCXB_4, TOCXA_4)$ , and ADTRG input to the A/D converter. Port B has the same set of pin functions in all operating modes. Figure 7.23 shows the pin configuration of port B.

Pins in port B can drive one TTL load and a 30-pF capacitive load. They can also drive an LED or a Darlington transistor pair. Pins  $PB_3$  to  $PB_0$  have Schmitt-trigger inputs.

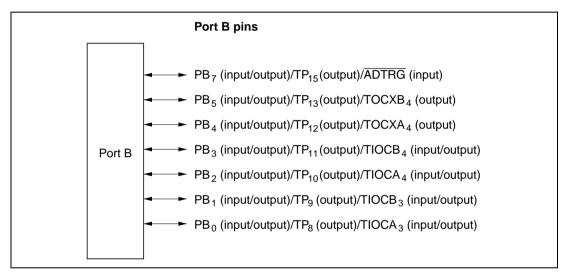


Figure 7.23 Port B Pin Configuration

#### 7.11.2 Register Descriptions

Table 7.17 summarizes the registers of port B.

#### Table 7.17 Port B Registers

Address*	Name	Abbreviation	R/W	Initial Value
H'FFD4	Port B data direction register	PBDDR	W	H'00
H'FFD6	Port B data register	PBDR	R/W	H'00
NI ( at )				

Note: \* Lower 16 bits of the address.

#### Port B Data Direction Register (PBDDR)

Bit	7	6	5	4	3	2	1	0
	PB7DDR	_	PB₅DDR	PB <sub>4</sub> DDR	PB₃DDR	PB <sub>2</sub> DDR	PB1 DDR	PB <sub>0</sub> DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W
					ta 7, 5 to s select inp		out for port	B pins

PBDDR is an 8-bit write-only register that can select input or output for each pin in port B.

A pin in port B becomes an output pin if the corresponding PBDDR bit is set to 1, and an input pin if this bit is cleared to 0.

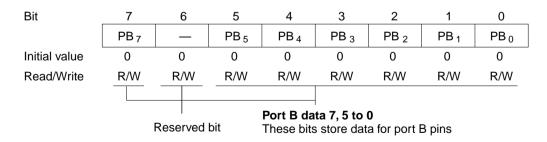
Bit 6 is reserved.

PBDDR is a write-only register. Its value cannot be read. All bits return 1 when read.

PBDDR is initialized to H'00 by a reset and in hardware standby mode. In software standby mode it retains its previous setting. If a PBDDR bit is set to 1, the corresponding pin maintains its output state in software standby mode.

#### Port B Data Register (PBDR)

PBDR is an 8-bit readable/writable register that stores data for pins PB<sub>7</sub>, PB<sub>5</sub> to PB<sub>0</sub>.



When a bit in PBDDR is set to 1, if port B is read the value of the corresponding PBDR bit is returned directly. When a bit in PBDDR is cleared to 0, if port B is read the corresponding pin level is read. Bit 6 is reserved. Bit 6 can be written and read, but cannot be used for a port input or output.

#### Renesas

If bit 6 in PBDDR is read while its value is 1, the value of bit 6 in PBDR will be read directly. If bit 6 in PBDDR is read while its value is 0, it will always be read as 1.

PBDR is initialized to H'00 by a reset and in hardware standby mode. In software standby mode it retains its previous setting.

When port B pins are used for TPC output, PBDR stores output data for TPC output groups 2 and 3. If a bit in the next data enable register (NDERB) is set to 1, the corresponding PBDR bit cannot be written. In this case, PBDR can be updated only when data is transferred from NDRB.

#### 7.11.3 Pin Functions

The port B pins are also used for TPC output  $(TP_{15}, TP_{13} \text{ to } TP_8)$ , ITU input/output  $(TIOCB_4, TIOCB_3, TIOCA_4, TIOCA_3)$  and output  $(TOCXB_4, TOCXA_4)$ , and  $\overline{ADTRG}$  input. Table 7.18 describes the selection of pin functions.

#### Table 7.18Port B Pin Functions

#### Pin Pin Functions and Selection Method

PB <sub>7</sub> / TP <sub>15</sub> / ADTRG	Bit TRGE in ADCR, bit NDER15 in NDERB and bit PB7DDR in PBDDR select the pin function as follows					
ADTRG	PB,DDR	1	1			
	NDER15	—	0	1		
	Pin function	PB <sub>7</sub> output	TP <sub>15</sub> output			
		ADTRG input*				

Notes: \*  $\overline{\text{ADTRG}}$  input when TRGE = 1.

PB,/ITU channel 4 settings (bit CMD1 in TFCR and bit EXB4 in TOER), bit NDER13 inTP1,/NDERB, and bit PB,DDR in PBDDR select the pin function as follows

EXB4, CMD1		Not both 1		Both 1
PB₅DDR	0	1	1	—
NDER13	_	0	1	—
Pin function	PB₅ input	PB₅ output	TP <sub>13</sub> output	TOCXB <sub>4</sub> output

PB₄/
TP <sub>12</sub> /
TOCXA

TOCXB,

# ITU channel 4 settings (bit CMD1 in TFCR and bit EXA4 in TOER), bit NDER12 in NDERB, and bit PB4DDR in PBDDR select the pin function as follows

EXA4, CMD1	Not both 1			Both 1
$PB_4DDR$	0	1	1	_
NDER12	_	0	1	_
Pin function	PB <sub>4</sub> input	PB <sub>4</sub> output	TP <sub>12</sub> output	TOCXA <sub>4</sub> output

PB,/ITU channel 4 settings (bit PWM4 in TMDR, bit CMD1 in TFCR, bit EB4 in TOER, and<br/>bits IOB2 to IOB0 in TIOR4), bit NDER11 in NDERB, and bit PB3DDR in PBDDR select<br/>the pin function as follows

ITU channel 4 settings	(1) in table below	(2) in table below			
PB <sub>3</sub> DDR	_	0	1	1	
NDER11	_	_	0	1	
Pin function	TIOCB <sub>4</sub> output	PB <sub>3</sub> input	PB <sub>3</sub> output	TP <sub>11</sub> output	
		TIOCB <sub>4</sub> input*			

Note: \* TIOCB<sub>4</sub> input when CMD1 = PWM4 = 0 and IOB2 = 1.

ITU channel 4 settings	(2)	(2)	(1)		(2)	(1)	
EB4	0		1				
CMD1	_		1				
IOB2	—	0	—				
IOB1	_	0	—				
IOB0		0	1			—	

PB<sub>2</sub>/ TP<sub>10</sub>/ TIOCA<sub>4</sub> ITU channel 4 settings (bit CMD1 in TFCR, bit EA4 in TOER, bit PWM4 in TMDR, and bits IOA2 to IOA0 in TIOR4), bit NDER10 in NDERB, and bit PB<sub>2</sub>DDR in PBDDR select the pin function as follows

ITU channel 4 settings	(1) in table below	(2) in table below				
PB <sub>2</sub> DDR	_	0	1	1		
NDER11	_	_	0	1		
Pin function	TIOCA <sub>4</sub> output	PB <sub>2</sub> input	PB <sub>2</sub> output	TP <sub>10</sub> output		
		TIOCA <sub>4</sub> input*				

Note: \* TIOCA<sub>4</sub> input when CMD1 = PWM4 = 0 and IOB2 = 1.

ITU channel 4 settings	(2)	(2) (1) (2) (1)				1)	
EA4	0		1				
CMD1	—		0				
PWM4	—	0 1					—
IOA2	—	0	0	0	1	_	—
IOA1	—	0 0 1 — —					—
IOA0		0	1				—

# PB,/TP,/ ITU channel 3 settings (bit PWM3 in TMDR, bit CMD1 in TFCR, bit EB3 in TOER, and bits IOB2 to IOB0 in TIOR3), bit NDER11 in NDERB, and bit PB,DDR in PBDDR select the pin function as follows

ITU channel 3 settings	(1) in table below	(2) in table below				
PB,DDR	—	0	1	1		
NDER9	_	_	0	1		
Pin function	TIOCB <sub>3</sub> output	PB <sub>1</sub> input	PB <sub>1</sub> output	TP <sub>9</sub> output		
		TIOCB <sub>3</sub> input*				

Note: \* TIOCB3 input when CMD1 = PWM3 = 0 and IOB2 = 1.

ITU channel 3 settings	(2)	(2)	(1)		(2)	(1)		
EB3	0		1					
CMD1	_		1					
IOB2	_	0	—					
IOB1	_	0	—					
IOB0		0	1			_		

PB\_/TP\_/ ITU channel 3 settings (bit CMD1 in TFCR, bit EA3 in TOER, bit PWM3 in TMDR, and bits IOA2 to IOA0 in TIOR3), bit NDER8 in NDERB, and bit PB\_DDR in PBDDR select TIČCA

the pin function as follows

ITU channel 3 settings	(1) in table below	(2) in table below				
PB₀DDR	_	0	1	1		
NDER8	_	_	0	1		
Pin function	unction TIOCA <sub>3</sub> output		PB <sub>o</sub> input PB <sub>o</sub> output TP <sub>s</sub> output			
		TIOCA <sub>3</sub> input*				

TIOCA, input when CMD1 = PWM3 = 0 and IOA2 = 1. Note:

ITU channel 3 settings	(2)	(2) (1) (2) (1)			)		
EA3	0		1				
CMD1	_		0				
PWM3	_	0 1					—
IOA2	_	0	0	0	1		—
IOA1	—	0 0 1 — —					—
IOA0		0	1	_			—

## Section 8 16-Bit Integrated Timer Unit (ITU)

## 8.1 Overview

The H8/3039 Group has a built-in 16-bit integrated timer-pulse unit (ITU) with five 16-bit timer channels.

#### 8.1.1 Features

ITU features are listed below.

- Capability to process up to 12 pulse outputs or 10 pulse inputs
- Ten general registers (GRs, two per channel) with independently-assignable output compare or input capture functions
- Selection of eight counter clock sources for each channel: Internal clocks: φ, φ/2, φ/4, φ/8 External clocks: TCLKA, TCLKB, TCLKC, TCLKD
- Five operating modes selectable in all channels:
  - Waveform output by compare match
    - Selection of 0 output, 1 output, or toggle output (only 0 or 1 output in channel 2)
  - Input capture function

Rising edge, falling edge, or both edges (selectable)

- Counter clearing function

Counters can be cleared by compare match or input capture

- Synchronization

Two or more timer counters (TCNTs) can be preset simultaneously, or cleared simultaneously by compare match or input capture. Counter synchronization enables synchronous register input and output.

— PWM mode

PWM output can be provided with an arbitrary duty cycle. With synchronization, up to five-phase PWM output is possible

• Phase counting mode selectable in channel 2

Two-phase encoder output can be counted automatically.

### Renesas

- Three additional modes selectable in channels 3 and 4
  - Reset-synchronized PWM mode

If channels 3 and 4 are combined, three-phase PWM output is possible with three pairs of complementary waveforms.

- Complementary PWM mode

If channels 3 and 4 are combined, three-phase PWM output is possible with three pairs of non-overlapping complementary waveforms.

— Buffering

Input capture registers can be double-buffered. Output compare registers can be updated automatically.

• High-speed access via internal 16-bit bus

The 16-bit timer counters, general registers, and buffer registers can be accessed at high speed via a 16-bit bus.

• Fifteen interrupt sources

Each channel has two compare match/input capture interrupts and an overflow interrupt. All interrupts can be requested independently.

• Output triggering of programmable pattern controller (TPC)

Compare match/input capture signals from channels 0 to 3 can be used as TPC output triggers.

Table 8.1 summarizes the ITU functions.

Item		Channel 0	Channel 1	Channel 2	Channel 3	Channel 4		
Clock sour	ces		Internal clocks: φ, φ/2, φ/4, φ/8 External clocks: TCLKA, TCLKB, TCLKC, TCLKD, selectable independently					
General re (output cor input captu	0	GRA0, GRB0	GRA1, GRB1	GRA2, GRB2	GRA3, GRB3	GRA4, GRB4		
Buffer regi	sters	_	_	_	BRA3, BRB3	BRA4, BRB4		
Input/outpu	ut pins	TIOCA <sub>0</sub> , TIOCB <sub>0</sub>	TIOCA,, TIOCB	TIOCA <sub>2</sub> , TIOCB <sub>2</sub>	TIOCA <sub>3</sub> , TIOCB <sub>3</sub>	$TIOCA_4$ , $TIOCB_4$		
Output pin	S	—	—	—	—	TOCXA₄, TOCXB₄		
Counter cle function	earing	GRA0/GRB0 compare match or input capture	•	GRA2/GRB2 compare match or input capture	GRA3/GRB3 compare match or input capture	GRA4/GRB4 compare match or input capture		
Compare	0	0	0	0	0	0		
match	1	0	0	0	0	0		
output	Toggle	0	0		0	0		
Input captu	ure function	0	0	0	0	0		
Synchroniz	zation	0	0	0	0	0		
PWM mod	е	0	0	0	0	0		
Reset-syno PWM mod		_	_	_	0	0		
Compleme mode	entary PWM	_	_	_	0	0		
Phase cou	nting mode	_	_	0	_	_		
Buffering		_	_		0	0		
Interrupt sources		Three sources	Three sources	Three sources	Three sources	Three sources		
		Compare match/input capture A0	Compare match/input capture A1	<ul> <li>Compare match/input capture A2</li> </ul>	<ul> <li>Compare match/input capture A3</li> </ul>	<ul> <li>Compare match/input capture A4</li> </ul>		
		Compare match/input capture B0	<ul> <li>Compare match/input capture B1</li> </ul>	<ul> <li>Compare match/input capture B2</li> </ul>	<ul> <li>Compare match/input capture B3</li> </ul>	<ul> <li>Compare match/input capture B4</li> </ul>		
		<ul> <li>Overflow</li> </ul>	<ul> <li>Overflow</li> </ul>	Overflow	Overflow	Overflow		
Legend:								

#### Table 8.1ITU Functions

Legend:

O: Available

-: Not available

## 8.1.2 Block Diagrams

## ITU Block Diagram (Overall)

Figure 8.1 is a block diagram of the ITU.

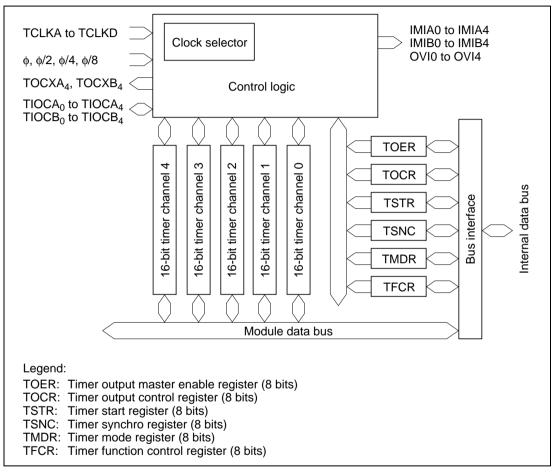


Figure 8.1 ITU Block Diagram (Overall)

#### **Block Diagram of Channels 0 and 1**

ITU channels 0 and 1 are functionally identical. Both have the structure shown in figure 8.2.

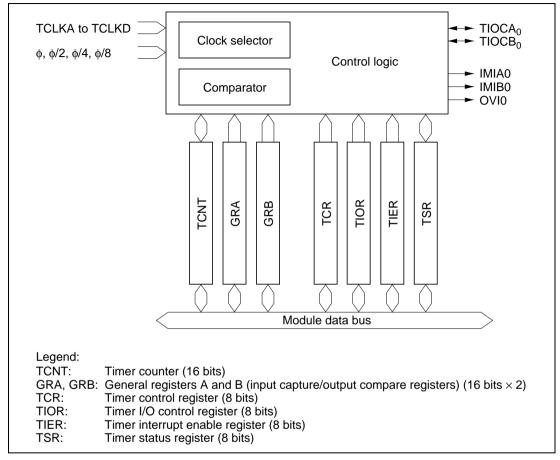


Figure 8.2 Block Diagram of Channels 0 and 1 (for Channel 0)

## **Block Diagram of Channel 2**

Figure 8.3 is a block diagram of channel 2. This is the channel that provides only 0 output and 1 output.

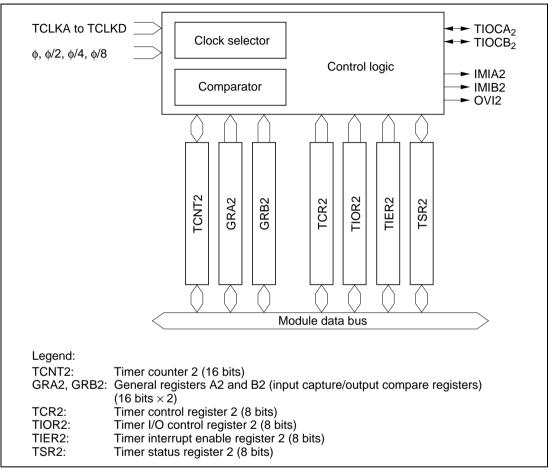


Figure 8.3 Block Diagram of Channel 2

#### **Block Diagrams of Channels 3 and 4**

Figure 8.4 is a block diagram of channel 3. Figure 8.5 is a block diagram of channel 4.

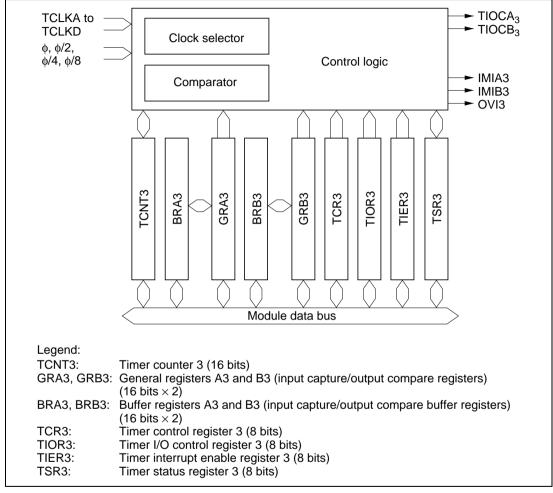


Figure 8.4 Block Diagram of Channel 3

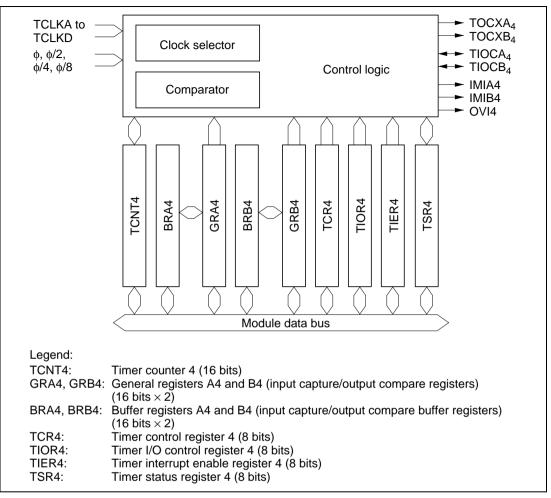


Figure 8.5 Block Diagram of Channel 4

## 8.1.3 Input/Output Pins

Table 8.2 summarizes the ITU pins.

## Table 8.2 ITU Pins

Channel	Name	Abbre- viation	Input/ Output	Function
Common	Clock input A	TCLKA	Input	External clock A input pin (phase-A input pin in phase counting mode)
	Clock input B	TCLKB	Input	External clock B input pin (phase-B input pin in phase counting mode)
	Clock input C	TCLKC	Input	External clock C input pin
	Clock input D	TCLKD	Input	External clock D input pin
0	Input capture/output compare A0		Input/ output	GRA0 output compare or input capture pin PWM output pin in PWM mode
	Input capture/output compare B0	TIOCB <sub>0</sub>	Input/ output	GRB0 output compare or input capture pin
1	Input capture/output compare A1	TIOCA <sub>1</sub>	Input/ output	GRA1 output compare or input capture pin PWM output pin in PWM mode
	Input capture/output compare B1	TIOCB <sub>1</sub>	Input/ output	GRB1 output compare or input capture pin
2	Input capture/output compare A2	TIOCA <sub>2</sub>	Input/ output	GRA2 output compare or input capture pin PWM output pin in PWM mode
	Input capture/output compare B2	TIOCB <sub>2</sub>	Input/ output	GRB2 output compare or input capture pin
3	Input capture/output compare A3	TIOCA <sub>3</sub>	Input/ output	GRA3 output compare or input capture pin PWM output pin in PWM mode, complementary PWM mode, or reset- synchronized PWM mode
	Input capture/output compare B3	TIOCB <sub>3</sub>	Input/ output	GRB3 output compare or input capture pin PWM output pin in complementary PWM mode or reset-synchronized PWM mode

Channel	Name	Abbre- viation	Input/ Output	Function
4	Input capture/output compare A4	TIOCA <sub>4</sub>	Input/ output	GRA4 output compare or input capture pin PWM output pin in PWM mode, complementary PWM mode, or reset- synchronized PWM mode
	Input capture/output compare B4	TIOCB <sub>4</sub>	Input/ output	GRB4 output compare or input capture pin PWM output pin in complementary PWM mode or reset-synchronized PWM mode
	Output compare XA4	TOCXA <sub>4</sub>	Output	PWM output pin in complementary PWM mode or reset-synchronized PWM mode
_	Output compare XB4	TOCXB₄	Output	PWM output pin in complementary PWM mode or reset-synchronized PWM mode

## 8.1.4 Register Configuration

Table 8.3 summarizes the ITU registers.

## Table 8.3ITU Registers

Channel	Address*1	Name	Abbre- viation	R/W	Initial Value
Common	H'FF60	Timer start register	TSTR	R/W	H'E0
	H'FF61	Timer synchro register	TSNC	R/W	H'E0
	H'FF62	Timer mode register	TMDR	R/W	H'80
	H'FF63	Timer function control register	TFCR	R/W	H'C0
	H'FF90	Timer output master enable register	TOER	R/W	H'FF
	H'FF91	Timer output control register	TOCR	R/W	H'FF
0	H'FF64	Timer control register 0	TCR0	R/W	H'80
	H'FF65	Timer I/O control register 0	TIOR0	R/W	H'88
	H'FF66	Timer interrupt enable register 0	TIER0	R/W	H'F8
	H'FF67	Timer status register 0	TSR0	R/(W)* <sup>2</sup>	H'F8
	H'FF68	Timer counter 0 (high)	TCNT0H	R/W	H'00
	H'FF69	Timer counter 0 (low)	<b>TCNT0L</b>	R/W	H'00
	H'FF6A	General register A0 (high)	GRA0H	R/W	H'FF
	H'FF6B	General register A0 (low)	GRA0L	R/W	H'FF
	H'FF6C	General register B0 (high)	GRB0H	R/W	H'FF
	H'FF6D	General register B0 (low)	GRB0L	R/W	H'FF
1	H'FF6E	Timer control register 1	TCR1	R/W	H'80
	H'FF6F	Timer I/O control register 1	TIOR1	R/W	H'88
	H'FF70	Timer interrupt enable register 1	TIER1	R/W	H'F8
	H'FF71	Timer status register 1	TSR1	R/(W)* <sup>2</sup>	H'F8
	H'FF72	Timer counter 1 (high)	TCNT1H	R/W	H'00
	H'FF73	Timer counter 1 (low)	TCNT1L	R/W	H'00
	H'FF74	General register A1 (high)	GRA1H	R/W	H'FF
	H'FF75	General register A1 (low)	GRA1L	R/W	H'FF
	H'FF76	General register B1 (high)	GRB1H	R/W	H'FF
	H'FF77	General register B1 (low)	GRB1L	R/W	H'FF

Channel	Address*1	Name	Abbre- viation	R/W	Initial Value
2	H'FF78	Timer control register 2	TCR2	R/W	H'80
	H'FF79	Timer I/O control register 2	TIOR2	R/W	H'88
	H'FF7A	Timer interrupt enable register 2	TIER2	R/W	H'F8
	H'FF7B	Timer status register 2	TSR2	R/(W)* <sup>2</sup>	H'F8
	H'FF7C	Timer counter 2 (high)	TCNT2H	R/W	H'00
	H'FF7D	Timer counter 2 (low)	TCNT2L	R/W	H'00
	H'FF7E	General register A2 (high)	GRA2H	R/W	H'FF
	H'FF7F	General register A2 (low)	GRA2L	R/W	H'FF
	H'FF80	General register B2 (high)	GRB2H	R/W	H'FF
	H'FF81	General register B2 (low)	GRB2L	R/W	H'FF
3	H'FF82	Timer control register 3	TCR3	R/W	H'80
	H'FF83	Timer I/O control register 3	TIOR3	R/W	H'88
	H'FF84	Timer interrupt enable register 3	TIER3	R/W	H'F8
	H'FF85	Timer status register 3	TSR3	R/(W)* <sup>2</sup>	H'F8
	H'FF86	Timer counter 3 (high)	TCNT3H	R/W	H'00
	H'FF87	Timer counter 3 (low)	TCNT3L	R/W	H'00
	H'FF88	General register A3 (high)	GRA3H	R/W	H'FF
	H'FF89	General register A3 (low)	GRA3L	R/W	H'FF
	H'FF8A	General register B3 (high)	GRB3H	R/W	H'FF
	H'FF8B	General register B3 (low)	GRB3L	R/W	H'FF
	H'FF8C	Buffer register A3 (high)	<b>BRA3H</b>	R/W	H'FF
	H'FF8D	Buffer register A3 (low)	BRA3L	R/W	H'FF
	H'FF8E	Buffer register B3 (high)	BRB3H	R/W	H'FF
	H'FF8F	Buffer register B3 (low)	BRB3L	R/W	H'FF

Channel	Address*1	Name	Abbre- viation	R/W	Initial Value
4	H'FF92	Timer control register 4	TCR4	R/W	H'80
	H'FF93	Timer I/O control register 4	TIOR4	R/W	H'88
	H'FF94	Timer interrupt enable register 4	TIER4	R/W	H'F8
	H'FF95	Timer status register 4	TSR4	R/(W)* <sup>2</sup>	H'F8
	H'FF96	Timer counter 4 (high)	TCNT4H	R/W	H'00
	H'FF97	Timer counter 4 (low)	TCNT4L	R/W	H'00
	H'FF98	General register A4 (high)	GRA4H	R/W	H'FF
	H'FF99	General register A4 (low)	GRA4L	R/W	H'FF
	H'FF9A	General register B4 (high)	GRB4H	R/W	H'FF
	H'FF9B	General register B4 (low)	GRB4L	R/W	H'FF
	H'FF9C	Buffer register A4 (high)	BRA4H	R/W	H'FF
	H'FF9D	Buffer register A4 (low)	BRA4L	R/W	H'FF
	H'FF9E	Buffer register B4 (high)	BRB4H	R/W	H'FF
	H'FF9F	Buffer register B4 (low)	BRB4L	R/W	H'FF

Notes: 1. The lower 16 bits of the address are indicated.

2. Only 0 can be written, to clear flags.

# 8.2 **Register Descriptions**

# 8.2.1 Timer Start Register (TSTR)

TSTR is an 8-bit readable/writable register that starts and stops the timer counter (TCNT) in channels 0 to 4.

Bit	7	6	5	4	3	2	1	0
	_	_	—	STR4	STR3	STR2	STR1	STR0
Initial value	1	1	1	0	0	0	0	0
Read/Write	—	—	—	R/W	R/W	R/W	R/W	R/W
Reserved bits			Thes	nter start 4 e bits start TCNT4 to	t and			

TSTR is initialized to H'E0 by a reset and in standby mode.

Bits 7 to 5—Reserved: These bits cannot be modified and are always read as 1.

Bit 4—Counter Start 4 (STR4): Starts and stops timer counter 4 (TCNT4).

Bit4 STR4	Description	
0	TCNT4 is halted	(Initial value)
1	TCNT4 is counting	

Bit 3—Counter Start 3 (STR3): Starts and stops timer counter 3 (TCNT3).

Bit 3 STR3	Description	
0	TCNT3 is halted	(Initial value)
1	TCNT3 is counting	

Bit 2—Counter Start 2 (STR2): Starts and stops timer counter 2 (TCNT2).

Bit 2 STR2	Description	
0	TCNT2 is halted	(Initial value)
1	TCNT2 is counting	

Bit 1—Counter Start 1 (STR1): Starts and stops timer counter 1 (TCNT1).

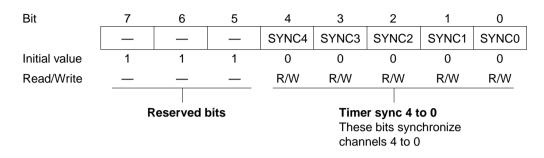
Bit 1 STR1	Description	
0	TCNT1 is halted	(Initial value)
1	TCNT1 is counting	

Bit 0—Counter Start 0 (STR0): Starts and stops timer counter 0 (TCNT0).

Bit 0 STR0	Description	
0	TCNT0 is halted	(Initial value)
1	TCNT0 is counting	

## 8.2.2 Timer Synchro Register (TSNC)

TSNC is an 8-bit readable/writable register that selects whether channels 0 to 4 operate independently or synchronously. Channels are synchronized by setting the corresponding bits to 1.



TSNC is initialized to H'E0 by a reset and in standby mode.

Bits 7 to 5—Reserved: These bits cannot be modified and are always read as 1.

**Bit 4—Timer Sync 4 (SYNC4):** Selects whether channel 4 operates independently or synchronously.

Bit 4 SYNC4	Description	
0	Channel 4's timer counter (TCNT4) operates independently TCNT4 cleared independently of other channels	is preset and (Initial value)
1	Channel 4 operates synchronously	
	TCNT4 can be synchronously preset and cleared	

**Bit 3—Timer Sync 3 (SYNC3):** Selects whether channel 3 operates independently or synchronously.

Bit 3 SYNC3	Description	
0	Channel 3's timer counter (TCNT3) operates independently TCNT3 is preset and cleared independently of other channels (Initial value)	¥)
1	Channel 3 operates synchronously	
	TCNT3 can be synchronously preset and cleared	

**Bit 2—Timer Sync 2 (SYNC2):** Selects whether channel 2 operates independently or synchronously.

#### Bit 2 SYNC2 D

SYNC2	Description		
0	Channel 2's timer counter (TCNT2) operates independently TCNT2 is preset and		
	cleared independently of other channels	(Initial value)	
1	Channel 2 operates synchronously		
	TCNT2 can be synchronously preset and cleared		

**Bit 1—Timer Sync 1 (SYNC1):** Selects whether channel 1 operates independently or synchronously.

## Bit 1

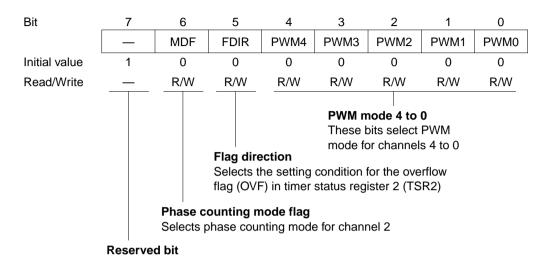
SYNC1	Description	
0	Channel 1's timer counter (TCNT1) operates independently TCNT1 cleared independently of other channels	is preset and (Initial value)
1	Channel 1 operates synchronously	
	TCNT1 can be synchronously preset and cleared	

**Bit 0—Timer Sync 0 (SYNC0):** Selects whether channel 0 operates independently or synchronously.

Bit 0 SYNC0	Description	
0	Channel 0's timer counter (TCNT0) operates independently TCNT0 is preset and cleared independently of other channels (Initial value)	_
1	Channel 0 operates synchronously	_
	TCNT0 can be synchronously preset and cleared	

## 8.2.3 Timer Mode Register (TMDR)

TMDR is an 8-bit readable/writable register that selects PWM mode for channels 0 to 4. It also selects phase counting mode and the overflow flag (OVF) setting conditions for channel 2.



TMDR is initialized to H'80 by a reset and in standby mode.

Bit 7—Reserved: This bit cannot be modified and is always read as 1.

**Bit 6—Phase Counting Mode Flag (MDF):** Selects whether channel 2 operates normally or in phase counting mode.

Bit 6 MDF	Description	
0	Channel 2 operates normally	(Initial value)
1	Channel 2 operates in phase counting mode	

When MDF is set to 1 to select phase counting mode, timer counter 2 (TCNT2) operates as an up/down-counter and pins TCLKA and TCLKB become counter clock input pins. TCNT2 counts both rising and falling edges of TCLKA and TCLKB, and counts up or down as follows.

Counting Direction	Down-Counting			Down-Counting Up-Counting				
TCLKA pin	-	High	₹	Low	_ <b>-</b>	Low	₹	High
TCLKB pin	Low	–	High	<b> </b> ▼_	High	_ <b>_</b>	Low	┣

In phase counting mode channel 2 operates as above regardless of the external clock edges selected by bits CKEG1 and CKEG0 and the clock source selected by bits TPSC2 to TPSC0 in timer control register 2 (TCR2). Phase counting mode takes precedence over these settings.

The counter clearing condition selected by the CCLR1 and CCLR0 bits in TCR2 and the compare match/input capture settings and interrupt functions of timer I/O control register 2 (TIOR2), timer interrupt enable register 2 (TIER2), and timer status register 2 (TSR2) remain effective in phase counting mode.

**Bit 5—Flag Direction (FDIR):** Designates the setting condition for the overflow flag (OVF) in timer status register 2 (TSR2). The FDIR designation is valid in all modes in channel 2.

Bit 5 FDIR	Description	
0	OVF is set to 1 in TSR2 when TCNT2 overflows or underflows	(Initial value)
1	OVF is set to 1 in TSR2 when TCNT2 overflows	

Bit 4—PWM Mode 4 (PWM4): Selects whether channel 4 operates normally or in PWM mode.

Bit 4 PWM4	Description	
0	Channel 4 operates normally	(Initial value)
1	Channel 4 operates in PWM mode	

When bit PWM4 is set to 1 to select PWM mode, pin TIOCA4 becomes a PWM output pin. The output goes to 1 at compare match with general register A4 (GRA4), and to 0 at compare match with general register B4 (GRB4).

If complementary PWM mode or reset-synchronized PWM mode is selected by bits CMD1 and CMD0 in the timer function control register (TFCR), the CMD1 and CMD0 setting takes precedence and the PWM4 setting is ignored.

Bit 3—PWM Mode 3 (PWM3): Selects whether channel 3 operates normally or in PWM mode.

Bit 3 PWM3	Description	
0	Channel 3 operates normally	(Initial value)
1	Channel 3 operates in PWM mode	

#### Renesas

When bit PWM3 is set to 1 to select PWM mode, pin TIOCA3 becomes a PWM output pin. The output goes to 1 at compare match with general register A3 (GRA3), and to 0 at compare match with general register B3 (GRB3).

If complementary PWM mode or reset-synchronized PWM mode is selected by bits CMD1 and CMD0 in the timer function control register (TFCR), the CMD1 and CMD0 setting takes precedence and the PWM3 setting is ignored.

Bit 2—PWM Mode 2 (PWM2): Selects whether channel 2 operates normally or in PWM mode.

Bit 2 PWM2	Description	
0	Channel 2 operates normally	(Initial value)
1	Channel 2 operates in PWM mode	

When bit PWM2 is set to 1 to select PWM mode, pin TIOCA2 becomes a PWM output pin. The output goes to 1 at compare match with general register A2 (GRA2), and to 0 at compare match with general register B2 (GRB2).

Bit 1—PWM Mode 1 (PWM1): Selects whether channel 1 operates normally or in PWM mode.

Bit 1 PWM1	Description	
0	Channel 1 operates normally	(Initial value)
1	Channel 1 operates in PWM mode	

When bit PWM1 is set to 1 to select PWM mode, pin TIOCA1 becomes a PWM output pin. The output goes to 1 at compare match with general register A1 (GRA1), and to 0 at compare match with general register B1 (GRB1).

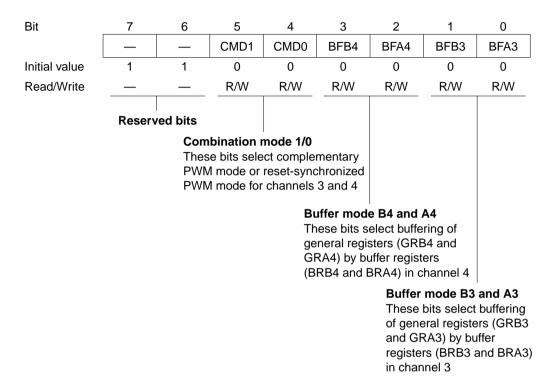
Bit 0—PWM Mode 0 (PWM0): Selects whether channel 0 operates normally or in PWM mode.

Bit 0 PWM0	Description	
0	Channel 0 operates normally	(Initial value)
1	Channel 0 operates in PWM mode	

When bit PWM0 is set to 1 to select PWM mode, pin TIOCA0 becomes a PWM output pin. The output goes to 1 at compare match with general register A0 (GRA0), and to 0 at compare match with general register B0 (GRB0).

## 8.2.4 Timer Function Control Register (TFCR)

TFCR is an 8-bit readable/writable register that selects complementary PWM mode, resetsynchronized PWM mode, and buffering for channels 3 and 4.



TFCR is initialized to H'C0 by a reset and in standby mode.

Bits 7 and 6—Reserved: These bits cannot be modified and are always read as 1.

**Bits 5 and 4—Combination Mode 1 and 0 (CMD1, CMD0):** These bits select whether channels 3 and 4 operate in normal mode, complementary PWM mode, or reset-synchronized PWM mode.

Bit 5 CMD1	Bit 4 CMD0	Description	
0	0	Channels 3 and 4 operate normally	(Initial value)
	1		
1	0	Channels 3 and 4 operate together in complementary PWM mode	
	1	Channels 3 and 4 operate together in reset-synchronized	d PWM mode

Before selecting reset-synchronized PWM mode or complementary PWM mode, halt the timer counter or counters that will be used in these modes.

When these bits select complementary PWM mode or reset-synchronized PWM mode, they take precedence over the setting of the PWM mode bits (PWM4 and PWM3) in TMDR. Settings of timer sync bits SYNC4 and SYNC3 in the timer synchro register (TSNC) are valid in complementary PWM mode and reset-synchronized PWM mode, however. When complementary PWM mode is selected, channels 3 and 4 must not be synchronized (do not set bits SYNC3 and SYNC4 both to 1 in TSNC).

**Bit 3—Buffer Mode B4 (BFB4):** Selects whether GRB4 operates normally in channel 4, or whether GRB4 is buffered by BRB4.

Bit 3 BFB4	Description	
0	GRB4 operates normally	(Initial value)
1	GRB4 is buffered by BRB4	

**Bit 2—Buffer Mode A4 (BFA4):** Selects whether GRA4 operates normally in channel 4, or whether GRA4 is buffered by BRA4.

Bit 2 BFA4	Description	
0	GRA4 operates normally	(Initial value)
1	GRA4 is buffered by BRA4	

**Bit 1—Buffer Mode B3 (BFB3):** Selects whether GRB3 operates normally in channel 3, or whether GRB3 is buffered by BRB3.

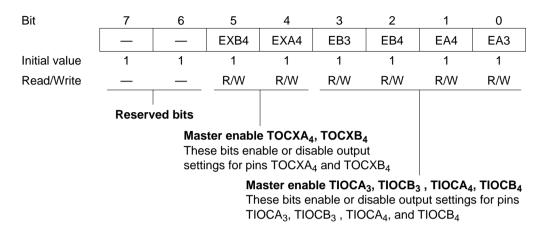
Bit 1 BFB3	Description	
0	GRB3 operates normally	(Initial value)
1	GRB3 is buffered by BRB3	

**Bit 0—Buffer Mode A3 (BFA3):** Selects whether GRA3 operates normally in channel 3, or whether GRA3 is buffered by BRA3.

Bit 0 BFA3	Description	
0	GRA3 operates normally	(Initial value)
1	GRA3 is buffered by BRA3	

## 8.2.5 Timer Output Master Enable Register (TOER)

TOER is an 8-bit readable/writable register that enables or disables output settings for channels 3 and 4.



TOER is initialized to H'FF by a reset and in standby mode.

Bits 7 and 6—Reserved: These bits cannot be modified and are always read as 1.

Bit 5—Master Enable TOCXB<sub>4</sub> (EXB4): Enables or disables ITU output at pin TOCXB<sub>4</sub>.

Bit 5 EXB4	Description	
0	TOCXB <sub>4</sub> output is disabled regardless of TFCR settings (TOCXB <sub>4</sub> input/output pin). If XTGD = 0, EXB4 is cleared to 0 when input ca channel 1.	
1	TOCXB <sub>4</sub> is enabled for output according to TFCR settings	(Initial value)

#### Bit 4—Master Enable TOCXA<sub>4</sub> (EXA4): Enables or disables ITU output at pin TOCXA<sub>4</sub>.

Bit 4 EXA4	Description	
0	TOCXA <sub>4</sub> output is disabled regardless of TFCR settings (TOCXA <sub>4</sub> oper input/output pin).	ates as a generic
	If XTGD = 0, EXA4 is cleared to 0 when input capture A occurs in char	nnel 1.
1	$TOCXA_4$ is enabled for output according to TFCR settings	(Initial value)

Bit 3—Master Enable TIOCI	(EB3): Enables or disables IT	U output at pin TIOCB <sub>3</sub> .
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Bit 3 EB3	Description	
0	TIOCB <sub>3</sub> output is disabled regardless of TIOR3 and TFCR settings (TIC a generic input/output pin).	OCB <sub>3</sub> operates as
	If XTGD = 0, EB3 is cleared to 0 when input capture A occurs in chann	el 1.
1	$TIOCB_{\scriptscriptstyle 3}$ is enabled for output according to $TIOR3$ and $TFCR$ settings	(Initial value)

## Bit 2—Master Enable TIOCB<sub>4</sub> (EB4): Enables or disables ITU output at pin TIOCB<sub>4</sub>.

Bit 2 EB4	Description	
0	TIOCB <sub>4</sub> output is disabled regardless of TIOR4 and TFCR settings (TIC a generic input/output pin).	CB <sub>4</sub> operates as
	If XTGD = 0, EB4 is cleared to 0 when input capture A occurs in channel	el 1.
1	$TIOCB_4$ is enabled for output according to TIOR4 and TFCR settings	(Initial value)

## Bit 1—Master Enable TIOCA<sub>4</sub> (EA4): Enables or disables ITU output at pin TIOCA<sub>4</sub>.

## Bit 1

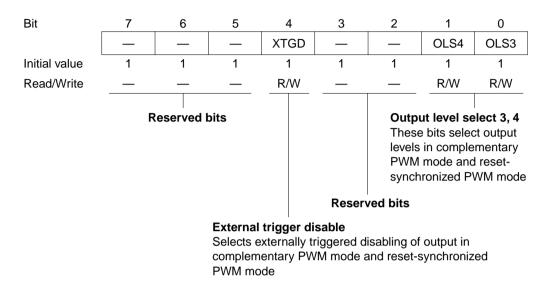
EA4	Description
0	TIOCA <sub>4</sub> output is disabled regardless of TIOR4, TMDR, and TFCR settings (TIOCA <sub>4</sub> operates as a generic input/output pin).
	If XTGD = 0, EA4 is cleared to 0 when input capture A occurs in channel 1.
1	$TIOCA_4$ is enabled for output according to $TIOR4, TMDR, and TFCR$ settings (Initial value)

## Bit 0—Master Enable TIOCA<sub>3</sub> (EA3): Enables or disables ITU output at pin TIOCA<sub>3</sub>.

Bit 0 EA3	Description
0	TIOCA <sub>3</sub> output is disabled regardless of TIOR3, TMDR, and TFCR settings (TIOCA <sub>3</sub> operates as a generic input/output pin).
	If XTGD = 0, EA3 is cleared to 0 when input capture A occurs in channel 1.
1	$TIOCA_{\!_3}$ is enabled for output according to TIOR3, TMDR, and TFCR settings (Initial value)

## 8.2.6 Timer Output Control Register (TOCR)

TOCR is an 8-bit readable/writable register that selects externally triggered disabling of output in complementary PWM mode and reset-synchronized PWM mode, and inverts the output levels.



The settings of the XTGD, OLS4, and OLS3 bits are valid only in complementary PWM mode and reset-synchronized PWM mode. These settings do not affect other modes.

TOCR is initialized to H'FF by a reset and in standby mode.

Bits 7 to 5—Reserved: These bits cannot be modified and are always read as 1.

**Bit 4—External Trigger Disable (XTGD):** Selects externally triggered disabling of ITU output in complementary PWM mode and reset-synchronized PWM mode.

Bit 4 XTGD	Description	
0	Input capture A in channel 1 is used as an external trigger signal in cor PWM mode and reset-synchronized PWM mode.	mplementary
	When an external trigger occurs, bits 5 to 0 in the timer output master e (TOER) are cleared to 0, disabling ITU output.	enable register
1	External triggering is disabled	(Initial value)

Bits 3 and 2—Reserved: These bits cannot be modified and are always read as 1.

**Bit 1—Output Level Select 4 (OLS4):** Selects output levels in complementary PWM mode and reset-synchronized PWM mode.

Bit 1 OLS4	Description	
0	TIOCA <sub>3</sub> , TIOCA <sub>4</sub> , and TIOCB <sub>4</sub> pin outputs are inverted	
1	$TIOCA_3$ , $TIOCA_4$ , and $TIOCB_4$ pin outputs are not inverted	(Initial value)

**Bit 0—Output Level Select 3 (OLS3):** Selects output levels in complementary PWM mode and reset-synchronized PWM mode.

## Bit 0

OLS3	Description	
0	$TIOCB_3$ , $TOCXA_4$ , and $TOCXB_4$ pin outputs are inverted	
1	$TIOCB_3$ , $TOCXA_4$ , and $TOCXB_4$ pin outputs are not inverted	(Initial value)

## 8.2.7 Timer Counters (TCNT)

TCNT is a 16-bit counter. The ITU has five TCNTs, one for each channel.

Channel	Abbreviation	Function
0	TCNT0	Up-counter
1	TCNT1	_
2	TCNT2	Phase counting mode: up/down-counter Other modes: up-counter
3	TCNT3	Complementary PWM mode: up/down-counter
4	TCNT4	Other modes: up-counter

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Read/Write	R/W															

Each TCNT is a 16-bit readable/writable register that counts pulse inputs from a clock source. The clock source is selected by bits TPSC2 to TPSC0 in the timer control register (TCR).

TCNT0 and TCNT1 are up-counters. TCNT2 is an up/down-counter in phase counting mode and an up-counter in other modes. TCNT3 and TCNT4 are up/down-counters in complementary PWM mode and up-counters in other modes.

TCNT can be cleared to H'0000 by compare match with general register A or B (GRA or GRB) or by input capture to GRA or GRB (counter clearing function) in the same channel.

When TCNT overflows (changes from H'FFFF to H'0000), the overflow flag (OVF) is set to 1 in the timer status register (TSR) of the corresponding channel.

When TCNT underflows (changes from H'0000 to H'FFFF), the overflow flag (OVF) is set to 1 in TSR of the corresponding channel.

The TCNTs are linked to the CPU by an internal 16-bit bus and can be written or read by either word access or byte access.

Each TCNT is initialized to H'0000 by a reset and in standby mode.

## 8.2.8 General Registers (GRA, GRB)

The general registers are 16-bit registers. The ITU has 10 general registers, two in each channel.

Channel	Abbreviation	Function
0	GRA0, GRB0	Output compare/input capture register
1	GRA1, GRB1	_
2	GRA2, GRB2	_
3	GRA3, GRB3	Output compare/input capture register; can be by buffer registers
4	GRA4, GRB4	BRA and BRB
Rit	15 14	13 12 11 10 9 8 7 6 5 4 3 2 1 0

Bit	15	14	13	12	11	10	9	8	1	6	5	4	3	2	1	0
Initial value	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
Read/Write	R/W															

A general register is a 16-bit readable/writable register that can function as either an output compare register or an input capture register. The function is selected by settings in the timer I/O control register (TIOR).

When a general register is used as an output compare register, its value is constantly compared with the TCNT value. When the two values match (compare match), the IMFA or IMFB flag is set to 1 in the timer status register (TSR). Compare match output can be selected in TIOR.

When a general register is used as an input capture register, rising edges, falling edges, or both edges of an external input capture signal are detected and the current TCNT value is stored in the general register. The corresponding IMFA or IMFB flag in TSR is set to 1 at the same time. The valid edge or edges of the input capture signal are selected in TIOR.

TIOR settings are ignored in PWM mode, complementary PWM mode, and reset-synchronized PWM mode.

General registers are linked to the CPU by an internal 16-bit bus and can be written or read by either word access or byte access.

General registers are initialized to the output compare function (with no output signal) by a reset and in standby mode. The initial value is H'FFFF.

## Renesas

## 8.2.9 Buffer Registers (BRA, BRB)

The buffer registers are 16-bit registers. The ITU has four buffer registers, two each in channels 3 and 4.

Channel	Abb	revia	tion	Fu	nctio	on											
3	BRA	\3, BF	RB3	Us	ed fo	or buf	fering	3									
4	BRA	4, BF	RB4	•	com buff	npare er reg	regis gister	ster, l :: the	BRA	or BR or Bl	RB ca RB va	n fun alue i		as a	n out	put c	out ompare sferred
				•	cap buff	ture r er reg	egist gister	er, B :: the	RA o	r BRE	3 can RB v	func	funct tion a is aut	as an	inpu	t capt	
Bit		15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Initial va	lue	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
Read/W	rite	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

A buffer register is a 16-bit readable/writable register that is used when buffering is selected. Buffering can be selected independently by bits BFB4, BFA4, BFB3, and BFA3 in TFCR.

The buffer register and general register operate as a pair. When the general register functions as an output compare register, the buffer register functions as an output compare buffer register. When the general register functions as an input capture register, the buffer register functions as an input capture buffer register.

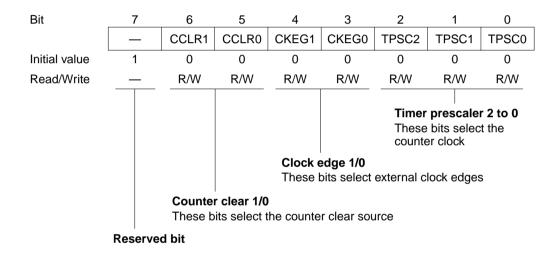
The buffer registers are linked to the CPU by an internal 16-bit bus and can be written or read by either word or byte access.

Buffer registers are initialized to H'FFFF by a reset and in standby mode.

## 8.2.10 Timer Control Registers (TCR)

TCR is an 8-bit register. The ITU has five TCRs, one in each channel.

0         TCR0         TCR controls the timer counter. The TCRs in all channels are functionally identical. When phase counting mode is selected in functionally identical. When phase counting mode is selected in	
channel 2, the settings of bits CKEG1 and CKEG0 and TPSC2	
2 TCR2 TPSC0 in TCR2 are ignored.	
3 TCR3	
4 TCR4	



Each TCR is an 8-bit readable/writable register that selects the timer counter clock source, selects the edge or edges of external clock sources, and selects how the counter is cleared.

TCR is initialized to H'80 by a reset and in standby mode.

Bit 7—Reserved: This bit cannot be modified and is always read as 1.

Bits 6 and 5—Counter Clear 1/0 (CCLR1, CCLR0): These bits select how TCNT is cleared.

Bit 6 CCLR1	Bit 5 CCLR0	Description	
0	0	TCNT is not cleared	(Initial value)
	1	TCNT is cleared by GRA compare match or input capture*1	
1	0	TCNT is cleared by GRB compare match or input capture*	
	1	Synchronous clear: TCNT is cleared in synchronization with synchronized timers* <sup>2</sup>	other

Notes: 1. TCNT is cleared by compare match when the general register functions as an output compare match register, and by input capture when the general register functions as an input capture register.

2. Selected in the timer synchro register (TSNC).

Bits 4 and 3—Clock Edge 1/0 (CKEG1, CKEG0): These bits select external clock input edges when an external clock source is used.

Bit 4 CKEG1	Bit 3 CKEG0	Description	
0	0	Count rising edges	(Initial value)
	1	Count falling edges	
1	_	Count both edges	

When channel 2 is set to phase counting mode, bits CKEG1 and CKEG0 in TCR2 are ignored. Phase counting takes precedence.

Bits 2 to 0—Timer Prescaler 2 to 0 (TPSC2 to TPSC0): These bits select the counter clock source.

Bit 1 TPSC1	Bit 0 TPSC0	Function	
0	0	Internal clock:	(Initial value)
	1	Internal clock: φ/2	
1	0	Internal clock: ø/4	
	1	Internal clock: ø/8	
0	0	External clock A: TCLKA input	
	1	External clock B: TCLKB input	
1	0	External clock C: TCLKC input	
	1	External clock D: TCLKD input	
	TPSC1           0           1	TPSC1         TPSC0           0         0           1         1           1         0           1         1           0         0           1         1           0         1           1         1           1         1           1         1           1         1	TPSC1         TPSC0         Function           0         Internal clock: φ           1         Internal clock: φ/2           1         Internal clock: φ/4           1         Internal clock: φ/4           1         Internal clock: φ/8           0         External clock A: TCLKA input           1         External clock B: TCLKB input           1         0

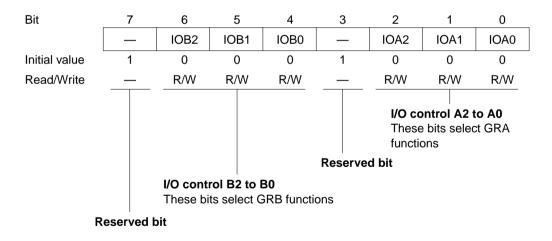
When bit TPSC2 is cleared to 0 an internal clock source is selected, and the timer counts only falling edges. When bit TPSC2 is set to 1 an external clock source is selected, and the timer counts the edge or edges selected by bits CKEG1 and CKEG0.

When channel 2 is set to phase counting mode (MDF = 1 in TMDR), the settings of bits TPSC2 to TPSC0 in TCR2 are ignored. Phase counting takes precedence.

#### 8.2.11 Timer I/O Control Register (TIOR)

TIOR is an 8-bit register. The ITU has five TIORs, one in each channel.

Channel	Abbreviation	Function
0	TIOR0	TIOR controls the general registers. Some functions differ in PWM
1	TIOR1	<ul> <li>mode. TIOR3 and TIOR4 settings are ignored when complementary</li> <li>PWM mode or reset-synchronized PWM mode is selected in</li> </ul>
2	TIOR2	channels 3 and 4.
3	TIOR3	_
4	TIOR4	_



Each TIOR is an 8-bit readable/writable register that selects the output compare or input capture function for GRA and GRB, and specifies the functions of the TIOCA and TIOCB pins. If the output compare function is selected, TIOR also selects the type of output. If input capture is selected, TIOR also selects the edge or edges of the input capture signal.

TIOR is initialized to H'88 by a reset and in standby mode.

Bit 7—Reserved: This bit cannot be modified and is always read as 1.

Bit 6 IOB2	Bit 5 IOB1	Bit 4 IOB0	Function			
0	0	0	GRB is an output	No output at compare match (Initial value)		
		1	compare register	0 output at GRB compare match*1		
	1	0		1 output at GRB compare match*1		
		1	_	Output toggles at GRB compare match (1 output in channel 2)* <sup>1</sup> * <sup>2</sup>		
1	0	0	GRB is an input	GRB captures rising edge of input		
		1	capture register	GRB captures falling edge of input		
	1	0	_	GRB captures both edges of input		
		1	_			

Bits 6 to 4—I/O Control B2 to B0 (IOB2 to IOB0): These bits select the GRB function.

Notes: 1. After a reset, the output is 0 until the first compare match.

2. Channel 2 output cannot be toggled by compare match. This setting selects 1 output instead.

Bit 3—Reserved: This bit cannot be modified and is always read as 1.

Bits 2 to 0—I/O Control A2 to A0 (IOA2 to IOA0): These bits select the GRA function.

Bit 2 IOA2	Bit 1 IOA1	Bit 0 IOA0	Function			
0	0	0	GRA is an output	No output at compare match (Initial value)		
	1		compare register	0 output at GRA compare match*1		
	1	0	_	1 output at GRA compare match*1		
		1		Output toggles at GRA compare match (1 output in channel 2)* <sup>1</sup> * <sup>2</sup>		
1	0	0	GRA is an input	GRA captures rising edge of input		
		1	capture register	GRA captures falling edge of input		
	1	0	_	GRA captures both edges of input		
_		1	_			

Notes: 1. After a reset, the output is 0 until the first compare match.

2. Channel 2 output cannot be toggled by compare match. This setting selects 1 output instead.

## 8.2.12 Timer Status Register (TSR)

TSR is an 8-bit register. The ITU has five TSRs, one in each channel.

Channel	Abb	reviation	Functior	Function					
0	TSR	0	Indicate	Indicates input capture, compare match, and overflow status					
1	TSR	1							
2	TSR	2							
3	TSR	3							
4	TSR	4							
Bit		7	6	5	4	3	2	1	0
		_			_		OVF	IMFB	IMFA
Initial va	lue	1	1	1	1	1	0	0	0
Read/W	rite	_	—		—		R/(W)*	R/(W)*	R/(W)*
			Re	∣ served b	its				
Overflow flag Status flag indicating overflow or underflow									
Input capture/compare match flag B Status flag indicating GRB compare match or input capture									
Input capture/compare match flag A Status flag indicating GRA compare match or input capture									

Note: \* Only 0 can be written to clear the flag.

Each TSR is an 8-bit readable/writable register containing flags that indicate TCNT overflow or underflow and GRA or GRB compare match or input capture. These flags are interrupt sources and generate CPU interrupts if enabled by corresponding bits in the timer interrupt enable register (TIER).

TSR is initialized to H'F8 by a reset and in standby mode.

Bits 7 to 3—Reserved: These bits cannot be modified and are always read as 1.

#### Section 8 16-Bit Integrated Timer Unit (ITU)

Bit 2—Overflow Flag (OVF): This status flag indicates TCNT overflow or underflow.

Bit 2 OVF	Description							
0	[Clearing condition]	(Initial value)						
	Read OVF when OVF = 1, then write 0 in OVF							
1	[Setting condition]							
	TCNT overflowed from H'FFFF to H'0000, or underflowed from H'0000	) to H'FFFF*						
Notes: *	<ul> <li>TCNT underflow occurs when TCNT operates as an up/down-counter. Underflow occurs only under the following conditions:</li> <li>1. Channel 2 operates in phase counting mode (MDF = 1 in TMDR)</li> <li>2. Channels 3 and 4 operate in complementary PWM mode (CMD1 = 1 and CMD0 = 0 in TFCR)</li> </ul>							

## Bit 1—Input Capture/Compare Match Flag B (IMFB): This status flag indicates GRB

compare match or input capture events.

Bit 1 IMFB	Description					
0	[Clearing condition]					
	Read IMFB when IMFB = 1, then write 0 in IMFB	(Initial value)				
1	[Setting conditions]					
	TCNT = GRB when GRB functions as a compare match re	egister.				
	<ul> <li>TCNT value is transferred to GRB by an input capture sign as an input capture register.</li> </ul>	nal, when GRB functions				

**Bit 0—Input Capture/Compare Match Flag A (IMFA):** This status flag indicates GRA compare match or input capture events.

Bit 0 IMFA	Description					
0	[Clearing condition]					
	Read IMFA when IMFA = 1, then write 0 in IMFA.	(Initial value)				
1	[Setting conditions]					
	• TCNT = GRA when GRA functions as a compare match re	gister.				
	<ul> <li>TCNT value is transferred to GRA by an input capture sign as an input capture register.</li> </ul>	al, when GRA functions				

## 8.2.13 Timer Interrupt Enable Register (TIER)

TIER is an 8-bit register. The ITU has five TIERs, one in each channel.

Channel	Abb	reviation	Functio	on					
0	TIE	٦0	Enables or disables interrupt requests.						
1	TIEF	٦1	_						
2	TIEF	२2							
3	TIEF	२३							
4	TIEF	٦4							
Bit		7	6	5	4	3	2	1	0
			—	—	—	—	OVIE	IMIEB	IMIEA
Initial va	lue	1	1	1	1	1	0	0	0
Read/W	rite		—	_	—		R/W	R/W	R/W
			R	eserved b	its				
	Overflow interrupt enable Enables or disables OVF interrupts								
Input capture/compare match interrupt enable B Enables or disables IMFB interrupts									
Input capture/compare match interrupt enable A Enables or disables IMFA interrupts									

Each TIER is an 8-bit readable/writable register that enables and disables overflow interrupt requests and general register compare match and input capture interrupt requests.

TIER is initialized to H'F8 by a reset and in standby mode.

Bits 7 to 3—Reserved: These bits cannot be modified and are always read as 1.

**Bit 2—Overflow Interrupt Enable (OVIE):** Enables or disables the interrupt requested by the overflow flag (OVF) in TSR when OVF is set to 1.

Bit 2 OVIE	Description	
0	OVI interrupt requested by OVF is disabled	(Initial value)
1	OVI interrupt requested by OVF is enabled	

**Bit 1—Input Capture/Compare Match Interrupt Enable B (IMIEB):** Enables or disables the interrupt requested by the IMFB flag in TSR when IMFB is set to 1.

Bit 1 IMIEB	Description	
0	IMIB interrupt requested by IMFB is disabled	(Initial value)
1	IMIB interrupt requested by IMFB is enabled	

**Bit 0—Input Capture/Compare Match Interrupt Enable A (IMIEA):** Enables or disables the interrupt requested by the IMFA flag in TSR when IMFA is set to 1.

Bit 0 IMIEA	Description	
0	IMIA interrupt requested by IMFA is disabled	(Initial value)
1	IMIA interrupt requested by IMFA is enabled	

# 8.3 CPU Interface

## 8.3.1 16-Bit Accessible Registers

The timer counters (TCNTs), general registers A and B (GRAs and GRBs), and buffer registers A and B (BRAs and BRBs) are 16-bit registers, and are linked to the CPU by an internal 16-bit data bus. These registers can be written or read a word at a time, or a byte at a time.

Figures 8.6 and 8.7 show examples of word access to a timer counter (TCNT). Figures 8.8, 8.9, 8.10, and 8.11 show examples of byte access to TCNTH and TCNTL.

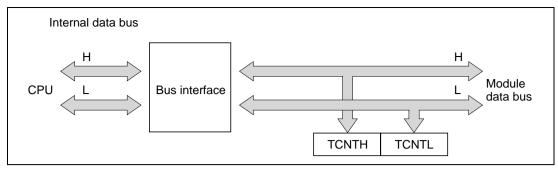


Figure 8.6 Access to Timer Counter (CPU Writes to TCNT, Word)

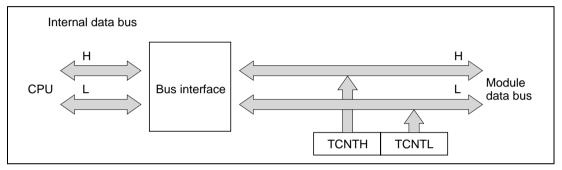


Figure 8.7 Access to Timer Counter (CPU Reads TCNT, Word)

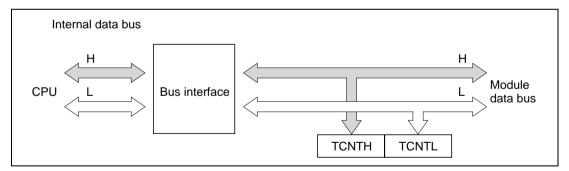


Figure 8.8 Access to Timer Counter (CPU Writes to TCNT, Upper Byte)

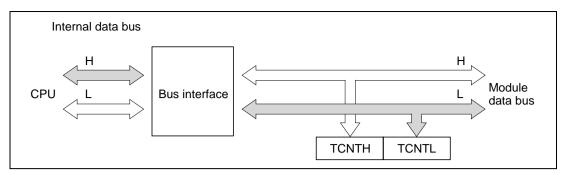


Figure 8.9 Access to Timer Counter (CPU Writes to TCNT, Lower Byte)

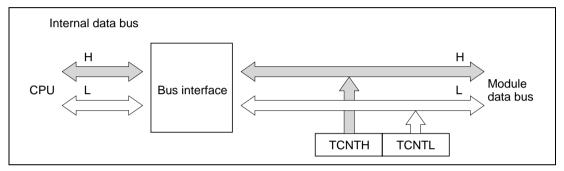


Figure 8.10 Access to Timer Counter (CPU Reads TCNT, Upper Byte)

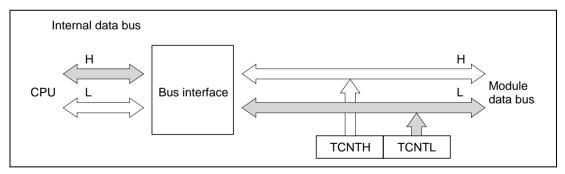


Figure 8.11 Access to Timer Counter (CPU Reads TCNT, Lower Byte)

#### 8.3.2 8-Bit Accessible Registers

The registers other than the timer counters, general registers, and buffer registers are 8-bit registers. These registers are linked to the CPU by an internal 8-bit data bus.

Figures 8.12 and 8.13 show examples of byte read and write access to a TCR.

If a word-size data transfer instruction is executed, two byte transfers are performed.

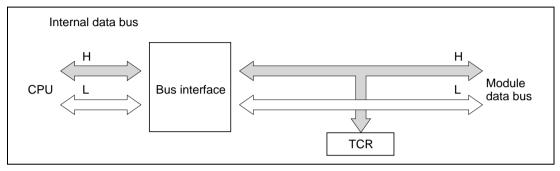


Figure 8.12 TCR Access (CPU Writes to TCR)

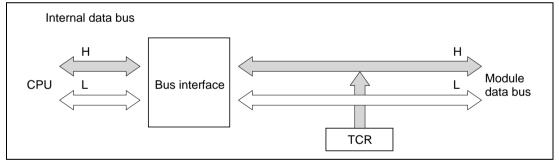


Figure 8.13 TCR Access (CPU Reads TCR)

# 8.4 Operation

## 8.4.1 Overview

A summary of operations in the various modes is given below.

**Normal Operation:** Each channel has a timer counter and general registers. The timer counter counts up, and can operate as a free-running counter, periodic counter, or external event counter. General registers A and B can be used for input capture or output compare.

**Synchronous Operation:** The timer counters in designated channels are preset synchronously. Data written to the timer counter in any one of these channels is simultaneously written to the timer counters in the other channels as well. The timer counters can also be cleared synchronously if so designated by the CCLR1 and CCLR0 bits in the TCRs.

**PWM Mode:** A PWM waveform is output from the TIOCA pin. The output goes to 1 at compare match A and to 0 at compare match B. The duty cycle can be varied from 0% to 100% depending on the settings of GRA and GRB. When a channel is set to PWM mode, its GRA and GRB automatically become output compare registers.

**Reset-Synchronized PWM Mode:** Channels 3 and 4 are paired for three-phase PWM output with complementary waveforms. (The three phases are related by having a common transition point.) When reset-synchronized PWM mode is selected GRA3, GRB3, GRA4, and GRB4 automatically function as output compare registers, TIOCA<sub>3</sub>, TIOCB<sub>3</sub>, TIOCA<sub>4</sub>, TOCXA<sub>4</sub>, TIOCB<sub>4</sub>, and TOCXB<sub>4</sub> function as PWM output pins, and TCNT<sub>3</sub> operates as an up-counter. TCNT4 operates independently, and is not compared with GRA4 or GRB4.

**Complementary PWM Mode:** Channels 3 and 4 are paired for three-phase PWM output with non-overlapping complementary waveforms. When complementary PWM mode is selected GRA3, GRB3, GRA4, and GRB4 automatically function as output compare registers, and TIOCA<sub>3</sub>, TIOCB<sub>3</sub>, TIOCA<sub>4</sub>, TOCXA<sub>4</sub>, TIOCB<sub>4</sub>, and TOCXB<sub>4</sub> function as PWM output pins. TCNT3 and TCNT4 operate as up/down-counters.

**Phase Counting Mode:** The phase relationship between two clock signals input at TCLKA and TCLKB is detected and TCNT2 counts up or down accordingly. When phase counting mode is selected TCLKA and TCLKB become clock input pins and TCNT2 operates as an up/down-counter.

#### **Buffering:**

- If the general register is an output compare register When compare match occurs the buffer register value is transferred to the general register.
- If the general register is an input capture register When input capture occurs the TCNT value is transferred to the general register, and the previous general register value is transferred to the buffer register.
- Complementary PWM mode The buffer register value is transferred to the general register when TCNT3 and TCNT4 change counting direction.
- Reset-synchronized PWM mode The buffer register value is transferred to the general register at GRA3 compare match.

### 8.4.2 Basic Functions

## **Counter Operation**

When one of bits STR0 to STR4 is set to 1 in the timer start register (TSTR), the timer counter (TCNT) in the corresponding channel starts counting. The counting can be free-running or periodic.

**Sample setup procedure for counter:** Figure 8.14 shows a sample procedure for setting up a counter.

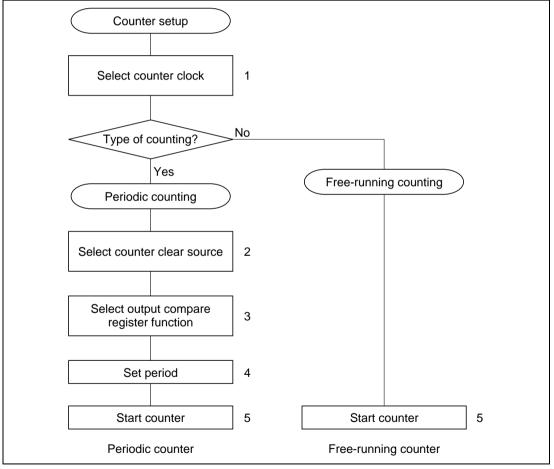
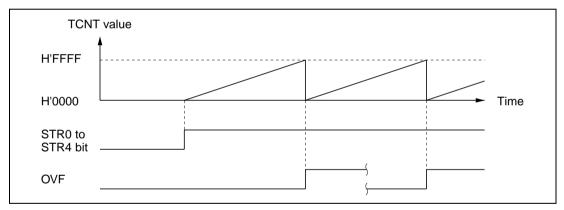
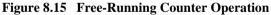


Figure 8.14Counter Setup Procedure (Example)

- 1. Set bits TPSC2 to TPSC0 in TCR to select the counter clock source. If an external clock source is selected, set bits CKEG1 and CKEG0 in TCR to select the desired edge(s) of the external clock signal.
- 2. For periodic counting, set CCLR1 and CCLR0 in TCR to have TCNT cleared at GRA compare match or GRB compare match.
- 3. Set TIOR to select the output compare function of GRA or GRB, whichever was selected in step 2.
- 4. Write the count period in GRA or GRB, whichever was selected in step 2.
- 5. Set the STR bit to 1 in TSTR to start the timer counter.

**Free-running and periodic counter operation:** A reset leaves the counters (TCNTs) in ITU channels 0 to 4 all set as free-running counters. A free-running counter starts counting up when the corresponding bit in TSTR is set to 1. When the count overflows from H'FFFF to H'0000, the overflow flag (OVF) is set to 1 in the timer status register (TSR). If the corresponding OVIE bit is set to 1 in the timer interrupt enable register, a CPU interrupt is requested. After the overflow, the counter continues counting up from H'0000. Figure 8.15 illustrates free-running counting.





When a channel is set to have its counter cleared by compare match, in that channel TCNT operates as a periodic counter. Select the output compare function of GRA or GRB, set bit CCLR1 or CCLR0 in the timer control register (TCR) to have the counter cleared by compare match, and set the count period in GRA or GRB. After these settings, the counter starts counting up as a periodic counter when the corresponding bit is set to 1 in TSTR. When the count matches GRA or GRB, the IMFA or IMFB flag is set to 1 in TSR and the counter is cleared to H'0000. If the corresponding IMIEA or IMIEB bit is set to 1 in TIER, a CPU interrupt is requested at this time. After the compare match, TCNT continues counting up from H'0000. Figure 8.16 illustrates periodic counting.

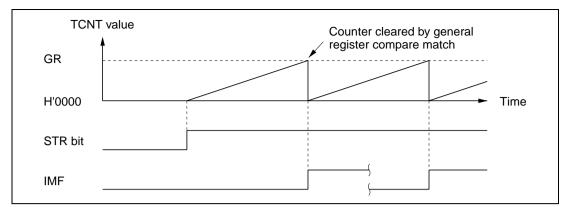


Figure 8.16 Periodic Counter Operation

## **Count timing:**

Internal clock source

Bits TPSC2 to TPSC0 in TCR select the system clock ( $\phi$ ) or one of three internal clock sources obtained by prescaling the system clock ( $\phi/2$ ,  $\phi/4$ ,  $\phi/8$ ). Figure 8.17 shows the timing.

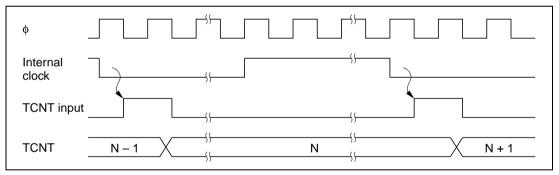


Figure 8.17 Count Timing for Internal Clock Sources

External clock source

Bits TPSC2 to TPSC0 in TCR select an external clock input pin (TCLKA to TCLKD), and its valid edge or edges are selected by bits CKEG1 and CKEG0. The rising edge, falling edge, or both edges can be selected.

The pulse width of the external clock signal must be at least 1.5 system clocks when a single edge is selected, and at least 2.5 system clocks when both edges are selected. Shorter pulses will not be counted correctly.

Figure 8.18 shows the timing when both edges are detected.

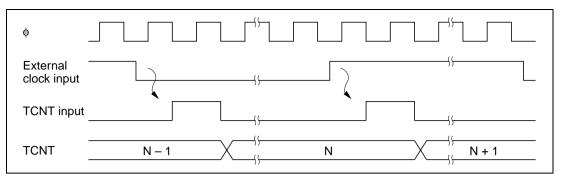


Figure 8.18 Count Timing for External Clock Sources (when Both Edges are Detected)

#### Waveform Output by Compare Match

In ITU channels 0, 1, 3, and 4, compare match A or B can cause the output at the TIOCA or TIOCB pin to go to 0, go to 1, or toggle. In channel 2 the output can only go to 0 or go to 1.

**Sample setup procedure for waveform output by compare match:** Figure 8.19 shows a sample procedure for setting up waveform output by compare match.

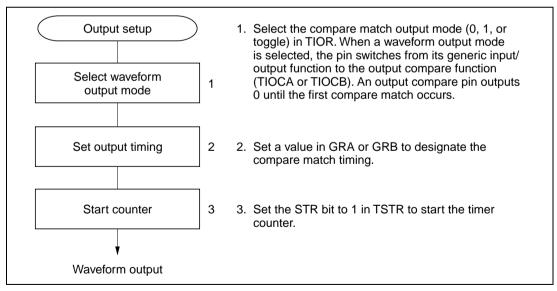


Figure 8.19 Setup Procedure for Waveform Output by Compare Match (Example)

**Examples of waveform output:** Figure 8.20 shows examples of 0 and 1 output. TCNT operates as a free-running counter, 0 output is selected for compare match A, and 1 output is selected for compare match B. When the pin is already at the selected output level, the pin level does not change.

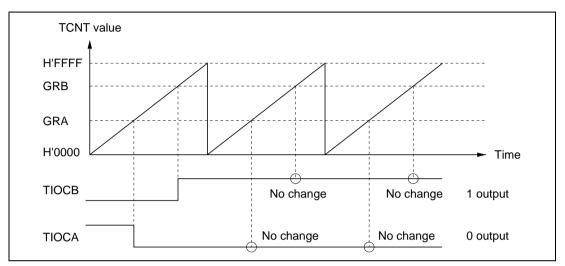
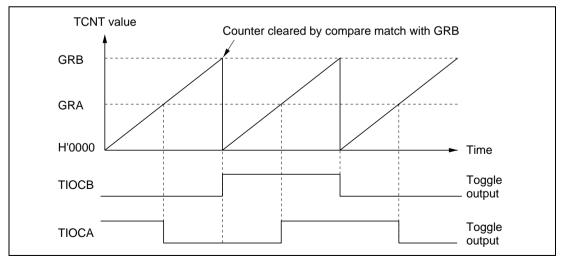


Figure 8.20 0 and 1 Output (Examples)

Figure 8.21 shows examples of toggle output. TCNT operates as a periodic counter, cleared by compare match B. Toggle output is selected for both compare match A and B.





**Output compare timing:** The compare match signal is generated in the last state in which TCNT and the general register match (when TCNT changes from the matching value to the next value). When the compare match signal is generated, the output value selected in TIOR is output at the output compare pin (TIOCA or TIOCB). When TCNT matches a general register, the compare match signal is not generated until the next counter clock pulse.

Figure 8.22 shows the output compare timing.

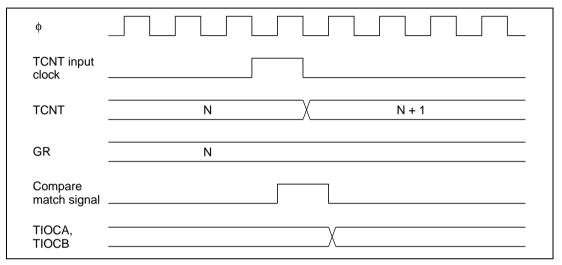
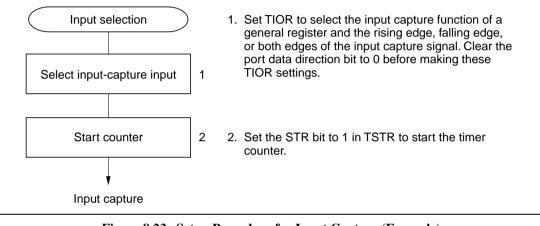


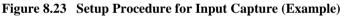
Figure 8.22 Output Compare Timing

## Input Capture Function

The TCNT value can be captured into a general register when a transition occurs at an input capture/output compare pin (TIOCA or TIOCB). Capture can take place on the rising edge, falling edge, or both edges. The input capture function can be used to measure pulse width or period.

**Sample setup procedure for input capture:** Figure 8.23 shows a sample procedure for setting up input capture.





**Examples of input capture:** Figure 8.24 illustrates input capture when the falling edge of TIOCB and both edges of TIOCA are selected as capture edges. TCNT is cleared by input capture into GRB.

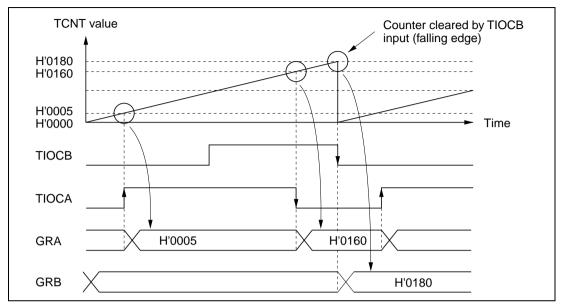


Figure 8.24 Input Capture (Example)

**Input capture signal timing:** Input capture on the rising edge, falling edge, or both edges can be selected by settings in TIOR. Figure 8.25 shows the timing when the rising edge is selected. The pulse width of the input capture signal must be at least 1.5 system clocks for single-edge capture, and 2.5 system clocks for capture of both edges.

φ		
Input-capture input		
Internal input capture signal		
TCNT	N	
GRA, GRB		Ν

Figure 8.25 Input Capture Signal Timing

#### 8.4.3 Synchronization

The synchronization function enables two or more timer counters to be synchronized by writing the same data to them simultaneously (synchronous preset). With appropriate TCR settings, two or more timer counters can also be cleared simultaneously (synchronous clear). Synchronization enables additional general registers to be associated with a single time base. Synchronization can be selected for all channels (0 to 4).

#### Sample Setup Procedure for Synchronization

Figure 8.26 shows a sample procedure for setting up synchronization.

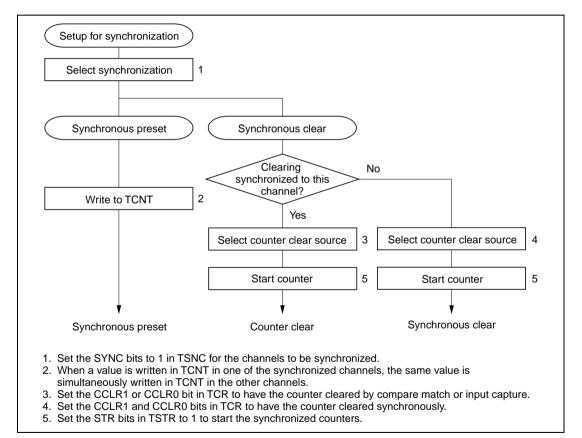


Figure 8.26 Setup Procedure for Synchronization (Example)

### Example of Synchronization

Figure 8.27 shows an example of synchronization. Channels 0, 1, and 2 are synchronized, and are set to operate in PWM mode. Channel 0 is set for counter clearing by compare match with GRB0. Channels 1 and 2 are set for synchronous counter clearing. The timer counters in channels 0, 1, and 2 are synchronously preset, and are synchronously cleared by compare match with GRB0. A three-phase PWM waveform is output from pins  $TIOCA_0$ ,  $TIOCA_1$ , and  $TIOCA_2$ . For further information on PWM mode, see section 8.4.4, PWM Mode.

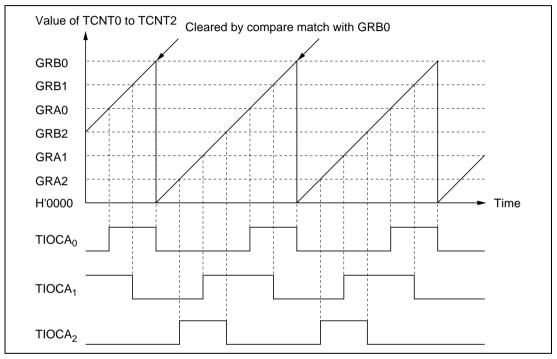


Figure 8.27 Synchronization (Example)

### 8.4.4 PWM Mode

In PWM mode GRA and GRB are paired and a PWM waveform is output from the TIOCA pin. GRA specifies the time at which the PWM output changes to 1. GRB specifies the time at which the PWM output changes to 0. If either GRA or GRB is selected as the counter clear source, a PWM waveform with a duty cycle from 0% to 100% is output at the TIOCA pin. PWM mode can be selected in all channels (0 to 4).

Table 8.4 summarizes the PWM output pins and corresponding registers. If the same value is set in GRA and GRB, the output does not change when compare match occurs.

Channel	Output Pin	1 Output	0 Output
0		GRA0	GRB0
1	TIOCA <sub>1</sub>	GRA1	GRB1
2	TIOCA <sub>2</sub>	GRA2	GRB2
3	TIOCA <sub>3</sub>	GRA3	GRB3
4	TIOCA <sub>4</sub>	GRA4	GRB4

#### Table 8.4 PWM Output Pins and Registers

## Sample Setup Procedure for PWM Mode

Figure 8.28 shows a sample procedure for setting up PWM mode.

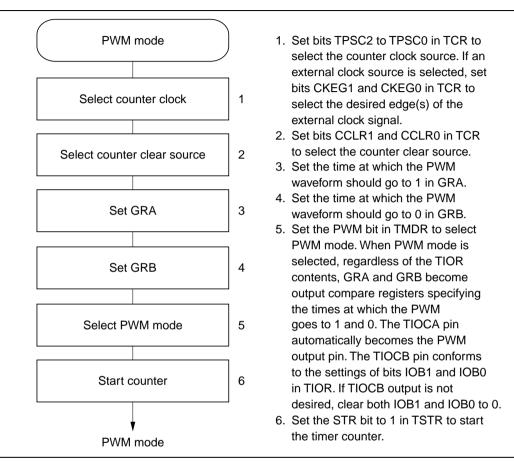


Figure 8.28 Setup Procedure for PWM Mode (Example)

#### **Examples of PWM Mode**

Figure 8.29 shows examples of operation in PWM mode. The PWM waveform is output from the TIOCA pin. The output goes to 1 at compare match with GRA, and to 0 at compare match with GRB.

In the examples shown, TCNT is cleared by compare match with GRA or GRB. Synchronized operation and free-running counting are also possible.

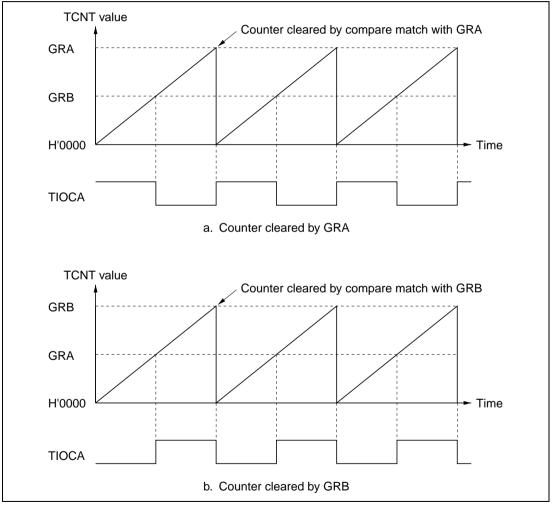




Figure 8.30 shows examples of the output of PWM waveforms with duty cycles of 0% and 100%. If the counter is cleared by compare match with GRB, and GRA is set to a higher value than GRB, the duty cycle is 0%. If the counter is cleared by compare match with GRA, and GRB is set to a higher value than GRA, the duty cycle is 100%.

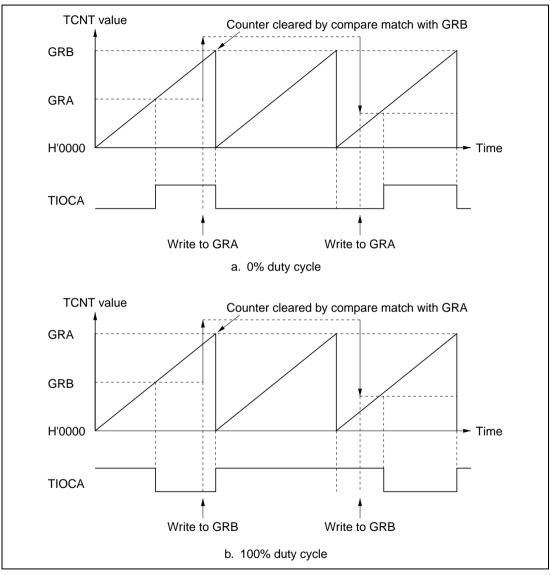


Figure 8.30 PWM Mode (Example 2)

### 8.4.5 Reset-Synchronized PWM Mode

In reset-synchronized PWM mode channels 3 and 4 are combined to produce three pairs of complementary PWM waveforms, all having one waveform transition point in common.

When reset-synchronized PWM mode is selected  $TIOCA_3$ ,  $TIOCB_3$ ,  $TIOCA_4$ ,  $TOCXA_4$ ,  $TIOCB_4$ , and  $TOCXB_4$  automatically become PWM output pins, and  $TCNT_3$  functions as an up-counter.

Table 8.5 lists the PWM output pins. Table 8.6 summarizes the register settings.

Channel	Output Pin	Description
3	TIOCA <sub>3</sub>	PWM output 1
	TIOCB <sub>3</sub>	PWM output 1' (complementary waveform to PWM output 1)
4	TIOCA <sub>4</sub>	PWM output 2
	TOCXA <sub>4</sub>	PWM output 2' (complementary waveform to PWM output 2)
	TIOCB <sub>4</sub>	PWM output 3
	TOCXB <sub>4</sub>	PWM output 3' (complementary waveform to PWM output 3)

 Table 8.5
 Output Pins in Reset-Synchronized PWM Mode

### Table 8.6 Register Settings in Reset-Synchronized PWM Mode

Register	Setting
TCNT3	Initially set to H'0000
TCNT4	Not used (operates independently)
GRA3	Specifies the count period of TCNT3
GRB3	Specifies a transition point of PWM waveforms output from TIOCA $_{\rm 3}$ and TIOCB $_{\rm 3}$
GRA4	Specifies a transition point of PWM waveforms output from TIOCA <sub>4</sub> and TOCXA <sub>4</sub>
GRB4	Specifies a transition point of PWM waveforms output from $TIOCB_4$ and $TOCXB_4$

### Sample Setup Procedure for Reset-Synchronized PWM Mode

Figure 8.31 shows a sample procedure for setting up reset-synchronized PWM mode.

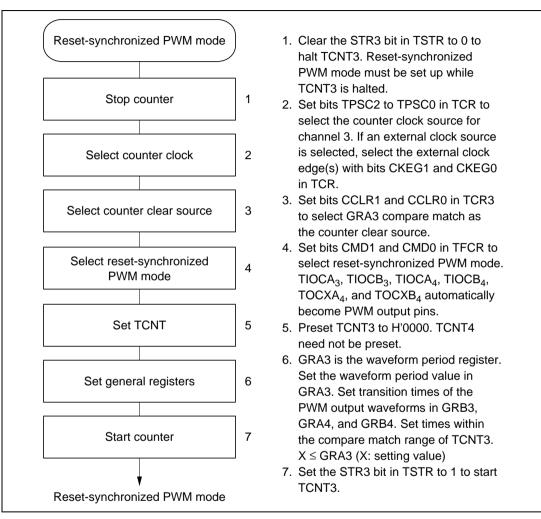


Figure 8.31 Setup Procedure for Reset-Synchronized PWM Mode (Example)

### Example of Reset-Synchronized PWM Mode

Figure 8.32 shows an example of operation in reset-synchronized PWM mode. TCNT3 operates as an up-counter in this mode. TCNT4 operates independently, detached from GRA4 and GRB4. When TCNT3 matches GRA3, TCNT3 is cleared and resumes counting from H'0000. The PWM outputs toggle at compare match with GRB3, GRA4, GRB4, and TCNT3 respectively, and when the counter is cleared.

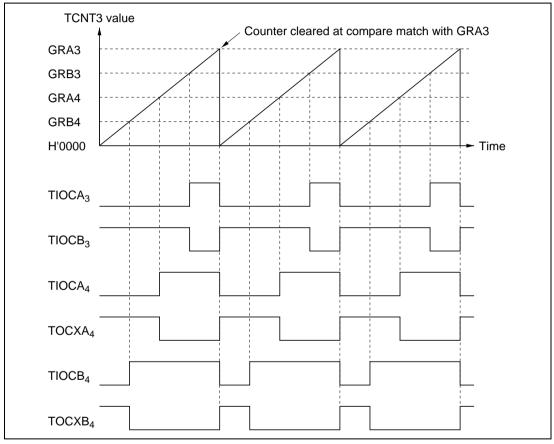


Figure 8.32 Operation in Reset-Synchronized PWM Mode (Example) (when OLS3 = OLS4 = 1)

For the settings and operation when reset-synchronized PWM mode and buffer mode are both selected, see section 8.4.8, Buffering.

### 8.4.6 Complementary PWM Mode

In complementary PWM mode channels 3 and 4 are combined to output three pairs of complementary, non-overlapping PWM waveforms.

When complementary PWM mode is selected  $TIOCA_3$ ,  $TIOCB_3$ ,  $TIOCA_4$ ,  $TOCXA_4$ ,  $TIOCB_4$ , and  $TOCXB_4$  automatically become PWM output pins, and TCNT3 and TCNT4 function as up/down-counters.

Table 8.7 lists the PWM output pins. Table 8.8 summarizes the register settings.

Channel	Output Pin	Description
3		PWM output 1
	TIOCB <sub>3</sub>	PWM output 1' (non-overlapping complementary waveform to PWM output 1)
4	TIOCA <sub>4</sub>	PWM output 2
	TOCXA <sub>4</sub>	PWM output 2' (non-overlapping complementary waveform to PWM output 2)
	TIOCB <sub>4</sub>	PWM output 3
		PWM output 3' (non-overlapping complementary waveform to PWM output 3)

 Table 8.7
 Output Pins in Complementary PWM Mode

#### Table 8.8 Register Settings in Complementary PWM Mode

Register	Setting
TCNT3	Initially specifies the non-overlap margin (difference to TCNT4)
TCNT4	Initially set to H'0000
GRA3	Specifies the upper limit value of TCNT3 minus 1
GRB3	Specifies a transition point of PWM waveforms output from TIOCA <sub>3</sub> and TIOCB <sub>3</sub>
GRA4	Specifies a transition point of PWM waveforms output from TIOCA <sub>4</sub> and TOCXA <sub>4</sub>
GRB4	Specifies a transition point of PWM waveforms output from $TIOCB_4$ and $TOCXB_4$

#### Setup Procedure for Complementary PWM Mode

Figure 8.33 shows a sample procedure for setting up complementary PWM mode.

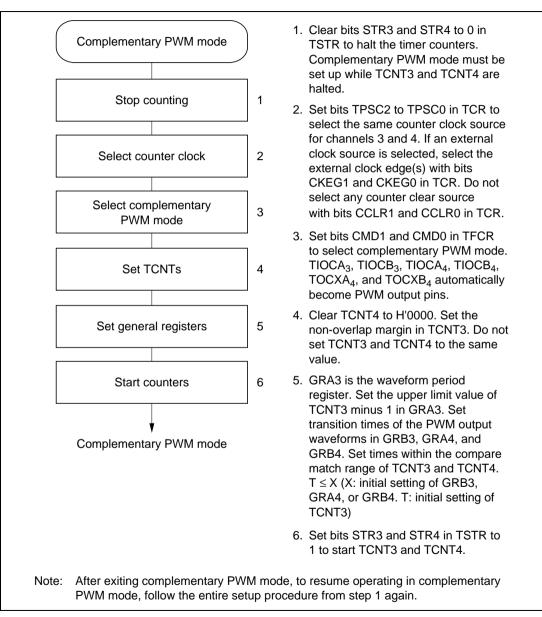


Figure 8.33 Setup Procedure for Complementary PWM Mode (Example)

## **Clearing Complementary PWM Mode**

Figure 8.34 shows a sample procedure for clearing complementary PWM mode.

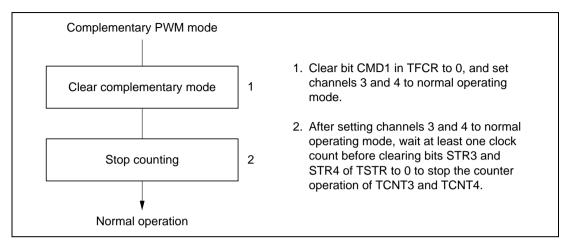


Figure 8.34 Clearing Procedure for Complementary PWM Mode (Example)

### **Examples of Complementary PWM Mode**

Figure 8.35 shows an example of operation in complementary PWM mode. TCNT3 and TCNT4 operate as up/down-counters, counting down from compare match between TCNT3 and GRA3 and counting up from the point at which TCNT4 underflows. During each up-and-down counting cycle, PWM waveforms are generated by compare match with general registers GRB3, GRA4, and GRB4. Since TCNT3 is initially set to a higher value than TCNT4, compare match events occur in the sequence TCNT3, TCNT4, TCNT4, TCNT3.

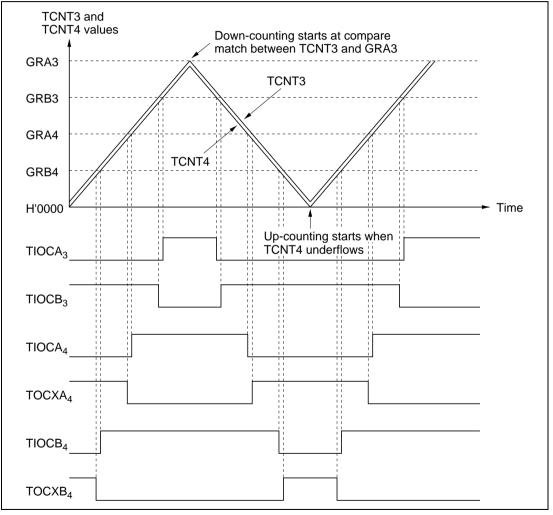


Figure 8.35 Operation in Complementary PWM Mode (Example 1) (when OLS3 = OLS4 = 1)

Figure 8.36 shows examples of waveforms with 0% and 100% duty cycles (in one phase) in complementary PWM mode. In this example the outputs change at compare match with GRB3, so waveforms with duty cycles of 0% or 100% can be output by setting GRB3 to a value larger than GRA3. The duty cycle can be changed easily during operation by use of the buffer registers. For further information see section 8.4.8, Buffering.

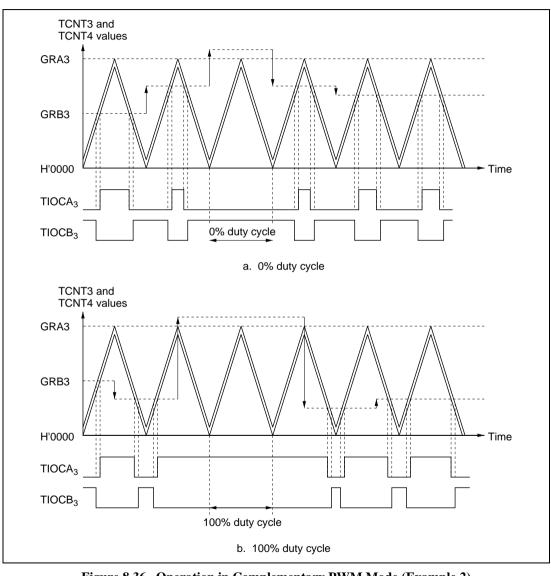


Figure 8.36 Operation in Complementary PWM Mode (Example 2) (when OLS3 = OLS4 = 1)

In complementary PWM mode, TCNT3 and TCNT4 overshoot and undershoot at the transitions between up-counting and down-counting. The setting conditions for the IMFA bit in channel 3 and the OVF bit in channel 4 differ from the usual conditions. In buffered operation the buffer transfer conditions also differ. Timing diagrams are shown in figures 8.37 and 8.38.

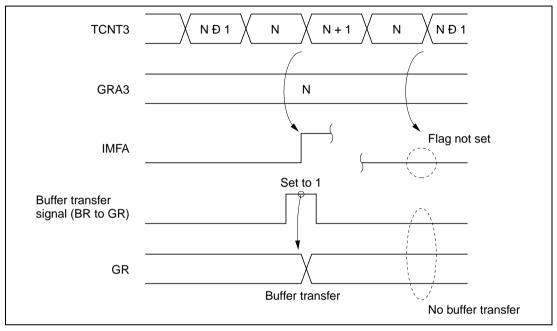


Figure 8.37 Overshoot Timing

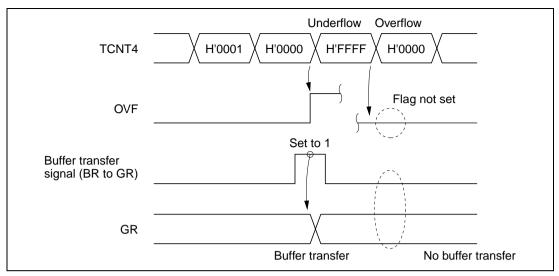


Figure 8.38 Undershoot Timing

In channel 3, IMFA is set to 1 only during up-counting. In channel 4, OVF is set to 1 only when an underflow occurs. When buffering is selected, buffer register contents are transferred to the general register at compare match A3 during up-counting, and when TCNT4 underflows.

### General Register Settings in Complementary PWM Mode

When setting up general registers for complementary PWM mode or changing their settings during operation, note the following points.

• Initial settings

Do not set values from H'0000 to T - 1 (where T is the initial value of TCNT3). After the counters start and the first compare match A3 event has occurred, however, settings in this range also become possible.

• Changing settings

Use the buffer registers. Correct waveform output may not be obtained if a general register is written to directly.

• Cautions on changes of general register settings

Figure 8.39 shows six correct examples and one incorrect example.

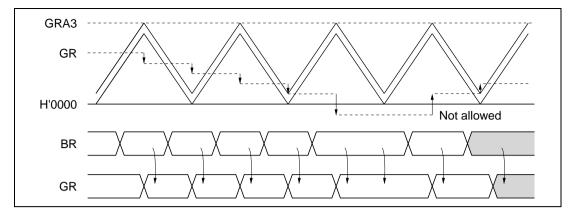


Figure 8.39 Changing a General Register Setting by Buffer Transfer (Example 1)

- Buffer transfer at transition from up-counting to down-counting

If the general register value is in the range from GRA3 - T + 1 to GRA3, do not transfer a buffer register value outside this range. Conversely, if the general register value is outside this range, do not transfer a value within this range. See figure 8.40.

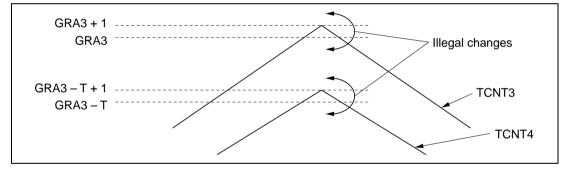


Figure 8.40 Changing a General Register Setting by Buffer Transfer (Caution 1)

- Buffer transfer at transition from down-counting to up-counting

If the general register value is in the range from H'0000 to T - 1, do not transfer a buffer register value outside this range. Conversely, when a general register value is outside this range, do not transfer a value within this range. See figure 8.41.

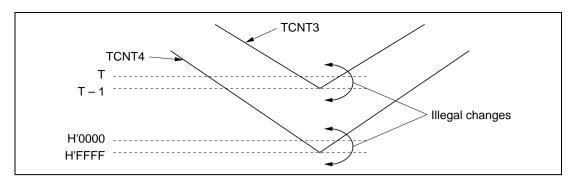


Figure 8.41 Changing a General Register Setting by Buffer Transfer (Caution 2)

— General register settings outside the counting range (H'0000 to GRA3) Waveforms with a duty cycle of 0% or 100% can be output by setting a general register to a value outside the counting range. When a buffer register is set to a value outside the counting range, then later restored to a value within the counting range, the counting direction (up or down) must be the same both times. See figure 8.42.

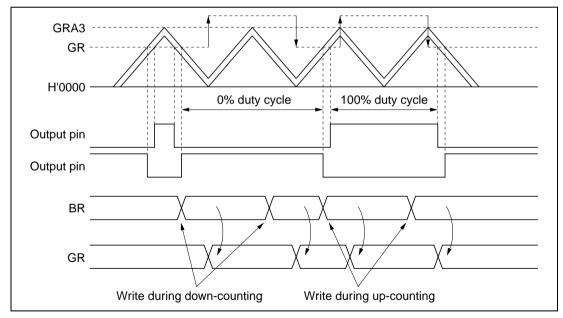


Figure 8.42 Changing a General Register Setting by Buffer Transfer (Example 2)

Settings can be made in this way by detecting GRA3 compare match or TCNT4 underflow before writing to the buffer register.

### 8.4.7 Phase Counting Mode

In phase counting mode the phase difference between two external clock inputs (at the TCLKA and TCLKB pins) is detected, and TCNT2 counts up or down accordingly.

In phase counting mode, the TCLKA and TCLKB pins automatically function as external clock input pins and TCNT2 becomes an up/down-counter, regardless of the settings of bits TPSC2 to TPSC0, CKEG1, and CKEG0 in TCR2. Settings of bits CCLR1, CCLR0 in TCR2, and settings in TIOR2, TIER2, TSR2, GRA2, and GRB2 are valid. The input capture and output compare functions can be used, and interrupts can be generated.

Phase counting is available only in channel 2.

### Sample Setup Procedure for Phase Counting Mode

Figure 8.43 shows a sample procedure for setting up phase counting mode.

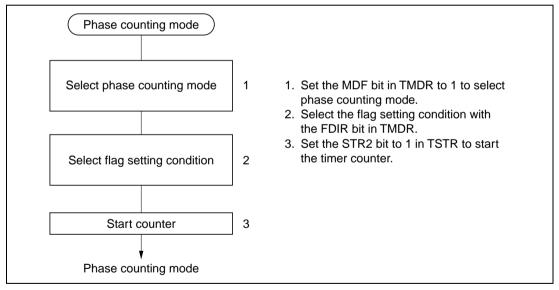


Figure 8.43 Setup Procedure for Phase Counting Mode (Example)

### **Example of Phase Counting Mode**

Figure 8.44 shows an example of operations in phase counting mode. Table 8.9 lists the upcounting and down-counting conditions for TCNT2.

In phase counting mode both the rising and falling edges of TCLKA and TCLKB are counted. The phase difference between TCLKA and TCLKB must be at least 1.5 states, the phase overlap must also be at least 1.5 states, and the pulse width must be at least 2.5 states. See figure 8.45.

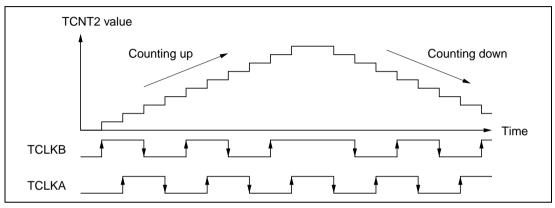


Figure 8.44 Operation in Phase Counting Mode (Example)

Table 8.9 Up/Down Counting Conditions	Table 8.9	Up/Down	Counting	Conditions
---------------------------------------	-----------	---------	----------	------------

Counting Direction Up-Counting Down-Counting				Up-Counting				
TCLKA pin	⊸	High	₹	Low	High	₹	Low	_ <b>F</b>
TCLKB pin	Low	_ <b>_</b>	High	┍╋┤	┝╋┤	Low	_ <b>F</b>	High

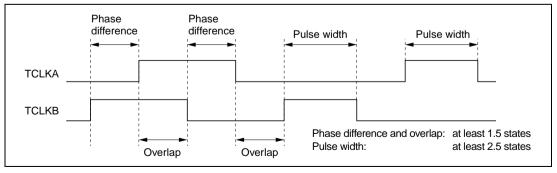


Figure 8.45 Phase Difference, Overlap, and Pulse Width in Phase Counting Mode

### 8.4.8 Buffering

Buffering operates differently depending on whether a general register is an output compare register or an input capture register, with further differences in reset-synchronized PWM mode and complementary PWM mode. Buffering is available only in channels 3 and 4. Buffering operations under the conditions mentioned above are described next.

• General register used for output compare

The buffer register value is transferred to the general register at compare match. See figure 8.46.

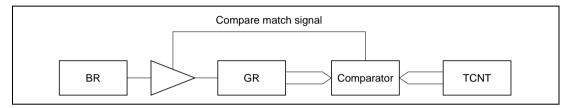


Figure 8.46 Compare Match Buffering

• General register used for input capture

The TCNT value is transferred to the general register at input capture. The previous general register value is transferred to the buffer register.

See figure 8.47.

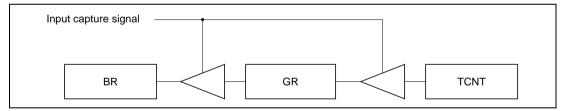


Figure 8.47 Input Capture Buffering

• Complementary PWM mode

The buffer register value is transferred to the general register when TCNT3 and TCNT4 change counting direction. This occurs at the following two times:

- When TCNT3 matches GRA3
- When TCNT4 underflows
- Reset-synchronized PWM mode

The buffer register value is transferred to the general register at compare match A3.

## Sample Buffering Setup Procedure

Figure 8.48 shows a sample buffering setup procedure.

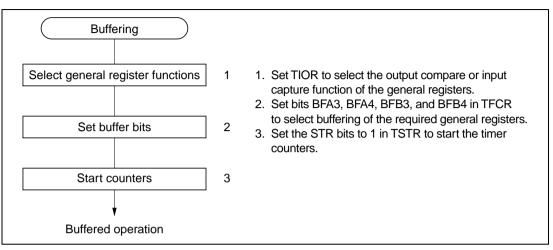


Figure 8.48 Buffering Setup Procedure (Example)

### **Examples of Buffering**

Figure 8.49 shows an example in which GRA is set to function as an output compare register buffered by BRA, TCNT is set to operate as a periodic counter cleared by GRB compare match, and TIOCA and TIOCB are set to toggle at compare match A and B. Because of the buffer setting, when TIOCA toggles at compare match A, the BRA value is simultaneously transferred to GRA. This operation is repeated each time compare match A occurs. Figure 8.50 shows the transfer timing.

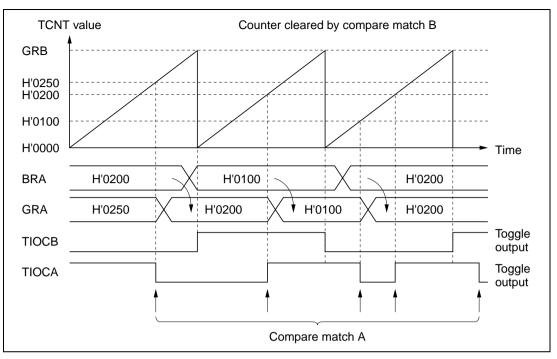


Figure 8.49 Register Buffering (Example 1: Buffering of Output Compare Register)

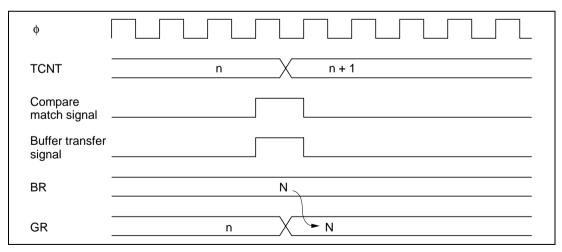


Figure 8.50 Compare Match and Buffer Transfer Timing (Example)

Figure 8.51 shows an example in which GRA is set to function as an input capture register buffered by BRA, and TCNT is cleared by input capture B. The falling edge is selected as the input capture edge at TIOCB. Both edges are selected as input capture edges at TIOCA. Because of the buffer setting, when the TCNT value is captured into GRA at input capture A, the previous GRA value is simultaneously transferred to BRA. Figure 8.52 shows the transfer timing.

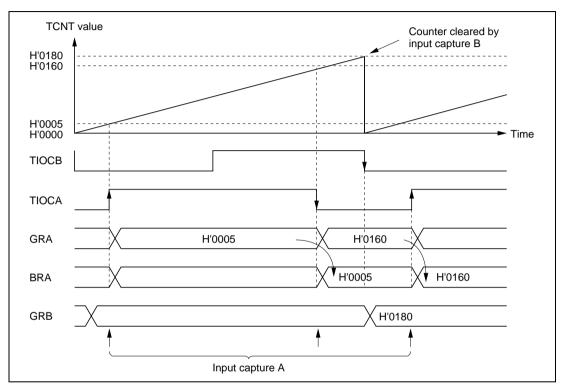


Figure 8.51 Register Buffering (Example 2: Buffering of Input Capture Register)

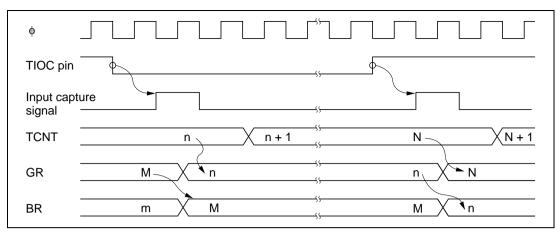


Figure 8.52 Input Capture and Buffer Transfer Timing (Example)

Figure 8.53 shows an example in which GRB3 is buffered by BRB3 in complementary PWM mode. Buffering is used to set GRB3 to a higher value than GRA3, generating a PWM waveform with 0% duty cycle. The BRB3 value is transferred to GRB3 when TCNT3 matches GRA3, and when TCNT4 underflows.

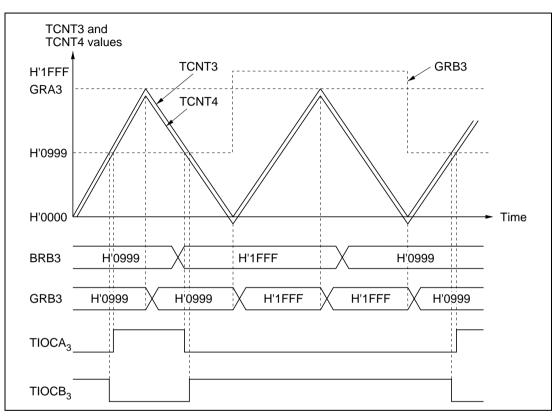


Figure 8.53 Register Buffering (Example 4: Buffering in Complementary PWM Mode)

# 8.4.9 ITU Output Timing

The ITU outputs from channels 3 and 4 can be disabled by bit settings in TOER or by an external trigger, or inverted by bit settings in TOCR.

### Timing of Enabling and Disabling of ITU Output by TOER

In this example an ITU output is disabled by clearing a master enable bit to 0 in TOER. An arbitrary value can be output by appropriate settings of the data register (DR) and data direction register (DDR) of the corresponding input/output port. Figure 8.54 illustrates the timing of the enabling and disabling of ITU output by TOER.

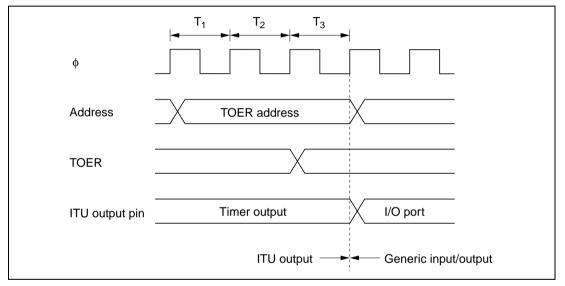


Figure 8.54 Timing of Disabling of ITU Output by Writing to TOER (Example)

# Timing of Disabling of ITU Output by External Trigger

If the XTGD bit is cleared to 0 in TOCR in reset-synchronized PWM mode or complementary PWM mode, when an input capture A signal occurs in channel 1, the master enable bits are cleared to 0 in TOER, disabling ITU output. Figure 8.55 shows the timing.

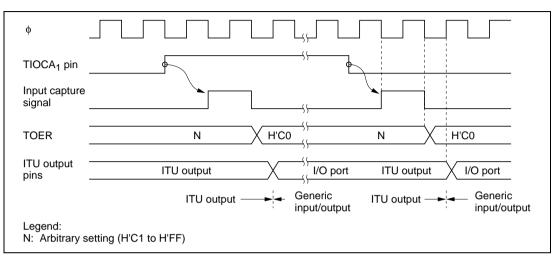


Figure 8.55 Timing of Disabling of ITU Output by External Trigger (Example)

# Timing of Output Inversion by TOCR

The output levels in reset-synchronized PWM mode and complementary PWM mode can be inverted by inverting the output level select bits (OLS4 and OLS3) in TOCR. Figure 8.56 shows the timing.

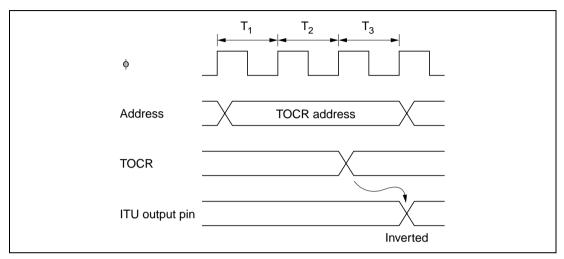


Figure 8.56 Timing of Inverting of ITU Output Level by Writing to TOCR (Example)

# 8.5 Interrupts

The ITU has two types of interrupts: input capture/compare match interrupts, and overflow interrupts.

# 8.5.1 Setting of Status Flags

# Timing of Setting of IMFA and IMFB at Compare Match

IMFA and IMFB are set to 1 by a compare match signal generated when TCNT matches a general register (GR). The compare match signal is generated in the last state in which the values match (when TCNT is updated from the matching count to the next count). Therefore, when TCNT matches a general register, the compare match signal is not generated until the next timer clock input. Figure 8.57 shows the timing of the setting of IMFA and IMFB.

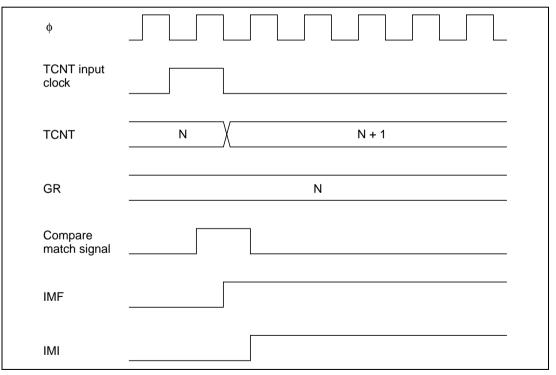


Figure 8.57 Timing of Setting of IMFA and IMFB by Compare Match

### Timing of Setting of IMFA and IMFB by Input Capture

IMFA and IMFB are set to 1 by an input capture signal. The TCNT contents are simultaneously transferred to the corresponding general register. Figure 8.58 shows the timing.

φ					
Input captur signal	e 				
IMF		 			
TCNT		 	N		 
GR				Ν	
IMI					 

Figure 8.58 Timing of Setting of IMFA and IMFB by Input Capture

# Timing of Setting of Overflow Flag (OVF)

OVF is set to 1 when TCNT overflows from H'FFFF to H'0000 or underflows from H'0000 to H'FFFF. Figure 8.59 shows the timing.



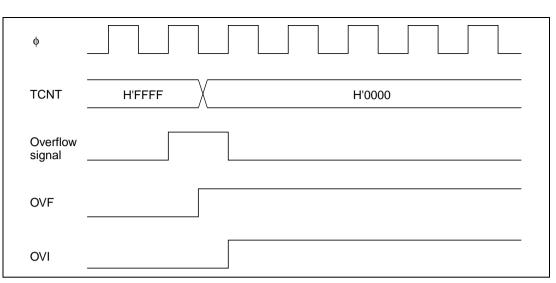


Figure 8.59 Timing of Setting of OVF

# 8.5.2 Clearing of Status Flags

If the CPU reads a status flag while it is set to 1, then writes 0 in the status flag, the status flag is cleared. Figure 8.60 shows the timing.

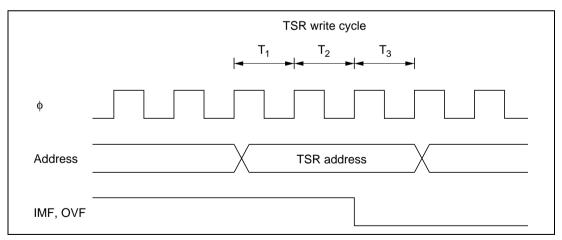


Figure 8.60 Timing of Clearing of Status Flags

### 8.5.3 Interrupt Sources

Each ITU channel can generate a compare match/input capture A interrupt, a compare match/input capture B interrupt, and an overflow interrupt. In total there are 15 interrupt sources, all independently vectored. An interrupt is requested when the interrupt request flag and interrupt enable bit are both set to 1.

The priority order of the channels can be modified in interrupt priority registers A and B (IPRA and IPRB). For details see section 5, Interrupt Controller.

Table 8.10 lists the interrupt sources.

Channel	Interrupt Source	Description	<b>Priority</b> *
0	IMIA0	Compare match/input capture A0	High
	IMIB0	Compare match/input capture B0	<b>↑</b>
	OVI0	Overflow 0	
1	IMIA1	Compare match/input capture A1	
	IMIB1	Compare match/input capture B1	
	OVI1	Overflow 1	
2	IMIA2	Compare match/input capture A2	
	IMIB2	Compare match/input capture B2	
	OVI2	Overflow 2	
3	IMIA3	Compare match/input capture A3	
	IMIB3	Compare match/input capture B3	
	OVI3	Overflow 3	
4	IMIA4	Compare match/input capture A4	
	IMIB4	Compare match/input capture B4	
	OVI4	Overflow 4	Low

#### Table 8.10 ITU Interrupt Sources

Note: \* The priority immediately after a reset is indicated. Inter-channel priorities can be changed by settings in IPRA and IPRB.

# Renesas

# 8.6 Usage Notes

This section describes contention and other matters requiring special attention during ITU operations.

# Contention between TCNT Write and Clear

If a counter clear signal occurs in the  $T_3$  state of a TCNT write cycle, clearing of the counter takes priority and the write is not performed. See figure 8.61.

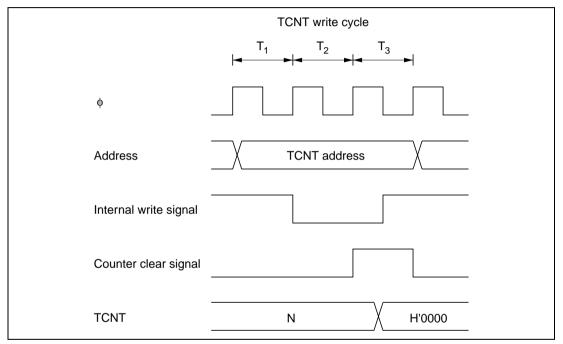


Figure 8.61 Contention between TCNT Write and Clear

#### **Contention between TCNT Word Write and Increment**

If an increment pulse occurs in the  $T_3$  state of a TCNT word write cycle, writing takes priority and TCNT is not incremented. See figure 8.62.

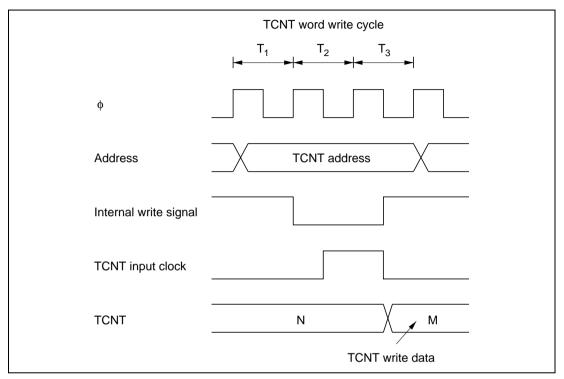


Figure 8.62 Contention between TCNT Word Write and Increment

# Contention between TCNT Byte Write and Increment

If an increment pulse occurs in the  $T_2$  or  $T_3$  state of a TCNT byte write cycle, writing takes priority and TCNT is not incremented. The TCNT byte that was not written retains its previous value. See figure 8.63, which shows an increment pulse occurring in the  $T_2$  state of a byte write to TCNTH.

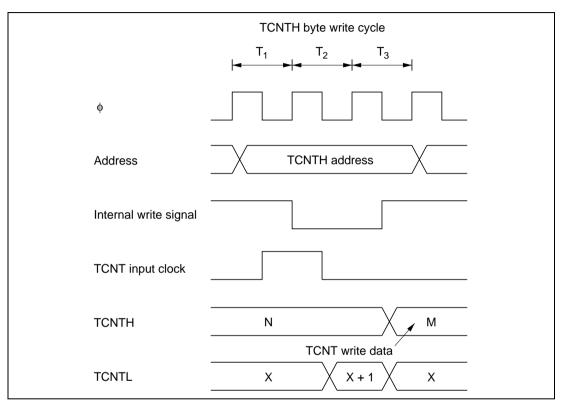


Figure 8.63 Contention between TCNT Byte Write and Increment

# Contention between General Register Write and Compare Match

If a compare match occurs in the  $T_3$  state of a general register write cycle, writing takes priority and the compare match signal is inhibited. See figure 8.64.

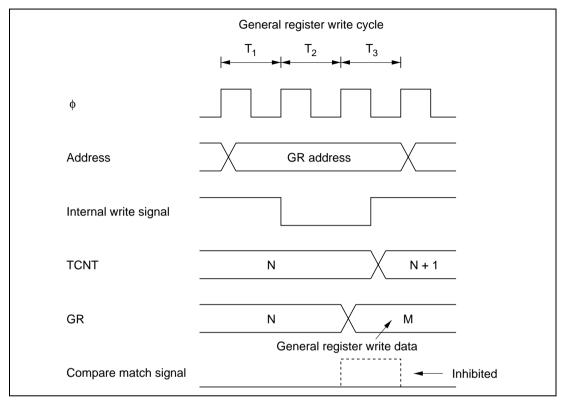


Figure 8.64 Contention between General Register Write and Compare Match

# Contention between TCNT Write and Overflow or Underflow

If an overflow occurs in the  $T_3$  state of a TCNT write cycle, writing takes priority and the counter is not incremented. OVF is set to 1. The same holds for underflow. See figure 8.65.

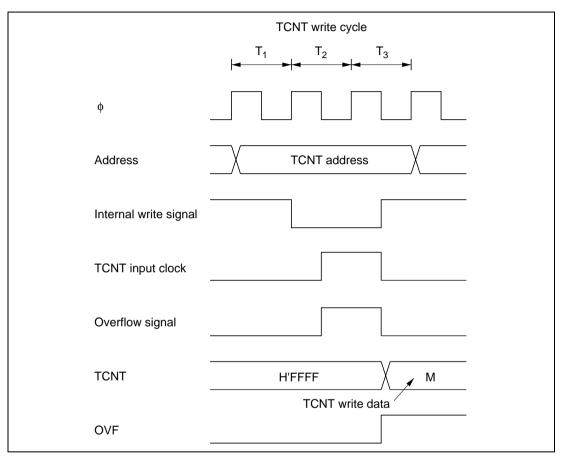


Figure 8.65 Contention between TCNT Write and Overflow

### Contention between General Register Read and Input Capture

If an input capture signal occurs during the  $T_3$  state of a general register read cycle, the value before input capture is read. See figure 8.66.

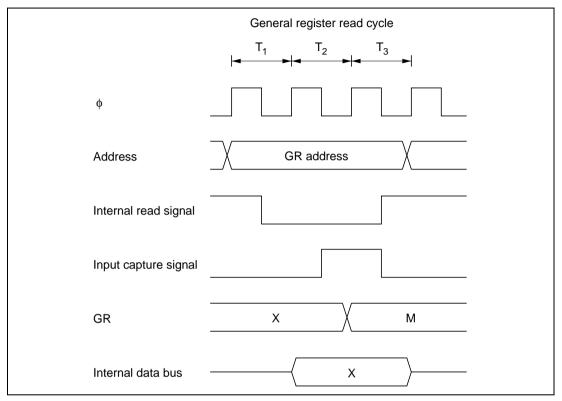


Figure 8.66 Contention between General Register Read and Input Capture

# Renesas

# Contention between Counter Clearing by Input Capture and Counter Increment

If an input capture signal and counter increment signal occur simultaneously, the counter is cleared according to the input capture signal. The counter is not incremented by the increment signal. The value before the counter is cleared is transferred to the general register. See figure 8.67.

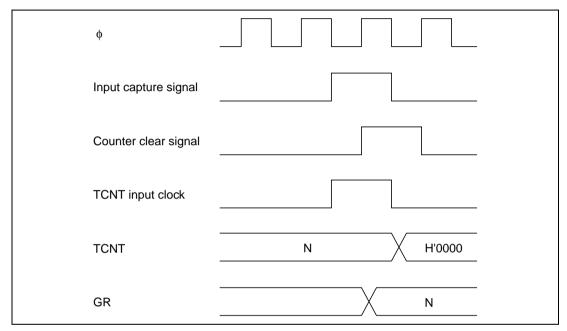


Figure 8.67 Contention between Counter Clearing by Input Capture and Counter Increment

# Contention between General Register Write and Input Capture

If an input capture signal occurs in the  $T_3$  state of a general register write cycle, input capture takes priority and the write to the general register is not performed. See figure 8.68.

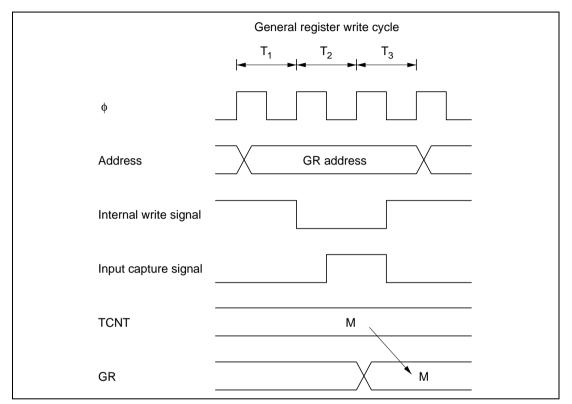


Figure 8.68 Contention between General Register Write and Input Capture

# Note on Waveform Period Setting

When a counter is cleared by compare match, the counter is cleared in the last state at which the TCNT value matches the general register value, at the time when this value would normally be updated to the next count. The actual counter frequency is therefore given by the following formula:

$$f = \frac{\phi}{(N+1)}$$

(f: counter frequency.  $\phi$ : system clock frequency. N: value set in general register.)

# Renesas

# Contention between Buffer Register Write and Input Capture

If a buffer register is used for input capture buffering and an input capture signal occurs in the  $T_3$  state of a write cycle, input capture takes priority and the write to the buffer register is not performed. See figure 8.69.

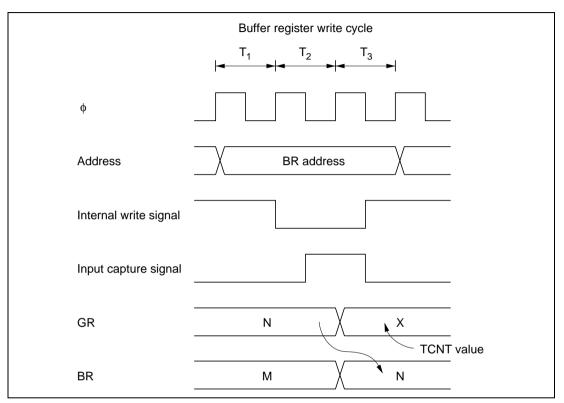


Figure 8.69 Contention between Buffer Register Write and Input Capture

#### Note on Synchronous Preset

When channels are synchronized, if a TCNT value is modified by byte write access, all 16 bits of all synchronized counters assume the same value as the counter that was addressed.

(Example) When channels 2 and 3 are synchronized

• Byte write to channel 2 or byte write to channel 3

			Write A to upper byte			
TCNT2	W	Х	of channel 2	TCNT2	А	Х
TCNT3	Y	Z		TCNT3	А	Х
	Upper byte	Lower byte	Write A to lower byte of channel 3		Upper byte	Lower byte
				TCNT2	Y	A
				TCNT3	Y	А
					Upper byte	Lower byte
Word writ	te to channe	el 2 or word	write to channel 3			
TCNT2	W	Х	<b>&gt;</b>	TCNT2	A	В
TCNT3	Y	Z	Write AB word to channel 2 or 3	TCNT3	А	В
	Upper byte	Lower byte			Upper byte	Lower byte

#### Note on Setup of Reset-Synchronized PWM Mode and Complementary PWM Mode

When setting bits CMD1 and CMD0 in TFCR, take the following precautions:

- Write to bits CMD1 and CMD0 only when TCNT3 and TCNT4 are stopped.
- Do not switch directly between reset-synchronized PWM mode and complementary PWM mode. First switch to normal mode (by clearing bit CMD1 to 0), then select reset-synchronized PWM mode or complementary PWM mode.

# ITU Operating Modes

# Table 8.11 (a) ITU Operating Modes (Channel 0)

								Registe	<b>Register Settings</b>	s					
		TSNC		TMDR			TFCR		10	TOCR	TOER	TIC	TIOR0	TCR0	
Operating Mode	g Mode	Synchro- nization	MDF	FDIR	PWM	Comple- mentary PWM	Reset- Synchro- nized PWM	Buffer- ing	ХТСР	Output Level Select	Master Enable	VOI	IOB	Clear Select	Clock Select
Synchronous preset	us preset	SYNC0 = 1	I	Ι	0	I	I	I	I	I	I	0	0	0	0
PWM mode		0	Ι	I	PWM0 = 1	Ι	Ι	Ι	I			I	*	0	0
Output compare A	npare A	0	I	I	PWM0 = 0	I	I		I		I	IOA2 = 0 Other bits unrestricted	0	0	0
Output compare B	Ipare B	0	ļ	I	0	I	I	I	I	I	I	0	IOB2 = 0 Other bits unrestricted	0	0
Input capture A	re A	0		Ι	PWM0 = 0	I	I	I	I	I	I	IOA2 = 1 Other bits unrestricted	0	0	0
Input capture B	re B	0		_	PWM0 = 0		I	Ι	l	I	I	0	IOB2 = 1 Other bits unrestricted	0	0
Counter By compare clearing match/input capture A	By compare match/input capture A	0	I	I	0	I	Ι	I	I	I	I	0	0	CCLR1 = 0 CCLR0 = 1	0
<u>s z g</u>	By compare match/input capture B	0	Ι		0	Ι	Ι	Ι	I	-	-	0	0	CCLR1 = 1 CCLR0 = 0	0
ର୍ଟ୍ ଅ	Syn- chronous clear	SYNC0 = 1	I	I	0	I	I	I	I	I	I	0	0	CCLR1 = 1 CCLR0 = 1	0
Legend: C	): Setting ar	Legend: 〇: Setting available (valid). —: Setting does not affect this mode. Note: The input capture function cannot be used in PWM mode. If compare match A and compare match signal is inhibited.	). —: Sett nnot be us	ing doe	s not affect 1 WM mode. It	this mode.	match A an	id compai	re match	B occur s	simultaneo	ously, the con	ipare match	signal is inhib	ited.

							Register	Register Settings	s					
	TSNC		TMDR	~		TFCR		2	TOCR	TOER	TIC	TIOR1	TCR1	
Operating Mode	de Synchro- nization	n MDF	FDIR	PWM	Comple- mentary PWM	Reset- Synchro- nized PWM	Buffer- ing	XTGD	Output Level Select	Master Enable	IOA	IOB	Clear Select	Clock Select
Synchronous preset	set SYNC1 = 1	-	1	0	1		I	I	1	1	0	0	0	0
PWM mode	0	Ι	Ι	PWM1 = 1	Ι	Ι	Ι	Ι	Ι	Ι	Ι	0*1	0	0
Output compare A	0	Ι	l	PWM1 = 0	I	I	I			I	IOA2 = 0 Other bits unrestricted	0	0	0
Output compare B	0 x	I	I	0	I	I	I	I	I	I	0	IOB2 = 0 Other bits unrestricted	0	0
Input capture A	0	I		PWM1 = 0	I	I	I	$O^{*2}$	I	I	IOA2 = 1 Other bits unrestricted	0	0	0
Input capture B	0	I	I	PWM1 = 0	I	I	I			I	0	IOB2 = 1 Other bits unrestricted	0	0
Counter By compare clearing match/input capture A	but O	I		0	I	I	I	_	Ι	Ι	0	0	CCLR1 = 0 CCLR0 = 1	0
By compare match/input capture B	but O	Ι	l	0	I	I	I			I	0	0	CCLR1 = 1 CCLR0 = 0	0
Syn- chronous clear	s SYNC1 = 1	-	I	0	I	I	I	I	I	I	0	0	CCLR1 = 1 CCLR0 = 1	0
Legend: O: Setting available (valid). —: Setting does not affect this mode. Notes: 1. The input capture function cannot be used in PWM mode. If compare match A and compare match B occur simultaneously, the compare match signal is inhibited. 2. Valid only when channels 3 and 4 are operating in complementary PWM mode or reset-synchronized PWM mode.	<ol> <li>O: Setting available (valid). —: Setting does not affect this mode.</li> <li>The input capture function cannot be used in PVMM mode. If compare match A and compare match B occur simult.</li> <li>Valid only when channels 3 and 4 are operating in complementary PVMM mode or reset-synchronized PVMM mode.</li> </ol>	/alid). —: Se tion cannot l iels 3 and 4 a	etting do be used are oper	es not affect t in PWM mod ating in comp	his mode. e. If compa lementary l	Fre match /	A and con le or reset	npare ma ⊱synchro	atch B occ nized PW	ur simult	aneously, the	compare mat	ch signal is ir	hibited.

# Table 8.11 (b) ITU Operating Modes (Channel 1)

# Renesas

								Registe	Register Settings	s					
		TSNC		TMDR			TFCR		10	TOCR	TOER	TIC	TIOR2	TCR2	
Opera	Operating Mode	Synchro- nization	MDF	FDIR	MWd	Comple- mentary PWM	Reset- Synchro- Buffer- nized ing	Buffer- ing	XTGD	Output Level Select	Master Enable	POI	IOB	Clear Select	Clock Select
Synchror	Synchronous preset	SYNC2 = 1	0	Ι	0	1			Ι			0	0	0	0
PWM mode	ode	0	0	I	PWM2 = 1	1	I	I	I	I	1	I	*0	0	0
Output c	Output compare A	0	0	I	PWM2 = 0	I	I	I	I			IOA2 = 0 Other bits unrestricted	0	0	0
Output c	Output compare B	0	0	I	0	I		I	I		I	0	IOB2 = 0 Other bits unrestricted	0	0
Input capture A	oture A	0	0	I	PWM2 = 0	I		I	Ι		I	IOA2 = 1 Other bits unrestricted	0	0	0
Input capture B	oture B	0	0	I	PWM2 = 0	I	I	I	I	I	I	0	IOB2 = 1 Other bits unrestricted	0	0
Counter clearing	Counter By compare clearing match/input capture A	0	0		0	I	I	Ι	Ι	I	I	0	0	CCLR1 = 0 CCLR0 = 1	0
	By compare match/input capture B	0	0		0	I	I	Ι		I	I	0	0	CCLR1 = 1 CCLR0 = 0	0
	Syn- chronous clear	SYNC2 = 1	0	I	0	I	-	Ι	Ι		I	0	0	CCLR1 = 1 CCLR0 = 1	0
Phase counting mode	ounting	0	MDF = 1	0	0	I	ļ	I	Ι	ļ	I	0	0	0	I
Legend: ( Note: The	Legend: 〇: Setting available (valid). —: Setting does not affect this mode. Note: The input capture function cannot be used in PWM mode. If compare match A and compare match B occur simultaneously , the compare match signal is inhibited.	$\bigcirc$ : Setting available (valid). —: Setting does not affect this mode. Input capture function cannot be used in PWM mode. If compare r	t). —: Sett not be use	ing doe ∍d in PV	s not affect th VM mode. If	his mode. compare m	atch A and	d compar	e match	B occur si	multaneo	usly , the cor	mpare match	signal is inhit	bited.

# Table 8.11 (c) ITU Operating Modes (Channel 2)

Section 8 16-Bit Integrated Timer Unit (ITU)

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								Register Settings	ettings						
		TSNC		TMDR			TFCR		5	TOCR	TOER	TIOR3	<b>č</b> 3	TCR3	
Opera	Operating Mode	Synchro- nization	MDF	FDIR	PWM	Comple- mentary PWM	Reset- Synchro- nized PWM	Buffering	XTGD	Output Level Select	Master Enable	POA	IOB	Clear Select	Clock Select
Synchron	Synchronous preset	SYNC3 = 1	1	1	0	0*3	0	0	1	I	0*1	0	0	0	0
PWM mode	de	0	Ι		PWM3 = 1 CMD1 = 0	CMD1 = 0	CMD1 = 0	0	Ι	Ι	0	Ι	$O^{*2}$	0	0
Output co	Output compare A	0	I	1	PWM3 = 0 CMD1 = 0	CMD1 = 0	CMD1 = 0	0	I		0	IOA2 = 0 Other bits unrestricted	0	0	0
Output co	Output compare B	0	I	I	0	CMD1 = 0	CMD1 = 0	0	I	I	0	0	IOB2 = 0 Other bits unrestricted	0	0
Input capture A	ture A	0	I		PWM3 = 0 CMD1 = 0	CMD1 = 0	CMD1 = 0	0	I		EA3 ignored Other bits unrestricted	IOA2 = 1 Other bits unrestricted	0	0	0
Input capture B	ture B	0	I		PWM3 = 0 CMD1 = 0	CMD1 = 0	CMD1 = 0	0	I		EB3 ignored Other bits unrestricted	0	IOB2 = 1 Other bits unrestricted	0	0
Counter clearing	By compare match/input capture A	0	Ι	I	0	Illegal setting: CMD1 = 1 CMD0 = 0	0*4	0	I	I	0*1	0	0	CCLR1 = 0 CCLR0 = 1	0
	By compare match/input capture B	0	I	1	0	CMD1 = 0	CMD1 = 0	0	I		°*1	0	0	CCLR1 = 1 CCLR0 = 0	0
	Syn- chronous clear	SYNC3 = 1	I	I	0	Illegal setting: CMD1 = 1 CMD0 = 0	0	0	I		0*1	0	0	CCLR1 = 1 CCLR0 = 1	0
Complementary PWM mode	ientary de	0*3		I	I	CMD1 = 1 CMD0 = 0	CMD1 = 1 CMD0 = 0	0	9*0	0	0	Ι	I	CCLR1 = 0 CCLR0 = 0	0*5
Reset-synch PWM mode	Reset-synchronized PWM mode	0		I	I	CMD1 = 1 CMD0 = 1	CMD1 = 1 CMD0 = 1	0	0*6	0	0	I	l	CCLR1 = 0 CCLR0 = 1	0
Buf fering (BRA)		0	I		0	0	0	BFA3 = 1 Other bits unrestricted		I	0*1	0	0	0	0
Buffering (BRB)		0	I		0	0	0	BFB3 = 1 Other bits unrestricted	I	I	0 <u>*</u>	0	0	0	0
Legend: Notes: -1. 21. 6. 54.		C): Setting available (valid). —: Setting does Master enable bit strings are valid only durin The input capture function cannot be used in Do not set both channels 3 and 4 for synchron The counter cannot be cleared by input captu In complementary PWM mode, select the sam of the input capture A function in channel 1.	<ul> <li>d). —: So are valion cannot and cannot and a sand 4 leared by mode, se unction ir unction ir unction ir inction ir inct</li></ul>	etting dc d only c be user for sync input c ilect the	pes not affi during wavu d in PWM I chronous o apture A w same cloc el 1.	○: Setting available (valid). —: Setting does not affect this mode. Master enable bit settings are valid only during waveform output. The intracapture transmiss are valid only during waveform output. Do not set both channels 3 and 4 for synchronous operation when complementary PWM mode is selected. The counter cannot be clarend by input capture A when reset-synchronoizad PWM mode is selected. In commemorary PWM mode. select the same clock source for channels 3 and 4. Use the input capture A function in channel 1.	oare match / n compleme nchronized P channels 3 a	A and compé ntary PWM i 'WM mode i	are matc mode is s selecte	sh B occu selected ed.	ur simultaneo	usly, the corr	ipare match	dni si lengis r	libited.

 Table 8.11 (d)
 ITU Operating Modes (Channel 3)

Rev.3.00 Mar. 26, 2007 Page 289 of 682 REJ09B0353-0300

								Register Settings	ttings						
		TSNC		TMDR			TFCR		5	TOCR	TOER	TIOR4	R4	TCR4	
Operating Mode	Mode	Synchro- nization	MDF	FDIR	PWM	Comple- mentary PWM	Reset- Synchro- nized PWM	Buffering	хтбр	Output Level Select	Master Enable	POA	BOI	Clear Select	Clock Select
Synchronous preset	preset	SYNC4 = 1	I	I	0	0*3	0	0	I	I	0*1	0	0	0	0
PWM mode		0	I	I	PWM4 = 1	PWM4 = 1 CMD1 = 0	CMD1 = 0	0	I	I	0	I	0*2	0	0
Output compare A	are A	0	I	I	PWM4 = 0	PWM4 = 0 CMD1 = 0	CMD1 = 0	0	Ι		0	IOA2 = 0 Other bits unrestricted	0	0	0
Output compare B	are B	0		I	0	CMD1 = 0	CMD1 = 0	0	Ι		0	0	IOB2 = 0 Other bits unrestricted	0	0
Input capture A	A	0	I		PWM4 = 0	PWM4 = 0 CMD1 = 0	CMD1 = 0	0	-	I	EA4 ignored Other bits unrestricted	IOA2 = 1 Other bits unrestricted	0	0	0
Input capture B	В	0	I		PWM4 = 0	CMD1 = 0	CMD1 = 0	0		I	EB4 ignored Other bits unrestricted	0	IOB2 = 1 Other bits unrestricted	0	0
Counter By clearing ma	By compare match/input capture A	0	I	I	0	Illegal setting: CMD1 = 1 CMD0 = 0	0 <sup>*4</sup>	0	I	I	0*1	0	0	CCLR1 = 0 CCLR0 = 1	0
By ma car	By compare match/input capture B	0	I	I	0	Illegal setting: CMD1 = 1 CMD0 = 0	* <sup>4</sup>	0	I	I	0*1	0	0	CCLR1 = 1 CCLR0 = 0	0
Syn- chron clear	Syn- chronous clear	SYNC4 = 1	I	I	0	Illegal setting: CMD1 = 1 CMD0 = 0	0*4	0	I		0*1	0	0	CCLR1 = 1 CCLR0 = 1	0
Complementary PWM mode	ary	0*3	I	I	I	CMD1 = 1 CMD0 = 0	CMD1 = 1 CMD0 = 0	0	0	0	0	I	I	CCLR1 = 0 CCLR0 = 0	¥2
Reset-synchronized PWM mode	onized	0	I	I	I	CMD1 = 1 CMD0 = 1	CMD1 = 1 CMD0 = 1	0	0	0	0	l	I	9*6	0*6
Buffering (BRA)		0	I	I	0	0	0	BFA4 = 1 Other bits unrestricted	I		0*1	0	0	0	0
Buffering (BRB)		0	I		0	0	0	BFB4 = 1 Other bits unrestricted	I	I	0*1	0	0	0	0
Legend: O: Notes: 1. Ms 2. Th 3. Do 3. Do 5. In 6. TO	Setting a aster enak aster enak ie input ca not set b on not set b complemet CR4 settin	vailable (vali ble bit setting: pture functio oth channels synchronized antary PWM i gs are valid ii	<ul> <li>d). —: Se are valid and a cannot</li> <li>and 4</li> <li>and 4</li> <li>PWM m</li> <li>mode, se</li> <li>n reset-s)</li> </ul>	atting d a only c be use for sync lode is nchror	bes not aff( luring wave hin PWM r hronous o selected, T same cloc same cloc	C): Setting available (valid). —: Setting does not affect this mode. <ol> <li>Anster enable bit settings are valid only during waveform output.</li> <li>Anster enable bit settings are and only during waveform output.</li> <li>The input capture function cannot be used in PWM mode. If compare match A and compare match B occur simultaneously, the compare match signal is inhibited.</li> <li>Do not set both channels 3 and 4 for synchronous operation when complementary PWM mode is selected.</li> <li>When reset-synchronized PWM mode is selected. TNT4 operates independently and the counter clearing function is available. W aveform output is not affected.</li> <li>Incomplementary PWM mode, selected, TNT4 operates independently and the counter clearing function is available. W aveform output is not affected.</li> <li>TCR4 settings are valid in reset-synchronized PWM mode, but TCNT4 operates independently, without affecting waveform output.</li> </ol>	bare match A n compleme es independ channels 3 a CNT4 operat	, and compa ntary PWM r ently and th nd 4. es indepenc	rre matc mode is e counte lently , v	th B occu selected er clearin vithout af	ir simultaneo g function is fecting wave	usly, the cor available. W form output.	npare match aveform ou	r signal is inf tput is not aff	iibited. fected.

 Table 8.11 (e)
 ITU Operating Modes (Channel 4)

Section 8

Rev.3.00 Mar. 26, 2007 Page 290 of 682 REJ09B0353-0300

# Section 9 Programmable Timing Pattern Controller

# 9.1 Overview

The H8/3039 Group has a built-in programmable timing pattern controller (TPC)\* that provides pulse outputs by using the 16-bit integrated timer-pulse unit (ITU) as a time base. The TPC pulse outputs are divided into 4-bit groups (group 3 to group 0) that can operate simultaneously and independently.

# 9.1.1 Features

TPC features are listed below.

• 15-bit output data

Maximum 15-bit data can be output. TPC output can be enabled on a bit-by-bit basis.

- Four output groups and one 3-bit output. Output trigger signals can be selected in 4-bit groups to provide up to three different 4-bit outputs and one 3-bit output.
- Selectable output trigger signals Output trigger signals can be selected for each group from the compare-match signals of four ITU channels.
- Non-overlap mode

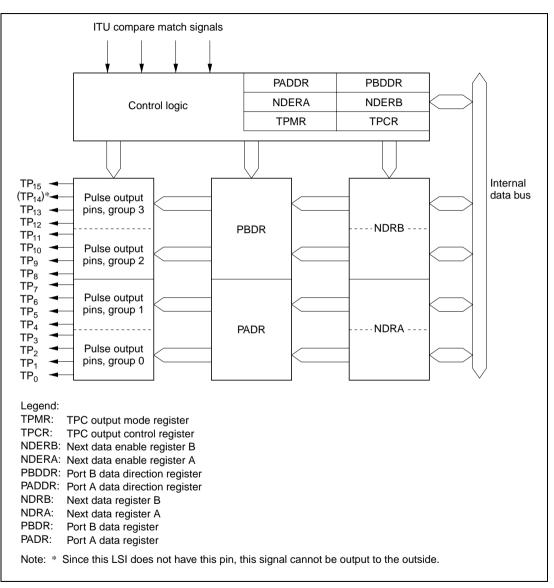
A non-overlap margin can be provided between pulse outputs.

Note: \* Note that since this LSI does not have a  $TP_{14}$  pin, it is a 15-bit programmable timing pattern controller (TPC).

# Renesas

### 9.1.2 Block Diagram

Figure 9.1 shows a block diagram of the TPC.



# Figure 9.1 TPC Block Diagram

Rev.3.00 Mar. 26, 2007 Page 292 of 682 REJ09B0353-0300

#### 9.1.3 TPC Pins

Table 9.1 summarizes the TPC output pins.

# Table 9.1 TPC Pins

Name	Symbol	I/O	Function
TPC output 0	TP <sub>o</sub>	Output	Group 0 pulse output
TPC output 1	TP <sub>1</sub>	Output	
TPC output 2		Output	
TPC output 3	TP <sub>3</sub>	Output	
TPC output 4	TP <sub>4</sub>	Output	Group 1 pulse output
TPC output 5	TP <sub>5</sub>	Output	
TPC output 6	TP <sub>6</sub>	Output	
TPC output 7	TP <sub>7</sub>	Output	
TPC output 8	TP <sub>8</sub>	Output	Group 2 pulse output
TPC output 9	TP <sub>9</sub>	Output	
TPC output 10	TP <sub>10</sub>	Output	
TPC output 11	TP <sub>11</sub>	Output	
TPC output 12	TP <sub>12</sub>	Output	Group 3 pulse output
TPC output 13	TP <sub>13</sub>	Output	
(TPC output 14)*	(TP <sub>14</sub> )*	(Output)*	
TPC output 15	TP <sub>15</sub>	Output	

Note: \* Since this LSI does not have this pin, this signal cannot be output to the outside.

### 9.1.4 Registers

Table 9.2 summarizes the TPC registers.

### Table 9.2TPC Registers

Address*1	Name	Abbreviation	R/W	Initial Value
H'FFD1	Port A data direction register	PADDR	W	H'00
H'FFD3	Port A data register	PADR	R/(W)* <sup>2</sup>	H'00
H'FFD4	Port B data direction register	PBDDR	W	H'00
H'FFD6	Port B data register	PBDR	R/(W)* <sup>2</sup>	H'00
H'FFA0	TPC output mode register	TPMR	R/W	H'F0
H'FFA1	TPC output control register	TPCR	R/W	H'FF
H'FFA2	Next data enable register B	NDERB	R/W	H'00
H'FFA3	Next data enable register A	NDERA	R/W	H'00
H'FFA5/ H'FFA7* <sup>3</sup>	Next data register A	NDRA	R/W	H'00
H'FFA4/ H'FFA6* <sup>3</sup>	Next data register B	NDRB	R/W	H'00

Notes: 1. Lower 16 bits of the address.

2. Bits used for TPC output cannot be written.

3. The NDRA address is H'FFA5 when the same output trigger is selected for TPC output groups 0 and 1 by settings in TPCR. When the output triggers are different, the NDRA address is H'FFA7 for group 0 and H'FFA5 for group 1. Similarly, the address of NDRB is H'FFA4 when the same output trigger is selected for TPC output groups 2 and 3 by settings in TPCR. When the output triggers are different, the NDRB address is H'FFA6 for group 2 and H'FFA4 for group 3.

# 9.2 Register Descriptions

### 9.2.1 Port A Data Direction Register (PADDR)

PADDR is an 8-bit write-only register that selects input or output for each pin in port A.

Bit	7	6	5	4	3	2	1	0
	PA7DDR	PA <sub>6</sub> DDR	PA5DDR	PA <sub>4</sub> DDR	PA3DDR	PA <sub>2</sub> DDR	PA1DDR	PA <sub>0</sub> DDR
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W
				Port A da These bits				

output for port A pins

Port A is multiplexed with pins  $TP_7$  to  $TP_0$ . Bits corresponding to pins used for TPC output must be set to 1. For further information about PADDR, see section 7.10, Port A.

### 9.2.2 Port A Data Register (PADR)

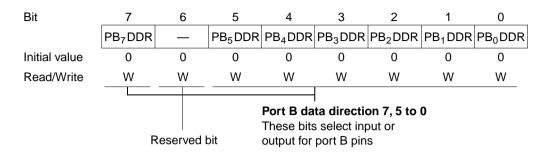
PADR is an 8-bit readable/writable register that stores TPC output data for groups 0 and 1, when these TPC output groups are used.

7	6	5	4	3	2	1	0
PA <sub>7</sub>	PA <sub>6</sub>	$PA_5$	PA <sub>4</sub>	$PA_3$	PA <sub>2</sub>	PA <sub>1</sub>	PA <sub>0</sub>
0	0	0	0	0	0	0	0
R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*
	0	PA <sub>7</sub> PA <sub>6</sub> 0 0	PA <sub>7</sub> PA <sub>6</sub> PA <sub>5</sub> 0         0         0           R/(W)*         R/(W)*         R/(W)*	PA <sub>7</sub> PA <sub>6</sub> PA <sub>5</sub> PA <sub>4</sub> 0         0         0         0           R/(W)*         R/(W)*         R/(W)*         R/(W)*           Port A da These bits         Port A da	PA <sub>7</sub> PA <sub>6</sub> PA <sub>5</sub> PA <sub>4</sub> PA <sub>3</sub> 0         0         0         0         0           R/(W)*         R/(W)*         R/(W)*         R/(W)*         R/(W)*           Port A data 7 to 0 These bits store out         Port Store out         Pa	PA <sub>7</sub> PA <sub>6</sub> PA <sub>5</sub> PA <sub>4</sub> PA <sub>3</sub> PA <sub>2</sub> 0         0         0         0         0         0         0           R/(W)*         R/(W)*         R/(W)*         R/(W)*         R/(W)*         R/(W)*         R/(W)*           Port A data 7 to 0         These bits store output data         PA         PA         PA         PA	PA <sub>7</sub> PA <sub>6</sub> PA <sub>5</sub> PA <sub>4</sub> PA <sub>3</sub> PA <sub>2</sub> PA <sub>1</sub> 0         0         0         0         0         0         0         0           R/(W)*         R/(W)*         R/(W)*         R/(W)*         R/(W)*         R/(W)*         R/(W)*         R/(W)*           Port A data 7 to 0

Note: \* Bits selected for TPC output by NDERA settings become read-only bits.

For further information about PADR, see section 7.10, Port A.

# 9.2.3 Port B Data Direction Register (PBDDR)

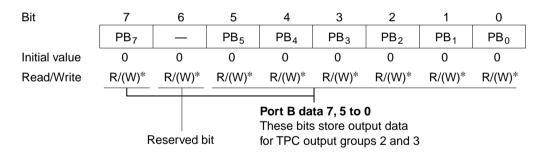


PBDDR is an 8-bit write-only register that selects input or output for each pin in port B.

Port B is multiplexed with pins  $TP_{15}$ ,  $TP_{13}$  to  $TP_8$ . Bits corresponding to pins used for TPC output must be set to 1. For further information about PBDDR, see section 7.11, Port B.

# 9.2.4 Port B Data Register (PBDR)

PBDR is an 8-bit readable/writable register that stores TPC output data for groups 2 and 3, when these TPC output groups are used.



Note: \* Bits selected for TPC output by NDERB settings become read-only bits.

For further information about PBDR, see section 7.11, Port B.

### 9.2.5 Next Data Register A (NDRA)

NDRA is an 8-bit readable/writable register that stores the next output data for TPC output groups 1 and 0 (pins  $TP_7$  to  $TP_0$ ). During TPC output, when an ITU compare match event specified in TPCR occurs, NDRA contents are transferred to the corresponding bits in PADR. The address of NDRA differs depending on whether TPC output groups 0 and 1 have the same output trigger or different output triggers.

NDRA is initialized to H'00 by a reset and in hardware standby mode. It is not initialized in software standby mode.

#### Same Trigger for TPC Output Groups 0 and 1

If TPC output groups 0 and 1 are triggered by the same compare match event, the NDRA address is H'FFA5. The upper 4 bits belong to group 1 and the lower 4 bits to group 0. Address H'FFA7 consists entirely of reserved bits that cannot be modified and always read 1.

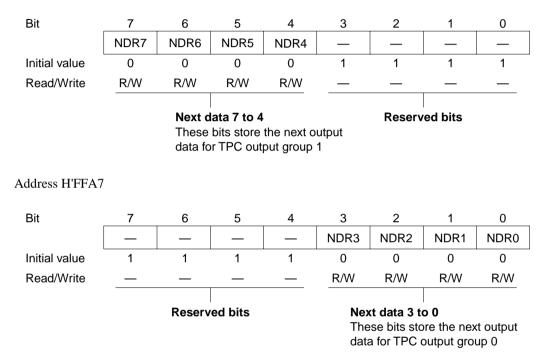
Address H'FFA5

Bit	7	6	5	4	3	2	1	0		
	NDR7	NDR6	NDR5	NDR4	NDR3	NDR2	NDR1	NDR0		
Initial value	0	0	0	0	0	0	0	0		
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W		
	Th	ext data 7 lese bits st ta for TPC	ore the ne		<b>Next data 3 to 0</b> These bits store the next output data for TPC output group 0					
Address H'FFA7										
Bit	7	6	5	4	3	2	1	0		
	_	_	—	—	_	_	—	_		
Initial value	1	1	1	1	1	1	1	1		
Read/Write		—	—	—	—	—	—	_		
		Reserved bits								

# Different Triggers for TPC Output Groups 0 and 1

If TPC output groups 0 and 1 are triggered by different compare match events, the address of the upper 4 bits of NDRA (group 1) is H'FFA5 and the address of the lower 4 bits (group 0) is H'FFA7. Bits 3 to 0 of address H'FFA5 and bits 7 to 4 of address H'FFA7 are reserved bits that cannot be modified and always read 1.

### Address H'FFA5



### 9.2.6 Next Data Register B (NDRB)

NDRB is an 8-bit readable/writable register that stores the next output data for TPC output groups 3 and 2 (pins  $TP_{15}$  to  $TP_{3}$ )\*. During TPC output, when an ITU compare match event specified in TPCR occurs, NDRB contents are transferred to the corresponding bits in PBDR. The address of NDRB differs depending on whether TPC output groups 2 and 3 have the same output trigger or different output triggers.

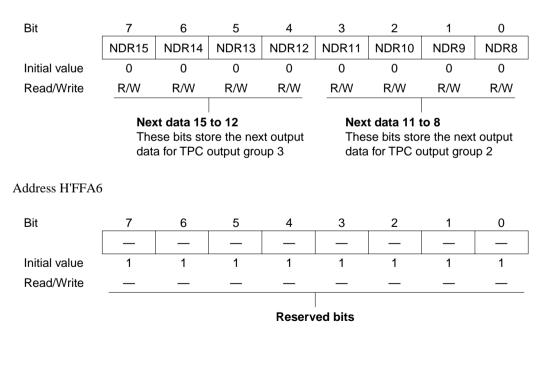
NDRB is initialized to H'00 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Note: \* Since this LSI does not have a  $TP_{14}$  pin, the  $TP_{14}$  signal cannot be output to the outside.

#### Same Trigger for TPC Output Groups 2 and 3

If TPC output groups 2 and 3 are triggered by the same compare match event, the NDRB address is H'FFA4. The upper 4 bits belong to group 3 and the lower 4 bits to group 2. Address H'FFA6 consists entirely of reserved bits that cannot be modified and always read 1.

Address H'FFA4



# Different Triggers for TPC Output Groups 2 and 3

If TPC output groups 2 and 3 are triggered by different compare match events, the address of the upper 4 bits of NDRB (group 3)\* is H'FFA4 and the address of the lower 4 bits (group 2) is H'FFA6. Bits 3 to 0 of address H'FFA4 and bits 7 to 4 of address H'FFA6 are reserved bits that cannot be modified and always read 1.

Note: \* Since this LSI does not have a  $TP_{14}$  pin, the  $TP_{14}$  signal cannot be output off-chip.

### Address H'FFA4

Bit	7	6	5	4	3	2	1	0	
	NDR15	NDR14	NDR13	NDR12	—	_	—	_	
Initial value	0	0	0	0	1	1	1	1	
Read/Write	R/W	R/W	R/W	R/W	_	—	—		
	Next data 15 to 12Reserved bitsThese bits store the next output data for TPC output group 3								
Address H'FFA6	i								
Bit	7	6	5	4	3	2	1	0	
	_	_	_	—	NDR11	NDR10	NDR9	NDR8	
Initial value	1	1	1	1	0	0	0	0	
Read/Write	_		_		R/W	R/W	R/W	R/W	
		Reserv	ed bits		<b>Next data 11 to 8</b> These bits store the next output data for TPC output group 2				

#### 9.2.7 Next Data Enable Register A (NDERA)

----

NDERA is an 8-bit readable/writable register that enables or disables TPC output groups 1 and 0 (TP<sub>7</sub> to TP<sub>0</sub>) on a bit-by-bit basis.

Bit	7	6	5	4	3	2	1	0
	NDER7	NDER6	NDER5	NDER4	NDER3	NDER2	NDER1	NDER0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

**Next data enable 7 to 0** These bits enable or disable TPC output groups 1 and 0

If a bit is enabled for TPC output by NDERA, then when the ITU compare match event selected in the TPC output control register (TPCR) occurs, the NDRA value is automatically transferred to the corresponding PADR bit, updating the output value. If TPC output is disabled, the bit value is not transferred from NDRA to PADR and the output value does not change.

NDERA is initialized to H'00 by a reset and in hardware standby mode. It is not initialized in software standby mode.

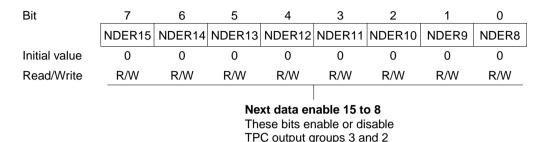
**Bits 7 to 0—Next Data Enable 7 to 0 (NDER7 to NDER0):** These bits enable or disable TPC output groups 1 and 0 ( $TP_7$  to  $TP_0$ ) on a bit-by-bit basis.

Bits 7 to 0 NDER7 to NDER0	Description	
0	TPC outputs $TP_7$ to $TP_0$ are disabled (NDR <sub>7</sub> to NDR0 are not transferred to PA <sub>7</sub> to PA <sub>0</sub> )	(Initial value)
1	TPC outputs $TP_7$ to $TP_0$ are enabled (NDR <sub>7</sub> to NDR <sub>0</sub> are transferred to PA <sub>7</sub> to PA <sub>0</sub> )	

# Bits 7 to 0—Next Data Enable 7 to 0 (NDER7 to NDER0): These bits enable or disable T

#### 9.2.8 Next Data Enable Register B (NDERB)

NDERB is an 8-bit readable/writable register that enables or disables TPC output groups 3 and 2  $(TP_{15} \text{ to } TP_8)^*$  on a bit-by-bit basis.



If a bit is enabled for TPC output by NDERB, then when the ITU compare match event selected in the TPC output control register (TPCR) occurs, the NDRB value is automatically transferred to the corresponding PBDR bit, updating the output value. If TPC output is disabled, the bit value is not transferred from NDRB to PBDR and the output value does not change.

NDERB is initialized to H'00 by a reset and in hardware standby mode. It is not initialized in software standby mode.

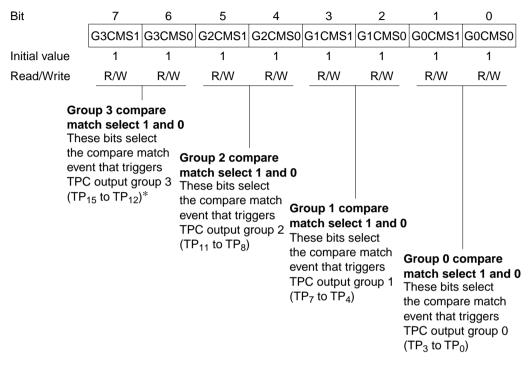
**Bits 7 to 0—Next Data Enable 15 to 8 (NDER15 to NDER8):** These bits enable or disable TPC output groups 3 and 2 ( $TP_{15}$  to  $TP_{8}$ )\* on a bit-by-bit basis.

Bits 7 to 0 NDER15 to NDER8	Description	
0	TPC outputs $TP_{15}$ to $TP_8$ are disabled (NDR15 to NDR8 are not transferred to PB <sub>7</sub> to PB <sub>0</sub> )	(Initial value)
1	TPC outputs $TP_{15}$ to $TP_8$ are enabled (NDR15 to NDR8 are transferred to $PB_7$ to $PB_0$ )	

Note: \* Since this LSI does not have a  $TP_{14}$  pin, the  $TP_{14}$  signal cannot be output to the outside.

#### 9.2.9 TPC Output Control Register (TPCR)

TPCR is an 8-bit readable/writable register that selects output trigger signals for TPC outputs on a group-by-group basis.



Note: \* Since this LSI does not have a TP<sub>14</sub> pin, the TP<sub>14</sub> signal cannot be output to the outside.

TPCR is initialized to H'FF by a reset and in hardware standby mode. It is not initialized in software standby mode.

**Bits 7 and 6—Group 3 Compare Match Select 1 and 0 (G3CMS1, G3CMS0):** These bits select the compare match event that triggers TPC output group 3  $(TP_{15} \text{ to } TP_{12})^*$ .

Bit 7 G3CMS1	Bit6 G3CMS0	Description
0	0	TPC output group 3 (TP $_{15}$ to TP $_{12}$ )* is triggered by compare match in ITU channel 0
	1	TPC output group 3 $(TP_{15} \text{ to } TP_{12})^*$ is triggered by compare match in ITU channel 1
1	0	TPC output group 3 (TP <sub>15</sub> to TP <sub>12</sub> )* is triggered by compare match in ITU channel 2
	1	TPC output group 3 $(TP_{15} \text{ to } TP_{12})^*$ is triggered by compare match in ITU channel 3 (Initial value)
Note: *	Since this LSI doe	es not have a $TP_{14}$ pin, the $TP_{14}$ signal cannot be output off-chip.

Bits 5 and 4—Group 2 Compare Match Select 1 and 0 (G2CMS1, G2CMS0): These bits
select the compare match event that triggers TPC output group 2 ( $TP_{11}$ to $TP_8$ ).

Bit 5 G2CMS1	Bit4 G2CMS0	Description
0	0	TPC output group 2 (TP $_{11}$ to TP $_{8}$ ) is triggered by compare match in ITU channel 0
	1	TPC output group 2 (TP <sub>11</sub> to TP <sub>8</sub> ) is triggered by compare match in ITU channel 1
1	0	TPC output group 2 (TP <sub>11</sub> to TP <sub>8</sub> ) is triggered by compare match in ITU channel 2
	1	TPC output group 2 (TP <sub>11</sub> to TP <sub>8</sub> ) is triggered by compare match in ITU channel 3 (Initial value)

Bits 3 and 2—Group 1 Compare Match Select 1 and 0 (G1CMS1, G1CMS0): These bits	
select the compare match event that triggers TPC output group 1 (TP <sub>7</sub> to TP <sub>4</sub> ).	

Bit 3 G1CMS1	Bit2 G1CMS0	Description
0	0	TPC output group 1 (TP $_7$ to TP $_4$ ) is triggered by compare match in ITU channel 0
	1	TPC output group 1 (TP $_7$ to TP $_4$ ) is triggered by compare match in ITU channel 1
1	0	TPC output group 1 (TP $_7$ to TP $_4$ ) is triggered by compare match in ITU channel 2
	1	TPC output group 1 (TP $_7$ to TP $_4$ ) is triggered by compare match in ITU channel 3 (Initial value)

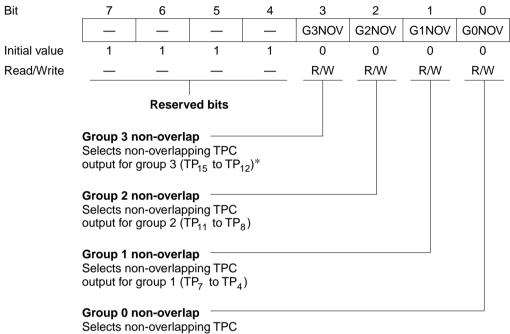
### Bits 1 and 0—Group 0 Compare Match Select 1 and 0 (G0CMS1, G0CMS0): These bits

select the compare match event	that triggers TPC output	group 0 ( $TP_3$ to $TP_0$ ).
--------------------------------	--------------------------	-------------------------------

Bit1 G0CMS1	Bit0 G0CMS0	Description
0	0	TPC output group 0 (TP $_{3}$ to TP $_{0}$ ) is triggered by compare match in ITU channel 0
	1	TPC output group 0 (TP $_3$ to TP $_0$ ) is triggered by compare match in ITU channel 1
1	0	TPC output group 0 (TP $_{3}$ to TP $_{0}$ ) is triggered by compare match in ITU channel 2
	1	TPC output group 0 (TP $_3$ to TP $_0$ ) is triggered by compare match in ITU channel 3 (Initial value)

#### 9.2.10 TPC Output Mode Register (TPMR)

TPMR is an 8-bit readable/writable register that selects normal or non-overlapping TPC output for each group.



output for group 0 ( $TP_3$  to  $TP_0$ )

Note: \* Since this LSI does not have a TP<sub>14</sub> pin, the TP<sub>14</sub> signal cannot be output to the outside.

The output trigger period of a non-overlapping TPC output waveform is set in general register B (GRB) in the ITU channel selected for output triggering. The non-overlap margin is set in general register A (GRA). The output values change at compare match A and B. For details see section 9.3.4, Non-Overlapping TPC Output.

TPMR is initialized to H'F0 by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 to 4—Reserved: These bits cannot be modified and are always read as 1.

**Bit 3—Group 3 Non-Overlap (G3NOV):** Selects normal or non-overlapping TPC output for group 3  $(TP_{15} \text{ to } TP_{12})^*$ .

Note: \* Since this LSI does not have a  $TP_{14}$  pin, the  $TP_{14}$  signal cannot be output off-chip.

Bit 3 G3NOV	Description
0	Normal TPC output in group 3 (output values change at compare match A in the selected ITU channel) (Initial value)
1	Non-overlapping TPC output in group 3 (independent 1 and 0 output at compare match A and B in the selected ITU channel)

**Bit 2—Group 2 Non-Overlap (G2NOV):** Selects normal or non-overlapping TPC output for group 2 ( $TP_{11}$  to  $TP_{8}$ ).

Bit 2 G2NOV	Description
0	Normal TPC output in group 2 (output values change at compare match A in the selected ITU channel) (Initial value)
1	Non-overlapping TPC output in group 2 (independent 1 and 0 output at compare match A and B in the selected ITU channel)

**Bit 1—Group 1 Non-Overlap (G1NOV):** Selects normal or non-overlapping TPC output for group 1 ( $TP_7$  to  $TP_4$ ).

# Bit 1 G1NOV Description 0 Normal TPC output in group 1 (output values change at compare match A in the selected ITU channel) 1 Non-overlapping TPC output in group 1 (independent 1 and 0 output at compare match A and B in the selected ITU channel)

#### Renesas

**Bit 0—Group 0 Non-Overlap (G0NOV):** Selects normal or non-overlapping TPC output for group 0 ( $TP_3$  to  $TP_0$ ).

Bit 0 G0NOV	Description
0	Normal TPC output in group 0 (output values change at compare match A in the selected ITU channel) (Initial value)
1	Non-overlapping TPC output in group 0 (independent 1 and 0 output at compare match A and B in the selected ITU channel)

#### 9.3 Operation

#### 9.3.1 Overview

When corresponding bits in PADDR or PBDDR and NDERA or NDERB are set to 1, TPC output is enabled. The TPC output initially consists of the corresponding PADR or PBDR contents. When a compare-match event selected in TPCR occurs, the corresponding NDRA or NDRB bit contents are transferred to PADR or PBDR to update the output values.

Figure 9.2 illustrates the TPC output operation. Table 9.3 summarizes the TPC operating conditions.

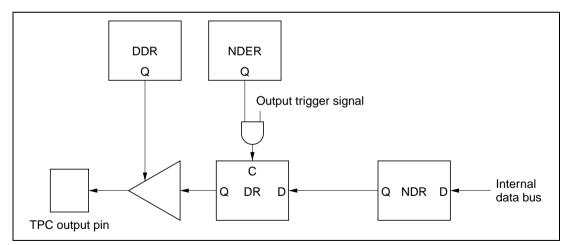


Figure 9.2 TPC Output Operation

NDER	DDR	Pin Function
0	0	Generic input port
	1	Generic output port
1	0	Generic input port (but the DR bit is a read-only bit, and when compare match occurs, the NDR bit value is transferred to the DR bit)
	1	TPC pulse output

Table 9.5 IFC Operating Conditions	Table 9.3	<b>TPC Operating</b>	Conditions
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Sequential output of up to 16-bit patterns is possible by writing new output data to NDRA and NDRB before the next compare match. For information on non-overlapping operation, see section 9.3.4, Non-Overlapping TPC Output.

#### 9.3.2 Output Timing

If TPC output is enabled, NDRA/NDRB contents are transferred to PADR/PBDR and output when the selected compare match event occurs. Figure 9.3 shows the timing of these operations for the case of normal output in groups 0 and 1, triggered by compare match A.

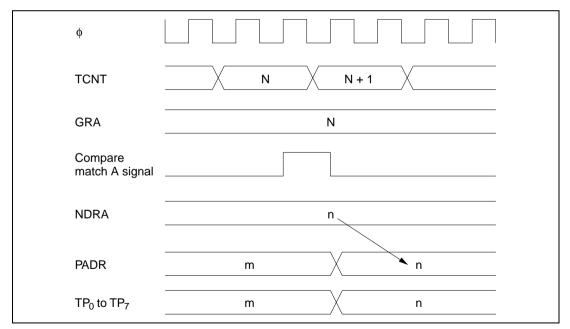
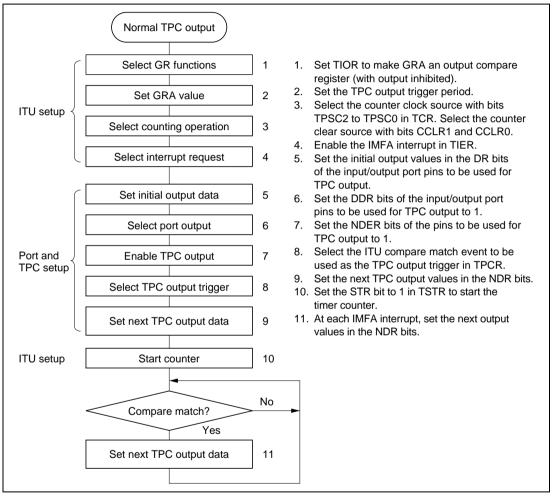


Figure 9.3 Timing of Transfer of Next Data Register Contents and Output (Example)

#### 9.3.3 Normal TPC Output

#### Sample Setup Procedure for Normal TPC Output

Figure 9.4 shows a sample procedure for setting up normal TPC output.





#### Example of Normal TPC Output (Example of Five-Phase Pulse Output)

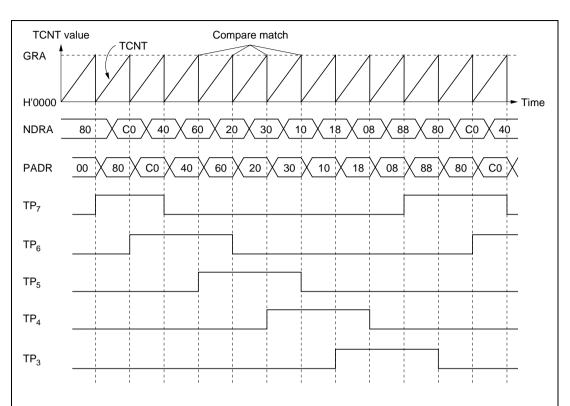


Figure 9.5 shows an example in which the TPC is used for cyclic five-phase pulse output.

- The ITU channel to be used as the output trigger channel is set up so that GRA is an output compare register and the counter will be cleared by compare match A. The trigger period is set in GRA. The IMIEA bit is set to 1 in TIER to enable the compare match A interrupt.
- H'F8 is written in PADDR and NDERA, and bits G1CMS1, G1CMS0, G0CMS1, and G0CMS0 are set in TPCR to select compare match in the ITU channel set up in step 1 as the output trigger. Output data H'80 is written in NDRA.
- The timer counter in this ITU channel is started. When compare match A occurs, the NDRA contents are transferred to PADR and output. The compare match/input capture A (IMFA) interrupt service routine writes the next output data (H'C0) in NDRA.
- Five-phase overlapping pulse output (one or two phases active at a time) can be obtained by writing H'40, H'60, H'20, H'30, H'10, H'18, H'08, H'88... at successive IMFA interrupts.

Figure 9.5 Normal TPC Output Example (Five-Phase Pulse Output)

#### 9.3.4 Non-Overlapping TPC Output

#### Sample Setup Procedure for Non-Overlapping TPC Output

Figure 9.6 shows a sample procedure for setting up non-overlapping TPC output.

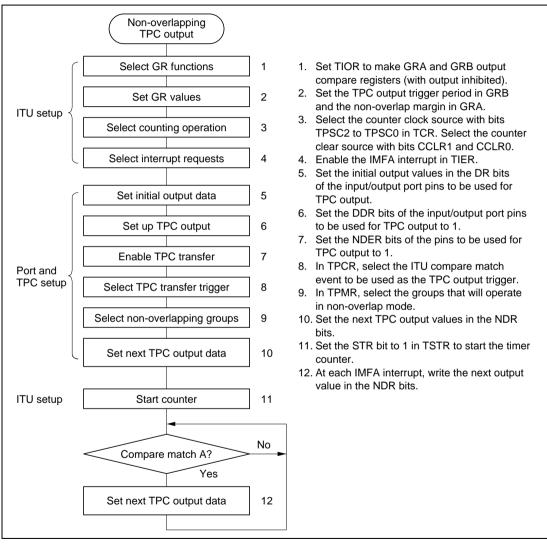
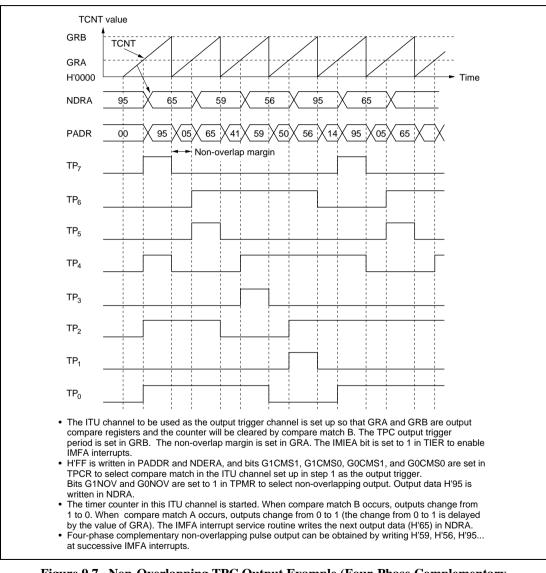


Figure 9.6 Setup Procedure for Non-Overlapping TPC Output (Example)

#### Example of Non-Overlapping TPC Output (Example of Four-Phase Complementary Non-Overlapping Output)

Figure 9.7 shows an example of the use of TPC output for four-phase complementary nonoverlapping pulse output.



# Figure 9.7 Non-Overlapping TPC Output Example (Four-Phase Complementary Non-Overlapping Pulse Output)

#### 9.3.5 TPC Output Triggering by Input Capture

TPC output can be triggered by ITU input capture as well as by compare match. If GRA functions as an input capture register in the ITU channel selected in TPCR, TPC output will be triggered by the input capture signal. Figure 9.8 shows the timing.

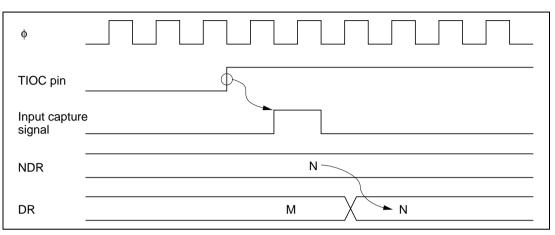


Figure 9.8 TPC Output Triggering by Input Capture (Example)

#### 9.4 Usage Notes

#### 9.4.1 Operation of TPC Output Pins

 $TP_0$  to  $TP_{15}^*$  are multiplexed with ITU pin functions. When ITU output is enabled, the corresponding pins cannot be used for TPC output. The data transfer from NDR bits to DR bits takes place, however, regardless of the usage of the pin.

Pin functions should be changed only under conditions in which the output trigger event will not occur.

Note: \* Since this LSI does not have a  $TP_{14}$  pin, the  $TP_{14}$  signal cannot be output to the outside.

#### 9.4.2 Note on Non-Overlapping Output

During non-overlapping operation, the transfer of NDR bit values to DR bits takes place as follows.

- 1. NDR bits are always transferred to DR bits at compare match A.
- 2. At compare match B, NDR bits are transferred only if their value is 0. Bits are not transferred if their value is 1.

Figure 9.9 illustrates the non-overlapping TPC output operation.

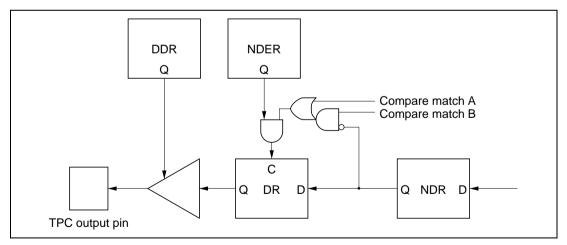


Figure 9.9 Non-Overlapping TPC Output

Therefore, 0 data can be transferred ahead of 1 data by making compare match B occur before compare match A. NDR contents should not be altered during the interval from compare match B to compare match A (the non-overlap margin).

This can be accomplished by having the IMFA interrupt service routine write the next data in NDR, or by having the IMFA interrupt activate the DMAC. The next data must be written before the next compare match B occurs.

Figure 9.10 shows the timing relationships.

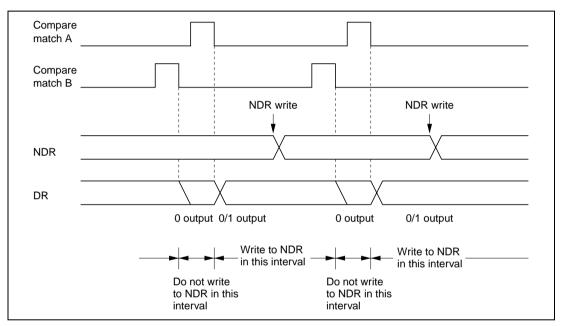


Figure 9.10 Non-Overlapping Operation and NDR Write Timing

## Section 10 Watchdog Timer

#### 10.1 Overview

The H8/3039 Group has an on-chip watchdog timer (WDT). The WDT has two selectable functions: it can operate as a watchdog timer to supervise system operation, or it can operate as an interval timer. As a watchdog timer, it generates a reset signal for the H8/3039 Group chip if a system crash allows the timer counter (TCNT) to overflow before being rewritten. In interval timer operation, an interval timer interrupt is requested at each TCNT overflow.

#### 10.1.1 Features

WDT features are listed below.

- Selection of eight counter clock sources φ/2, φ/32, φ/64, φ/128, φ/256, φ/512, φ/2048, or φ/4096
- Interval timer option
- Timer counter overflow generates a reset signal or interrupt.

The reset signal is generated in watchdog timer operation. An interval timer interrupt is generated in interval timer operation.

• Watchdog timer reset signal resets the entire H8/3039 Group chip internally, and can also be output externally.\*

The reset signal generated by timer counter overflow during watchdog timer operation resets the entire H8/3039 Group internally. An external reset signal can be output from the  $\overline{\text{RESO}}$  pin to reset other system devices simultaneously.

Note: \* The RESO pin of the mask ROM version is the dedicated FWE input pin of the F-ZTAT version. Therefore, the F-ZTAT version cannot output the reset signal to the outside.

#### Renesas

#### 10.1.2 Block Diagram

Figure 10.1 shows a block diagram of the WDT.

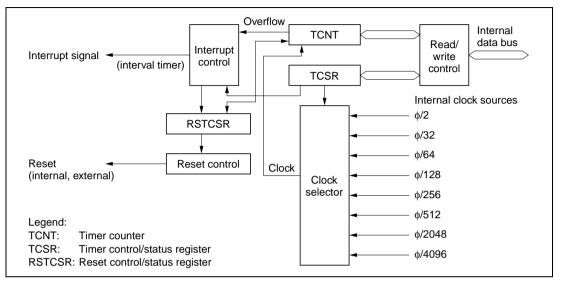


Figure 10.1 WDT Block Diagram

#### 10.1.3 Pin Configuration

Table 10.1 describes the WDT output pin.\*

Note: \* Shows the mask ROM version pin. The F-ZTAT does not have any pins used by the WDT. For F-ZTAT version, see section 15.9, Notes on Flash Memory Programming/Erasing.

#### Table 10.1 WDT Pin

Name	Abbreviation	I/O	Function		
Reset output	RESO	Output*	External output of the watchdog timer reset signal		
Note: * Open-drain output. Externally pull-up to Vcc whether or not the reset output is used					

#### 10.1.4 Register Configuration

Table 10.2 summarizes the WDT registers.

#### Table 10.2 WDT Registers

Address\*1

Write* <sup>2</sup>	Read	Name	Abbreviation	R/W	Initial Value
H'FFA8	H'FFA8	Timer control/status register	TCSR	R/(W)* <sup>3</sup>	H'18
	H'FFA9	Timer counter	TCNT	R/W	H'00
H'FFAA	H'FFAB	Reset control/status register	RSTCSR	R/(W)* <sup>3</sup>	H'3F

Notes: 1. Lower 16 bits of the address.

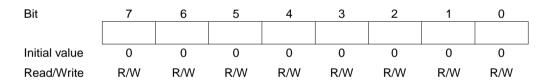
2. Write word data starting at this address.

3. Only 0 can be written in bit 7 to clear the flag.

#### **10.2** Register Descriptions

#### **10.2.1** Timer Counter (TCNT)

TCNT is an 8-bit readable and writable\* up-counter.



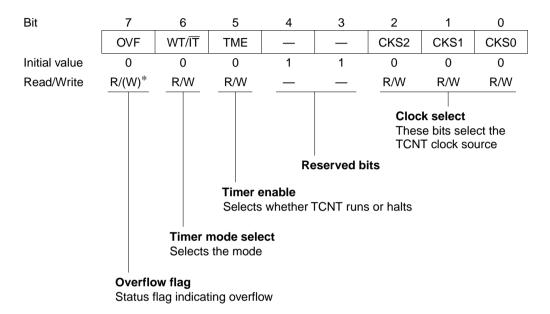
When the TME bit is set to 1 in TCSR, TCNT starts counting pulses generated from an internal clock source selected by bits CKS2 to CKS0 in TCSR. When the count overflows (changes from H'FF to H'00), the OVF bit is set to 1 in TCSR. TCNT is initialized to H'00 by a reset and when the TME bit is cleared to 0.

Note: \* TCNT is write-protected by a password. For details see section 10.2.4, Notes on Register Access.

#### 10.2.2 Timer Control/Status Register (TCSR)

TCSR is an 8-bit readable and writable\* register. Its functions include selecting the timer mode and clock source.

Note: \* TCSR is write-protected by a password. For details see section 10.2.4, Notes on Register Access.



Note: \* Only 0 can be written to clear the flag.

Bits 7 to 5 are initialized to 0 by a reset and in standby mode. Bits 2 to 0 are initialized to 0 by a reset. In software standby mode bits 2 to 0 are not initialized, but retain their previous values.

**Bit 7—Overflow Flag (OVF):** This status flag indicates that the timer counter has overflowed from H'FF to H'00.

Bit 7 OVF	Description	
0	[Clearing condition]	
	Cleared by reading OVF when OVF = 1, then writing 0 in OVF	(Initial value)
1	[Setting condition]	
	Set when TCNT changes from H'FF to H'00	

**Bit 6—Timer Mode Select (WT/IT):** Selects whether to use the WDT as a watchdog timer or interval timer. If used as an interval timer, the WDT generates an interval timer interrupt request when TCNT overflows. If used as a watchdog timer, the WDT generates a reset signal when TCNT overflows.

Bit 6 WT/IT	Description	
0	Interval timer: requests interval timer interrupts	(Initial value)
1	Watchdog timer: generates a reset signal	

Bit 5—Timer Enable (TME): Selects whether TCNT runs or is halted.

Bit 5 TME	Description	
0	TCNT is initialized to H'00 and halted	(Initial value)
1	TCNT is counting	

Bits 4 and 3—Reserved: These bits cannot be modified and are always read as 1.

Bits 2 to 0—Clock Select 2 to 0 (CKS2/1/0): These bits select one of eight internal clock sources, obtained by prescaling the system clock ( $\phi$ ), for input to TCNT.

Bit 2 CKS2	Bit 1 CKS1	Bit 0 CKS0	Description	
0	0	0	ф/2	(Initial value)
		1	ф/32	
	1	0	ф/64	
		1	ф/128	
1	0	0	ф/256	
		1	φ/512	
	1	0	ф/2048	
		1	ф/4096	

#### Section 10 Watchdog Timer

#### 10.2.3 Reset Control/Status Register (RSTCSR)

RSTCSR is an 8-bit readable and writable\* register that indicates when a reset signal has been generated by watchdog timer overflow, and controls external output of the reset signal.

Note: \* RSTCSR is write-protected by a password. For details see section 10.2.4, Notes on Register Access.

Bit	7	6	5	4	3	2	1	0
	WRST	RSTOE	—	—	—	—	—	—
Initial value	0	0	1	1	1	1	1	1
Read/Write	R/(W)*1	R/W	—		—	—	—	—
					Reserv	ed bits		
		eset outpunables or d			put of the	reset signa	al	
	atchdog ti dicates tha			been gene	erated			

- Notes: 1. Only 0 can be written in bit 7 to clear the flag.
  - 2. With the mask ROM version, enable and disable can be set. With the F-ZTAT version, do not set enable.

Bits 7 and 6 are initialized by input of a reset signal at the  $\overline{\text{RES}}$  pin. They are not initialized by reset signals generated by watchdog timer overflow.

**Bit 7—Watchdog Timer Reset (WRST):** During watchdog timer operation, this bit indicates that TCNT has overflowed and generated a reset signal. This reset signal resets the entire chip internally. If bit RSTOE is set to 1, this reset signal is also output (low) at the  $\overline{\text{RESO}}$  pin\*<sup>1</sup> to initialize external system devices.

Bit 7 WRST	Description	
0	[Clearing conditions]	(Initial value)
	(1) Cleared to 0 by reset signal input at $\overline{\text{RES}}$ pin	
	(2) Cleared by reading WRST when WRST = 1, then writing 0 in WER	RST
1	[Setting condition]	
	Set when TCNT overflow generates a reset signal during watchdog tir	mer operation

**Bit 6—Reset Output Enable (RSTOE):** Enables or disables external output at the  $\overline{\text{RESO}}$  pin<sup>\*1</sup> of the reset signal generated if TCNT overflows during watchdog timer operation.

Bit 6 RSTO	Description	
0	Reset signal is not output externally	(Initial value)
1	Reset signal is output externally* <sup>2</sup>	
Notes:	1. Mask ROM version. Dedicated FWE input pin for F-ZTAT version.	
	2. Mask ROM version. Do not set to 1 with the F-ZTAT version.	

Bits 5 to 0—Reserved: These bits cannot be modified and are always read as 1.

#### Renesas

#### 10.2.4 Notes on Register Access

The watchdog timer's TCNT, TCSR, and RSTCSR registers differ from other registers in being more difficult to write. The procedures for writing and reading these registers are given below.

#### Writing to TCNT and TCSR

These registers must be written by a word transfer instruction. They cannot be written by byte instructions. Figure 10.2 shows the format of data written to TCNT and TCSR. TCNT and TCSR both have the same write address. The write data must be contained in the lower byte of the written word. The upper byte must contain H'5A (password for TCNT) or H'A5 (password for TCSR). This transfers the write data from the lower byte to TCNT or TCSR.

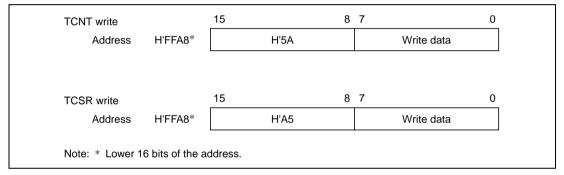
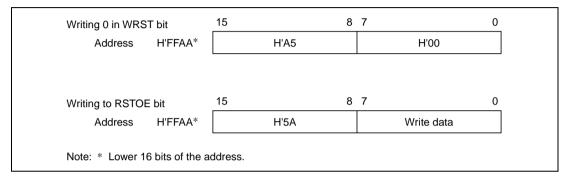


Figure 10.2 Format of Data Written to TCNT and TCSR

#### Writing to RSTCSR

RSTCSR must be written by a word transfer instruction. It cannot be written by byte transfer instructions. Figure 10.3 shows the format of data written to RSTCSR. To write 0 in the WRST bit, the write data must have H'A5 in the upper byte and H'00 in the lower byte. The H'00 in the lower byte clears the WRST bit in RSTCSR to 0. To write to the RSTOE bit, the upper byte must contain H'5A and the lower byte must contain the write data. Writing this word transfers a write data value into the RSTOE bit.





#### Reading TCNT, TCSR, and RSTCSR

These registers are read like other registers. Byte access instructions can be used. The read addresses are H'FFA8 for TCSR, H'FFA9 for TCNT, and H'FFAB for RSTCSR, as listed in table 10.3.

#### Table 10.3 Read Addresses of TCNT, TCSR, and RSTCSR

Address*	Register	
H'FFA8	TCSR	
H'FFA9	TCNT	
H'FFAB	RSTCSR	
Nata: V Laura 44		

Note: \* Lower 16 bits of the address.

#### Renesas

#### 10.3 Operation

Operations when the WDT is used as a watchdog timer and as an interval timer are described below.

#### 10.3.1 Watchdog Timer Operation

Figure 10.4 illustrates watchdog timer operation. To use the WDT as a watchdog timer, set the WT/IT and TME bits to 1 in TCSR. Software must prevent TCNT overflow by rewriting the TCNT value (normally by writing H'00) before overflow occurs. If TCNT fails to be rewritten and overflows due to a system crash etc., the H8/3039 Group is internally reset for a duration of 518 states.

The watchdog reset signal can be externally output from the  $\overline{\text{RESO}}$  pin\* to reset external system devices. The reset signal is output externally for 132 states. External output can be enabled or disabled by the RSTOE bit in RSTCSR.

A watchdog reset has the same vector as a reset generated by input at the  $\overline{\text{RES}}$  pin. Software can distinguish a  $\overline{\text{RES}}$  reset from a watchdog reset by checking the WRST bit in RSTCSR.

If a  $\overline{\text{RES}}$  reset and a watchdog reset occur simultaneously, the  $\overline{\text{RES}}$  reset takes priority.

Note: \* Mask ROM version.

Since the RES pin is a dedicated FWE input pin with the F-ZTAT version, the reset signal cannot be output to the outside.

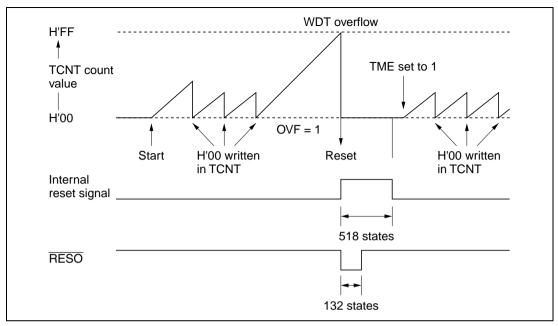
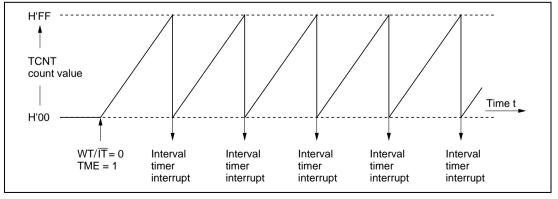
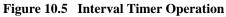


Figure 10.4 Watchdog Timer Operation (Mask ROM Version)

#### 10.3.2 Interval Timer Operation

Figure 10.5 illustrates interval timer operation. To use the WDT as an interval timer, clear bit  $WT/\overline{IT}$  to 0 and set bit TME to 1 in TCSR. An interval timer interrupt request is generated at each TCNT overflow. This function can be used to generate interval timer interrupts at regular intervals.





#### 10.3.3 Timing of Setting of Overflow Flag (OVF)

Figure 10.6 shows the timing of setting of the OVF flag in TCSR. The OVF flag is set to 1 when TCNT overflows. At the same time, a reset signal is generated in watchdog timer operation, or an interval timer interrupt is generated in interval timer operation.

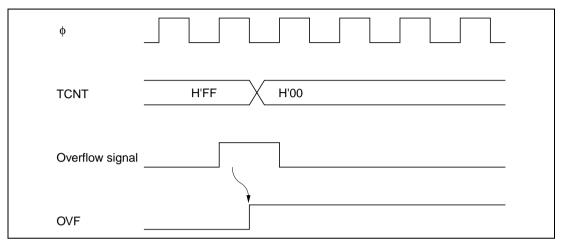
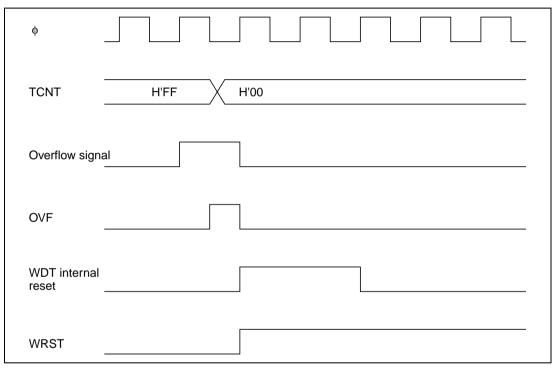


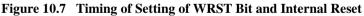
Figure 10.6 Timing of Setting of OVF

#### 10.3.4 Timing of Setting of Watchdog Timer Reset Bit (WRST)

The WRST bit in RSTCSR is valid when bits  $WT/\overline{IT}$  and TME are both set to 1 in TCSR.

Figure 10.7 shows the timing of setting of WRST and the internal reset timing. The WRST bit is set to 1 when TCNT overflows and OVF is set to 1. At the same time an internal reset signal is generated for the entire H8/3039 Group chip. This internal reset signal clears OVF to 0, but the WRST bit remains set to 1. The reset routine must therefore clear the WRST bit.





#### **10.4** Interrupts

During interval timer operation, an overflow generates an interval timer interrupt (WOVI). The interval timer interrupt is requested whenever the OVF bit is set to 1 in TCSR.

#### 10.5 Usage Notes

#### **Contention between TCNT Write and Increment**

If a timer counter clock pulse is generated during the T3 state of a write cycle to TCNT, the write takes priority and the timer count is not incremented. See figure 10.8.

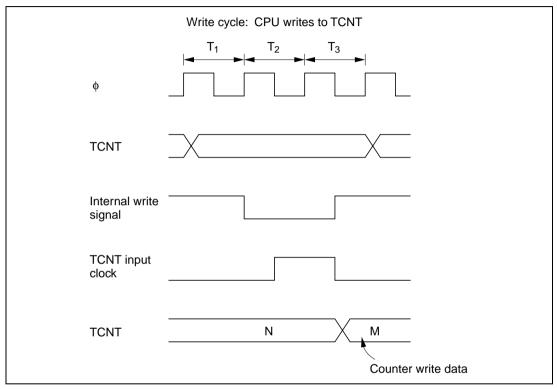


Figure 10.8 Contention between TCNT Write and Increment

#### Changing CKS2 to CKS0 Values

Halt TCNT by clearing the TME bit to 0 in TCSR before changing the values of bits CKS2 to CKS0.

## Section 11 Serial Communication Interface

#### 11.1 Overview

The H8/3039 Group has a serial communication interface (SCI) with two independent channels. The two channels are functionally identical. The SCI can communicate in asynchronous or synchronous mode. It also has a multiprocessor communication function for serial communication among two or more processors.

When the SCI is not used, it can be halted to conserve power. Each SCI channel can be halted independently. For details see section 17.6, Module Standby Function.

Channel 0 (SCI0) also has a smart card interface function conforming to the ISO/IEC7816-3 (Identification Card) standard. This function supports serial communication with a smart card. For details, see section 12, Smart Card Interface.

#### 11.1.1 Features

SCI features are listed below.

- Selection of asynchronous or synchronous mode for serial communication
- a. Asynchronous mode

Serial data communication is synchronized one character at a time. The SCI can communicate with a universal asynchronous receiver/transmitter (UART), asynchronous communication interface adapter (ACIA), or other chip that employs standard asynchronous serial communication. It can also communicate with two or more other processors using the multiprocessor communication function. There are twelve selectable serial data communication formats.

- Data length: 7 or 8 bits
- Stop bit length: 1 or 2 bits
- Parity bit: even, odd, or none
- Multiprocessor bit: 1 or 0
- Receive error detection: parity, overrun, and framing errors
- Break detection: by reading the RxD level directly when a framing error occurs

#### Renesas

b. Synchronous mode

Serial data communication is synchronized with a clock signal. The SCI can communicate with other chips having a synchronous communication function. There is one serial data communication format.

- Data length: 8 bits
- Receive error detection: overrun errors
- Full-duplex communication

The transmitting and receiving sections are independent, so the SCI can transmit and receive simultaneously. The transmitting and receiving sections are both double-buffered, so serial data can be transmitted and received continuously.

- Built-in baud rate generator with selectable bit rates
- Selectable transmit/receive clock sources: internal clock from baud rate generator, or external clock from the SCK pin.
- Four types of interrupts

Transmit-data-empty, transmit-end, receive-data-full, and receive-error interrupts are requested independently.

#### 11.1.2 Block Diagram

Figure 11.1 shows a block diagram of the SCI.

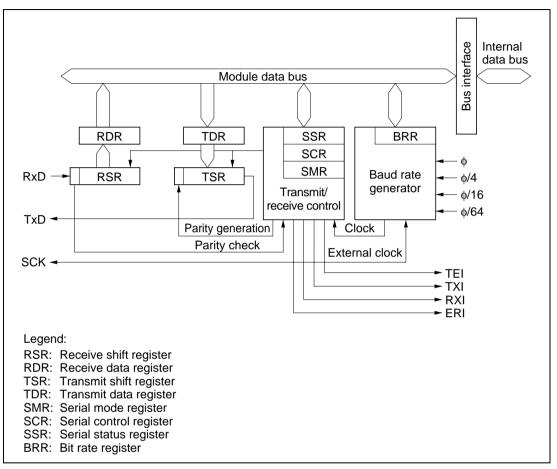


Figure 11.1 SCI Block Diagram

#### 11.1.3 Input/Output Pins

The SCI has the serial pins for each channel as listed in table 11.1.

#### Table 11.1 SCI Pins

Channel	Name	Abbreviation	I/O	Function
0	Serial clock pin	SCK	Input/output	SCI <sub>0</sub> clock input/output
	Receive data pin	RxD <sub>0</sub>	Input	$SCI_{0}$ receive data input
	Transmit data pin	TxD <sub>0</sub>	Output	SCI <sub>0</sub> transmit data output
1	Serial clock pin	SCK,	Input/output	SCI1 clock input/output
	Receive data pin	RxD <sub>1</sub>	Input	SCI, receive data input
	Transmit data pin	TxD <sub>1</sub>	Output	SCI, transmit data output

#### 11.1.4 Register Configuration

The SCI has the internal registers as listed in table 11.2. These registers select asynchronous or synchronous mode, specify the data format and bit rate, and control the transmitter and receiver sections.

Channel	Address*1	Name	Abbreviation	R/W	Initial Value
0	H'FFB0	Serial mode register	SMR	R/W	H'00
	H'FFB1	Bit rate register	BRR	R/W	H'FF
	H'FFB2	Serial control register	SCR	R/W	H'00
	H'FFB3	Transmit data register	TDR	R/W	H'FF
	H'FFB4	Serial status register	SSR	R/(W)* <sup>2</sup>	H'84
	H'FFB5	Receive data register	RDR	R	H'00
1	H'FFB8	Serial mode register	SMR	R/W	H'00
	H'FFB9	Bit rate register	BRR	R/W	H'FF
	H'FFBA	Serial control register	SCR	R/W	H'00
	H'FFBB	Transmit data register	TDR	R/W	H'FF
	H'FFBC	Serial status register	SSR	R/(W)* <sup>2</sup>	H'84
	H'FFBD	Receive data register	RDR	R	H'00

#### Table 11.2 Registers

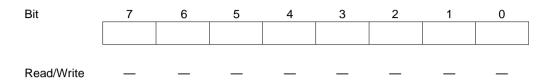
Notes: 1. Lower 16 bits of the address.

2. Only 0 can be written to clear flags.

#### **11.2** Register Descriptions

#### 11.2.1 Receive Shift Register (RSR)

RSR is an 8-bit register that receives serial data.



The SCI loads serial data input at the RxD pin into RSR in the order received, LSB (bit 0) first, thereby converting the data to parallel data. When 1 byte has been received, it is automatically transferred to RDR. The CPU cannot read or write RSR directly.

#### 11.2.2 Receive Data Register (RDR)

RDR is an 8-bit register that stores received serial data.

Bit	7	6	5	4	3	2	1	0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R

When the SCI finishes receiving 1 byte of serial data, it transfers the received data from RSR into RDR for storage. RSR is then ready to receive the next data. This double buffering allows data to be received continuously.

RDR is a read-only register. Its contents cannot be modified by the CPU. RDR is initialized to H'00 by a reset and in standby mode.

### 11.2.3 Transmit Shift Register (TSR)

TSR is an 8-bit register used to transmit serial data.

Bit	7	6	5	4	3	2	1	0
Read/Write			_	_		_	_	

The SCI loads transmit data from TDR into TSR, then transmits the data serially from the TxD pin, LSB (bit 0) first. After transmitting one data byte, the SCI automatically loads the next transmit data from TDR into TSR and starts transmitting it. If the TDRE flag is set to 1 in SSR, however, the SCI does not load the TDR contents into TSR. The CPU cannot read or write TSR directly.

### 11.2.4 Transmit Data Register (TDR)

TDR is an 8-bit register that stores data for serial transmission.

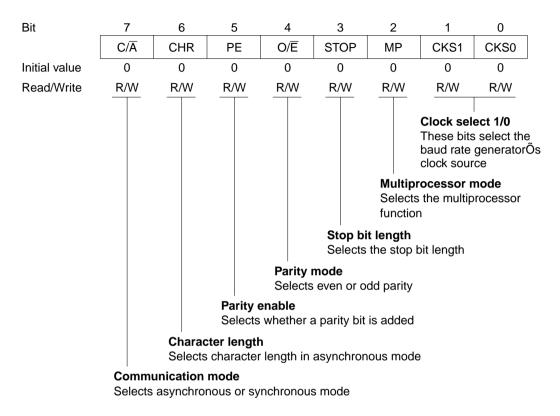
Bit	7	6	5	4	3	2	1	0
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W							

When the SCI detects that TSR is empty, it moves transmit data written in TDR from TDR into TSR and starts serial transmission. Continuous serial transmission is possible by writing the next transmit data in TDR during serial transmission from TSR.

The CPU can always read and write TDR. TDR is initialized to H'FF by a reset and in standby mode.

## 11.2.5 Serial Mode Register (SMR)

SMR is an 8-bit register that specifies the SCI serial communication format and selects the clock source for the baud rate generator.



The CPU can always read and write SMR. SMR is initialized to H'00 by a reset and in standby mode.

Bit 7—Communication Mode  $(C/\overline{A})$ : Selects whether the SCI operates in asynchronous or synchronous mode.

Bit 7 C/Ā	Description	
0	Asynchronous mode	(Initial value)
1	Synchronous mode	

**Bit 6—Character Length (CHR):** Selects 7-bit or 8-bit data length in asynchronous mode. In synchronous mode the data length is 8 bits regardless of the CHR setting.

Bit 6 CHR		Description	
0		8-bit data	(Initial value)
1		7-bit data*	
Note:	*	When 7-bit data is selected, the MSB (bit 7) in TDR is not transmitted.	

**Bit 5—Parity Enable (PE):** In asynchronous mode, this bit enables or disables the addition of a parity bit to transmit data, and the checking of the parity bit in receive data. In synchronous mode

parity bit to transmit data, and the checking of the parity bit in receive data. In synchronous mode the parity bit is neither added nor checked, regardless of the PE setting.

Bit 5 PE		Description	
0		Parity bit not added or checked	(Initial value)
1		Parity bit added and checked*	
Note:	*	When PE is set to 1, an even or odd parity bit is added to trar even or odd parity mode selected by the $O/\overline{E}$ bit, and the pari checked to see that it matches the even or odd mode selecte	ity bit in receive data is

**Bit 4—Parity Mode (O/Ē):** Selects even or odd parity. The  $O/\overline{E}$  bit setting is valid in asynchronous mode when the PE bit is set to 1 to enable the adding and checking of a parity bit. The  $O/\overline{E}$  setting is ignored in synchronous mode, or when parity adding and checking is disabled in asynchronous mode.

Bit 4 O/E		Description
0		Even parity*1 (Initial value)
1		Odd parity* <sup>2</sup>
Notes:	1.	When even parity is selected, the parity bit added to transmit data makes an even number of 1s in the transmitted character and parity bit combined. Receive data must have an even number of 1s in the received character and parity bit combined.
	2.	When odd parity is selected, the parity bit added to transmit data makes an odd number of 1s in the transmitted character and parity bit combined. Receive data must have an

odd number of 1s in the received character and parity bit combined.

**Bit 3—Stop Bit Length (STOP):** Selects one or two stop bits in asynchronous mode. This setting is used only in asynchronous mode. In synchronous mode no stop bit is added, so the STOP bit setting is ignored.

## Renesas

Bit 3 STOP	Description				
0	One stop bit*1	(Initial value)			
1	Two stop bits* <sup>2</sup>				
Notes:	es: 1. One stop bit (with value 1) is added at the end of each transmitted character.				
	2 Two stop hits (with value 1) are added at the	and of each transmitted character			

2. I wo stop bits (with value 1) are added at the end of each transmitted character.

In receiving, only the first stop bit is checked, regardless of the STOP bit setting. If the second stop bit is 1 it is treated as a stop bit. If the second stop bit is 0 it is treated as the start bit of the next incoming character.

Bit 2-Multiprocessor Mode (MP): Selects a multiprocessor format. When a multiprocessor format is selected, parity settings made by the PE and  $O\overline{E}$  bits are ignored. The MP bit setting is valid only in asynchronous mode. It is ignored in synchronous mode.

For further information on the multiprocessor communication function, see section 11.3.3, Multiprocessor Communication.

Bit 2 MP	Description				
0	Multiprocessor function disabled	(Initial value)			
1	Multiprocessor format selected				

Bits 1 and 0—Clock Select 1 and 0 (CKS1/0): These bits select the clock source of the on-chip baud rate generator. Four clock sources are available:  $\phi$ ,  $\phi/4$ ,  $\phi/16$ , and  $\phi/64$ .

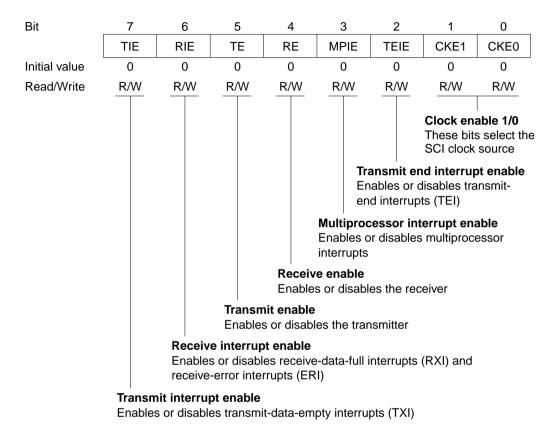
For the relationship between the clock source, bit rate register setting, and baud rate, see section 11.2.8, Bit Rate Register (BRR).

Bit 1 CKS1	Bit 0 CKS0	Description	
0	0	$\phi$ clock selected	(Initial value)
0	1	φ/4 clock selected	
1	0	¢/16 clock selected	
1	1	φ/64 clock selected	

Section 11 Serial Communication Interface

### 11.2.6 Serial Control Register (SCR)

SCR enables the SCI transmitter and receiver, enables or disables serial clock output in asynchronous mode, enables or disables interrupts, and selects the transmit/receive clock source.



The CPU can always read and write SCR. SCR is initialized to H'00 by a reset and in standby mode.

## Renesas

**Bit 7—Transmit Interrupt Enable (TIE):** Enables or disables the transmit-data-empty interrupt (TXI) requested when the TDRE flag in SSR is set to 1 due to transfer of serial transmit data from TDR to TSR.

## Bit 7

TIE	Description		
0	Transmit-data-empty interrupt request (TXI) is disabled* (Initial value)		
1	Transmit-data-empty interrupt request (TXI) is enabled		
Note:	TXI interrupt requests can be cleared by reading the value 1 from the TDRE flag, the clearing it to 0; or by clearing the TIE bit to 0.		

**Bit 6—Receive Interrupt Enable (RIE):** Enables or disables the receive-data-full interrupt (RXI) requested when the RDRF flag is set to 1 in SSR due to transfer of serial receive data from RSR to RDR; also enables or disables the receive-error interrupt (ERI).

Bit 6 RIE		Description	
0 Receive-end (RX		Receive-end (RXI) and receive-error (ERI) interrupt requests are disabled* (Initial value)	
1	Receive-end (RXI) and receive-error (ERI) interrupt requests are enabled		
Note:	* RXI and ERI interrupt requests can be cleared by reading the value 1 from the RDR FER, PER, or ORER flag, then clearing it to 0; or by clearing the RIE bit to 0.		

Bit 5—Transmit Enable (TE): Enables or disables the start of SCI serial transmitting operations.

Bit 5		
TE	Description	
0	Transmitting disabled*1	(Initial value)
1	Transmitting enabled* <sup>2</sup>	
Notos: 1	The TDPE flag is fixed at 1 in SSP	

Notes: 1. The TDRE flag is fixed at 1 in SSR.

 In the enabled state, serial transmitting starts when the TDRE flag in SSR is cleared to 0 after writing of transmit data into TDR. Select the transmit format in SMR before setting the TE bit to 1. Bit 4—Receive Enable (RE): Enables or disables the start of SCI serial receiving operations.

Bit 4 RE	Description	
0	Receiving disabled*1	(Initial value)
1	Receiving enabled* <sup>2</sup>	
Notes:	1. Clearing the RE bit to 0 does not affect the R	DRF. FER. PER. and ORER flags. These

Notes: 1. Clearing the RE bit to 0 does not affect the RDRF, FER, PER, and ORER flags. These flags retain their previous values.

 In the enabled state, serial receiving starts when a start bit is detected in asynchronous mode, or serial clock input is detected in synchronous mode. Select the receive format in SMR before setting the RE bit to 1.

**Bit 3—Multiprocessor Interrupt Enable (MPIE):** Enables or disables multiprocessor interrupts. The MPIE setting is valid only in asynchronous mode, and only if the MP bit is set to 1 in SMR. The MPIE setting is ignored in synchronous mode or when the MP bit is cleared to 0.

Bit 3 MPIE	Description		
0	Multiprocessor interrupts are disabled (normal receive operation)	(Initial value)	
	[Clearing conditions]		
	The MPIE bit is cleared to 0		
	• MPB = 1 in received data		
1	Multiprocessor interrupts are enabled*		
	Receive-data-full interrupts (RXI), receive-error interrupts (ERI), and setting of the RDRF, FER, and ORER status flags in SSR are disabled until data with the multiprocessor bit set to 1 is received.		
Note: *	* The SCI does not transfer receive data from RSR to RDR, does not detect receive errors, and does not set the RDRF, FER, and ORER flags in SSR. When it receives data in which MPB = 1, the SCI sets the MPB bit to 1 in SSR, automatically clears the MPIE bit to 0, and enables RXI and ERI interrupts (if the RIE bit is set to 1 in SCR) and setting of the FER and ORER flags.		

### Renesas

**Bit 2—Transmit-End Interrupt Enable (TEIE):** Enables or disables the transmit-end interrupt (TEI) requested if TDR does not contain new transmit data when the MSB is transmitted.

Bit 2 TEIE	Description	
0	Transmit-end interrupt requests (	EI) are disabled* (Initial value
1	Transmit-end interrupt requests (	EI) are enabled*
Note:		ed by reading the value 1 from the TDRE flag in to 0, thereby also clearing the TEND flag to 0; or b

**Bits 1 and 0—Clock Enable 1 and 0 (CKE1/0):** These bits select the SCI clock source and enable or disable clock output from the SCK pin. Depending on the settings of CKE1 and CKE0, the SCK pin can be used for generic input/output, serial clock output, or serial clock input.

The CKE0 setting is valid only in asynchronous mode, and only when the SCI is internally clocked (CKE1 = 0). The CKE0 setting is ignored in synchronous mode, or when an external clock source is selected (CKE1 = 1). After setting the CKE1 and CKE0 bits, select the SCI operating mode in SMR. For further details on selection of the SCI clock source, see table 11.9.

Bit 1 CKE1	Bit 0 CKE0	Description		
0	0	Asynchronous mode	Internal clock, SCK pin available for generic input/output*1	
		Synchronous mode	Internal clock, SCK pin used for serial clock output*1	
0	1	Asynchronous mode	Internal clock, SCK pin used for clock output* <sup>2</sup>	
		Synchronous mode	Internal clock, SCK pin used for serial clock output	
1	0	Asynchronous mode	External clock, SCK pin used for clock input*3	
		Synchronous mode	External clock, SCK pin used for serial clock input	
1	1	Asynchronous mode	External clock, SCK pin used for clock input* <sup>3</sup>	
		Synchronous mode	External clock, SCK pin used for serial clock input	

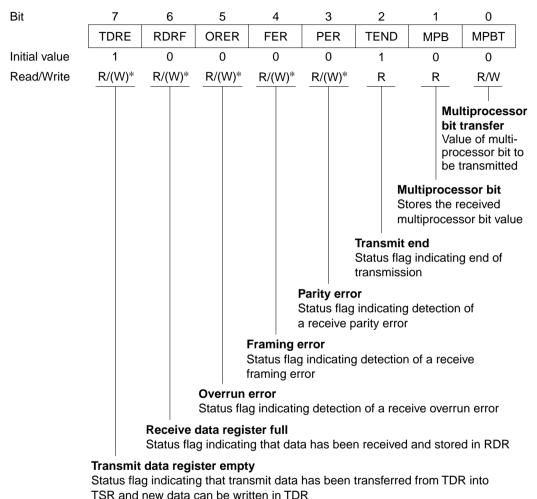
Notes: 1. Initial value

2. The output clock frequency is the same as the bit rate.

3. The input clock frequency is 16 times the bit rate.

### 11.2.7 Serial Status Register (SSR)

SSR is an 8-bit register containing multiprocessor bit values, and status flags that indicate the SCI operating status.



TSR and new data can be written in TL

Note: \* Only 0 can be written to clear the flag.

The CPU can always read and write SSR, but cannot write 1 in the TDRE, RDRF, ORER, PER, and FER flags. These flags can be cleared to 0 only if they have first been read while set to 1. The TEND and MPB flags are read-only bits that cannot be written.

SSR is initialized to H'84 by a reset and in standby mode.

**Bit 7—Transmit Data Register Empty (TDRE):** Indicates that the SCI has loaded transmit data from TDR into TSR and the next serial transmit data can be written in TDR.

Bit 7 TDRE	Description	
0	TDR contains valid transmit data	
	[Clearing condition]	
	Software reads TDRE while it is set to 1, then writes 0	
1	TDR does not contain valid transmit data	(Initial value)
	[Setting conditions]	
	The chip is reset or enters standby mode	
	The TE bit in SCR is cleared to 0	
	• TDR contents are loaded into TSR, so new data can be write	tten in TDR

#### Bit 6—Receive Data Register Full (RDRF): Indicates that RDR contains new receive data.

Bit 6 RDRF	Description		
0	RDR does not contain new receive data (Initial value)		
	[Clearing conditions]		
	The chip is reset or enters standby mode		
	• Software reads RDRF while it is set to 1, then writes 0		
	The DMAC reads data from RDR		
1	RDR contains new receive data		
	[Setting condition]		
	When serial data is received normally and transferred from RSR to RDR		

Note: The RDR contents and RDRF flag are not affected by detection of receive errors or by clearing of the RE bit to 0 in SCR. They retain their previous values. If the RDRF flag is still set to 1 when reception of the next data ends, an overrun error occurs and receive data is lost.

**Bit 5—Overrun Error (ORER):** Indicates that data reception ended abnormally due to an overrun error.

Bit 5 ORER	R Description	
0	Receiving is in progress or has ended normally	(Initial value)*1
	[Clearing conditions]	
	The chip is reset or enters standby mode	
	• Software reads ORER while it is set to 1, then writes 0	
1	A receive overrun error occurred* <sup>2</sup>	
	[Setting condition]	
	Reception of the next serial data ends when RDRF = 1	
Notes:	: 1. Clearing the RE bit to 0 in SCR does not affect the ORER flag, previous value.	which retains its
	<ol> <li>RDR continues to hold the receive data before the overrun error data is lost. Serial receiving cannot continue while the ORER fla</li> </ol>	, ,

data is lost. Serial receiving cannot continue while the ORER flag is set to 1. In synchronous mode, serial transmitting is also disabled.

**Bit 4—Framing Error (FER):** Indicates that data reception ended abnormally due to a framing error in asynchronous mode.

Bit 4	Description	
FER	Description	
0	Receiving is in progress or has ended normally	(Initial value)*1
	[Clearing conditions]	
	The chip is reset or enters standby mode	
	Software reads FER while it is set to 1, then writes 0	
1	A receive framing error occurred* <sup>2</sup>	
	[Setting condition]	
	The stop bit at the end of receive data is checked and found to be 0	
Notes: 1. Clearing the RE bit to 0 in SCR does not affect the FER flag, which retains its prvalue.		ains its previous
2.	When the stop bit length is 2 bits, only the first bit is checked. The second checked. When a framing error occurs the SCI transfers the receive date does not set the RDRF flag. Serial receiving cannot continue while the	ata into RDR but

to 1. In synchronous mode, serial transmitting is also disabled.

**Bit 3—Parity Error (PER):** Indicates that data reception ended abnormally due to a parity error in asynchronous mode.

Bit 3 PER		Description	
0		Receiving is in progress or has ended normally*1	(Initial value)
		[Clearing condition]	
		The chip is reset or enters standby mode. Software reads PER while it is writes 0	s set to 1, then
1		A receive parity error occurred* <sup>2</sup>	
		[Setting condition]	
		The number of 1s in receive data, including the parity bit, does not matc odd parity setting of O/ $\overline{E}$ in SMR	h the even or
Notes:	1.	Clearing the RE bit to 0 in SCR does not affect the PER flag, which retain value.	ns its previous
	2.	When a parity error occurs the SCI transfers the receive data into RDR b the RDRF flag. Serial receiving cannot continue while the PER flag is set	

**Bit 2—Transmit End (TEND):** Indicates that when the last bit of a serial character was transmitted TDR did not contain new transmit data, so transmission has ended. The TEND flag is a read-only bit and cannot be written.

synchronous mode, serial transmitting is also disabled.

Bit 2 TEND	Description		
0	Transmission is in progress		
	[Clearing condition]		
	Software reads TDRE while it is set to 1, then writes 0 in the TDRE flag		
1	End of transmission	(Initial value)	
	[Setting conditions]		
	The chip is reset or enters standby mode. The TE bit is cleared to 0 in SCR		
	<ul> <li>TDRE is 1 when the last bit of a serial character is transmitted</li> </ul>		

**Bit 1—Multiprocessor Bit (MPB):** Stores the value of the multiprocessor bit in receive data when a multiprocessor format is used in asynchronous mode. MPB is a read-only bit and cannot be written.

Bit 1 MPB	Description	
0	Multiprocessor bit value in receive data is 0*	(Initial value)
1	Multiprocessor bit value in receive data is 1	
Note: *	If the RE bit is cleared to 0 when a multiprocessor format is selected, previous value.	MPB retains its

**Bit 0—Multiprocessor Bit Transfer (MPBT):** Stores the value of the multiprocessor bit added to transmit data when a multiprocessor format is selected for transmitting in asynchronous mode. The MPBT setting is ignored in synchronous mode, when a multiprocessor format is not selected, or when the SCI is not transmitting.

Bit 0 MPBT	Description	
0	Multiprocessor bit value in transmit data is 0	(Initial value)
1	Multiprocessor bit value in transmit data is 1	

### 11.2.8 Bit Rate Register (BRR)

BRR is an 8-bit register that, together with the CKS1 and CKS0 bits in SMR that select the baud rate generator clock source, determines the serial communication bit rate.

Bit	7	6	5	4	3	2	1	0
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W							

The CPU can always read and write BRR. BRR is initialized to H'FF by a reset and in standby mode.

The baud rate generator is controlled separately for the individual channels, so different values may be set for each.

Table 11.3 shows examples of BRR settings in asynchronous mode. Table 11.4 shows examples of BRR settings in synchronous mode.

						φ (N	(Hz)					
		2	2		2.09	7152		2.4	576	3		
Bit Rate (bits/s)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)
110	1	141	0.03	1	148	-0.04	1	174	-0.26	1	212	0.03
150	1	103	0.16	1	108	0.21	1	127	0	1	155	0.16
300	0	207	0.16	0	217	0.21	0	255	0	1	77	0.16
600	0	103	0.16	0	108	0.21	0	127	0	0	155	0.16
1200	0	51	0.16	0	54	-0.70	0	63	0	0	77	0.16
2400	0	25	0.16	0	26	1.14	0	31	0	0	38	0.16
4800	0	12	0.16	0	13	-2.48	0	15	0	0	19	-2.34
9600	0	6	-6.99	0	6	-2.48	0	7	0	0	9	-2.34
19200	0	2	8.51	0	2	13.78	0	3	0	0	4	-2.34
31250	0	1	0	0	1	4.86	0	1	22.88	0	2	0
38400	0	1	-18.62	0	1	-14.67	0	1	0	_	_	_

 Table 11.3
 Examples of Bit Rates and BRR Settings in Asynchronous Mode

φ (MHz)

		3.6	864			4		4.91	52			5
Bit Rate (bits/s)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)
110	2	64	0.70	2	70	0.03	2	86	0.31	2	88	-0.25
150	1	191	0	1	207	0.16	1	255	0	2	64	0.16
300	1	95	0	1	103	0.16	1	127	0	1	129	0.16
600	0	191	0	0	207	0.16	0	255	0	1	64	0.16
1200	0	95	0	0	103	0.16	0	127	0	0	129	0.16
2400	0	47	0	0	51	0.16	0	63	0	0	64	0.16
4800	0	23	0	0	25	0.16	0	31	0	0	32	-1.36
9600	0	11	0	0	12	0.16	0	15	0	0	15	1.73
19200	0	5	0	0	6	-6.99	0	7	0	0	7	1.73
31250	—		_	0	3	0	0	4	-1.70	0	4	0
38400	0	2	0	0	2	8.51	0	3	0	0	3	1.73

						φ(Ιν	I <b>H</b> Z)					
			6		6.1	44		7.37	28			8
Bit Rate (bits/s)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)
110	2	106	-0.44	2	108	0.08	2	130	-0.07	2	141	0.03
150	1	77	0.16	2	79	0	2	95	0	2	103	0.16
300	1	155	0.16	1	159	0	1	191	0	1	207	0.16
600	1	77	0.16	1	79	0	1	95	0	1	103	0.16
1200	0	155	0.16	0	159	0	0	191	0	0	207	0.16
2400	0	77	0.16	0	79	0	0	95	0	0	103	0.16
4800	0	38	0.16	0	39	0	0	47	0	0	51	0.16
9600	0	19	-2.34	0	19	0	0	23	0	0	25	0.16
19200	0	9	-2.34	0	9	0	0	11	0	0	12	0.16
31250	0	5	0	0	5	2.40	0	6	5.33	0	7	0
38400	0	4	-2.34	0	4	0	0	5	0	0	6	-6.99

φ (MHz)

## φ (MHz)

						Ψ ("						
		9.8	304		1	10		12	2		12	.288
Bit Rate (bits/s)	n	N	Error (%)									
110	2	174	-0.26	2	177	-0.25	2	212	0.03	2	217	0.08
150	2	127	0	2	129	0.16	2	155	0.16	2	159	0
300	1	255	0	2	64	0.16	2	77	0.16	2	79	0
600	1	127	0	1	129	0.16	1	155	0.16	1	159	0
1200	0	255	0	1	64	0.16	1	77	0.16	1	79	0
2400	0	127	0	0	129	0.16	0	155	0.16	0	159	0
4800	0	63	0	0	64	0.16	0	77	0.16	0	79	0
9600	0	31	0	0	32	-1.36	0	38	0.16	0	39	0
19200	0	15	0	0	15	1.73	0	19	-2.34	0	19	0
31250	0	9	-1.70	0	9	0	0	11	0	0	11	2.40
38400	0	7	0	0	7	1.73	0	9	-2.34	0	9	0

						φ (N	/Hz)						
		1	4		14.7	7456	16				18		
Bit Rate (bits/s)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	
110	2	248	-0.17	3	64	0.70	3	70	0.03	3	79	-0.12	
150	2	181	0.16	2	191	0	2	207	0.16	2	233	0.16	
300	2	90	0.16	2	95	0	2	103	0.16	2	116	0.16	
600	1	181	0.16	1	191	0	1	207	0.16	1	233	0.16	
1200	1	90	0.16	1	95	0	1	103	0.16	1	116	0.16	
2400	0	181	0.16	0	191	0	0	207	0.16	0	233	0.16	
4800	0	90	0.16	0	95	0	0	103	0.16	0	116	0.16	
9600	0	45	-0.93	0	47	0	0	51	0.16	0	58	-0.69	
19200	0	22	-0.93	0	23	0	0	25	0.16	0	28	1.02	
31250	0	11	0	0	14	-1.70	0	15	0	0	17	0.00	
38400	0	10	3.57	0	11	0	0	12	0.16	0	14	-2.34	

					φ (I	MHz)					
	2		4		8		10		16		18
n	Ν	n	Ν	n	Ν	n	Ν	n	Ν	n	Ν
3	70	_	_	_	_	_	_		_	_	_
2	124	2	249	3	124	_	—	3	249	_	—
1	249	2	124	2	249	_	_	3	124	3	140
1	124	1	249	2	124	_	—	2	249	3	69
0	199	1	99	1	199	1	249	2	99	2	112
0	99	0	199	1	99	1	124	1	199	1	224
0	49	0	99	0	199	0	249	1	99	1	112
0	19	0	39	0	79	0	99	0	159	0	179
0	9	0	19	0	39	0	49	0	79	0	89
0	4	0	9	0	19	0	24	0	39	0	44
0	1	0	3	0	7	0	9	0	15	0	17
0	0*	0	1	0	3	0	4	0	7	0	8
		0	0*	0	1	_	_	0	3	0	4
				0	0*			0	1		
				_	_	0	0*	_	_	_	_
								0	0*	_	_
	3 2 1 0 0 0 0 0 0 0 0 0 0	n         N           3         70           2         124           1         249           1         124           0         199           0         99           0         49           0         19           0         9           0         49           0         1           0         1           0         1           0         1           0         1           0         1	N         n           3         70            2         124         2           1         249         2           1         124         1           0         199         1           0         99         0           0         49         0           0         99         0           0         99         0           0         19         0           0         19         0           0         19         0           0         19         0           0         9         0           0         9         0           0         9         0           0         9         0           0         9         0           0         1         0           0         0*         0	n         N         n         N           3         70             2         124         2         249           1         249         2         124           1         124         1         249           1         124         1         249           0         199         1         99           0         99         0         199           0         49         0         99           0         9         0         19           0         49         0         99           0         19         0         39           0         9         0         19           0         4         0         9           0         1         0         3           0         0*         0         1	N         n         N         n           3         70             2         124         2         249         3           1         249         2         124         2           1         124         1         249         2           1         124         1         249         2           0         199         1         99         1           0         99         0         199         1           0         49         0         99         0           0         19         0         39         0           0         99         0         199         0           0         19         0         39         0           0         9         0         19         0         0           0         4         0         9         0         0           0         1         0         3         0         0           0         0*         0         1         0         0         0	248nNnNnN37021242249312412492124224911241249212401991991199099019919904909901990190390790901903901030700*010300*0103	nNnNnNn3702124224931241249212422491124124921240199199119910990199199001903901990019039019900190390199001030790010307000*0103000*01030	24810nNnNnN370212422493124124921242249112412492124019919911991249099019919911240990990990990990199199999909901990249019039099090190240103079900*01030400*01030400*010304	$ \begin{array}{ c c c c c c c c c c c c c c c c c c c$	$ \begin{array}{c c c c c c c c c c c c c c c c c c c $	$ \begin{array}{c c c c c c c c c c c c c c c c c c c $

Legend:

Blank: No setting available

-: Setting possible, but error occurs

\*: Continuous transmission/reception not possible

Note: Settings with an error of 1% or less are recommended.

The BRR setting is calculated as follows:

Asynchronous mode:

$$N = \frac{\phi}{64 \times 2^{2n-1} \times B} \times 10^6 - 1$$

Synchronous mode:

$$N = \frac{\varphi}{-8 \times 2^{2n-1} \times B} \times 10^6 - 1$$

- B: Bit rate (bits/s)
- N: BRR setting for baud rate generator ( $0 \le N \le 255$ )
- φ: System clock frequency (MHz)
- n: Baud rate generator clock source (n = 0, 1, 2, 3)(For the clock sources and values of n, see the following table.)

		SMR Settings						
n	Clock Source	CKS1	CKS0					
0	φ	0	0					
1	ф/4	0	1					
2	φ/16	1	0					
3	φ/64	1	1					

The bit rate error in asynchronous mode is calculated as follows.

Error (%) = { $\frac{\phi \times 10^{6}}{(N + 1) \times B \times 64 \times 2^{2n-1}} - 1$ } × 100

**.** ...

Table 11.5 indicates the maximum bit rates in asynchronous mode for various system clock frequencies. Tables 11.6 and 11.7 indicate the maximum bit rates with external clock input.

		Settings	
Maximum Bit Rate (bits/s)	n	Ν	
62500	0	0	
65536	0	0	
76800	0	0	
93750	0	0	
115200	0	0	
125000	0	0	
153600	0	0	
156250	0	0	
187500	0	0	
192000	0	0	
230400	0	0	
250000	0	0	
307200	0	0	
312500	0	0	
375000	0	0	
384000	0	0	
437500	0	0	
460800	0	0	
500000	0	0	
537600	0	0	
562500	0	0	
	62500         65536         76800         93750         115200         125000         153600         156250         187500         230400         250000         307200         312500         384000         460800         500000         537600	62500       0         65536       0         76800       0         93750       0         115200       0         125000       0         153600       0         156250       0         187500       0         192000       0         230400       0         307200       0         312500       0         375000       0         460800       0         500000       0         537600       0	Maximum Bit Rate (bits/s)         n         N           62500         0         0         0           65536         0         0         0           76800         0         0         0           93750         0         0         0           115200         0         0         0           125000         0         0         0           156250         0         0         0           187500         0         0         0           192000         0         0         0           230400         0         0         0           307200         0         0         0           312500         0         0         0           384000         0         0         0           437500         0         0         0           500000         0         0         0           500000         0         0         0

Table 11.5	Maximum	Bit Rates f	or Various	<b>Frequencies</b>	(Asynchronous Mode)
------------	---------	-------------	------------	--------------------	---------------------

φ (MHz)	External Input Clock (MHz)	Maximum Bit Rate (bits/s)
2	0.5000	31250
2.097152	0.5243	32768
2.4576	0.6144	38400
3	0.7500	46875
3.6864	0.9216	57600
4	1.0000	62500
4.9152	1.2288	76800
5	1.2500	78125
6	1.5000	93750
6.144	1.5360	96000
7.3728	1.8432	115200
8	2.0000	125000
9.8304	2.4576	153600
10	2.5000	156250
12	3.0000	187500
12.288	3.0720	192000
14	3.5000	218750
14.7456	3.6864	230400
16	4.0000	250000
17.2032	4.3008	268800
18	4.5000	281250

Table 11.6	Maximum Bit Rates with External Clock Input (Asynch	ronous Mode)
Table The	musimum bit Rutes with External Clock input (insyner	nonous mouc)

φ (MHz)	External Input Clock (MHz)	Maximum Bit Rate (bits/s)
2	0.3333	333333.3
4	0.6667	666666.7
6	1.0000	100000.0
8	1.3333	1333333.3
10	1.6667	1666666.7
12	2.0000	200000.0
14	2.3333	2333333.3
16	2.6667	2666666.7
18	3.0000	300000.0

 Table 11.7
 Maximum Bit Rates with External Clock Input (Synchronous Mode)

# 11.3 Operation

## 11.3.1 Overview

The SCI has an asynchronous mode in which characters are synchronized individually, and a synchronous mode in which communication is synchronized with clock pulses. Serial communication is possible in either mode. Asynchronous or synchronous mode and the communication format are selected in SMR, as shown in table 11.8. The SCI clock source is selected by the  $C/\overline{A}$  bit in SMR and the CKE1 and CKE0 bits in SCR, as shown in table 11.9.

Asynchronous Mode:

- Data length is selectable: 7 or 8 bits.
- Parity and multiprocessor bits are selectable, and so is the stop bit length (1 or 2 bits). These selections determine the communication format and character length.
- In receiving, it is possible to detect framing errors, parity errors, overrun errors, and the break state.
- An internal or external clock can be selected as the SCI clock source.
  - When an internal clock is selected, the SCI operates using the on-chip baud rate generator, and can output a serial clock signal with a frequency matching the bit rate.
  - When an external clock is selected, the external clock input must have a frequency 16 times the bit rate. (The on-chip baud rate generator is not used.)

Synchronous Mode:

- The communication format has a fixed 8-bit data length.
- In receiving, it is possible to detect overrun errors.
- An internal or external clock can be selected as the SCI clock source.
  - When an internal clock is selected, the SCI operates using the on-chip baud rate generator, and outputs a serial clock signal to external devices.
  - When an external clock is selected, the SCI operates on the input serial clock. The on-chip baud rate generator is not used.

SMR Settings						SCI Communication Format			nat
Bit 7 C/A	Bit 6 CHR	Bit 2 MP	Bit 5 PE	Bit 3 STOP	- Mode	Data Length	Multi- processor Bit	Parity Bit	Stop Bit Length
0	0	0	0	0	Asynchronous	8-bit data	Absent	Absent	1 bit
0	0	0	0	1	<sup>–</sup> mode				2 bits
0	0	0	1	0	_			Present	1 bit
0	0	0	1	1	_				2 bits
0	1	0	0	0	_	7-bit data	_	Absent	1 bit
0	1	0	0	1	_				2 bits
0	1	0	1	0	_			Present	1 bit
0	1	0	1	1	_				2 bits
0	0	1		0	Asynchronous	8-bit data	Present	Absent	1 bit
0	0	1	_	1	mode (multi- processor				2 bits
0	1	1		0	format)	7-bit data	_		1 bit
0	1	1		1	_				2 bits
1	_				Synchronous mode	8-bit data	Absent	_	None

#### Table 11.8 SMR Settings and Serial Communication Formats

## Table 11.9 SMR and SCR Settings and SCI Clock Source Selection

SMR	SCR Settings					
Bit 7	Bit 1 Bit 0		-	SCI Communication Format		
C/Ā	CKE1	CKE0	Mode	Clock Source	SCK Pin Function	
0	0	0	Asynchronous	Internal	SCI does not use the SCK pin	
0	0	1	<sup>–</sup> mode		Outputs a clock with frequency matching the bit rate	
0	1	0	_	External	Inputs a clock with frequency	
0	1	1	_		16 times the bit rate	
1	0	0	Synchronous	Internal	Outputs the serial clock	
1	0	1	mode			
1	1	0	_	External	Inputs the serial clock	
1	1	1	_			

## 11.3.2 Operation in Asynchronous Mode

In asynchronous mode each transmitted or received character begins with a start bit and ends with a stop bit. Serial communication is synchronized one character at a time.

The transmitting and receiving sections of the SCI are independent, so full-duplex communication is possible. The transmitter and receiver are both double buffered, so data can be written and read while transmitting and receiving are in progress, enabling continuous transmitting and receiving.

Figure 11.2 shows the general format of asynchronous serial communication. In asynchronous serial communication the communication line is normally held in the mark (high) state. The SCI monitors the line and starts serial communication when the line goes to the space (low) state, indicating a start bit. One serial character consists of a start bit (low), data (LSB first), parity bit (high or low), and stop bit (high), in that order.

When receiving in asynchronous mode, the SCI synchronizes at the falling edge of the start bit. The SCI samples each data bit on the eighth pulse of a clock with a frequency 16 times the bit rate. Receive data is latched at the center of each bit.

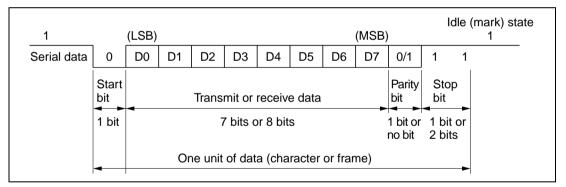


Figure 11.2 Data Format in Asynchronous Communication (Example: 8-Bit Data with Parity and 2 Stop Bits)

#### **Communication Formats**

Table 11.10 shows the 12 communication formats that can be selected in asynchronous mode. The format is selected by settings in SMR.

SMR Settings				Serial Communication Format and Frame Length			
CHR	PE	MP	STOP	1 _ 2 _ 3 _ 4 _ 5 _ 6 _ 7 _ 8 _ 9 _ 10 _ 11 _ 12 _			
0	0	0	0	S 8-bit data STOP			
0	0	0	1	S 8-bit data STOP STOP			
0	1	0	0	S 8-bit data P STOP			
0	1	0	1	S 8-bit data P STOP STOP			
1	0	0	0	S 7-bit data STOP			
1	0	0	1	S 7-bit data STOP STOP			
1	1	0	0	S 7-bit data P STOP			
1	1	0	1	S 7-bit data P STOP STOP			
0	_	1	0	S 8 bit data MPB STOP			
0	_	1	1	S 8 bit data MPB STOP STOP			
1	_	1	0	S 7-bit data MPB STOP			
1	_	1	1	S 7-bit data MPB STOP STOP			

## Table 11.10 Serial Communication Formats (Asynchronous Mode)

Legend:

S: Start bit

STOP: Stop bit

P: Parity bit

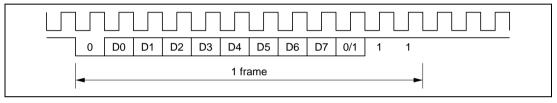
MPB: Multiprocessor bit

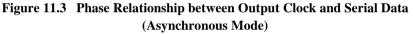
## Clock

An internal clock generated by the on-chip baud rate generator or an external clock input from the SCK pin can be selected as the SCI transmit/receive clock. The clock source is selected by the  $C/\overline{A}$  bit in SMR and bits CKE1 and CKE0 in SCR. See table 11.9.

When an external clock is input at the SCK pin, it must have a frequency equal to 16 times the desired bit rate.

When the SCI operates on an internal clock, it can output a clock signal at the SCK pin. The frequency of this output clock is equal to the bit rate. The phase is aligned as in figure 11.3 so that the rising edge of the clock occurs at the center of each transmit data bit.





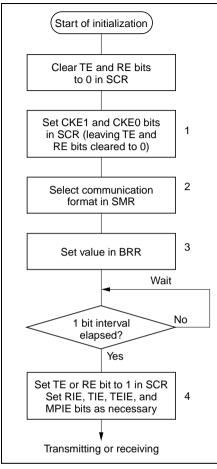
## **Transmitting and Receiving Data**

**SCI Initialization (Asynchronous Mode):** Before transmitting or receiving, clear the TE and RE bits to 0 in SCR, then initialize the SCI as follows.

When changing the communication mode or format, always clear the TE and RE bits to 0 before following the procedure given below. Clearing TE to 0 sets the TDRE flag to 1 and initializes TSR. Clearing RE to 0, however, does not initialize the RDRF, PER, FER, and ORER flags and RDR, which retain their previous contents.

When an external clock is used, the clock should not be stopped during initialization or subsequent operation. SCI operation becomes unreliable if the clock is stopped.

Figure 11.4 shows a sample flowchart for initializing the SCI.



- Select the clock source in SCR. Clear the RIE, TIE, TEIE, MPIE, TE, and RE bits to 0. If clock output is selected in asynchronous mode, clock output starts immediately after the setting is made in SCR.
- 2. Select the communication format in SMR.
- 3. Write the value corresponding to the bit rate in BRR. This step is not necessary when an external clock is used.
- 4. Wait for at least the interval required to transmit or receive 1 bit, then set the TE or RE bit to 1 in SCR. Set the RIE, TIE, TEIE, and MPIE bits as necessary. Setting the TE or RE bit enables the SCI to use the TxD or RxD pin.

Figure 11.4 Sample Flowchart for SCI Initialization

**Transmitting Serial Data (Asynchronous Mode):** Figure 11.5 shows a sample flowchart for transmitting serial data and indicates the procedure to follow.

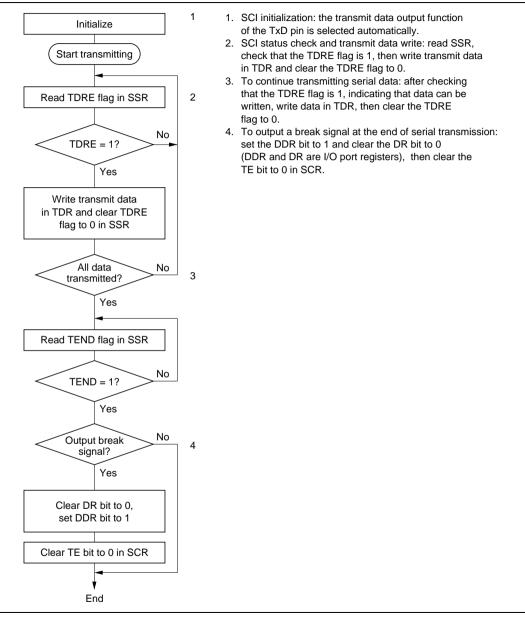


Figure 11.5 Sample Flowchart for Transmitting Serial Data

In transmitting serial data, the SCI operates as follows.

- The SCI monitors the TDRE flag in SSR. When the TDRE flag is cleared to 0 the SCI recognizes that TDR contains new data, and loads this data from TDR into TSR.
- After loading the data from TDR into TSR, the SCI sets the TDRE flag to 1 and starts transmitting. If the TIE bit is set to 1 in SCR, the SCI requests a transmit-data-empty interrupt (TXI) at this time.

Serial transmit data is transmitted in the following order from the TxD pin:

- Start bit: One 0 bit is output.
   Transmit data: 7 or 8 bits are output, LSB first.
   Parity bit or multiprocessor bit: One parity bit (even or odd parity) or one multiprocessor bit is output. Formats in which neither a parity bit nor a multiprocessor bit is output can also be selected.
   Stop bit: One or two 1 bits (stop bits) are output.
   Mark state: Output of 1 continues until the start bit of the next transmit data.
- The SCI checks the TDRE flag when it outputs the stop bit. If the TDRE flag is 0, the SCI loads new data from TDR into TSR, outputs the stop bit, then begins serial transmission of the next frame. If the TDRE flag is 1, the SCI sets the TEND flag to 1 in SSR, outputs the stop bit, then continues output of 1 in the mark state. If the TEIE bit is set to 1 in SCR, a transmit-end interrupt (TEI) is requested at this time.

Figure 11.6 shows an example of SCI transmit operation in asynchronous mode.

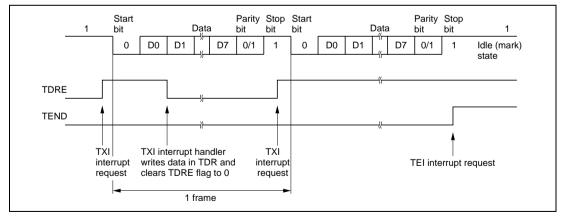


Figure 11.6 Example of SCI Transmit Operation in Asynchronous Mode (8-Bit Data with Parity and 1 Stop Bit)

### Renesas

**Receiving Serial Data (Asynchronous Mode):** Figure 11.7 shows a sample flowchart for receiving serial data and indicates the procedure to follow.

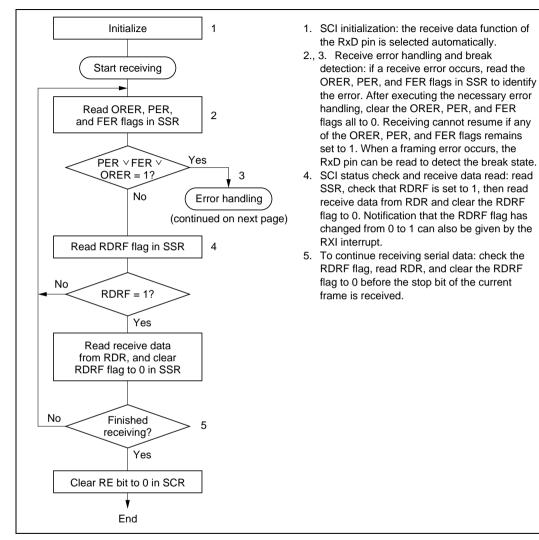


Figure 11.7 Sample Flowchart for Receiving Serial Data (1)

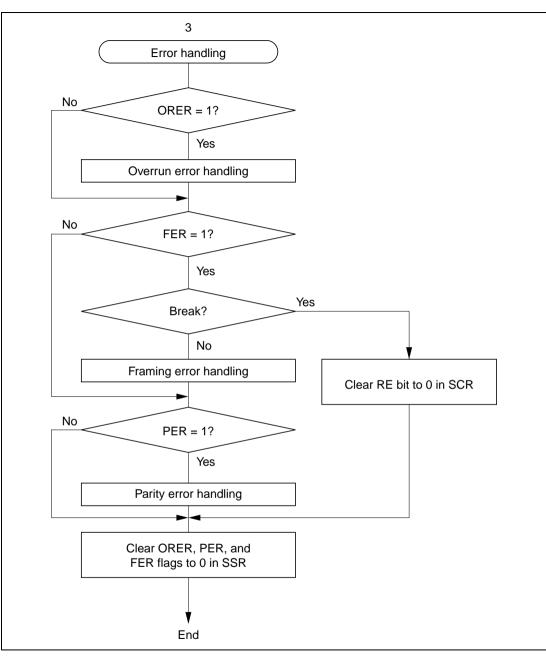


Figure 11.7 Sample Flowchart for Receiving Serial Data (2)

In receiving, the SCI operates as follows.

- The SCI monitors the receive data line. When it detects a start bit, the SCI synchronizes internally and starts receiving.
- Receive data is stored in RSR in order from LSB to MSB.
- The parity bit and stop bit are received.

After receiving data, the SCI makes the following checks:

— Parity check:	The number of 1s in the receive data must match the even or odd parity
	setting of the $O/\overline{E}$ bit in SMR.

- Stop bit check: The stop bit value must be 1. If there are two stop bits, only the first stop bit is checked.
- --- Status check: The RDRF flag must be 0 so that receive data can be transferred from RSR into RDR.

If these checks all pass, the RDRF flag is set to 1 and the received data is stored in RDR. If one of the checks fails (receive error)\*, the SCI operates as indicated in table 11.11.

- Note: \* When a receive error occurs, further receiving is disabled. In receiving, the RDRF flag is not set to 1. Be sure to clear the error flags.
- When the RDRF flag is set to 1, if the RIE bit is set to 1 in SCR, a receive-data-full interrupt (RXI) is requested. If the ORER, PER, or FER flag is set to 1 and the RIE bit in SCR is also set to 1, a receive-error interrupt (ERI) is requested.

<b>Receive Error</b>	Abbreviation	Condition	Data Transfer
Overrun error	ORER	Receiving of next data ends while RDRF flag is still set to 1 in SSR	Receive data not transferred from RSR to RDR
Framing error	FER	Stop bit is 0	Receive data transferred from RSR to RDR
Parity error	PER	Parity of receive data differs from even/odd parity setting in SMR	Receive data transferred from RSR to RDR

## **Table 11.11 Receive Error Conditions**

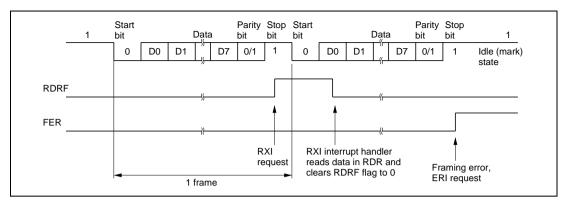


Figure 11.8 shows an example of SCI receive operation in asynchronous mode.

### Figure 11.8 Example of SCI Receive Operation (8-Bit Data with Parity and One Stop Bit)

### 11.3.3 Multiprocessor Communication

The multiprocessor communication function enables several processors to share a single serial communication line. The processors communicate in asynchronous mode using a format with an additional multiprocessor bit (multiprocessor format).

In multiprocessor communication, each receiving processor is addressed by an ID. A serial communication cycle consists of an ID-sending cycle that identifies the receiving processor, and a data-sending cycle. The multiprocessor bit distinguishes ID-sending cycles from data-sending cycles.

The transmitting processor starts by sending the ID of the receiving processor with which it wants to communicate as data with the multiprocessor bit set to 1. Next the transmitting processor sends transmit data with the multiprocessor bit cleared to 0.

Receiving processors skip incoming data until they receive data with the multiprocessor bit set to 1. When they receive data with the multiprocessor bit set to 1, receiving processors compare the data with their IDs. The receiving processor with a matching ID continues to receive further incoming data. Processors with IDs not matching the received data skip further incoming data until they again receive data with the multiprocessor bit set to 1. Multiple processors can send and receive data in this way.

Figure 11.9 shows an example of communication among different processors using a multiprocessor format.

### **Communication Formats**

Four formats are available. Parity-bit settings are ignored when a multiprocessor format is selected. For details see table 11.11.

### Clock

See the description of asynchronous mode.

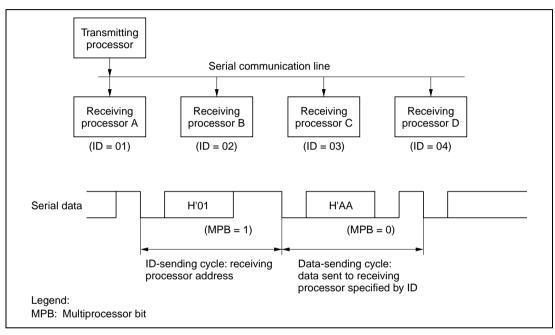


Figure 11.9 Example of Communication among Processors using Multiprocessor Format (Sending Data H'AA to Receiving Processor A)

#### **Transmitting and Receiving Data**

**Transmitting Multiprocessor Serial Data:** Figure 11.10 shows a sample flowchart for transmitting multiprocessor serial data and indicates the procedure to follow.

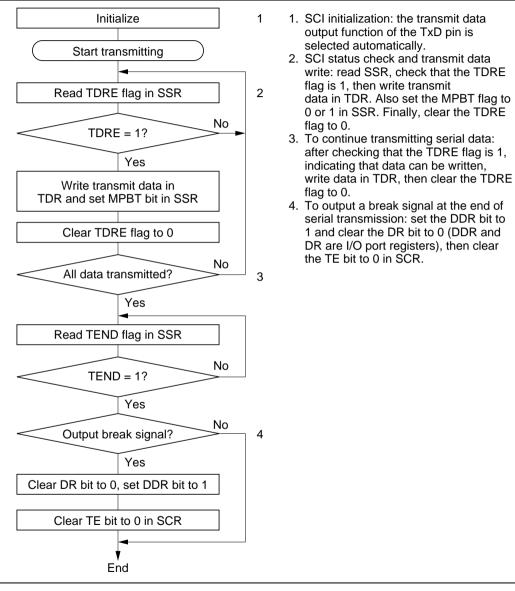


Figure 11.10 Sample Flowchart for Transmitting Multiprocessor Serial Data

In transmitting serial data, the SCI operates as follows.

- The SCI monitors the TDRE flag in SSR. When the TDRE flag is cleared to 0 the SCI recognizes that TDR contains new data, and loads this data from TDR into TSR.
- After loading the data from TDR into TSR, the SCI sets the TDRE flag to 1 and starts transmitting. If the TIE bit in SCR is set to 1, the SCI requests a transmit-data-empty interrupt (TXI) at this time.

Serial transmit data is transmitted in the following order from the TxD pin:

- Start bit: One 0 bit is output.
- Transmit data: 7 or 8 bits are output, LSB first.
- Multiprocessor bit: One multiprocessor bit (MPBT value) is output.
- Stop bit: One or two 1 bits (stop bits) are output.
- Mark state: Output of 1 bits continues until the start bit of the next transmit data.
- The SCI checks the TDRE flag when it outputs the stop bit. If the TDRE flag is 0, the SCI loads data from TDR into TSR, outputs the stop bit, then begins serial transmission of the next frame. If the TDRE flag is 1, the SCI sets the TEND flag in SSR to 1, outputs the stop bit, then continues output of 1 bits in the mark state. If the TEIE bit is set to 1 in SCR, a transmit-end interrupt (TEI) is requested at this time.

Figure 11.11 shows an example of SCI transmit operation using a multiprocessor format.

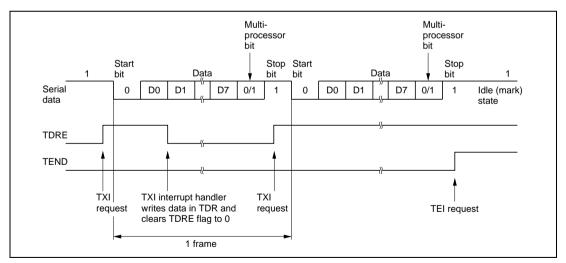
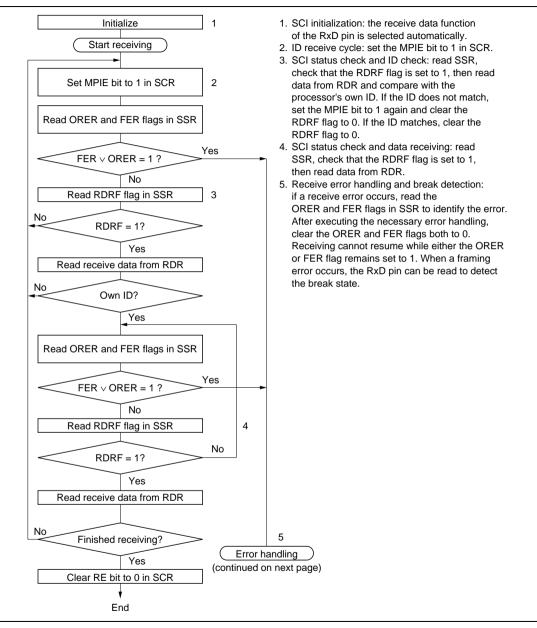


Figure 11.11 Example of SCI Transmit Operation (8-Bit Data with Multiprocessor Bit and One Stop Bit)

## Renesas

**Receiving Multiprocessor Serial Data:** Figure 11.12 shows a sample flowchart for receiving multiprocessor serial data and indicates the procedure to follow.





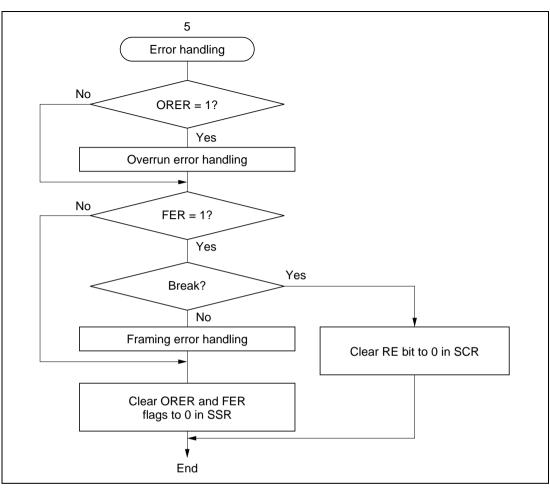


Figure 11.12 Sample Flowchart for Receiving Multiprocessor Serial Data (2)

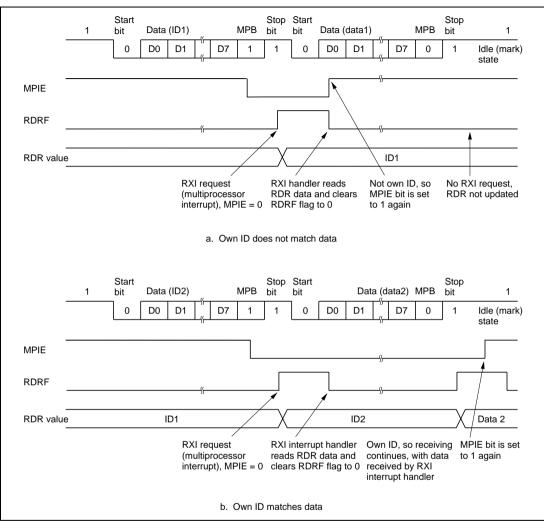


Figure 11.13 shows an example of SCI receive operation using a multiprocessor format.

Figure 11.13 Example of SCI Receive Operation (8-Bit Data with Multiprocessor Bit and One Stop Bit)

## 11.3.4 Synchronous Operation

In synchronous mode, the SCI transmits and receives data in synchronization with clock pulses. This mode is suitable for high-speed serial communication.

The SCI transmitter and receiver share the same clock but are otherwise independent, so full duplex communication is possible. The transmitter and receiver are also double buffered, so continuous transmitting or receiving is possible by reading or writing data while transmitting or receiving is in progress.

Figure 11.14 shows the general format in synchronous serial communication.

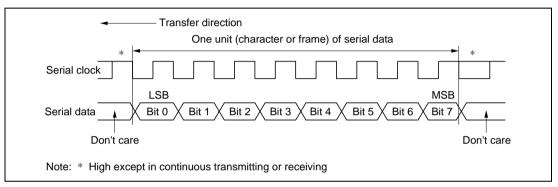


Figure 11.14 Data Format in Synchronous Communication

In synchronous serial communication, each data bit is placed on the communication line from one falling edge of the serial clock to the next. Data is guaranteed valid at the rise of the serial clock. In each character, the serial data bits are transmitted in order from LSB (first) to MSB (last). After output of the MSB, the communication line remains in the state of the MSB. In synchronous mode the SCI receives data by synchronizing with the rise of the serial clock.

## **Communication Format**

The data length is fixed at 8 bits. No parity bit or multiprocessor bit can be added.

## Clock

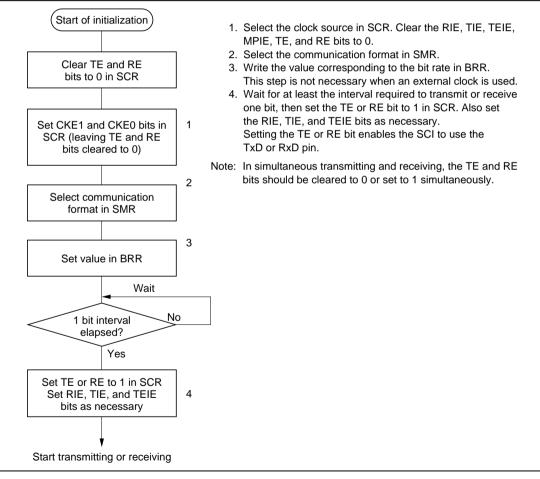
An internal clock generated by the on-chip baud rate generator or an external clock input from the SCK pin can be selected by clearing or setting the CKE1 and CKE0 bits in SCR and the C/ $\overline{A}$  bit in SMR. See table 11.9. When the SCI operates on an internal clock, it outputs the clock signal at the SCK pin. Eight clock pulses are output per transmitted or received character. When the SCI is not transmitting or receiving, the clock signal remains in the high state.

## **Transmitting and Receiving Data**

**SCI Initialization (Synchronous Mode):** Before transmitting or receiving, clear the TE and RE bits to 0 in SCR, then initialize the SCI as follows.

When changing the communication mode or format, always clear the TE and RE bits to 0 before following the procedure given below. Clearing the TE bit to 0 sets the TDRE flag to 1 and initializes TSR. Clearing the RE bit to 0, however, does not initialize the RDRF, PER, FER, and ORE flags and RDR, which retain their previous contents.

Figure 11.15 shows a sample flowchart for initializing the SCI.



## Figure 11.15 Sample Flowchart for SCI Initialization

**Transmitting Serial Data (Synchronous Mode):** Figure 11.16 shows a sample flowchart for transmitting serial data and indicates the procedure to follow.

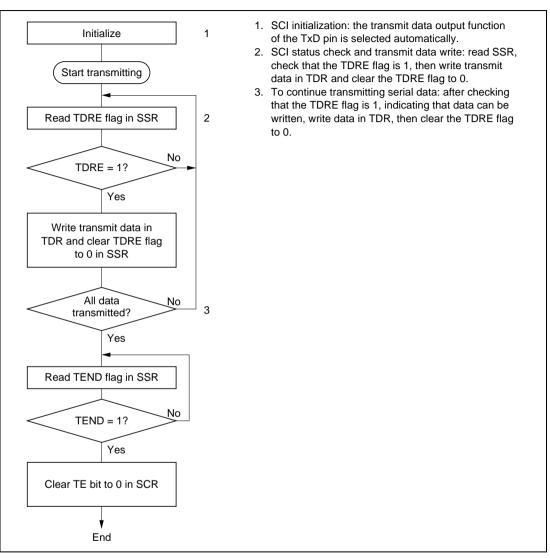


Figure 11.16 Sample Flowchart for Serial Transmitting

In transmitting serial data, the SCI operates as follows.

- The SCI monitors the TDRE flag in SSR. When the TDRE flag is cleared to 0 the SCI recognizes that TDR contains new data, and loads this data from TDR into TSR.
- After loading the data from TDR into TSR, the SCI sets the TDRE flag to 1 and starts transmitting. If the TIE bit is set to 1 in SCR, the SCI requests a transmit-data-empty interrupt (TXI) at this time.

If clock output is selected, the SCI outputs eight serial clock pulses. If an external clock source is selected, the SCI outputs data in synchronization with the input clock. Data is output from the TxD pin in order from LSB (bit 0) to MSB (bit 7).

- The SCI checks the TDRE flag when it outputs the MSB (bit 7). If the TDRE flag is 0, the SCI loads data from TDR into TSR and begins serial transmission of the next frame. If the TDRE flag is 1, the SCI sets the TEND flag to 1 in SSR, and after transmitting the MSB, holds the TxD pin in the MSB state. If the TEIE bit in SCR is set to 1, a transmit-end interrupt (TEI) is requested at this time.
- After the end of serial transmission, the SCK pin is held in a constant state.

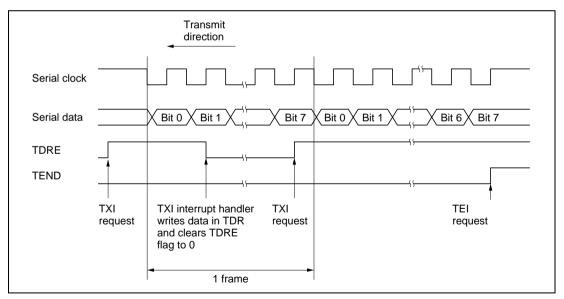


Figure 11.17 shows an example of SCI transmit operation.

Figure 11.17 Example of SCI Transmit Operation

**Receiving Serial Data (Synchronous Mode):** Figure 11.18 shows a sample flowchart for receiving serial data and indicates the procedure to follow. When switching from asynchronous mode to synchronous mode, make sure that the ORER, PER, and FER flags are cleared to 0. If the FER or PER flag is set to 1 the RDRF flag will not be set and both transmitting and receiving will be disabled.

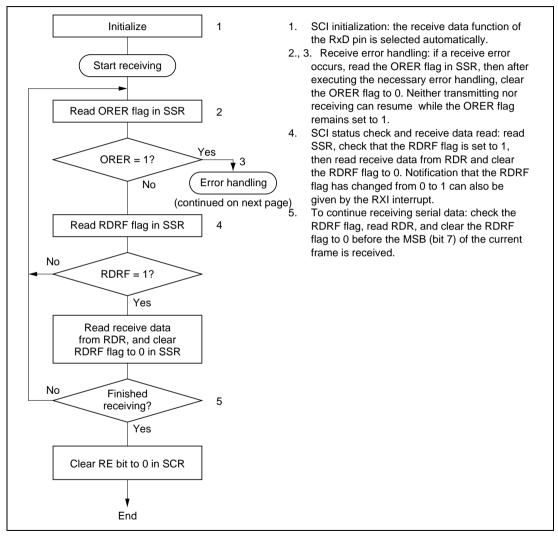


Figure 11.18 Sample Flowchart for Serial Receiving (1)

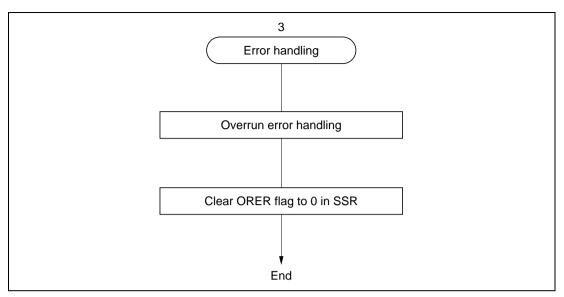


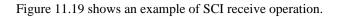
Figure 11.18 Sample Flowchart for Serial Receiving (2)

In receiving, the SCI operates as follows.

- The SCI synchronizes with serial clock input or output and initializes internally.
- Receive data is stored in RSR in order from LSB to MSB.

After receiving the data, the SCI checks that the RDRF flag is 0 so that receive data can be transferred from RSR to RDR. If this check passes, the RDRF flag is set to 1 and the received data is stored in RDR. If the check does not pass (receive error), the SCI operates as indicated in table 11.11. If any receive error is detected, the subsequent data transmission/reception is disabled.

• After setting the RDRF flag to 1, if the RIE bit is set to 1 in SCR, the SCI requests a receivedata-full interrupt (RXI). If the ORER flag is set to 1 and the RIE bit in SCR is also set to 1, the SCI requests a receive-error interrupt (ERI).



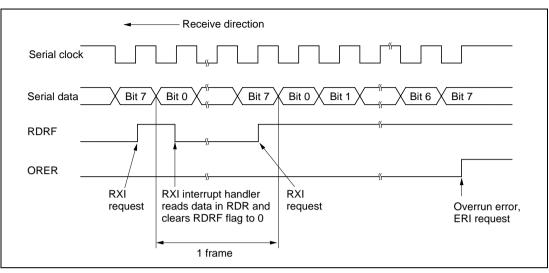
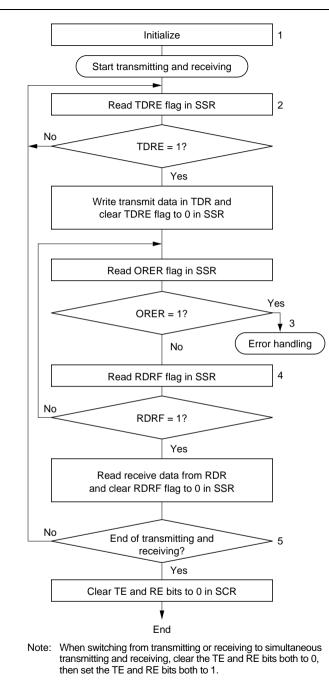


Figure 11.19 Example of SCI Receive Operation

**Transmitting and Receiving Serial Data Simultaneously (Synchronous Mode):** Figure 11.20 shows a sample flowchart for transmitting and receiving serial data simultaneously and indicates the procedure to follow.



- SCI initialization: the transmit data output function of the TxD pin and receive data input function of the RxD pin are selected, enabling simultaneous transmitting and receiving.
- SCI status check and transmit data write: read SSR, check that the TDRE flag is 1, then write transmit data in TDR and clear the TDRE flag to 0. Notification that the TDRE flag has changed from 0 to 1 can also be given by the TXI interrupt.
- 3. Receive error handling: if a receive error occurs, read the ORER flag in SSR, then after executing the necessary error handling, clear the ORER flag to 0.

Neither transmitting nor receiving can resume while the ORER flag remains set to 1.

- 4. SCI status check and receive data read: read SSR, check that the RDRF flag is 1, then read receive data from RDR and clear the RDRF flag to 0. Notification that the RDRF flag has changed from 0 to 1 can also be given by the RXI interrupt.
- 5. To continue transmitting and receiving serial data: check the RDRF flag, read RDR, and clear the RDRF flag to 0 before the MSB (bit 7) of the current frame is received. Also check that the TDRE flag is set to 1, indicating that data can be written, write data in TDR, then clear the TDRE flag to 0 before the MSB (bit 7) of the current frame is transmitted.

# Figure 11.20 Sample Flowchart for Serial Transmitting

# **11.4 SCI Interrupts**

The SCI has four interrupt request sources: TEI (transmit-end interrupt), ERI (receive-error interrupt), RXI (receive-data-full interrupt), and TXI (transmit-data-empty interrupt). Table 11.12 lists the interrupt sources and indicates their priority. These interrupts can be enabled and disabled by the TIE, RIE, and TEIE bits in SCR. Each interrupt request is sent separately to the interrupt controller.

The TXI interrupt is requested when the TDRE flag is set to 1 in SSR. The TEI interrupt is requested when the TEND flag is set to 1 in SSR.

The RXI interrupt is requested when the RDRF flag is set to 1 in SSR. The ERI interrupt is requested when the ORER, PER, or FER flag is set to 1 in SSR.

## Table 11.12 SCI Interrupt Sources

Interrupt	Description	Priority
ERI	Receive error (ORER, FER, or PER)	High
RXI	Receive data register full (RDRF)	↑
TXI	Transmit data register empty (TDRE)	
TEI	Transmit end (TEND)	Low

## 11.5 Usage Notes

Note the following points when using the SCI.

## **TDR Write and TDRE Flag**

The TDRE flag in SSR is a status flag indicating the loading of transmit data from TDR into TSR. The SCI sets the TDRE flag to 1 when it transfers data from TDR to TSR.

Data can be written into TDR regardless of the state of the TDRE flag. If new data is written in TDR when the TDRE flag is 0, the old data stored in TDR will be lost because this data has not yet been transferred to TSR. Before writing transmit data in TDR, be sure to check that the TDRE flag is set to 1.

## **Simultaneous Multiple Receive Errors**

Table 11.13 indicates the state of SSR status flags when multiple receive errors occur simultaneously. When an overrun error occurs the RSR contents are not transferred to RDR, so receive data is lost.

SSR Status Flags				Receive Data Transfer	
RDRF ORER FER PER		$RSR \rightarrow RDR$	Receive Errors		
1	1	0	0	×	Overrun error
0	0	1	0	0	Framing error
0	0	0	1	0	Parity error
1	1	1	0	×	Overrun error + framing error
1	1	0	1	×	Overrun error + parity error
0	0	1	1	0	Framing error + parity error
1	1	1	1	×	Overrun error + framing error + parity error

## Table 11.13 SSR Status Flags and Transfer of Receive Data

Notes: O: Receive data is transferred from RSR to RDR.

 $\times$ : Receive data is not transferred from RSR to RDR.

## **Break Detection and Processing**

Break signals can be detected by reading the RxD pin directly when a framing error (FER) is detected. In the break state the input from the RxD pin consists of all 0s, so the FER flag is set and the parity error flag (PER) may also be set. In the break state the SCI receiver continues to operate, so if the FER flag is cleared to 0 it will be set to 1 again.

## Sending a Break Signal

When the TE bit is cleared to 0 the TxD pin becomes an I/O port, the level and direction (input or output) of which are determined by DR and DDR bits. This feature can be used to send a break signal.

After the serial transmitter is initialized, the DR value substitutes for the mark state until the TE bit is set to 1 (the TxD pin function is not selected until the TE bit is set to 1). The DDR and DR bits should therefore both be set to 1 beforehand.

To send a break signal during serial transmission, clear the DR bit to 0, then clear the TE bit to 0. When the TE bit is cleared to 0 the transmitter is initialized, regardless of its current state, so the TxD pin becomes an input/output port outputting the value 0.

## **Receive Error Flags and Transmitter Operation (Synchronous Mode Only)**

When a receive error flag (ORER, PER, or FER) is set to 1 the SCI will not start transmitting, even if the TDRE flag is cleared to 0. Be sure to clear the receive error flags to 0 when starting to transmit. Note that clearing the RE bit to 0 does not clear the receive error flags to 0.

## Receive Data Sampling Timing in Asynchronous Mode and Receive Margin

In asynchronous mode the SCI operates on a base clock with 16 times the bit rate frequency. In receiving, the SCI synchronizes internally with the fall of the start bit, which it samples on the base clock. Receive data is latched at the rising edge of the eighth base clock pulse. See figure 11.21.

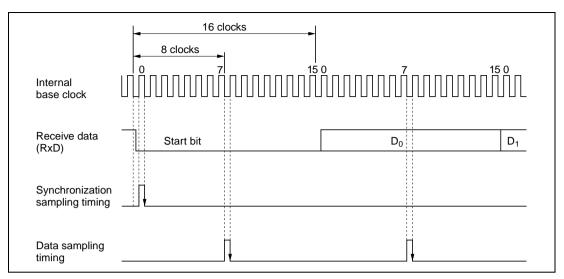


Figure 11.21 Receive Data Sampling Timing in Asynchronous Mode

The receive margin in asynchronous mode can therefore be expressed as shown in equation (1).

$$M = |(0.5 - \frac{1}{2N}) - (L - 0.5) F - \frac{|D - 0.5|}{N} (1 + F)| \times 100\% \dots (1)$$

- M: Receive margin (%)
- N: Ratio of clock frequency to bit rate (N = 16)
- D: Clock duty cycle (D = 0 to 1.0)
- L: Frame length (L = 9 to 12)
- F: Absolute deviation of clock frequency

From equation (1), if F = 0 and D = 0.5 the receive margin is 46.875%, as given by equation (2).

When D = 0.5, F = 0:

 $M = [0.5 - 1/(2 \times 16)] \times 100\%$ = 46.875%.....(2)

This is a theoretical value. A reasonable margin to allow in system design is 20% to 30%.

## **Restrictions in Synchronous Mode**

When data transmission is performed using an external clock source as the serial clock, an interval of at least 5 states is necessary between clearing the TDRE flag in SSR and the start (falling edge) of the first transmit clock pulse corresponding to each frame (figure 11.22). This interval is also necessary when performing continuous transmission. If this condition is not satisfied, an operation error may occur.

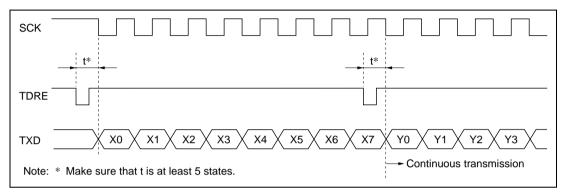


Figure 11.22 Transmission in Synchronous Mode (Example)

## Restrictions when Switching from SCK Pin to Port Function in Synchronous SCI

## 1. Problem in Operation

After setting DDR and DR to 1 and using synchronous SCI clock output, when the SCK pin is switched to the port function at the end of transmission, a low-level signal is output for one half-cycle before the port output state is established.

When switching to the port function by making the following settings while DDR = 1, DR = 1,  $C/\overline{A} = 1$ , CKE1 = 0, CKE0 = 0, and TE = 1, low-level output occurs for one half-cycle.

- (1) End of serial data transmission
- (2) TE bit = 0
- (3)  $C/\overline{A}$  bit = 0 ... switchover to port output
- (4) Occurrence of low-level output (see figure 11.23)

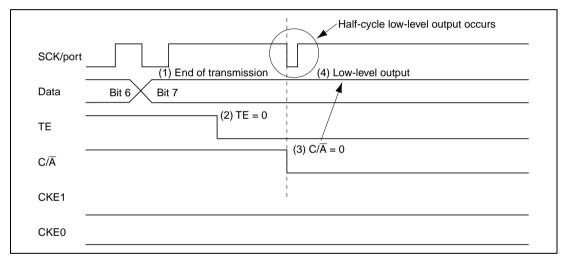


Figure 11.23 Operation when Switching from SCK Pin Function to Port Pin Function

## 2. Usage Note

The procedure shown below should be used to prevent low-level output when switching from the SCK pin function to the port function.

As this procedure temporarily places the SCK pin in the input state, the SCK/port pin should be pulled up beforehand with an external circuit. With DDR = 1, DR = 1,  $C/\overline{A} = 1$ , CKE1 = 0, CKE0 = 0, and TE = 1, make the following settings in the order shown.

- (1) End of serial data transmission
- (2) TE bit = 0
- (3) CKE1 bit = 1
- (4)  $C/\overline{A}$  bit = 0 ... switchover to port output
- (5) CKE1 bit = 0

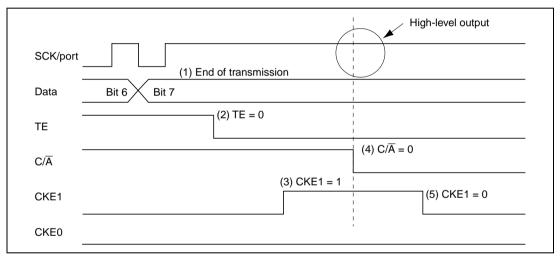


Figure 11.24 Operation when Switching from SCK Pin Function to Port Pin Function (Preventing Low-Level Output)

# Section 12 Smart Card Interface

## 12.1 Overview

SCI0 supports an IC card (Smart Card) interface conforming to ISO/IEC 7816-3 (Identification Card) as a serial communication interface extension function.

Switching between the normal serial communication interface and the Smart Card interface is carried out by means of a register setting.

## 12.1.1 Features

Features of the Smart Card interface supported by the H8/3039 Group are as follows.

- Asynchronous mode
  - Data length: 8 bits
  - Parity bit generation and checking
  - Transmission of error signal (parity error) in receive mode
  - Error signal detection and automatic data retransmission in transmit mode
  - Direct convention and inverse convention both supported
- On-chip baud rate generator allows any bit rate to be selected
- Three interrupt sources
  - Three interrupt sources (transmit data empty, receive data full, and transmit/receive error) that can issue requests independently

## 12.1.2 Block Diagram

Figure 12.1 shows a block diagram of the Smart Card interface.

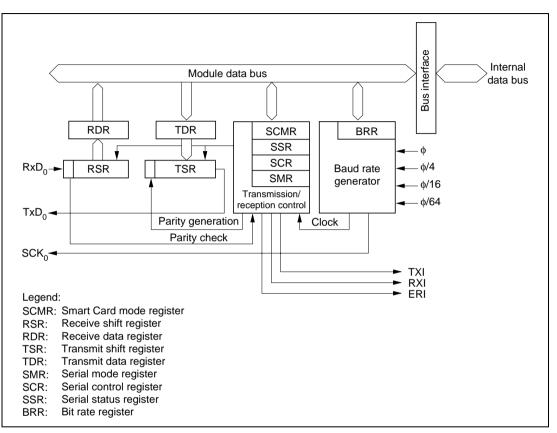


Figure 12.1 Block Diagram of Smart Card Interface

## 12.1.3 Pin Configuration

Table 12.1 shows the Smart Card interface pin configuration.

Table 12.1         Smart Card Interface Pins	<b>Table 12.1</b>	Smart	Card	Interface l	Pins
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Pin Name	Abbreviation	I/O	Function
Serial clock pin 0	SCK	Output	SCI <sub>0</sub> clock output
Receive data pin 0	RxD <sub>0</sub>	Input	SCI <sub>0</sub> receive data input
Transmit data pin 0	TxD <sub>o</sub>	Output	$SCI_{o}$ transmit data output

## 12.1.4 Register Configuration

Table 12.2 shows the registers used by the Smart Card interface. Details of SMR, BRR, SCR, TDR, and RDR are the same as for the normal SCI function: see the register descriptions in section 11, Serial Communication Interface.

<b>Table 12.2</b>	Smart	Card	Interface	Registers
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Address*1	Name	Abbreviation	R/W	Initial Value
H'FFB0	Serial mode register	SMR	R/W	H'00
H'FFB1	Bit rate register	BRR	R/W	H'FF
H'FFB2	Serial control register	SCR	R/W	H'00
H'FFB3	Transmit data register	TDR	R/W	H'FF
H'FFB4	Serial status register	SSR	R/(W)* <sup>2</sup>	H'84
H'FFB5	Receive data register	RDR	R	H'00
H'FFB6	Smart card mode register	SCMR	R/W	H'F2

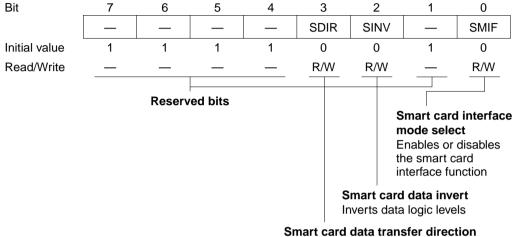
Notes: 1. Lower 16 bits of the address.

2. Can only be written with 0 for flag clearing.

## **12.2** Register Descriptions

Registers added with the Smart Card interface and bits for which the function changes are described here.

## 12.2.1 Smart Card Mode Register (SCMR)



Selects the serial/parallel conversion format

SCMR is an 8-bit readable/writable register that selects the Smart Card interface function.

SCMR is initialized to H'F2 by a reset, and in standby mode.

Bits 7 to 4—Reserved: These bits cannot be modified and are always read as 1.

**Bit 3—Smart Card Data Transfer Direction (SDIR):** Selects the serial/parallel conversion format.

Bit 3 SDIR	Description	
0	TDR contents are transmitted LSB-first	(Initial value)
	Receive data is stored in RDR LSB-first	
1	TDR contents are transmitted MSB-first	
	Receive data is stored in RDR MSB-first	

**Bit 2—Smart Card Data Invert (SINV):** Specifies inversion of the data logic level. This function is used together with the SDIR bit for communication with an inverse convention card. The SINV bit does not affect the logic level of the parity bit. For parity-related setting procedures, see section 12.3.4, Register Settings.

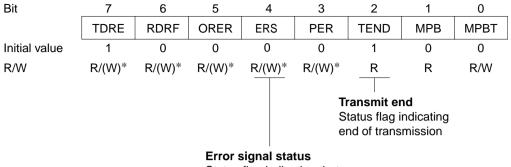
Bit 2 SINV	Description	
0	TDR contents are transmitted as they are	(Initial value)
	Receive data is stored as it is in RDR	
1	TDR contents are inverted before being transmitted	
	Receive data is stored in inverted form in RDR	

Bit 1—Reserved: This bit cannot be modified and is always read as 1.

Bit 0—Smart Card Interface Mode Select (SMIF): This bit enables or disables the Smart Card interface function.

Bit 0 SMIF	Description	
0	Smart Card interface function is disabled	(Initial value)
1	Smart Card interface function is enabled	

## 12.2.2 Serial Status Register (SSR)



Status flag indicating that an error signal has been received

Note: \* Only 0 can be written to bits 7 to 3, to clear these flags.

Bit 4 of SSR has a different function in Smart Card interface mode. Coupled with this, the setting conditions for bit 2, TEND, are also different.

**Bits 7 to 5**—Operate in the same way as for the normal SCI. For details, see section 11.2.7, Serial Status Register (SSR).

**Bit 4—Error Signal Status (ERS):** In Smart Card interface mode, bit 4 indicates the status of the error signal sent back from the receiving end in transmission. Framing errors are not detected in Smart Card interface mode.

Bit 4 ERS	Description
0	Indicates normal data transmission, with no error signal returned
	[Clearing conditions] (Initial value)
	Upon reset, in standby mode, or in module stop mode
	<ul> <li>When 0 is written to ERS after reading ERS = 1</li> </ul>
1	Indicates that the receiving device sent an error signal reporting a parity error
	[Setting condition]
	When the low level of the error signal is sampled
Note:	Clearing the TE bit in SCR to 0 does not affect the ERS flag, which retains its previous state.

**Bits 3 to 0**—Operate in the same way as for the normal SCI. For details, see section 11.2.7, Serial Status Register (SSR).

However, the setting conditions for the TEND bit are as shown below.

Bit 2 TEND	Description
0	Transmission is in progress
	[Clearing condition] (Initial value)
	When 0 is written to TDRE after reading TDRE = 1
1	End of transmission
	[Setting conditions]
	Upon reset and in standby mode
	<ul> <li>When the TE bit in SCR is 0 and the ERS bit is also 0</li> </ul>
	<ul> <li>When TDRE = 1 and ERS = 0 (normal transmission) 2.5 etu after transmission of a 1-byte serial character</li> </ul>

Note: etu: Elementary Time Unit (time for transfer of 1 bit)

## 12.3 Operation

#### 12.3.1 Overview

The main functions of the Smart Card interface are as follows.

- One frame consists of 8-bit and plus a parity bit.
- In transmission, a guard time of at least 2 etu (Elementary Time Unit: the time for transfer of one bit) is left between the end of the parity bit and the start of the next frame.
- If a parity error is detected during reception, a low error signal level is output for one etu period, 10.5 etu after the start bit.
- If the error signal is sampled during transmission, the same data is transmitted automatically after the elapse of 2 etu or longer.
- Only asynchronous communication is supported; there is no clocked synchronous communication function.

## 12.3.2 Pin Connections

Figure 12.2 shows a schematic diagram of Smart Card interface related pin connections.

In communication with an IC card, since both transmission and reception are carried out on a single data transmission line, the  $TxD_0$  pin and  $RxD_0$  pin should be connected with the LSI pin. The data transmission line should be pulled up to the  $V_{cc}$  power supply with a resistor.

When the clock generated on the Smart Card interface is used by an IC card, the  $SCK_0$  pin output is input to the CLK pin of the IC card. No connection is needed if the IC card uses an internal clock.

LSI port output is used as the reset signal.

Other pins must normally be connected to the power supply or ground.

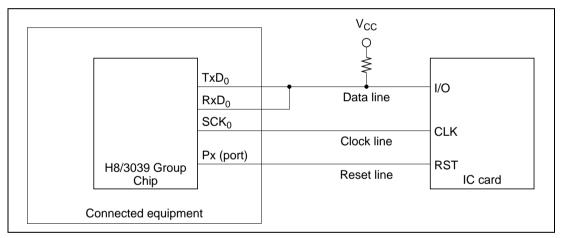


Figure 12.2 Schematic Diagram of Smart Card Interface Pin Connections

Note: If an IC card is not connected, and the TE and RE bits are both set to 1, closed transmission/reception is possible, enabling self-diagnosis to be carried out.

## 12.3.3 Data Format

Figure 12.3 shows the Smart Card interface data format. In reception in this mode, a parity check is carried out on each frame, and if an error is detected an error signal is sent back to the transmitting end, and retransmission of the data is requested. If an error signal is sampled during transmission, the same data is retransmitted.

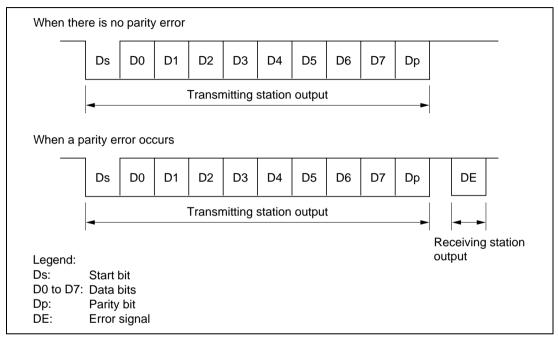


Figure 12.3 Smart Card Interface Data Format

The operation sequence is as follows.

- [1] When the data line is not in use it is in the high-impedance state, and is fixed high with a pullup resistor.
- [2] The transmitting station starts a date transfer of one frame. The data frame starts with a start bit (Ds, low-level). Then 8 data bits (D0 to D7) and a parity bit (Dp) follows.
- [3] With the Smart Card interface, the data line then returns to the high-impedance state. The data line is pulled high with a pull-up resistor.
- [4] The receiving station carries out a parity check.

If there is no parity error and the data is received normally, the receiving station waits for reception of the next data.

If a parity error occurs, however, the receiving station outputs an error signal (DE, low-level) to request retransmission of the data. After outputting the error signal for the prescribed length of time, the receiving station places the signal line in the high-impedance state again. The signal line is pulled high again by a pull-up resistor.

[5] If the transmitting station does not receive an error signal, it proceeds to transmit the next data frame.

If it does receive an error signal, however, it returns to step [2] and retransmits the erroneous data.

#### 12.3.4 Register Settings

Table 12.3 shows a bit map of the registers used by the smart card interface.

Bits indicated as 0 or 1 must be set to the value shown. The setting of other bits is described below.

	Bit							
Register	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
SMR	0	0	1	O/Ē	1	0	CKS1	CKS0
BRR	BRR7	BRR6	BRR5	BRR4	BRR3	BRR2	BRR1	BRR0
SCR	TIE	RIE	TE	RE	0	0	0	CKE0
TDR	TDR7	TDR6	TDR5	TDR4	TDR3	TDR2	TDR1	TDR0
SSR	TDRE	RDRF	ORER	ERS	PER	TEND	0	0
RDR	RDR7	RDR6	RDR5	RDR4	RDR3	RDR2	RDR1	RDR0
SCMR	_		_	_	SDIR	SINV	_	SMIF

 Table 12.3
 Smart Card Interface Register Settings

Note: —: Unused bit.

**SMR Setting:** The  $O/\overline{E}$  bit is cleared to 0 if the IC card is of the direct convention type, and set to 1 if of the inverse convention type.

Bits CKS1 and CKS0 select the clock source of the on-chip baud rate generator. See section 12.3.5, Clock.

**BRR Setting:** BRR is used to set the bit rate. See section 12.3.5, Clock, for the method of calculating the value to be set.

**SCR Setting:** The function of the TIE, RIE, TE, and RE bits is the same as for the normal SCI. For details, see section 11, Serial Communication Interface.

Bit CKE0 specifies the clock output. Set these bits to 0 if a clock is not to be output, or to 1 if a clock is to be output.

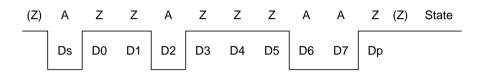
**SCMR Setting:** The SDIR bit is cleared to 0 if the IC card is of the direct convention type, and set to 1 if of the inverse convention type.

The SINV bit is cleared to 0 if the IC card is of the direct convention type, and set to 1 if of the inverse convention type.

The SMIF bit is set to 1 in the case of the Smart Card interface.

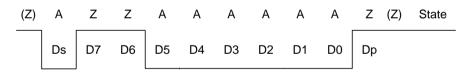
Examples of register settings and the waveform of the start character are shown below for the two types of IC card (direct convention and inverse convention).

• Direct convention (SDIR = SINV = 
$$O/\overline{E} = 0$$
)



With the direct convention type, the logic 1 level corresponds to state Z and the logic 0 level to state A, and transfer is performed in LSB-first order. The start character data above is H'3B. The parity bit is 1 since even parity is stipulated for the Smart Card.

• Inverse convention (SDIR = SINV = 
$$O/\overline{E} = 1$$
)



With the inverse convention type, the logic 1 level corresponds to state A and the logic 0 level to state Z, and transfer is performed in MSB-first order. The start character data above is H'3F. The parity bit is 0, corresponding to state Z, since even parity is stipulated for the Smart Card. With the H8/3039 Group, inversion specified by the SINV bit applies only to the data bits, D7 to D0. For parity bit inversion, the  $O/\overline{E}$  bit in SMR is set to odd parity mode (the same applies to both transmission and reception).

## 12.3.5 Clock

Only an internal clock generated by the on-chip baud rate generator can be used as the transmit/receive clock for the smart card interface. The bit rate is set with BRR and the CKS1 and CKS0 bits in SMR. The formula for calculating the bit rate is as shown below. Table 12.5 shows some sample bit rates.

If clock output is selected by setting CKE0 to 1, a clock with a frequency of 372 times the bit rate is output from the  $SCK_0$  pin.

$$B = \frac{\varphi}{1488 \times 2^{2n-1} \times (N+1)} \times 10^6$$

Where:  $N = Value \text{ set in } BRR \ (0 \le N \le 255)$ 

B = Bit rate (bit/s)

 $\phi = Operating frequency^* (MHz)$ 

n = See table 12.4

#### Table 12.4 Correspondence between n and CKS1, CKS0

n	CKS1	CKS0
0	0	0
1	-	1
2	1	0
3	-	1

Note: \* If the gear function is used to divide the system clock frequency, use the divided frequency to calculate the bit rate. The equation above applies directly to 1/1 frequency division.

Table 12.5 Examples of Bit Rate B (bit/s) for Various BRR Settings (When n = 0)	<b>Table 12.5</b>	<b>Examples of Bit Rate B</b>	(bit/s) for Various I	BRR Settings (When $n = 0$ )
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		φ (MHz)							
Ν	7.1424	10.00	10.7136	13.00	14.2848	16.00	18.00		
0	9600.0	13440.9	14400.0	17473.1	19200.0	21505.4	24193.5		
1	4800.0	6720.4	7200.0	8736.6	9600.0	10752.7	12096.8		
2	3200.0	4480.3	4800.0	5824.4	6400.0	7168.5	8064.5		

Note: Bit rates are rounded off to one decimal place.

The method of calculating the value from the operating frequency and bit rate, on the other hand, is shown below. N is an integer,  $0 \le N \le 255$ , and the smaller error is specified.

$$N = \frac{\phi}{1488 \times 2^{2n-1} \times B} \times 10^6 - 1$$

#### Table 12.6 Examples of BRR Settings for Bit Rate B (bit/s) (When n = 0)

							¢	(MHz)						
	7	.1424		10.00	10	0.7136		13.00	14	4.2848	1	16.00		18.00
bit/s	Ν	Error	Ν	Error	Ν	Error	Ν	Error	Ν	Error	Ν	Error	Ν	Error
9600	0	0.00	1	30	1	25	1	8.99	1	0.00	1	12.01	2	15.99

φ (MHz)	Maximum Bit Rate (bit/s)	Ν	n	
7.1424	9600	0	0	
10.00	13441	0	0	
10.7136	14400	0	0	
13.00	17473	0	0	
14.2848	19200	0	0	
16.00	21505	0	0	
18.00	24194	0	0	

The bit rate error is given by the following formula:

Error (%) = 
$$(\frac{\phi}{1488 \times 2^{2n-1} \times B \times (N+1)} \times 10^6 - 1) \times 100$$

## 12.3.6 Data Transfer Operations

## Initialization

Before transmitting and receiving data, initialize the SCI as described below. Initialization is also necessary when switching from transmit mode to receive mode, or vice versa.

- [1] Clear the TE and RE bits in SCR to 0.
- [2] Clear the error flags ERS, PER, and ORER in SSR to 0.
- [3] Set the O/ $\overline{E}$  bit and CKS1 and CKS0 bits in SMR. Clear the C/ $\overline{A}$ , CHR, and MP bits to 0, and set the STOP and PE bits to 1.
- [4] Set the SMIF, SDIR, and SINV bits in SCMR.When the SMIF bit is set to 1, the TxD<sub>0</sub> and RxD<sub>0</sub> pins are both switched from ports to SCI pins, and are placed in the high-impedance state.
- [5] Set the value corresponding to the bit rate in BRR.
- [6] Set the CKE0 bit in SCR. Clear the TIE, RIE, TE, RE, MPIE, TEIE and CKE1 bits to 0. If the CKE0 bit is set to 1, the clock is output from the SCK<sub>0</sub> pin.
- [7] Wait at least one bit interval, then set the TIE, RIE, TE, and RE bits in SCR. Do not set the TE bit and RE bit at the same time, except for self-diagnosis.

## Serial Data Transmission

As data transmission in smart card mode involves error signal sampling and retransmission processing, the processing procedure is different from that for the normal SCI. Figure 12.4 shows an example of the transmission processing flow, and figure 12.5 shows the relation between a transmit operation and the internal registers.

- [1] Perform Smart Card interface mode initialization as described above in Initialization.
- [2] Check that the ERS error flag in SSR is cleared to 0.
- [3] Repeat steps [2] and [3] until it can be confirmed that the TEND flag in SSR is set to 1.
- [4] Write the transmit data to TDR, clear the TDRE flag to 0, and perform the transmit operation. The TEND flag is cleared to 0.
- [5] When transmitting data continuously, go back to step [2].
- [6] To end transmission, clear the TE bit to 0.

With the above processing, interrupt servicing is possible.

If transmission ends and the TEND flag is set to 1 while the TIE bit is set to 1 and interrupt requests are enabled, a transmit data empty interrupt (TXI) request will be generated. If an error occurs in transmission and the ERS flag is set to 1 while the RIE bit is set to 1 and interrupt requests are enabled, a transfer error interrupt (ERI) request will be generated.

For details, see the following Interrupt Operations.

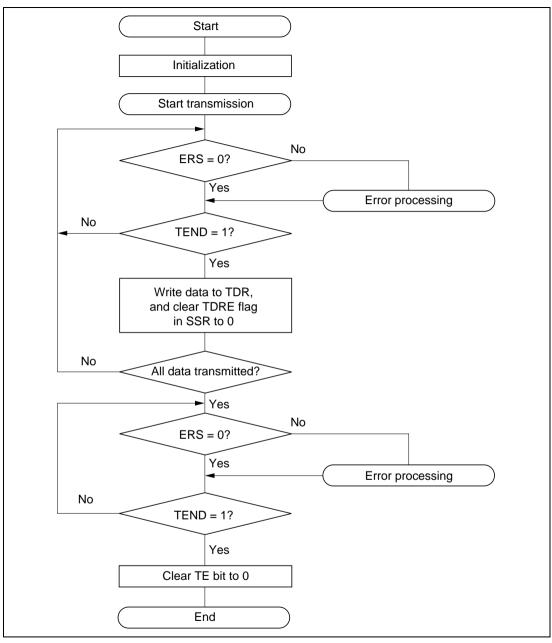


Figure 12.4 Example of Transmission Processing Flow

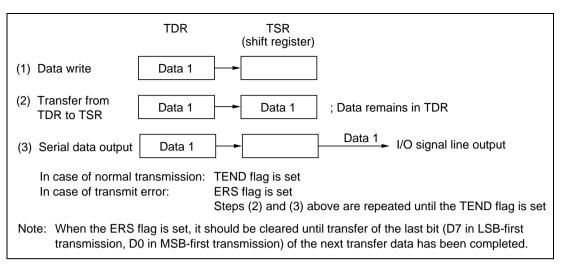


Figure 12.5 Relation Between Transmit Operation and Internal Registers

## **Serial Data Reception**

Data reception in Smart Card mode uses the same processing procedure as for the normal SCI. Figure 12.6 shows an example of the transmission processing flow.

- [1] Perform Smart Card interface mode initialization as described above in Initialization.
- [2] Check that the ORER flag and PER flag in SSR are cleared to 0. If either flag is set, perform the appropriate receive error processing, then clear both the ORER and the PER flag to 0.
- [3] Repeat steps [2] and [3] until it can be confirmed that the RDRF flag is set to 1.
- [4] Read the receive data from RDR.
- [5] When receiving data continuously, clear the RDRF flag to 0 and go back to step [2].
- [6] To end reception, clear the RE bit to 0.

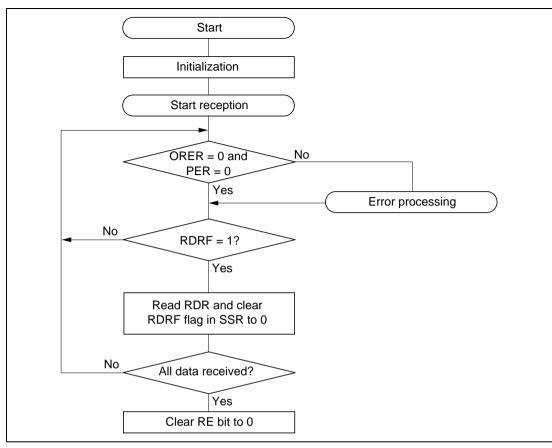


Figure 12.6 Example of Reception Processing Flow

With the above processing, interrupt servicing is possible.

If reception ends and the RDRF flag is set to 1 while the RIE bit is set to 1 and interrupt requests are enabled, a receive data full interrupt (RXI) request will be generated. If an error occurs in reception and either the ORER flag or the PER flag is set to 1, a transfer error interrupt (ERI) request will be generated.

For details, see Interrupt Operation below.

If a parity error occurs during reception and the PER is set to 1, the received data is still transferred to RDR, and therefore this data can be read.

## Mode Switching Operation

When switching from receive mode to transmit mode, first confirm that the receive operation has been completed, then start from initialization, clearing RE bit to 0 and setting TE bit to 1. The RDRF flag or the PER and ORER flags can be used to check that the receive operation has been completed.

When switching from transmit mode to receive mode, first confirm that the transmit operation has been completed, then start from initialization, clearing TE bit to 0 and setting RE bit to 1. The TEND flag can be used to check that the transmit operation has been completed.

## **Interrupt Operation**

There are three interrupt sources in smart card interface mode: transmit data empty interrupt (TXI) requests, transfer error interrupt (ERI) requests, and receive data full interrupt (RXI) requests. The transmit end interrupt (TEI) request is not used in this mode.

When the TEND flag in SSR is set to 1, a TXI interrupt request is generated.

When the RDRF flag in SSR is set to 1, an RXI interrupt request is generated.

When any of flags ORER, PER, and ERS in SSR is set to 1, an ERI interrupt request is generated. The relationship between the operating states and interrupt sources is shown in table 12.8.

Table 12.0 Small Call Moue Operating States and Interrupt Sources	<b>Table 12.8</b>	Smart Card Mode Operating States and Interrupt Sources
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<b>Operating State</b>		Flag	Mask Bit	Interrupt Source
Transmit Mode	Normal operation	TEND	TIE	TXI
	Error	ERS	RIE	ERI
Receive Mode	Normal operation	RDRF	RIE	RXI
	Error	PER, ORER	RIE	ERI

# 12.4 Usage Note

The following points should be noted when using the SCI as a smart card interface.

#### Receive Data Sampling Timing and Reception Margin in Smart Card Interface Mode

In smart card interface mode, the SCI operates on a basic clock with a frequency of 372 times the transfer rate.

In reception, the SCI samples the falling edge of the start bit using the basic clock, and performs internal synchronization. Receive data is latched internally at the rising edge of the 186th pulse of the basic clock. This is illustrated in figure 12.7.

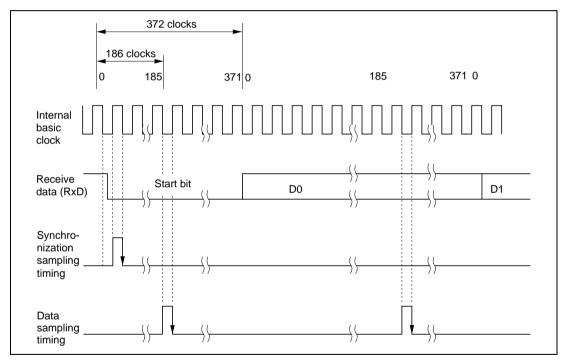


Figure 12.7 Receive Data Sampling Timing in Smart Card Mode

Thus the reception margin in smart card interface mode is given by the following formula.

$$M = \left| (0.5 - \frac{1}{2N}) - (L - 0.5) F - \frac{|D - 0.5|}{N} (1 + F) \right| \times 100\%$$

Where M: Reception margin (%)

- N: Ratio of bit rate to clock (N = 372)
- D: Clock duty (D = 0 to 1.0)
- L: Frame length (L = 10)
- F: Absolute value of clock frequency deviation

Assuming values of F = 0 and D = 0.5 in the above formula, the reception margin formula is as follows.

When D = 0.5 and F = 0,

#### **Retransfer Operations**

Retransfer operations are performed by the SCI in receive mode and transmit mode as described below.

• Retransfer operation when SCI is in receive mode

Figure 12.8 illustrates the retransfer operation when the SCI is in receive mode.

- [1] If an error is found when the received parity bit is checked, the PER bit in SSR is automatically set to 1. If the RIE bit in SCR is enabled at this time, an ERI interrupt request is generated. The PER bit in SSR should be kept cleared to 0 until the next parity bit is sampled.
- [2] The RDRF bit in SSR is not set for a frame in which an error has occurred.
- [3] If no error is found when the received parity bit is checked, the PER bit in SSR is not set to 1.
- [4] If no error is found when the received parity bit is checked, the receive operation is judged to have been completed normally, and the RDRF flag in SSR is automatically set to 1. If the RIE bit in SCR is enabled at this time, an RXI interrupt request is generated.
- [5] When a normal frame is received, the pin retains the high-impedance state at the timing for error signal transmission.

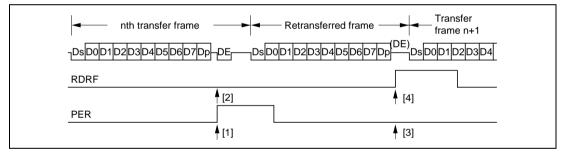


Figure 12.8 Retransfer Operation in SCI Receive Mode

- Retransfer operation when SCI is in transmit mode Figure 12.9 illustrates the retransfer operation when the SCI is in transmit mode.
- [6] If an error signal is sent back from the receiving end after transmission of one frame is completed, the ERS bit in SSR is set to 1. If the RIE bit in SCR is enabled at this time, an ERI interrupt request is generated. The ERS bit in SSR should be kept cleared to 0 until the next parity bit is sampled.
- [7] The TEND bit in SSR is not set for a frame for which an error signal indicating an abnormality is received.
- [8] If an error signal is not sent back from the receiving end, the ERS bit in SSR is not set.
- [9] If an error signal is not sent back from the receiving end, transmission of one frame, including a retransfer, is judged to have been completed, and the TEND bit in SSR is set to 1. If the TIE bit in SCR is enabled at this time, a TXI interrupt request is generated.

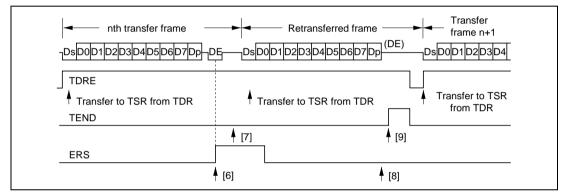


Figure 12.9 Retransfer Operation in SCI Transmit Mode

# Section 13 A/D Converter

# 13.1 Overview

The H8/3039 Group includes a 10-bit successive-approximations A/D converter with a selection of up to eight analog input channels.

When the A/D converter is not used, it can be halted independently to conserve power. For details see section 17.6, Module Standby Function.

#### 13.1.1 Features

A/D converter features are listed below.

- 10-bit resolution
- Eight input channels
- Selectable analog conversion voltage range The analog voltage conversion range can be programmed by input of an analog reference voltage at the AV<sub>CC</sub> pin.
- High-speed conversion Conversion time: minimum 7.4 µs per channel (with 18 MHz system clock)
- Two conversion modes

Single mode: A/D conversion of one channel

Scan mode: continuous conversion on one to four channels

• Four 16-bit data registers

A/D conversion results are transferred for storage into data registers corresponding to the channels.

- Sample-and-hold function
- A/D conversion can be externally triggered
- A/D interrupt requested at end of conversion At the end of A/D conversion, an A/D end interrupt (ADI) can be requested.

#### 13.1.2 Block Diagram

Figure 13.1 shows a block diagram of the A/D converter.

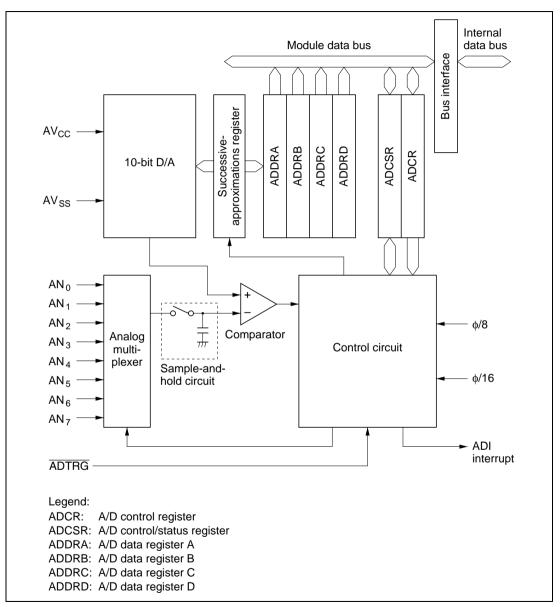


Figure 13.1 A/D Converter Block Diagram

#### 13.1.3 Input Pins

Table 13.1 summarizes the A/D converter's input pins. The eight analog input pins are divided into two groups: group 0 (AN<sub>0</sub> to AN<sub>3</sub>), and group 1 (AN<sub>4</sub> to AN<sub>7</sub>). AV<sub>cc</sub> and AV<sub>ss</sub> are the power supply for the analog circuits in the A/D converter.

# Table 13.1 A/D Converter Pins

Abbrevi- ation	I/O	Function
$AV_{cc}$	Input	Analog power supply and reference voltage
AV <sub>ss</sub>	Input	Analog ground and reference voltage
AN <sub>0</sub>	Input	Group 0 analog inputs
AN <sub>1</sub>	Input	
AN <sub>2</sub>	Input	
AN <sub>3</sub>	Input	
AN <sub>4</sub>	Input	Group 1 analog inputs
AN <sub>5</sub>	Input	
AN <sub>6</sub>	Input	
AN <sub>7</sub>	Input	
ADTRG	Input	External trigger input for starting A/D conversion
	ation           AV <sub>cc</sub> AV <sub>ss</sub> AN <sub>0</sub> AN <sub>1</sub> AN <sub>2</sub> AN <sub>3</sub> AN <sub>4</sub> AN <sub>5</sub> AN <sub>6</sub> AN <sub>7</sub>	ationI/O $AV_{cc}$ Input $AV_{ss}$ Input $AN_{o}$ Input $AN_{1}$ Input $AN_{1}$ Input $AN_{2}$ Input $AN_{3}$ Input $AN_{3}$ Input $AN_{4}$ Input $AN_{5}$ Input $AN_{6}$ Input $AN_{7}$ Input

## **13.1.4** Register Configuration

Table 13.2 summarizes the A/D converter's registers.

# Table 13.2 A/D Converter Registers

Address*1	Name	Abbreviation	R/W	Initial Value
H'FFE0	A/D data register A (high)	ADDRAH	R	H'00
H'FFE1	A/D data register A (low)	ADDRAL	R	H'00
H'FFE2	A/D data register B (high)	ADDRBH	R	H'00
H'FFE3	A/D data register B (low)	ADDRBL	R	H'00
H'FFE4	A/D data register C (high)	ADDRCH	R	H'00
H'FFE5	A/D data register C (low)	ADDRCL	R	H'00
H'FFE6	A/D data register D (high)	ADDRDH	R	H'00
H'FFE7	A/D data register D (low)	ADDRDL	R	H'00
H'FFE8	A/D control/status register	ADCSR	R/(W)* <sup>2</sup>	H'00
H'FFE9	A/D control register	ADCR	R/W	H'7F

Notes: 1. Lower 16 bits of the address

2. Only 0 can be written in bit 7 to clear the flag.

# **13.2** Register Descriptions

Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
ADDRn	AD9	AD8	AD7	AD6	AD5	AD4	AD3	AD2	AD1	AD0	—	—	_	_	_	—
Initial value	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0	0
Read/Write (n = A to D)	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R	R
	<b>A/D conversion data</b> 10-bit data giving an A/D conversion result							Re	eserv	ed bi	its					

## 13.2.1 A/D Data Registers A to D (ADDRA to ADDRD)

The four A/D data registers (ADDRA to ADDRD) are 16-bit read-only registers that store the results of A/D conversion.

An A/D conversion produces 10-bit data, which is transferred for storage into the A/D data register corresponding to the selected channel. The upper 8 bits of the result are stored in the upper byte of the A/D data register. The lower 2 bits are stored in the lower byte. Bits 5 to 0 of an A/D data register are reserved bits that always read 0. Table 13.3 indicates the pairings of analog input channels and A/D data registers.

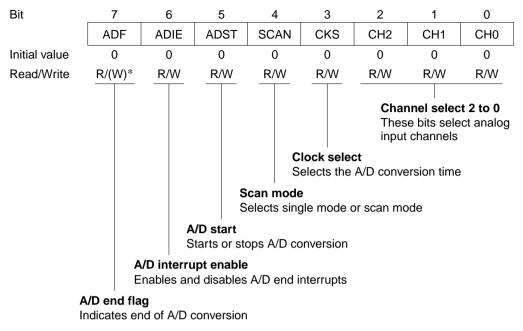
The CPU can always read the A/D data registers. The upper byte can be read directly, but the lower byte is read through a temporary register (TEMP). For details see section 13.3, CPU Interface.

The A/D data registers are initialized to H'0000 by a reset and in standby mode.

#### Table 13.3 Analog Input Channels and A/D Data Registers

Ai	nalog Input Channel		
Group 0	Group 1	A/D Data Register	
AN0	AN4	ADDRA	
AN1	AN5	ADDRB	
AN2	AN6	ADDRC	
AN3	AN7	ADDRD	

## 13.2.2 A/D Control/Status Register (ADCSR)



Note: \* Only 0 can be written to clear the flag.

ADCSR is an 8-bit readable/writable register that selects the mode and controls the A/D converter. ADCSR is initialized to H'00 by a reset and in standby mode.

Bit 7—A/D End Flag (ADF): Indicates the end of A/D conversion.

Bit 7 ADF	Description	
0	[Clearing condition]	(Initial value)
	Cleared by reading ADF while ADF = 1, then writing 0 in ADF	
1	[Setting conditions]	
	Single mode: A/D conversion ends	
	Scan mode: A/D conversion ends in all selected channels	

**Bit 6—A/D Interrupt Enable (ADIE):** Enables or disables the interrupt (ADI) requested at the end of A/D conversion.

Bit 6 ADIE	Description	
0	A/D end interrupt request (ADI) is disabled	(Initial value)
1	A/D end interrupt request (ADI) is enabled	

**Bit 5—A/D Start (ADST):** Starts or stops A/D conversion. The ADST bit remains set to 1 during A/D conversion. It can also be set to 1 by external trigger input at the ADTRG pin.

Bit 5 ADST	Description		
0	A/D conversion	on is stopped	(Initial value)
1	Single mode:	A/D conversion starts; ADST is automatically cleared to conversion ends.	0 when
	Scan mode:	A/D conversion starts and continues, cycling among the channels, until ADST is cleared to 0 by software, by a retransition to standby mode.	

**Bit 4—Scan Mode (SCAN):** Selects single mode or scan mode. For further information on operation in these modes, see section 13.4, Operation. Clear the ADST bit to 0 before switching the conversion mode.

Bit 4 SCAN	Description	
0	Single mode	(Initial value)
1	Scan mode	

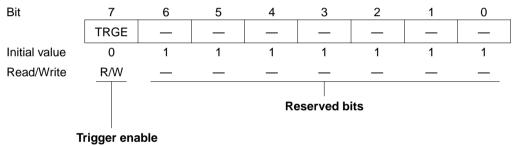
**Bit 3—Clock Select (CKS):** Selects the A/D conversion time. Clear the ADST bit to 0 before switching the conversion time.

Bit 3 CKS	Description	
0	Conversion time = 266 states (maximum)	(Initial value)
1	Conversion time = 134 states (maximum)	

**Bits 2 to 0—Channel Select 2 to 0 (CH2 to CH0):** These bits and the SCAN bit select the analog input channels. Clear the ADST bit to 0 before changing the channel selection.

Group Selection	Chan	nel Selection	Description			
CH2	CH1	CH0	Single Mode	Scan Mode		
0	0	0	$AN_{0}$ (Initial value)	AN <sub>o</sub>		
		1	AN <sub>1</sub>	AN <sub>0</sub> , AN <sub>1</sub>		
	1	0	AN <sub>2</sub>	AN <sub>0</sub> to AN <sub>2</sub>		
		1	AN <sub>3</sub>	$AN_0$ to $AN_3$		
1	0	0	AN <sub>4</sub>	AN <sub>4</sub>		
		1	AN <sub>5</sub>	AN <sub>4</sub> , AN <sub>5</sub>		
	1	0	AN <sub>6</sub>	$AN_4$ to $AN_6$		
		1	AN <sub>7</sub>	AN <sub>4</sub> to AN <sub>7</sub>		

#### 13.2.3 A/D Control Register (ADCR)



Enables or disables external triggering of A/D conversion

ADCR is an 8-bit readable/writable register that enables or disables external triggering of A/D conversion. ADCR is initialized to H'7F by a reset and in standby mode.

Bit 7—Trigger Enable (TRGE): Enables or disables external triggering of A/D conversion.

Bit 7 TRGE	Description	
0	A/D conversion cannot be externally triggered	(Initial value)
1	A/D conversion starts at the falling edge of the external trigger signal $\dot{a}$	ADTRG)

Bits 6 to 0—Reserved: These bits cannot be modified and are always read as 1.

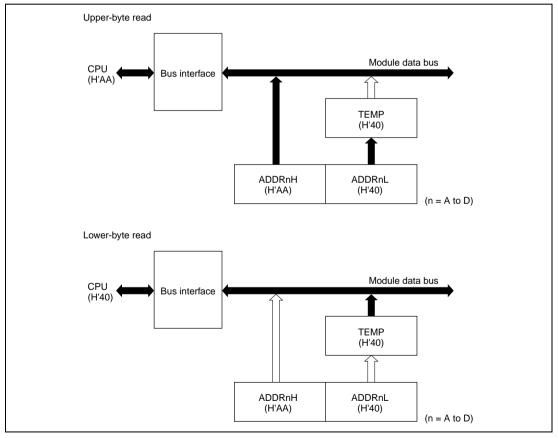
# 13.3 CPU Interface

ADDRA to ADDRD are 16-bit registers, but they are connected to the CPU by an 8-bit data bus. Therefore, although the upper byte can be accessed directly by the CPU, the lower byte is read through an 8-bit temporary register (TEMP).

An A/D data register is read as follows. When the upper byte is read, the upper-byte value is transferred directly to the CPU and the lower-byte value is transferred into TEMP. Next, when the lower byte is read, the TEMP contents are transferred to the CPU.

When reading an A/D data register, always read the upper byte before the lower byte. It is possible to read only the upper byte, but if only the lower byte is read, incorrect data may be obtained.

Figure 13.2 shows the data flow for access to an A/D data register.





# 13.4 Operation

The A/D converter operates by successive approximations with 10-bit resolution. It has two operating modes: single mode and scan mode.

# 13.4.1 Single Mode (SCAN = 0)

Single mode should be selected when only one A/D conversion on one channel is required. A/D conversion starts when the ADST bit is set to 1 by software, or by external trigger input. The ADST bit remains set to 1 during A/D conversion and is automatically cleared to 0 when conversion ends.

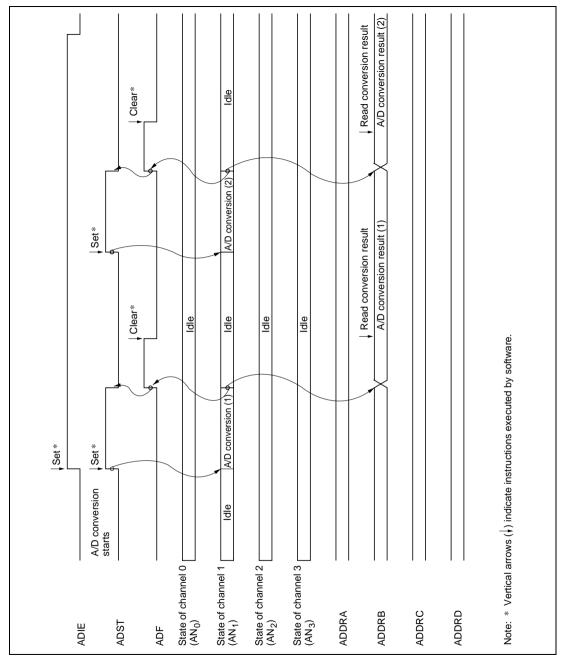
When conversion ends the ADF bit is set to 1. If the ADIE bit is also set to 1, an ADI interrupt is requested at this time. To clear the ADF flag to 0, first read ADCSR, then write 0 in ADF.

When the mode or analog input channel must be switched during analog conversion, to prevent incorrect operation, first clear the ADST bit to 0 in ADCSR to halt A/D conversion. After making the necessary changes, set the ADST bit to 1 to start A/D conversion again. The ADST bit can be set at the same time as the mode or channel is changed.

Typical operations when channel 1  $(AN_1)$  is selected in single mode are described next.

Figure 13.3 shows a timing diagram for this example.

- 1. Single mode is selected (SCAN = 0), input channel  $AN_1$  is selected (CH2 = CH1 = 0, CH0 = 1), the A/D interrupt is enabled (ADIE = 1), and A/D conversion is started (ADST = 1).
- 2. When A/D conversion is completed, the result is transferred into ADDRB. At the same time the ADF flag is set to 1, the ADST bit is cleared to 0, and the A/D converter becomes idle.
- 3. Since ADF = 1 and ADIE = 1, an ADI interrupt is requested.
- 4. The A/D interrupt handling routine starts.
- 5. The routine reads ADCSR, then writes 0 in the ADF flag.
- 6. The routine reads and processes the conversion result (ADDRB).
- 7. Execution of the A/D interrupt handling routine ends. After that, if the ADST bit is set to 1, A/D conversion starts again and steps 2 to 7 are repeated.



# Figure 13.3 Example of A/D Converter Operation (Single Mode, Channel 1 Selected)

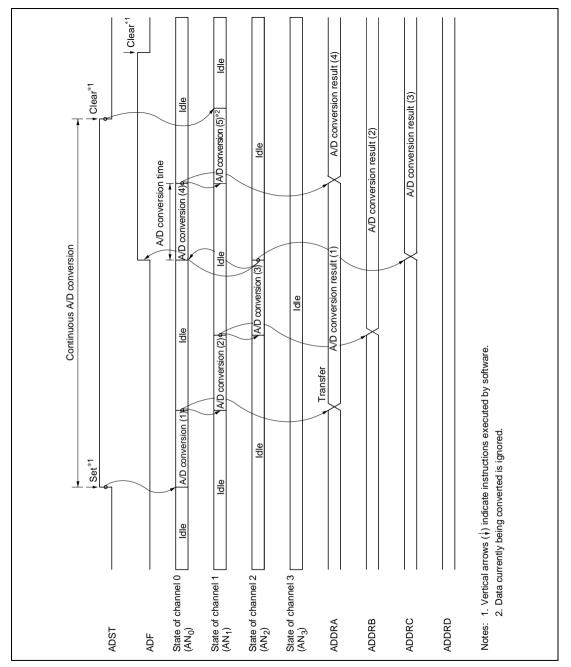
# **13.4.2** Scan Mode (SCAN = 1)

Scan mode is useful for monitoring analog inputs in a group of one or more channels. When the ADST bit is set to 1 by software or external trigger input, A/D conversion starts on the first channel in the group  $(AN_0 \text{ when } CH2 = 0, AN_4 \text{ when } CH2 = 1)$ . When two or more channels are selected, after conversion of the first channel ends, conversion of the second channel  $(AN_1 \text{ or } AN_5)$  starts immediately. A/D conversion continues cyclically on the selected channels until the ADST bit is cleared to 0. The conversion results are transferred for storage into the A/D data registers corresponding to the channels.

When the mode or analog input channel selection must be changed during analog conversion, to prevent incorrect operation, first clear the ADST bit to 0 in ADCSR to halt A/D conversion. After making the necessary changes, set the ADST bit to 1. A/D conversion will start again from the first channel in the group. The ADST bit can be set at the same time as the mode or channel selection is changed.

Typical operations when three channels in group 0 ( $AN_0$  to  $AN_2$ ) are selected in scan mode are described next. Figure 13.4 shows a timing diagram for this example.

- 1. Scan mode is selected (SCAN = 1), scan group 0 is selected (CH2 = 0), analog input channels  $AN_0$  to  $AN_2$  are selected (CH1 = 1, CH0 = 0), and A/D conversion is started (ADST = 1).
- 2. When A/D conversion of the first channel (AN<sub>0</sub>) is completed, the result is transferred into ADDRA. Next, conversion of the second channel (AN<sub>1</sub>) starts automatically.
- 3. Conversion proceeds in the same way through the third channel  $(AN_2)$ .
- 4. When conversion of all selected channels  $(AN_0 \text{ to } AN_2)$  is completed, the ADF flag is set to 1 and conversion of the first channel  $(AN_0)$  starts again. If the ADIE bit is set to 1, an ADI interrupt is requested at this time.
- 5. Steps 2 to 4 are repeated as long as the ADST bit remains set to 1. When the ADST bit is cleared to 0, A/D conversion stops. After that, if the ADST bit is set to 1, A/D conversion starts again from the first channel  $(AN_0)$ .



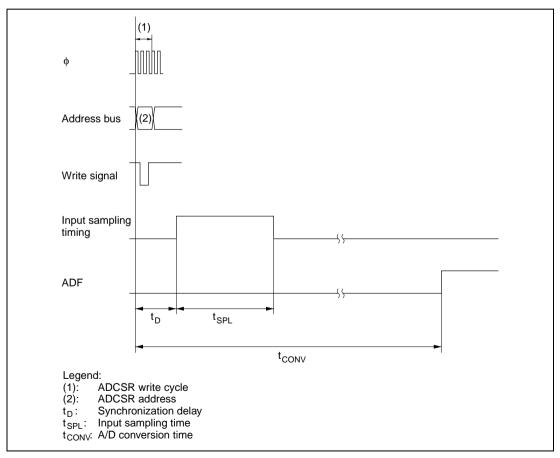
# Figure 13.4 Example of A/D Converter Operation (Scan Mode, Channels AN<sub>0</sub> to AN<sub>2</sub> Selected)

## 13.4.3 Input Sampling and A/D Conversion Time

The A/D converter has a built-in sample-and-hold circuit. The A/D converter samples the analog input at a time  $t_{\rm D}$  after the ADST bit is set to 1, then starts conversion. Figure 13.5 shows the A/D conversion timing. Table 13.4 indicates the A/D conversion time.

As indicated in figure 13.5, the A/D conversion time includes  $t_{D}$  and the input sampling time. The length of  $t_{D}$  varies depending on the timing of the write access to ADCSR. The total conversion time therefore varies within the ranges indicated in table 13.4.

In scan mode, the values given in table 13.4 apply to the first conversion. In the second and subsequent conversions the conversion time is fixed at 256 states when CKS = 0 or 128 states when CKS = 1.





Rev.3.00 Mar. 26, 2007 Page 428 of 682 REJ09B0353-0300

		CKS = 0			CKS = 1		
	Symbol	Min	Тур	Max	Min	Тур	Max
Synchronization delay	t <sub>D</sub>	10	_	17	6		9
Input sampling time	t <sub>spl</sub>	_	63	_	_	31	_
A/D conversion time	t <sub>conv</sub>	259	_	266	131	_	134

#### Table 13.4 A/D Conversion Time (Single Mode)

Note: Values in the table are numbers of states.

#### 13.4.4 External Trigger Input Timing

A/D conversion can be externally triggered. When the TRGE bit is set to 1 in ADCR, external trigger input is enabled at the  $\overline{\text{ADTRG}}$  pin. A high-to-low transition at the  $\overline{\text{ADTRG}}$  pin sets the ADST bit to 1 in ADCSR, starting A/D conversion. Other operations, in both single and scan modes, are the same as if the ADST bit had been set to 1 by software. Figure 13.6 shows the timing.

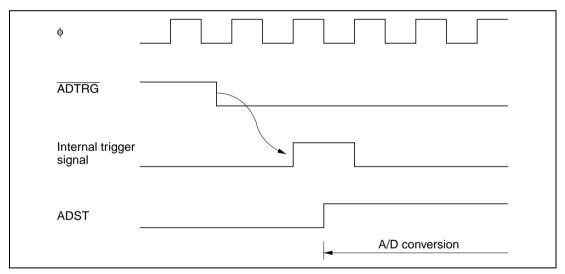


Figure 13.6 External Trigger Input Timing

# 13.5 Interrupts

The A/D converter generates an interrupt (ADI) at the end of A/D conversion. The ADI interrupt request can be enabled or disabled by the ADIE bit in ADCSR.

# 13.6 Usage Notes

The following points should be noted when using the A/D converter.

## Setting Range of Analog Power Supply and Other Pins

(1) Analog input voltage range

The voltage applied to analog input pins AN<sub>0</sub> to AN<sub>7</sub> during A/D conversion should be in the range AV<sub>ss</sub>  $\leq$  ANn  $\leq$  AV<sub>cc</sub>.

(2) Relation between AV $_{\rm CC},$  AV $_{\rm ss}$  and V $_{\rm CC},$  V $_{\rm ss}$ 

As the relationship between  $AV_{cc}$ ,  $AV_{ss}$  and  $V_{cc}$ ,  $V_{ss}$ , set  $AV_{ss} = V_{ss}$ . If the A/D converter is not used, the  $AV_{cc}$  and  $AV_{ss}$  pins must on no account be left open.

If conditions (1) and (2) above are not met, the reliability of the device may be adversely affected.

## Notes on Board Design

In board design, digital circuitry and analog circuitry should be as mutually isolated as possible, and layout in which digital circuit signal lines and analog circuit signal lines cross or are in close proximity should be avoided as far as possible. Failure to do so may result in incorrect operation of the analog circuitry due to inductance, adversely affecting A/D conversion values.

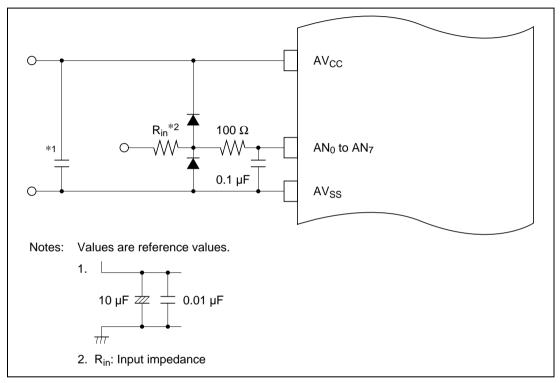
Also, digital circuitry must be isolated from the analog input signals ( $AN_0$  to  $AN_7$ ), and analog power supply and reference voltage ( $AV_{cc}$ ) by the analog ground ( $AV_{ss}$ ). Also, the analog ground ( $AV_{ss}$ ) should be connected at one point to a stable digital ground ( $V_{ss}$ ) on the board.

## Notes on Noise Countermeasures

A protection circuit connected to prevent damage due to an abnormal voltage such as an excessive surge at the analog input pins  $(AN_0 \text{ to } AN_7)$  and analog power supply  $(AV_{cc})$  should be connected between  $AV_{cc}$  and  $AV_{ss}$  as shown in figure 13.7.

Also, the bypass capacitors connected to  $AV_{cc}$  and the filter capacitor connected to  $AN_0$  to  $AN_7$  must be connected to  $AV_{ss}$ .

If a filter capacitor is connected as shown in figure 13.7, the input currents at the analog input pins  $(AN_0 \text{ to } AN_7)$  are averaged, and so an error may arise. Also, when A/D conversion is performed frequently, as in scan mode, if the current charged and discharged by the capacitance of the sample-and-hold circuit in the A/D converter exceeds the current input via the input impedance  $(R_{in})$ , an error will arise in the analog input pin voltage. Therefore careful consideration is required when deciding the circuit constants.

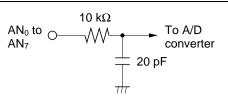


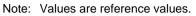


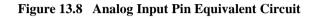
#### Table 13.5 Analog Pin Specifications

Item	Min	Max	Unit	
Analog input capacitance	—	20	pF	
Permissible signal source impedance	—	10*	kΩ	

Note: \* When  $V_{cc}$  = 4.0 V to 5.5 V and  $\phi \le$  12 MHz







#### **A/D Conversion Precision Definitions**

H8/3039 Group A/D conversion precision definitions are given below.

• Resolution

The number of A/D converter digital output codes

Offset error

The deviation of the analog input voltage value from the ideal A/D conversion characteristic when the digital output changes from the minimum voltage value B'0000000000 (H'000) to B'0000000001 (H'001) (see figure 13.10).

Full-scale error

The deviation of the analog input voltage value from the ideal A/D conversion characteristic when the digital output changes from B'111111110 (H'3FE) to B'1111111111 (H'3FF) (see figure 13.10).

• Quantization error

The deviation inherent in the A/D converter, given by 1/2 LSB (see figure 13.9).

• Nonlinearity error

The error with respect to the ideal A/D conversion characteristic between the zero voltage and the full-scale voltage. Does not include the offset error, full-scale error, or quantization error.

• Absolute precision

The deviation between the digital value and the analog input value. Includes the offset error, full-scale error, quantization error, and nonlinearity error.

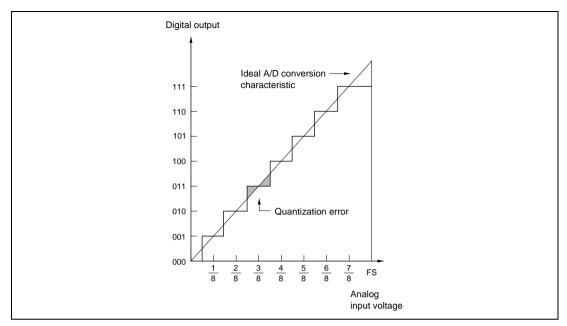


Figure 13.9 A/D Conversion Precision Definitions (1)

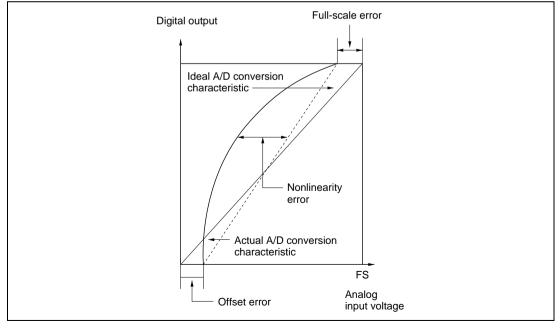


Figure 13.10 A/D Conversion Precision Definitions (2)

#### Permissible Signal Source Impedance

H8/3039 Group analog input is designed so that conversion precision is guaranteed for an input signal for which the signal source impedance is  $10 \text{ k}\Omega$  or less. This specification is provided to enable the A/D converter's sample-and-hold circuit input capacitance to be charged within the sampling time; if the sensor output impedance exceeds  $10 \text{ k}\Omega$ , charging may be insufficient and it may not be possible to guarantee the A/D conversion precision.

When converting in the single mode, if a large capacitance is provided externally, the input load will essentially comprise only the internal input resistance of  $10 \text{ k}\Omega$ , and the signal source impedance is ignored.

However, since a low-pass filter effect is obtained in this case, it may not be possible to follow an analog signal with a large differential coefficient (e.g., voltage regulation 5 mV/ $\mu$ s or greater).

When converting a high-speed analog signal and when performing conversion in the scan mode, a low-impedance buffer should be inserted.

#### **Influences on Absolute Precision**

Adding capacitance results in coupling with GND, and therefore noise in GND may adversely affect absolute precision. Be sure to make the connection to an electrically stable GND such as  $AV_{ss}$ .

Care is also required to insure that filter circuits do not communicate with digital signals on the mounting board, thus acting as antennas.

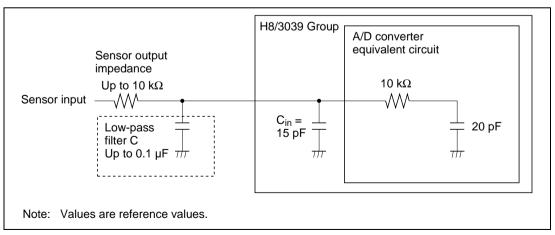


Figure 13.11 Example of Analog Input Circuit

# Section 14 RAM

# 14.1 Overview

The H8/3039 has 4 kbytes of on-chip static RAM, H8/3038 has 2 kbytes, H8/3037 has 1 kbyte, and H8/3036 has 512 bytes. The RAM is connected to the CPU by a 16-bit data bus. The CPU accesses both byte data and word data in two states, making the RAM suitable for rapid data transfer.

The RAM enable bit (RAME) in the system control register (SYSCR) can enable or disable the on-chip RAM.

Table 14.1 shows the address of the on-chip RAM in each operating mode.

Mode	H8/3039	H8/3038	H8/3037	H8/3036
	(4 kbytes)	(2 kbytes)	(1k byte)	(512 bytes)
Modes 1, 5, 7	H'FEF10 to	H'FF710 to	H'FFB10 to	H'FFD10 to
	H'FFF0F	H'FFF0F	H'FFF0F	H'FFF0F
Mode 3	H'FFEF10 to	H'FFF710 to	H'FFFB10 to	H'FFFD10 to
	H'FFFF0F	H'FFFF0F	H'FFFF0F	H'FFFF0F
Mode 6	H'F710 to	H'F710 to	H'FB10 to	H'FD10 to
	H'FF0F	H'FF0F	H'FF0F	H'FF0F

 Table 14.1
 The Address of the On-Chip RAM in Each Operating Mode

#### 14.1.1 Block Diagram

Figure 14.1 shows a block diagram of the on-chip RAM.

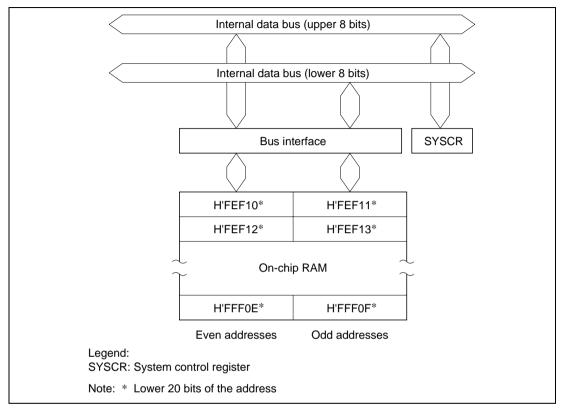


Figure 14.1 RAM Block Diagram (H8/3039 in Modes 1, 5 and 7)

## 14.1.2 Register Configuration

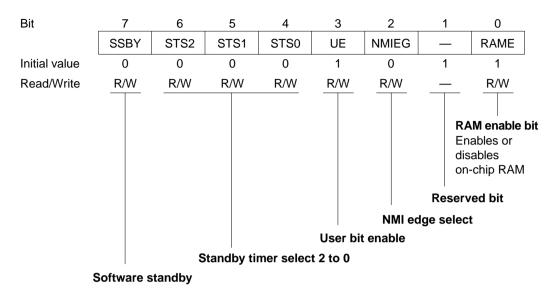
The on-chip RAM is controlled by the system control register (SYSCR). Table 14.2 gives the address and initial value of SYSCR.

#### Table 14.2 RAM Control Register

Address*	Name	Abbreviation	R/W	Initial Value
H'FFF2	System control register	SYSCR	R/W	H'0B
Noto: * I	ower 16 bits of the address			

Note: \* Lower 16 bits of the address

# 14.2 System Control Register (SYSCR)



One function of SYSCR is to enable or disable access to the on-chip RAM. The on-chip RAM is enabled or disabled by the RAME bit in SYSCR. For details about the other bits, see section 3.3, System Control Register (SYSCR).

**Bit 0—RAM Enable (RAME):** Enables or disables the on-chip RAM. The RAME bit is initialized at the rising edge of the input at the  $\overline{\text{RES}}$  pin. It is not initialized in software standby mode.

Bit 0 RAME	Description	
0	On-chip RAM is disabled	
1	On-chip RAM is enabled	(Initial value)

# 14.3 Operation

When the RAME bit is set to 1, the on-chip RAM is enabled. This LSI can access the on-chip RAM when addressing the addresses shown in table 14.1 in each operation mode. When the RAME bit is cleared to 0 in modes 1, 3, and 5 (expanded modes), external address space is accessed. When the RAME bit is cleared to 0 in modes 6 and 7 (single-chip modes), the on-chip RAM is not accessed. Read operation always reads H'FF and disables writing.

The on-chip RAM is connected to the CPU by a 16-bit wide data bus and can be read and written on a byte or a word basis.

Byte data can be accessed in two states using the higher 8 bits of the data bus. Word data beginning from an even address can be accessed in two states using the 16-bit data bus.

# Section 15 ROM

# 15.1 Overview

The H8/3039 has 128 kbytes of on-chip ROM (flash memory or mask ROM), the H8/3038 has 64 kbytes, the H8/3037 has 32 kbytes and H8/3036 has 16 kbytes. The ROM is connected to the CPU by a 16-bit data bus. The CPU accesses both byte and word data in two states, enabling rapid data transfer.

The mode pins  $(MD_2 \text{ to } MD_0)$  can be set to enable or disable the on-chip ROM. See table 15.1.

The on-chip flash memory product (H8/3039F-ZTAT) can be erased and programmed on-board as well as with a general-purpose PROM programmer.

## Table 15.1 Operating Mode and ROM

	I	Mode F		
Mode	MD <sub>2</sub>	$\mathbf{MD}_{1}$	MD₀	On-Chip ROM
Mode 1 (1-Mbyte expanded mode with on-chip ROM disabled)	0	0	1	Disabled (external address area)
Mode 2 (1-Mbyte expanded mode with on-chip ROM disabled)*	0	1	0	_
Mode 3 (16-Mbyte expanded mode with on-chip ROM disabled)	0	1	1	_
Mode 4 (16-Mbyte expanded mode with on-chip ROM disabled)*	1	0	0	_
Mode 5 (16-Mbyte expanded mode with on-chip ROM enabled)	1	0	1	Enabled
Mode 6 (single-chip normal mode)	1	1	0	_
Mode 7 (single-chip advanced mode)	1	1	1	_

Note: \* Modes 2 and 4 cannot be used with this LSI. Do not set the mode pin to mode 2 or 4.

# **15.2** Overview of Flash Memory

# 15.2.1 Features

The features of the flash memory are summarized below.

- Four flash memory operating modes
  - Program mode
  - Erase mode
  - Program-verify mode
  - Erase-verify mode
- Programming/erase methods

The flash memory is programmed 32 bytes at a time. Erasing is performed by block erase. The block to be erased can be specified by setting the corresponding bit. There are block areas of 32 kbytes  $\times$  3 blocks, 28 kbytes  $\times$  1 block, and 1 kbyte  $\times$  4 blocks.

• Programming/erase times

The flash memory programming time is 10 ms (typ.) for simultaneous 32-byte programming, equivalent to 300  $\mu$ s (typ.) per byte, and the erase time is 100 ms (typ.) per block.

• Reprogramming capability

The flash memory can be reprogrammed up to 100 times.

On-board programming modes

There are two modes in which flash memory can be programmed/erased/verified on-board:

- Boot mode
- User program mode
- Automatic bit rate adjustment

With data transfer in boot mode, the this LSI's bit rate can be automatically adjusted to match the transfer bit rate of the host (9600 bps and 4800 bps).

• Flash memory emulation by RAM

Part of the RAM area can be overlapped onto flash memory, to emulate flash memory updates in real time.

PROM mode

Flash memory can be programmed/erased in PROM mode, using a PROM programmer, as well as in on-board programming mode.

• Protect modes

There are three protect modes, hardware, software, and error protect, which allow protected status to be designated for flash memory program/erase/verify operations.

#### 15.2.2 Block Diagram

Figure 15.1 shows a block diagram of the flash memory.

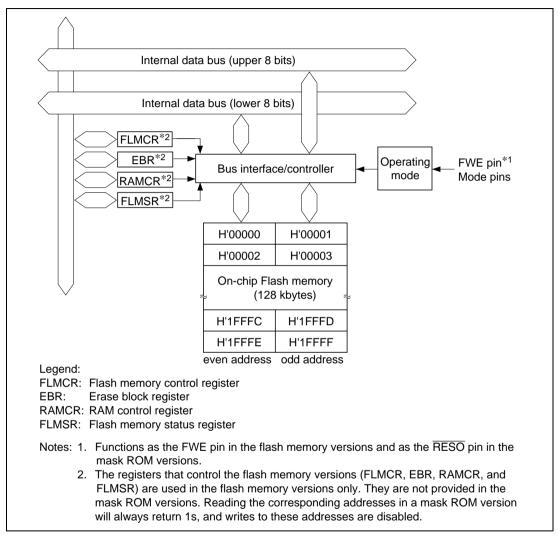


Figure 15.1 Block Diagram of Flash Memory

## 15.2.3 Pin Configuration

The flash memory is controlled by means of the pins shown in table 15.2.

<b>Table 15.2</b>	Flash Memory Pins
-------------------	-------------------

Pin Name	Abbreviation	I/O	Function
Reset	RES	Input	Reset
Flash write enable	FWE*	Input	Flash program/erase protection by hardware
Mode 2	MD <sub>2</sub>	Input	Sets this LSI operating mode
Mode 1	MD <sub>1</sub>	Input	Sets this LSI operating mode
Mode 0	MD <sub>o</sub>	Input	Sets this LSI operating mode
Transmit data	TxD <sub>1</sub>	Output	Serial transmit data output
Receive data	RxD <sub>1</sub>	Input	Serial receive data input

Notes: The transmit data and receive data pins are used in boot mode.

\* In the mask ROM versions, the FWE pin functions as the RESO pin.

#### 15.2.4 Register Configuration

The registers used to control the on-chip flash memory when enabled are shown in table 15.3.

#### Table 15.3 Flash Memory Registers

Register Name	Abbreviation	R/W	Initial Value	Address*1
Flash memory control register	FLMCR	R/W	H'00* <sup>2</sup>	H'FF40
Erase block register	EBR	R/W	H'00	H'FF42
RAM control register	RAMCR	R/W	H'F1	H'FF47
Flash memory status register	FLMSR	R	H'7F	H'FF4D

Notes: 1. Lower 16 bits of the address.

2. When a high level is input to the FWE pin, the initial value is H'80.

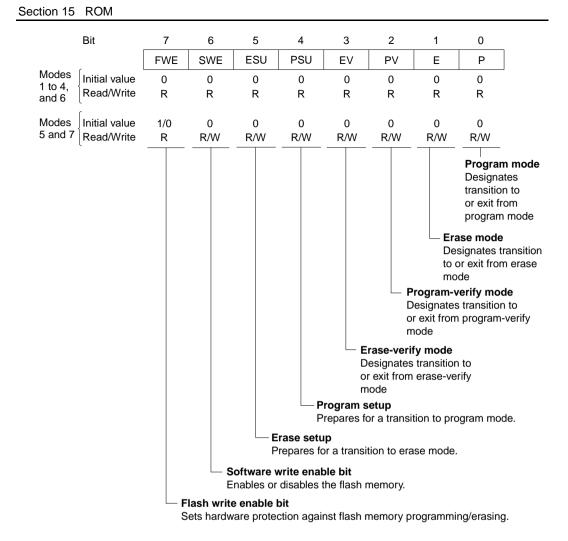
The registers in table 15.3 are used in the flash memory versions only. Reading the corresponding addresses in a mask ROM version will always return 1s, and writes to these addresses are disabled.

# **15.3** Register Descriptions

## 15.3.1 Flash Memory Control Register (FLMCR)

FLMCR is an 8-bit register used for flash memory operating mode control. Program-verify mode or erase-verify mode is entered by setting SWE to 1 when FWE = 1. Program mode is entered by setting SWE to 1 when FWE = 1, then setting the PSU bit, and finally setting the P bit. Erase mode is entered by setting SWE to 1 when FWE = 1, then setting the ESU bit, and finally setting the E bit. FLMCR is initialized by a reset, and in hardware standby mode and software standby mode. Its initial value is H'80 when a high level is input to the FWE pin, and H'00 when a low level is input. In mode 6 the FWE pin must be fixed low, as flash memory on-board programming is not supported. Therefore, bits in this register cannot be set to 1 in mode 6. When on-chip flash memory is disabled, a read will return H'00, and writes are invalid. When setting bits 6 to 0 in this register to 1, each bit should be set individually.

Writes to the ESU, PSU, EV and PV bits in FLMCR are enabled only when FWE = 1 and SWE = 1; writes to the E bit only when FWE = 1, SWE = 1, and ESU = 1; and writes to the P bit only when FWE = 1, SWE = 1, and PSU = 1.



**Bit 7—Flash Write Enable Bit (FWE):** Sets hardware protection against flash memory programming/erasing. When using this bit, refer to section 15.9, Notes on Flash Memory Programming/Erasing.

Bit 7 FWE	Description
0	When a low level is input to the FWE pin (hardware-protected state)
1	When a high level is input to the FWE pin

**Bit 6—Software Write Enable Bit (SWE)**\*<sup>1</sup>\*<sup>2</sup>**:** This bit enables/disables flash memory programming/erasing. This bit should be set before setting FLMCR bits 5 to 0, and EBR bits 7 to 0. Do not set the ESU, PSU, EV, PV, E, or P bits at the same time.

Bit 6 SWE	Description	
0	Program/erase disabled	(Initial value)
1	Program/erase enabled	
	[Setting condition]	
	When FWE = 1	

**Bit 5—Erase Setup Bit (ESU)**\*<sup>1</sup>: Prepares for a transition to erase mode. Do not set the SWE, PSU, EV, PV, E, or P bit at the same time.

Bit 5 ESU	Description	
0	Erase setup cleared	(Initial value)
1	Erase setup	
	[Setting condition]	
	When FWE = 1, and SWE = 1	

**Bit 4—Program Setup Bit (PSU)**\*<sup>1</sup>: Prepares for a transition to program mode. Do not set the SWE, ESU, EV, PV, E, or P bit at the same time.

Bit 4 PSU	Description	
0	Program setup cleared	(Initial value)
1	Program setup	
	[Setting condition]	
	When $FWE = 1$ , and $SWE = 1$	

**Bit 3—Erase-Verify (EV)**\*<sup>1</sup>: Selects erase-verify mode transition or clearing. Do not set the SWE, ESU, PSU, PV, E, or P bit at the same time.

Bit 3 EV	Description	
0	Erase-verify mode cleared	(Initial value)
1	Transition to erase-verify mode	
	[Setting condition]	
	When FWE = 1, and SWE = 1	

**Bit 2—Program-Verify** (**PV**)\*<sup>1</sup>: Selects program-verify mode transition or clearing. Do not set the SWE, ESU, PSU, EV, E, or P bit at the same time.

Bit 2 PV	Description	
0	Program-verify mode cleared	(Initial value)
1	Transition to program-verify mode	
	[Setting condition]	
	When $FWE = 1$ , and $SWE = 1$	

**Bit 1—Erase (E)**\*<sup>1</sup>\*<sup>3</sup>**:** Selects erase mode transition or clearing. Do not set the SWE, ESU, PSU, EV, PV, or P bit at the same time.

Bit 1 E		
	Description	
0	Erase mode cleared	(Initial value)
1	Transition to erase mode	
	[Setting condition]	
	When FWE = 1, SWE = 1, and ESU = 1	

**Bit 0—Program (P)**\*<sup>1</sup>\*<sup>3</sup>: Selects program mode transition or clearing. Do not set the SWE, ESU, PSU, EV, PV, or E bit at the same time.

Bit 0 P	Description	
0	Program mode cleared	(Initial value)
1	Transition to program mode	
	[Setting condition]	
	When $FWE = 1$ , $SWE = 1$ , and $PSU = 1$	

Notes: 1. Do not set two or more bits at the same time. Do not turn off  $V_{cc}$  when a bit is set.

- 2. Do not set/clear the SWE bit simultaneously with other bits (ESU, PSU, EV, PV, E, P).
- Set the P and E bits according to the program and erase algorithms shown in section 15.5, Programming/Erasing Flash Memory. For the usage precautions, see section 15.9, Notes on Flash Memory Programming/Erasing.

### 15.3.2 Erase Block Register (EBR)

EBR is an 8-bit register that designates the flash memory block for erasure. EBR is initialized to H'00 by a reset, in hardware standby mode, or software standby mode, when a high level is not input to the FWE terminal, or when the FLMCR SWE bit is 0 when a high level is applied to the FWE terminal. When a bit is set in EBR, the corresponding block can be erased. Other blocks are erase - protected. The blocks are erased block by block. Therefore, set only one bit in EBR; do not set bits in EBR to erase two or more blocks at the same time.

Each bit in EBR cannot be set until the SWE bit in FLMCR is set. The flash memory block configuration is shown in table 15.4. To erase all the blocks, erase each block sequentially.

This LSI does not support the on-board programming mode in mode 6, so bits in this register cannot be set to 1 in mode 6.

	Bit	7	6	5	4	3	2	1	0
		EB7	EB6	EB5	EB4	EB3	EB2	EB1	EB0
Modes 1 to 4, and 6	Initial value Read/Write	0 R							
Modes 5 and 7	Initial value Read/Write	0 R/W							

Bits 7 to 0—Block 7 to 0 (EB7 to EB0): These bits select blocks (EB7 to EB0) to be erased.

# Bits 7 to 0 Description 0 Block EB7 to EB0 is not selected. (Initial value) 1 Block EB7 to EB0 is selected.

Note: Set each bit of EBR to H'00 except when erasing.

### Table 15.4 Flash Memory Erase Blocks

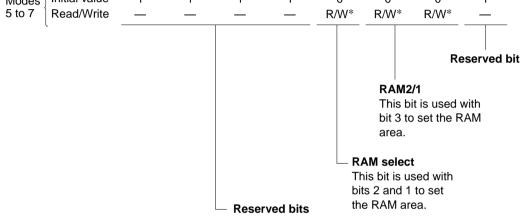
Block (Size)	Address
EB0 (1 kbyte)	H'00000 to H'003FF
EB1 (1 kbyte)	H'00400 to H'007FF
EB2 (1 kbyte)	H'00800 to H'00BFF
EB3 (1 kbyte)	H'00C00 to H'00FFF
EB4 (28 kbytes)	H'01000 to H'07FFF
EB5 (32 kbytes)	H'08000 to H'0FFFF
EB6 (32 kbytes)	H'10000 to H'17FFF
EB7 (32 kbytes)	H'18000 to H'1FFFF

# Section 15 ROM

### 15.3.3 RAM Control Register (RAMCR)

memory. Bit 7 4 3 2 1 6 5 0 RAMS RAM2 RAM1 \_\_\_\_ Initial value 1 1 0 0 0 Modes 1 1 1 1 to 4 Read/Write R R R Initial value 1 1 1 1 0 0 0 1 Modes 5 to 7 Read/Write R/W\* R/W\* R/W\*

RAMCR selects the RAM area used when emulating real-time reprogramming of the flash



Note: \* Cannot be set to 1 in mode 6.

Bits 7 to 4—Reserved: These bits cannot be modified and are always read as 1.

**Bit 3—RAM Select (RAMS):** Is used with bits 2 to 1 to reassign an area to RAM (see table 15.5). The initial setting for this bit is 0 in modes 5, 6, and 7 (internal flash memory enabled) and programming is enabled.\* In modes other than 5 to 7, 0 is always read and writing is disabled. It is initialized by a reset and in hardware standby mode. It is not initialized in software standby mode.

When bit 3 is set, all flash-memory blocks are protected from programming and erasing.

**Bits 2 to 1—RAM2 to RAM1:** These bits are used with bit 3 to reassign an area to RAM (see table 15.5). The initial setting for this bit is 0 in modes 5, 6, and 7 (internal flash memory enabled) and programming is enabled.\* In modes other than 5 to 7, 0 is always read and writing is disabled. They are initialized by a reset and in hardware standby mode. They are not initialized in software standby mode.

Bit 0—Reserved: This bit cannot be modified and is always read as 1.

Note: \* Flash memory emulation by RAM is not supported for Mode 6 (single chip normal mode), so programming is possible, but do not set 1.
 When performing flash memory emulation by RAM, the RAME bit in SYSCR must be set to 1.

	Bit 3	Bit 2	Bit 1	RAM
RAM Area	RAMS	RAM2	RAM1	Emulation State
H'FFF800 to H'FFFBFF	0	0/1	0/1	No emulation
H'000000 to H'0003FF	1	0	0	Mapping RAM
H'000400 to H'0007FF	1	0	1	
H'000800 to H'000BFF	1	1	0	
H'000C00 to H'000FFF	1	1	1	

#### Table 15.5 RAM Area Reassignment

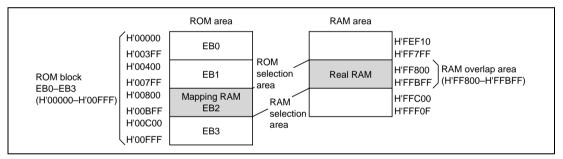
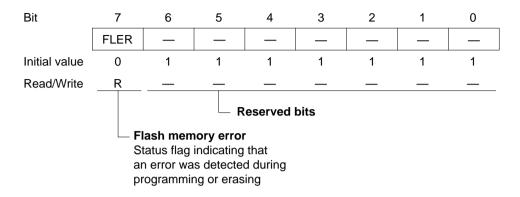


Figure 15.2 Example of Overlap ROM Area and RAM Area

#### 15.3.4 Flash Memory Status Register (FLMSR)

The flash memory status register (FLMSR) detects flash memory errors.



**Bit 7—Flash Memory Error (FLER):** Indicates that an error occurred while flash memory was being programmed or erased. When bit 7 is set, flash memory is placed in an error-protect mode.

Bit 7 FLER		Description
0		Flash memory program/erase protection (error protection* <sup>1</sup> ) is disabled (Initial value)
		[Clearing condition]
		WDT reset, reset by RES pin, or hardware standby mode
1		An error has occurred during flash memory programming/erasing Flash memory program/erase protection (error protection*1) is enabled
		[Setting conditions]
		<ol> <li>Flash memory was read*<sup>2</sup> while being programmed or erased (including vector or instruction fetch, but not including reading of a RAM area overlapped onto flash memory).</li> </ol>
		2. A hardware exception-handling sequence (other than a reset, invalid instruction, trap instruction, or zero-divide exception) was executed just before programming or erasing.* <sup>3</sup>
		<ol> <li>The SLEEP instruction (including software standby mode) was executed during programming or erasing.</li> </ol>
Notes:	1.	For details, see section 15.6.3, Error Protection.
	2.	The read data has undetermined values.
	3.	Before stack and vector read by exception handling.

Bits 6 to 0—Reserved: These bits cannot be modified and are always read as 1.

# 15.4 On-Board Programming Modes

When pins are set to on-board programming mode, program/erase/verify operations can be performed on the on-chip flash memory. There are two on-board programming modes: boot mode and user program mode. The pin settings for transition to each of these modes are shown in table 15.6. In mode 6 (on-chip ROM enabled) in this LSI, the boot mode and user program mode cannot be used. For the notes on FWE pin set/reset, see section 15.9, Notes on Flash Memory Programming/Erasing.

Mode		FWE	MD <sub>2</sub>	$\mathbf{MD}_{1}$	MD。	Notes
Boot mode	mode 5	<b>1</b> * <sup>1</sup>	0* <sup>2</sup>	0	1	0: V <sub>IL</sub>
	mode 7		0* <sup>2</sup>	1	1	1: V <sub>IH</sub>
User program mode	mode 5		1	0	1	
	mode 7		1	1	1	

<b>Table 15.6</b>	Setting On-Board	<b>Programming Modes</b>
	Seeming on Doma	

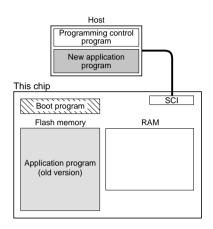
Notes: 1. For the High level input timing, see items (6) and (7) of Notes on Using the Boot Mode.2. In the boot mode, the MD, setting becomes inverted input.

In the boot mode, the mode control register (MDCR) can be used to monitor the status of modes 5 and 7 in the same way as in the normal mode.

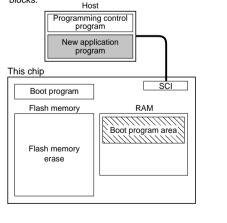
## **On-Board Programming Modes**

- Boot mode
  - 1. Initial state

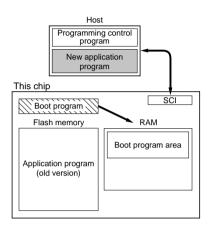
The flash memory is in the erased state when the device is shipped. The description here applies to the case where the old program version or data is being rewritten. The user should prepare the programming control program and new application program beforehand in the host.



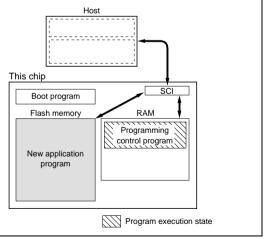
 Flash memory initialization The erase program in the boot program area (in RAM) is executed, and the flash memory is initialized (to H'FF). In boot mode, entire flash memory erasure is performed, without regard to blocks.



 Programming control program transfer When boot mode is entered, the boot program in this chip (originally incorporated in the chip) is started, an SCI communication check is carried out, and the boot program required for flash memory erasing is automatically transferred to the RAM boot program area.



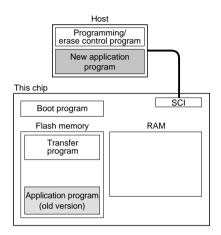
4. Writing new application program The programming control program transferred from the host to RAM by SCI communication is executed, and the new application program in the host is written into the flash memory.



#### Figure 15.3 Boot Mode

- User program mode
  - 1. Initial state

(1) The program that will transfer the programming/ erase control program to on-chip RAM should be written into the flash memory by the user beforehand. (2) The programming/erase control program should be prepared in the host or in the flash memory.



 Flash memory initialization The programming/erase program in RAM is executed, and the flash memory is initialized (to H'FF). Erasing can be performed in block units, but not in byte units.

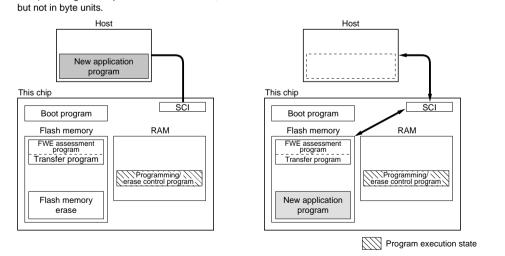
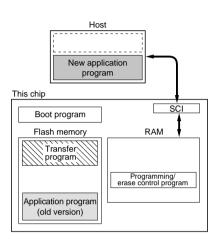


Figure 15.4 User Program Mode (Example)

Rev.3.00 Mar. 26, 2007 Page 454 of 682 REJ09B0353-0300  Programming/erase control program transfer When the FWE pin is driven high, user software confirms this fact, executes the transfer program in the flash memory, and transfers the programming/erase control program to RAM.



 Writing new application program Next, the new application program in the host is written into the erased flash memory blocks. Do not write to unerased blocks.



#### 15.4.1 Boot Mode

When boot mode is used, the flash memory programming control program must be prepared in the host beforehand. The channel 1 SCI to be used is set to asynchronous mode.

In reset start, after setting this LSI pin to the boot mode, start the microcomputer boot program, measure the Low period of the data sent from the host, and select the bit rate register (BRR) value beforehand. Then enable reception of the user program from the outside using the serial communication interface (SCI) on this LSI, and write the received user program to on-chip RAM.

After the program has been stored the end of writing, execution branches to the top address (H'FF300) of the on-chip RAM, execute the program written on the on-chip RAM, and enable flash memory program/erase.

The system configuration in boot mode is shown in figure 15.5, and the boot program mode execution procedure in figure 15.6.

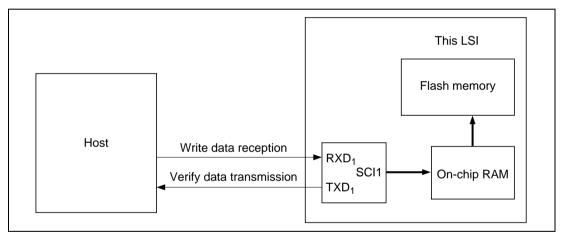
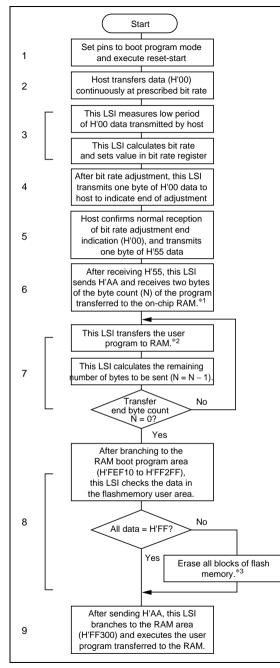


Figure 15.5 System Configuration in Boot Mode



- 1 Set this LSI to the boot mode and reset starts the LSI.
- 2 Set the host to the prescribed bit rate (4800, 9600) and consecutively send H'00 data in 8-bit data, 1 stop bit format.
- 3 This LSI repeatedly measures the RXD1 pin Low period and calculates the asynchronous communication bit rate at which the host performs transfer.
- 4 At the end of SCI bit rate adjustment, this LSI sends one byte of H'00 data to signal the end of adjustment.
- 5 Check if the host normally received the one byte bit rate adjustment end signal sent from this LSI and sent one byte of H'55 data.
- 6 After H'55 is sent, the host receives H'AA and sends the byte count of the user program that is transferred to this LSI. Send the 2-byte count in upper byte and lower byte order. Then sequentially send the program set by the user.

This LSI sequentially sends (echo back) each byte of the received byte count and user program to the host as verification data.

- 7 This LSI sequentially writes the received user program to the on-chip RAM area (H'FF300–H'FFEFF).
- 8 Before executing the transferred user program, this LSI checks if data was written to flash memory after control branched to the RAM boot program area (HFEF10–HFF2FF). If data was already written to flash memory, all the blocks are erased.
- 9 After sending H'AA, this LSI branches to the on-chip RAM area (H'FF300) and executes the user program written to that area.
- Notes: 1. The RAM area that can be used by the user is 3.0 kbytes. Set the transfer byte count to within 3.0 kbytes. Always send the 2-byte transfer byte count in upper byte and lower byte order. Transfer byte count example: For 256 bytes (H'0100), upper byte H'01, lower byte H'00.
  - Set the part that controls the user program flash memory at the program according to the flash memory programming/erase algorithms described later.
  - When a memory cell malfunctions and cannot be erased, this LSI sends one H'FF byte as an erase error and stops the erase operation and subsequent operations.

## Figure 15.6 Boot Mode Execution Procedure

#### Automatic SCI Bit Rate Adjustment

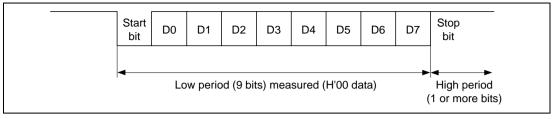


Figure 15.7 Measuring the Low Period of the Communication Data from the Host

When boot mode is initiated, this LSI measures the low period of the asynchronous SCI communication data (H'00) transmitted continuously from the host (figure 15.7). The SCI transmit/receive format should be set as follows: 8-bit data, 1 stop bit, no parity. This LSI calculates the bit rate of the transmission from the host from the measured low period, and transmits one H'00 byte to the host to indicate the end of bit rate adjustment. The host should confirm that this adjustment end indication (H'00) has been received normally, and transmit one H'55 byte to the LSI. If reception cannot be performed normally, initiate boot mode again (reset), and repeat the above operations. Depending on the host's transmission bit rate and the system clock frequency of this LSI, there will be a discrepancy between the bit rates of the host and the LSI. To ensure correct SCI operation, the host's transfer bit rate should be set to 4800 and 9600 bps\*<sup>1</sup>.

Table 15.7 shows typical host transfer bit rates and system clock frequencies for which automatic adjustment of this LSI bit rate is possible. The boot program should be executed within this system clock range\*<sup>2</sup>.

Host Bit Rate (bps)	System Clock Frequency for which Automatic Adjustment of This LSI Bit Rate Is Possible (MHz)
9600	8 to 18
4800	4 to 18

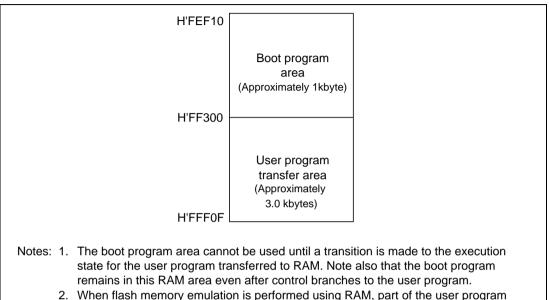
 
 Table 15.7
 System Clock Frequencies for which Automatic Adjustment of This LSI Bit Rate Is Possible

Notes: 1. The host bit rate settings are 4800 and 9600bps only. Do not use any other setting.

2. This LSI may automatically adjusts the bit rate except for bit rate and system clock combinations as shown in table 15.7. However, the bit rate of the host and this LSI will be different and subsequent transfers will not be carried out normally. Therefore, always execute the boot mode within the range of the bit rate and system clock combinations shown in table 15.7.

## **On-Chip RAM Area Divisions in Boot Mode**

In boot mode, the RAM area is divided into an area used by the boot program and an area to which the user program is transferred via the SCI, as shown in figure 15.8. The boot program area can be used when a transition is made to the execution state for the user program transferred to RAM.



 When flash memory emulation is performed using RAM, part of the user program transfer area (H'FF800 to H'FFBFF) is used as an area for carrying out emulation, and therefore user program transfer must not be performed to this area.

## Figure 15.8 RAM Areas in Boot Mode

#### Notes on using the boot mode

- (1) When this LSI comes out of reset in boot mode, it measures the low period the input at the SCI's RXD<sub>1</sub> pin. The reset should end with RXD<sub>1</sub> high. After the reset ends, it takes about 100 states for this LSI to get ready to measure the low period of the RXD<sub>1</sub> input.
- (2) If any data has been written to the flash memory (if all data is not H'FF), all flash memory blocks are erased when this mode is executed. Therefore, boot mode should be used for initial on-board programming, or for forced recovery if the program to be activated in user program mode is accidentally erased and user program mode cannot be executed, for example.
- (3) Interrupts cannot be used during programming or erasing of flash memory.

- (4) The  $RXD_1$  and  $TXD_1$  pins should be pulled up on the board.
- (5) This LSI terminates transmit and receive operations by the on-chip SCI(channel 1) (by clearing the RE and TE bits in serial control register (SCR)) before branching to the user program. However, the adjusted bit rate is held in the bit rate register (BRR). At this time, the TXD<sub>1</sub> pin is in the high level output state (P9DDR P9<sub>1</sub>DDR=1, P9DR P9<sub>1</sub>DR=1).

Before branching to the user program the value of the general registers in the CPU are also undefined. Therefore, the general registers must be initialized immediately after control branches to the user program. Since the stack pointer (SP) is implicitly used during subroutine call, etc., a stack area must be specified for use by the user program.

There are no other internal I/O registers in which the initial value is changed.

(6) Transition to the boot mode executes a reset-start of this LSI after setting the  $MD_0$  to  $MD_2$  and FWE pins according to the mode setting conditions shown in table 15.6.

At this time, this LSI latches the status of the mode pin inside the microcomputer to maintain the boot mode status at the reset clear (startup with Low  $\rightarrow$  High) timing<sup>\*1</sup>.

To clear boot mode, it is necessary to drive the FWE pin low during the reset, and then execute reset release\*<sup>1</sup>. The following points must be noted:

- (a) Before making a transition from the boot mode to the regular mode, the microcomputer boot mode must be reset by reset input via the  $\overline{\text{RES}}$  pin. At this time, the  $\overline{\text{RES}}$  pin must be hold at low level for at least 20 system clock.\*<sup>3</sup>
- (b) Do not change the input levels at the mode pins  $(MD_2 \text{ to } MD_0)$  or the FWE pin while in boot mode. When making a mode transition, first enter the reset state by inputting a low level to the  $\overline{RES}$  pin. When a watchdog timer reset was generated in the boot mode, the microcomputer mode is not reset and the on-chip boot program is restarted regardless of the state of the mode pin.
- (c) Do not input low level to the FWE pin while the boot program is executing and when programming/erasing flash memory.\*<sup>2</sup>
- (7) If the mode pin and FWE pin input levels are changed from 0 V to  $V_{cc}$  or from  $V_{cc}$  to 0V during a reset (while a low level is being input to the  $\overline{\text{RES}}$  pin), the microcomputer's operating mode will change.

Therefore, since the state of the address dual port and bus control output signals ( $\overline{AS}$ ,  $\overline{RD}$ ,  $\overline{WR}$ ) changes, use of these pins as output signals during reset must be disabled outside the microcomputer.

- Notes: 1. The mode pin and FWE pin input must satisfy the mode programming setup time  $(t_{MDS})$  relative to the reset clear timing.
  - 2. For notes on FWE pin High/Low, see section 15.9, Notes on Flash Memory Programming/Erasing.

 See section 4.2.2, Reset Sequence and 15.9, Notes on Flash Memory Programming/Erasing. With the mask ROM version of the H8/3039, H8/3038, H8/3037, and H8/3036, the minimum reset period during operation is 10 system clocks. However, the flash memory versions of the H8/3039 requires a minimum of 20 system clocks.

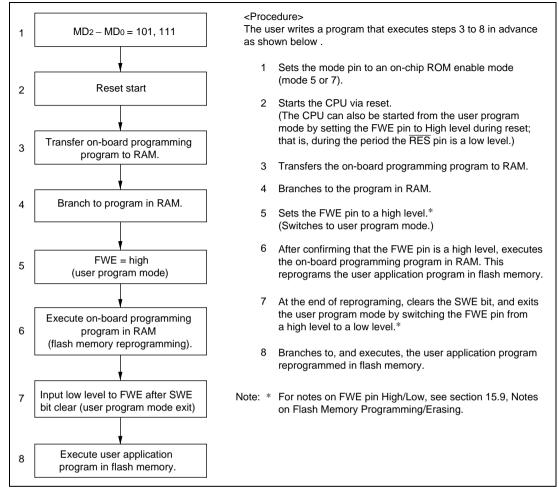
## 15.4.2 User Program Mode

When set to the user program mode, this LSI can erase and program its flash memory by executing a user program. Therefore, on-chip flash memory on-board programming can be performed by providing a means of controlling FWE and supplying the write data on the board and providing a write program in a part of the program area.

To select this mode, set the LSI to on-chip ROM enable modes 5 and 7 and apply a high level to the FWE pin. In this mode, the peripheral functions, other than flash memory, are performed the same as in modes 5 and 7.

Since the flash memory cannot be read while it is being programmed/erased, place a programming program on external memory, or transfer the programming program to RAM area, and execute it in the RAM. In mode 6, do not program/erase the flash memory. When setting mode 6, always input low level to the FWE pin.

Figure 15.9 shows the procedure for executing when transferred to on-chip RAM. During reset start, starting from the user program mode is possible.





Note: Normally do not apply a high level to the FWE pin. To prevent erroneous programming or erasing in the event of program runaway, etc., apply a high level to the FWE pin only when programming/erasing flash memory (including flash memory emulation by RAM). If program runaway, etc. causes overprogramming or overerasing of flash memory, the memory cells will not operate normally.

Also, while a high level is applied to the FWE pin, the watchdog timer should be activated to prevent overprogramming or overerasing due to program runaway, etc.

In mode 6, do not reprogram flash memory. When setting mode 6, always set the FWE pin to a low level.

# 15.5 Programming/Erasing Flash Memory

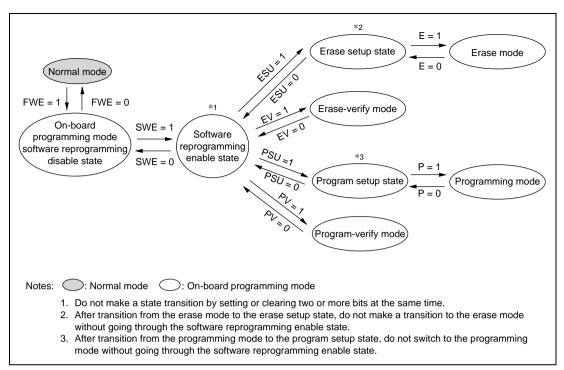
A software method, using the CPU, is employed to program and erase flash memory in the onboard programming modes. There are four flash memory operating modes: program mode, erase mode, program-verify mode, and erase-verify mode. Transitions to these modes can be made by setting the PSU, ESU, P, E, PV, and EV bits in FLMCR.

For a description of state transition by FLMCR bit setting, see figure 15.10.

The flash memory cannot be read while being programmed or erased. Therefore, the program that controls flash memory programming/erasing (the programming control program) should be located and executed in on-chip RAM or external memory.

For the programming/erasing notes, see section 15.9, Notes on Flash Memory Programming/Erasing. For the wait time after each bit in FLMCR is set or cleared, see section 18.2.5, Flash Memory Characteristics.

- Notes: 1. Operation is not guaranteed if setting/resetting of the SWE, ESU, PSU, EV, PV, E, and P bits in FLMCR is executed by a program in flash memory.
  - 2. When programming or erasing, set the FWE pin input level to the high level, and set FWE to 1. (programming/erasing will not be executed if FWE = 0).



## Figure 15.10 State Transition by Setting of Each Bit of FLMCR

#### 15.5.1 Program Mode

Follow the procedure shown in the program/program-verify flowchart in figure 15.11 to write data or programs to flash memory. Performing program operations according to this flowchart will enable data or programs to be written to flash memory without subjecting the device to voltage stress or sacrificing program data reliability. Programming should be carried out 32 bytes at a time.

For the wait time (x, y, z,  $\alpha$ ,  $\beta$ ,  $\gamma$ ,  $\varepsilon$ ,  $\eta$ ) after setting or clearing each bit in the flash memory control register (FLMCR) and the maximum programming count (N), see table 18.15.

Following the elapse of (x)  $\mu$ s or more after the SWE bit is set to 1 in flash memory control register (FLMCR), 32-byte program data is stored in the program data area and reprogram data area, and the 32-byte data in the reprogram data area written consecutively to the write addresses. (The lower 8 bits of the first address written to must be H'00, H'20, H'40, H'60, H'80, H'A0, H'C0, or H'E0.) 32 consecutive byte data transfers are performed. The program address and program data are latched in the flash memory. A 32-byte data transfer must be performed even if writing fewer than 32 bytes; in this case, H'FF data must be written to the extra addresses.

Next, the watchdog timer (WDT) is set to prevent overprogramming due to program runaway, etc. Set a value greater than  $(y + z + \alpha + \beta) \mu s$  as the WDT overflow period. Preparation for entering program mode (program setup) is performed next by setting the PSU bit in FLMCR. The operating mode is then switched to program mode by setting the P bit in FLMCR after the elapse of at least  $(y) \mu s$ .

The time while the P bit is set is the flash memory programming time. Make a program setting so that the time for one programming operation is within the range of (z)  $\mu$ s.

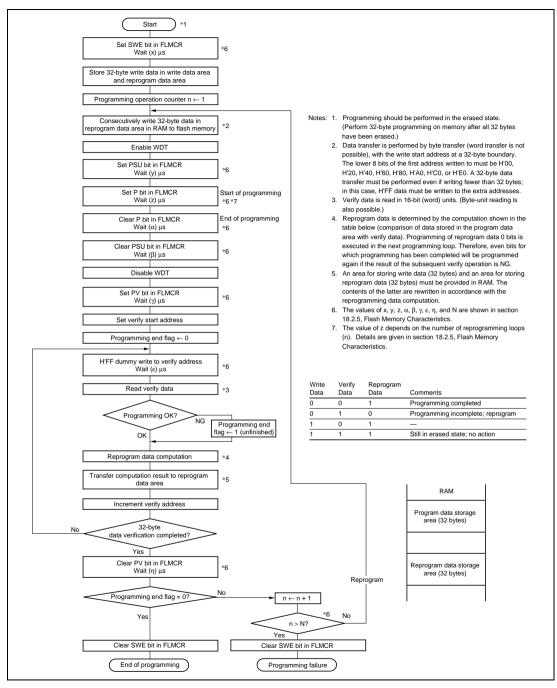
The wait time after P bit setting must be changed according to the number of reprogramming loops. For details, see section 18.2.5, Flash Memory Characteristics.

## 15.5.2 Program-Verify Mode

In program-verify mode, the data written in program mode is read to check whether it has been correctly written in the flash memory.

Clear the P bit in FLMCR, then wait for at least ( $\alpha$ ) µs before clearing the PSU bit to exit program mode. After exiting program mode, the watchdog timer setting is also cleared. Then the operating mode is switched to program-verify mode by setting the PV bit in FLMCR. Before reading in program-verify mode, a dummy write of H'FF data should be made to the addresses to be read. The dummy write should be executed after the elapse of ( $\gamma$ ) µs or more. When the flash memory is read in this state (verify data is read in 16-bit units), the data at the latched address is read. Wait at least ( $\varepsilon$ ) µs after the dummy write before performing this read operation. Next, the originally written data is compared with the verify data, and reprogram data is computed (see figure 15.11) and transferred to RAM. After verification of 32 bytes of data has been completed, exit programverify mode, wait for at least ( $\eta$ ) µs, then determine whether 32-byte programming has finished. If reprogramming is necessary, set program mode again, and repeat the program/program-verify sequence as before. However, ensure that the program/program-verify sequence is not repeated more than (N) times on the same bits.

Note: A 32-byte area to store program data and a 32-byte area to store reprogram data are required in RAM.





## 15.5.3 Erase Mode

Flash memory erasing should be performed block by block following the procedure shown in the erase/erase-verify flowchart (single-block erase) shown in figure 15.12.

For the wait time  $(x, y, z, \alpha, \beta, \gamma, \varepsilon, \eta)$  after setting or clearing of each bit in the flash memory control register (FLMCR) and the maximum erase count (N), see table 18.15.

To erase the contents of flash memory, make a 1 bit setting for the flash memory area to be erased in erase block register (EBR) at least (x)  $\mu$ s after setting the SWE bit to 1 in FLMCR. Next, the watchdog timer (WDT) is set to prevent overerasing due to program runaway, etc. Set a value greater than (z) ms + (y +  $\alpha$  +  $\beta$ )  $\mu$ s as the WDT overflow period. Preparation for entering erase mode (erase setup) is performed next by setting the ESU bit in FLMCR. The operating mode is then switched to erase mode by setting the E bit in FLMCR after the elapse of at least (y)  $\mu$ s.

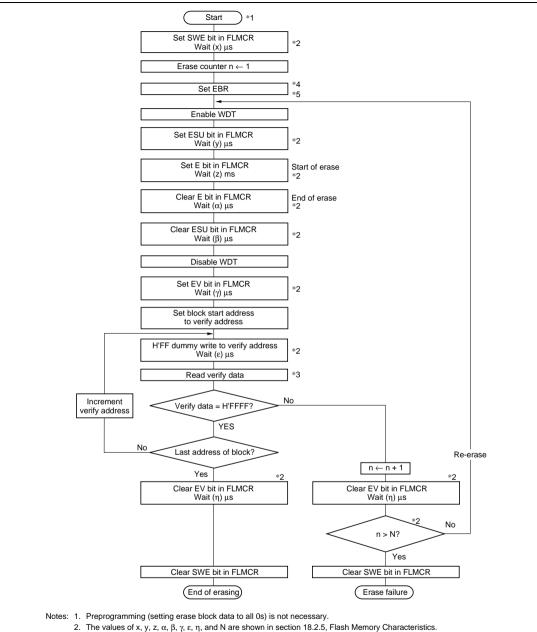
The time during which the E bit is set is the flash memory erase time. Ensure that the erase time does not exceed (z) ms.

Note: With flash memory erasing, preprogramming (setting all data in the memory to be erased to "0") is not necessary before starting the erase procedure.

## 15.5.4 Erase-Verify Mode

In erase-verify mode, data is read after memory has been erased to check whether it has been correctly erased.

After the elapse of the fixed erase time, clear the E bit in FLMCR, then wait for at least ( $\alpha$ )  $\mu$ s before clearing the ESU bit to exit erase mode. After exiting erase mode, the watchdog timer setting is also cleared. The operating mode is then switched to erase-verify mode by setting the EV bit in FLMCR. Before reading in erase-verify mode, a dummy write of H'FF data should be made to the addresses to be read. The dummy write should be executed after the elapse of ( $\gamma$ )  $\mu$ s or more. When the flash memory is read in this state (verify data is read in 16-bit units), the data at the latched address is read. Wait at least ( $\varepsilon$ )  $\mu$ s after the dummy write before performing this read operation. If the read data has been erased (all "1"), a dummy write is performed to the next address, and erase-verify sequence as before. However, do not repeat the erase/erase-verify sequence more than (N) times.



Verify data is read in 16-bit (word) units. (Byte-unit reading is also possible.)

- Set only one bit in EBR two or more bits must not be set simultaneously.
- 5. Erasing is performed in block units. To erase multiple blocks, each block must be erased in turn.

#### Figure 15.12 Erase/Erase-Verify Flowchart (Single-Block Erasing)

## **15.6** Flash Memory Protection

There are three kinds of flash memory program/erase protection: hardware protection, software protection, and error protection.

## 15.6.1 Hardware Protection

Hardware protection refers to a state in which programming/erasing of flash memory is forcibly disabled or aborted. Hardware protection is reset by settings in the flash memory control register (FLMCR) and erase block register (EBR). In the case of error protection, the P bit and E bit can be set, but a transition is not made to program mode or erase mode. (See table 15.8.)

			Functio	n
ltem	Description	Program	Erase	Verify*1
FWE pin protection	<ul> <li>When a low level is input to the FWE pin, FLMCR and EBR are initialized, and the program/erase-protected state is entered.*<sup>4</sup></li> </ul>	No*2	No* <sup>3</sup>	No
Reset/standby protection	<ul> <li>In a reset (including a WDT overflow reset) and in standby mode, FLMCR and EBR are initialized, and the program/erase-protected state is entered.</li> <li>In a reset via the RES pin, the reset</li> </ul>	No	No* <sup>3</sup>	No
	<ul> <li>In a reset via the RES pin, the reset state is not entered unless the RES pin is held low until oscillation stabilizes after powering on (The minimum oscillation stabilization time is 20ms). In the case of a reset during operation, hold the RES pin low for at least 20 system clock cycles.*<sup>5</sup></li> </ul>			
Error protection	• When a microcomputer operation error (error generation (FLER=1)) was detected while flash memory was being programmed/erased, error protection is enabled. At this time, the FLMCR and EBR settings are held, but programming/erasing is aborted at the time the error was generated. Error protection is released only by a reset via the RES pin or a WDT reset, or in the hardware standby mode.	No	No* <sup>3</sup>	Yes
	modes: program-verify and erase-verify.	ka al		
	RAM area that overlapped flash memory is delet ocks become unerasable and specification by b		sible	
		=		ing/Erocia
	nore information, see section 15.9, Notes on Fla		rogramm	my/⊏rasin

#### Table 15.8Hardware Protection

 See sections 4.2.2, Reset Sequence and 15.9, Notes on Flash Memory Programming/Erasing. This LSI requires a minimum reset time during operation of 20 system clocks.

## **15.6.2** Software Protection

Software protection can be implemented by setting the RAMS bit in RAM control register (RAMCR) and erase block register (EBR). When software protection is in effect, setting the P or E bit in flash memory control register (FLMCR) does not cause a transition to program mode or erase mode. (See table 15.9.)

<b>Table 15.9</b>	Software Protection
-------------------	---------------------

		Function				
Item	Description	Program	Erase	Verify*1		
Emulation protection* <sup>2</sup>	Setting the RAMS bit in RAMCR sets the program/erase-protected state for all blocks.	No* <sup>2</sup>	No* <sup>3</sup>	Yes		
Block specification protection	Erase protection can be set for individual blocks by settings in erase block register (EBR).* <sup>4</sup>	_	No	Yes		
	However, program protection is disabled.					
	Setting EBR to H'00 places all blocks in the erase-protected state.					
Notes: 1. Two	modes: program-verify mode and erase-verify m	node.				
2. Prog	ramming to the RAM area that overlaps flash me	emory is pos	ssible.			
2 416	aska haarma unarraachta, and an asification hy k	Jook in imme	aaibla			

3. All blocks become unerasable, and specification by block is impossible.

4. Set H'00 in the EBR bits, except for erase.

#### **15.6.3** Error Protection

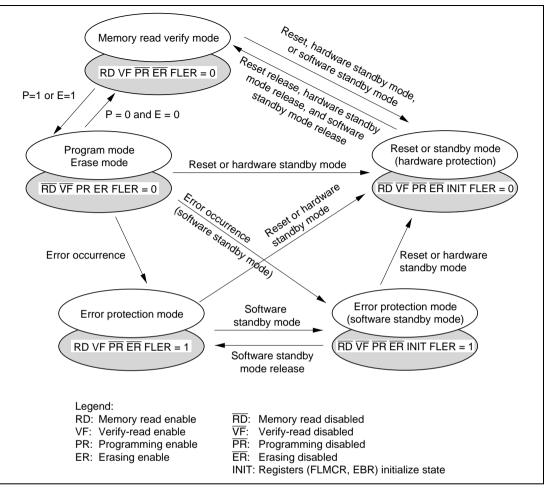
In error protection, an error is detected when this LSI runaway occurs during flash memory programming/erasing<sup>\*1</sup>, or operation is not performed in accordance with the program/erase algorithm, and the program/erase operation is aborted. Aborting the program/erase operation prevents damage to the flash memory due to overprogramming or overerasing.

If the LSI malfunctions during flash memory programming/erasing, the FLER bit\*<sup>2</sup> is set to 1 in flash memory status register (FLMSR) and the error protection state is entered. The FLMCR and EBR settings\*<sup>3</sup> are retained, but program mode or erase mode is aborted at the point at which the error occurred. When 1 is set in the FLER bit, transition to the program mode or erase mode cannot be made even by setting the P and E bits in FLMCR. However, PV and EV bit in FLMCR setting is enabled, and a transition can be made to verify mode.

Error protection is released only by a reset via the  $\overline{\text{RES}}$  pin or a WDT reset, or in the hardware standby mode.

Figure 15.13 shows the flash memory state transition diagram.

- Notes: 1. This is the state in which the P or E bit in FLMCR is set to 1. In this state, NMI input is disabled. For more information, see section 15.6.4, NMI Input Disable Conditions.
  - 2. For a detailed description of the FLER bits setting conditions, see section 15.3.4, Flash Memory Status Register (FLMSR).
  - 3. Data can be written to FLMCR and EBR. However, when transition to the software standby mode was made in the error protection state, the registers are initialized.



## Figure 15.13 Flash Memory State Transitions (When High Level Apply to FWE Pin in Modes 5 and 7 (On-Chip ROM Enabled))

The error protection function is disabled for errors other than the FLER bit set conditions. If considerable time elapses up to transit to this protection state, the flash memory may already be damaged. As a result, this function cannot completely protect the flash memory against damage.

Therefore, to prevent such erroneous operation, operation must be carried out correctly in according with the program/erase algorithms in the state that flash write enable (FWE) is set. In addition, the operation must be always carried out correctly by supervising microcomputer errors inside and outside the chip with the watchdog timer, etc. At transition to this protection mode, the flash memory may be erroneously programmed or erased, or its abort may result in incomplete

programming and erasing. In such cases, always forcibly return (reprogram) by boot mode. However, overprogramming and overerasing may prevent the boot mode from starting normally.

## 15.6.4 NMI Input Disable Conditions

While flash memory is being programmed/erased and the boot program is executing in the boot mode (however, period up to branching to on-chip RAM area)\*<sup>1</sup>, NMI input is disabled because the programming/erasing operations have priority.

This is done to avoid the following operation states:

- 1. Generation of an NMI input during programming/erasing violates the program/erase algorithms and normal operation can not longer be assured.
- 2. Vector-read cannot be carried out normally\*<sup>2</sup> during NMI exception handling during programming/erasing and the microcomputer runs away as a result.
- 3. If an NMI input is generated during boot program execution, the normal boot mode sequence cannot be executed.

Therefore, this LSI has conditions that exceptionally disable NMI inputs only in the on-board programming mode. However, this does not assure normal programming/erasing and microcomputer operation.

Thus, in the FWE application state, all requests, including NMI, inside and outside the microcomputer, exception handling, and bus release must be restricted. NMI inputs are also disabled in the error protection state and the state that holds the P or E bit in FLMCR during flash memory emulation by RAM.

Notes: 1. Indicates the period up to branching to the on-chip RAM boot program area (H'FEF10 to H'FF2FF). (This branch occurs immediately after user program transfer was completed.)

Therefore, after branching to RAM area, NMI input is enabled in states other than the program/erase state. Thus, interrupt requests inside and outside the microcomputer must be disabled until initial writing by user program (writing of vector table and NMI processing program, etc.) is completed.

- 2. In this case, vector read is not performed normally for the following two reasons:
  - a. The correct value cannot be read even by reading the flash memory during programming/erasing. (Value is undefined.)
  - b. If a value has not yet been written to the NMI vector table, NMI exception handling will not be performed correctly.

# 15.7 Flash Memory Emulation by RAM

Erasing and programming the flash memory takes time, which can make it difficult to tune parameters and other data in real time. In this case, overlapping part (H'FF800 to H'FFBFF) of RAM onto a small block area of flash memory can be performed to emulate real-time reprogramming of flash memory. This RAM reassignment is performed using bits 3 to 1 in the RAM control register (RAMCR).

After the RAM area change, two areas can be accessed: the overlapped flash memory area and the original RAM area (H'FF800 to H'FFBFF). For a description of the RAMCR and RAM area setting procedure, see section 15.3.3, RAM Control Register (RAMCR).

## Example of real-time emulation of flash memory

An example of RAM area H'FF800 to H'FFBFF overlapping EB2 (H'00800 to H'00BFF) flash memory area is shown below.

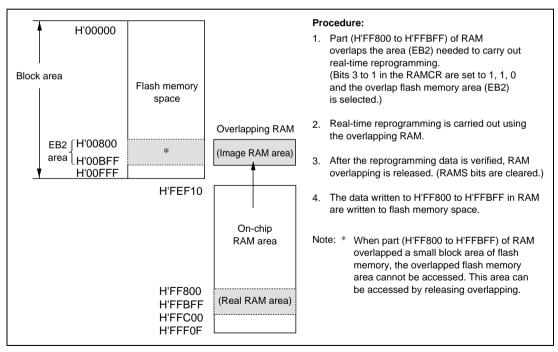


Figure 15.14 Example of RAM Overlapping Operation

#### Notes on use of the RAM emulation function

(1) Notes on flash write enable (FWE) high/low

Care is necessary to prevent erroneous programming/erasing at FWE = high/low, the same as in the on-board programming mode. To prevent erroneous programming and erasing due to program runaway, etc., during FWE application, in particular, the watchdog timer should be set when the P, or E bit is set to 1 in FLMCR, even while the emulation function is being used. For more information, see section 15.9, Notes on Flash Memory Programming/Erasing.

(2) NMI input disable conditions

When the P and E bits in FLMCR are set, NMI input is disabled, the same as normal program/erase even when using the emulation function.

NMI input is cleared when the P and E bits are reset (including watchdog timer reset), in the standby mode, when a high level is not applied to FWE, and when the SWE bit in FLMCR is 0 in state in which a high level is input to FWE.

## 15.8 Flash Memory PROM Mode

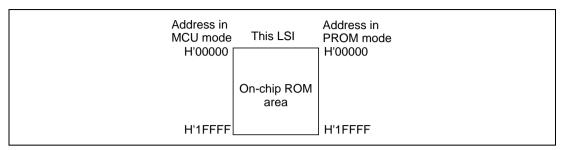
### 15.8.1 PROM Mode Setting

This LSI has a PROM mode, besides an on-board programming mode, as a flash memory program/erase mode. In the PROM mode, a program can be freely written to the on-chip ROM using a PROM programmer that supports the Renesas Technology 128 kbytes flash memory on-chip microcomputer device type.

For notes on PROM mode use, see sections 15.8.9, Notes on Memory Programming and 15.9, Notes on Flash Memory Programming/Erasing.

### 15.8.2 Memory Map

Figure 15.15 shows the PROM mode memory map.



## Figure 15.15 PROM Mode Memory Map

## 15.8.3 PROM Mode Operation

Table 15.10 shows how the different operating modes are set when using PROM mode, and table 15.11 lists the commands used in PROM mode. Details of each mode are given below.

Memory Read Mode

Memory read mode supports byte reads.

Auto-Program Mode

Auto-program mode supports programming of 128 bytes at a time. Status polling is used to confirm the end of auto-programming.

• Auto-Erase Mode

Auto-erase mode supports automatic erasing of the entire flash memory. Status polling is used to confirm the end of auto-erasing.

Status Read Mode

Status polling is used for auto-programming and auto-erasing, and normal termination can be confirmed by reading the I/O 6 signal. In status read mode, error information is output if an error occurs.

	Pin Names* <sup>3</sup>							
Mode	FWE	ĈĒ	ŌĒ	WE	D <sub>0</sub> to D <sub>7</sub>	$\mathbf{A}_{_{0}}$ to $\mathbf{A}_{_{17}}$		
Read	$V_{cc}$ or 0	L	L	Н	Data output	Ain		
Output disable	$V_{cc}$ or 0	L	Н	Н	Hi-Z	Х		
Command write	$V_{cc}$ or 0	L	Н	L	Data input	Ain* <sup>2</sup>		
Chip disable*1	$V_{cc}$ or 0	Н	Х	Х	Hi-Z	Х		

#### Table 15.10 Settings for Each Operating Mode in PROM Mode

Legend:

L: Low level

H: High level

X: Undefined

Hi-Z: High impedance

Notes: For command writes when making a transition to auto-program or auto-erase mode, input Vcc (V) to FWE.

- 1. Chip disable is not a standby state; internally, it is an operation state.
- 2. Ain indicates that there is also address input in auto-program mode.
- 3. The pin names are those assigned in this LSI PROM mode.

#### Table 15.11 PROM Mode Commands

Number		1st Cycle		2nd Cycle		
of Cycles	Mode	Address	Data	Mode	Address	Data
1	Write	Х	H'00	Read	RA	Dout
129	Write	Х	H'40	Write	WA	Din
2	Write	Х	H'20	Write	Х	H'20
2	Write	Х	H'71	Write	Х	H'71
	1 129 2	of CyclesMode1Write129Write2Write	Number of CyclesModeAddress1WriteX129WriteX2WriteX	of CyclesModeAddressData1WriteXH'00129WriteXH'402WriteXH'20	Number of CyclesModeAddressDataMode1WriteXH'00Read129WriteXH'40Write2WriteXH'20Write	Number of CyclesModeAddressDataModeAddress1WriteXH'00ReadRA129WriteXH'40WriteWA2WriteXH'20WriteX

Legend:

RA: Read address

WA: Program address

Dout: Read data

Din: Program data

Note: In auto-program mode, 129 cycles are required for command writing by a simultaneous 128-byte write.

## Table 15.12 DC Characteristics in Memory Read Mode

Conditions:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = 0 \text{ V}$ ,  $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ 

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Input high voltage	0 <sub>7</sub> -0 <sub>0</sub> , A <sub>16</sub> -A <sub>0</sub>	V <sub>IH</sub>	2.2	_	Vcc +0.3	V	
Input low voltage	0 <sub>7</sub> -0 <sub>0</sub> , A <sub>16</sub> -A <sub>0</sub>	V <sub>IL</sub>	0.3	—	0.8	V	
Schmitt trigger	OE, CE, WE	V <sub>T</sub>	1.0	_	2.5	V	
input voltage		$V_{T}^{+}$	2.0	_	3.5	V	
		$V_T^+ - V_T^-$	0.4	_	_	V	
Output high voltage	0,-0,	V <sub>oh</sub>	2.4	_	_	V	I <sub>oH</sub> = -200 μA
Output low voltage	0,-0,	V <sub>ol</sub>	_	_	0.45	V	I <sub>oL</sub> = 1.6 mA
Input leakage current	0 <sub>7</sub> -0 <sub>0</sub> , A <sub>16</sub> -A <sub>0</sub>	I <sub>u</sub>	_	_	2	μA	
V <sub>cc</sub> current	Reading	l <sub>cc</sub>		40	65	mA	
	Programming	I <sub>cc</sub>		50	85	mA	
	Erasing	I <sub>cc</sub>		50	85	mA	

Note: For the electrical characteristics of the flash memory version, see section 18.2.1, Absolute Maximum Ratings.

Exceeding the absolute maximum ratings may cause permanent damage to the chip.

#### 15.8.4 Memory Read Mode

## **AC Characteristics**

#### Table 15.13 AC Characteristics in Memory Read Mode Transition

Conditions:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = 0 \text{ V}$ ,  $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ 

Item	Symbol	Min	Max	Unit	Notes
Command write cycle	t <sub>nxtc</sub>	20	_	μs	
CE hold time	t <sub>ceh</sub>	0	_	ns	
CE setup time	t <sub>ces</sub>	0	_	ns	
Data hold time	t <sub>dh</sub>	50	_	ns	
Data setup time	t <sub>ds</sub>	50	_	ns	
Write pulse width	t <sub>wep</sub>	70	_	ns	
WE rise time	t <sub>r</sub>	_	30	ns	
WE fall time	t <sub>r</sub>		30	ns	

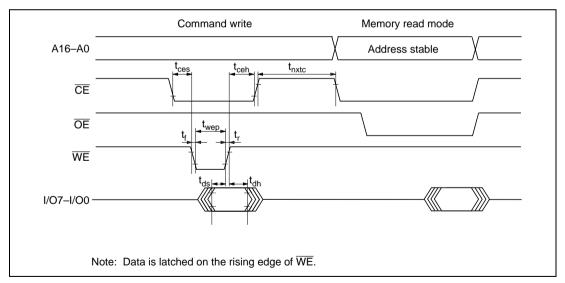


Figure 15.16 Timing Waveform in Memory Read Mode Transition

### Table 15.14 AC Characteristics in Memory Contents Read

Conditions:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = 0 \text{ V}$ ,  $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ 

Item	Symbol	Min	Max	Unit	Notes
Access time	t <sub>acc</sub>	_	20	μs	
CE output delay time	t <sub>ce</sub>	_	150	ns	
OE output delay time	t <sub>oe</sub>	_	150	ns	
Output disable delay time	t <sub>df</sub>	_	100	ns	
Data output hold time	t <sub>oh</sub>	5		ns	

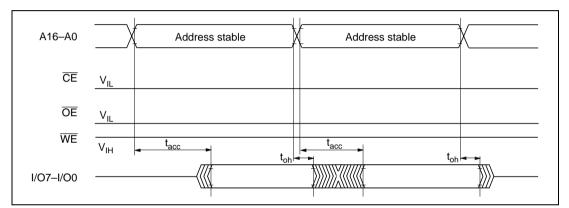


Figure 15.17 **CE/OE** Enable State Read

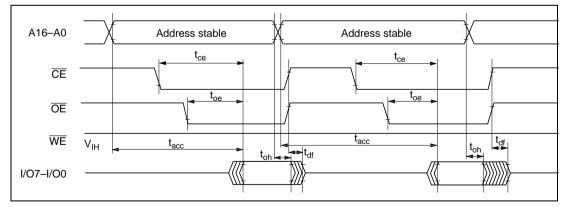


Figure 15.18 CE/OE Clock Read

Table 15.15         A	AC Characteristics in	<b>Transition from Memory</b>	<b>Read Mode to Another Mode</b>
-----------------------	-----------------------	-------------------------------	----------------------------------

Conditions:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = 0 \text{ V}$ ,  $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ 

Item	Symbol	Min	Max	Unit	Notes
Command write cycle	t <sub>nxtc</sub>	20	_	μs	
CE hold time	t <sub>ceh</sub>	0	_	ns	
CE setup time	t <sub>ces</sub>	0	_	ns	
Data hold time	t <sub>dh</sub>	50	_	ns	
Data setup time	t <sub>ds</sub>	50	_	ns	
Write pulse width	t <sub>wep</sub>	70	_	ns	
WE rise time	t <sub>r</sub>	_	30	ns	
WE fall time	t <sub>r</sub>	—	30	ns	

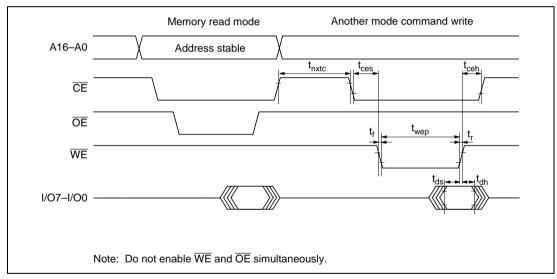


Figure 15.19 Transition From Memory Read Mode to Another Mode

### 15.8.5 Auto-Program Mode

# **AC Characteristics**

# Table 15.16 AC Characteristics in Auto-Program Mode

Conditions:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = 0 \text{ V}$ ,  $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ 

Item	Symbol	Min	Max	Unit	Notes
Command write cycle	t <sub>nxtc</sub>	20	_	μs	
CE hold time	t <sub>ceh</sub>	0		ns	
CE setup time	t <sub>ces</sub>	0	_	ns	
Data hold time	t <sub>dh</sub>	50	_	ns	
Data setup time	t <sub>ds</sub>	50	_	ns	
Write pulse width	t <sub>wep</sub>	70	_	ns	
Status polling start time	t <sub>wsts</sub>	1	_	ms	
Status polling access time	t <sub>spa</sub>	_	150	ns	
Address setup time	t <sub>as</sub>	0	_	ns	
Address hold time	t <sub>ah</sub>	60	_	ns	
Memory write time	t <sub>write</sub>	1	3000	ms	
WE rise time	t <sub>r</sub>	_	30	ns	
WE fall time	t <sub>r</sub>	_	30	ns	
Write setup time	t <sub>pns</sub>	100	_	ns	
Write end setup time	t <sub>pnh</sub>	100		ns	

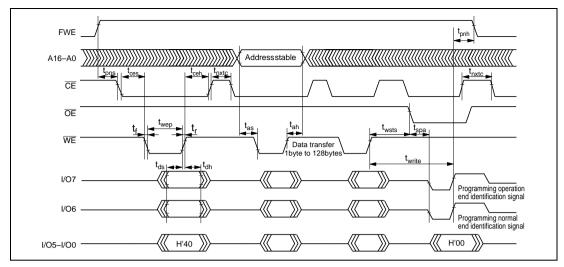


Figure 15.20 Auto-Program Mode Timing Waveforms

#### Cautions on Use of Auto-Program Mode

- In auto-program mode, 128 bytes are programmed simultaneously. This should be carried out by executing 128 consecutive byte transfers.
- A 128-byte data transfer is necessary even when programming fewer than 128 bytes. In this case, H'FF data must be written to the extra addresses.
- If a value other than an effective address is input, processing will switch to a memory write operation but a write error will be flagged.
- Memory address transfer is performed in the second cycle (figure 15.20). Do not perform transfer after the second cycle.
- Do not perform a command write during a programming operation.
- Perform one auto-programming operation for a 128-byte block for each address. Characteristics are not guaranteed for two or more programming operations.
- Confirm normal end of auto-programming by checking I/O 6. Alternatively, status read mode can also be used for this purpose.

# Renesas

#### 15.8.6 Auto-Erase Mode

# **AC Characteristics**

### Table 15.17 AC Characteristics in Auto-Erase Mode

Conditions:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = 0 \text{ V}$ ,  $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ 

Item	Symbol	Min	Max	Unit	Notes
Command write cycle	t <sub>nxtc</sub>	20	_	μs	
CE hold time	t <sub>ceh</sub>	0	_	ns	
CE setup time	t <sub>ces</sub>	0	_	ns	
Data hold time	t <sub>dh</sub>	50	_	ns	
Data setup time	t <sub>ds</sub>	50	_	ns	
Write pulse width	t <sub>wep</sub>	70	_	ns	
Status polling start time	t <sub>ests</sub>	1	_	ms	
Status polling access time	t <sub>spa</sub>	_	150	ns	
Memory erase time	t <sub>erase</sub>	100	40000	ms	
WE rise time	t,	_	30	ns	
WE fall time	t,	_	30	ns	
Erase setup time	t <sub>ens</sub>	100	_	ns	
Erase end setup time	t <sub>enh</sub>	100	_	ns	

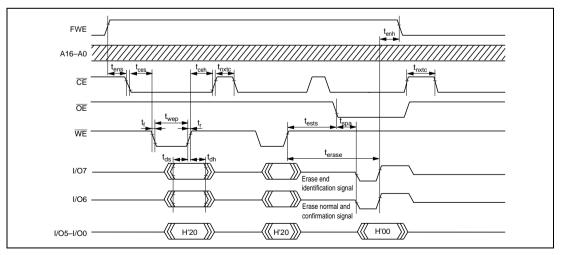


Figure 15.21 Auto-Erase Mode Timing Waveforms

#### Caution on Use of Erase-Program Mode

- Auto-erase mode supports only entire memory erasing.
- Do not perform a command write during auto-erasing.
- Confirm normal end of auto-erasing by checking I/O 6. Alternatively, status read mode can also be used for this purpose.

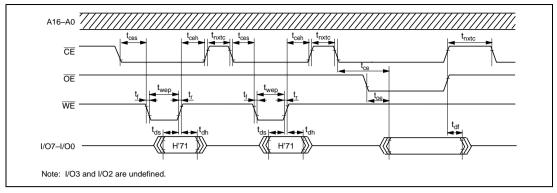
#### 15.8.7 Status Read Mode

#### **AC Characteristics**

#### Table 15.18 AC Characteristics in Status Read Mode

Conditions:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = 0 \text{ V}$ ,  $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ 

Item	Symbol	Min	Max	Unit	Notes
Command write cycle	t <sub>nxtc</sub>	20		μs	
CE hold time	t <sub>ceh</sub>	0	_	ns	
CE setup time	t <sub>ces</sub>	0	_	ns	
Data hold time	t <sub>dh</sub>	50	_	ns	
Data setup time	t <sub>ds</sub>	50	_	ns	
Write pulse width	t <sub>wep</sub>	70	_	ns	
OE output delay time	t <sub>oe</sub>	_	150	ns	
Disable delay time	t <sub>df</sub>	_	100	ns	
CE output delay time	t <sub>ce</sub>	_	150	ns	
WE rise time	t <sub>r</sub>	_	30	ns	
WE fall time	t <sub>r</sub>	_	30	ns	





### Table 15.19 Status Read Mode Return Commands

Pin Name	I/07	I/O6	I/O5	I/O4	I/O3	I/O2	I/O1	I/O0
Attribute	Normal end identification	Command error	Program- ming error	Erase error	_	_	Program- ming or erase count exceeded	Effective address error
Initial value	0	0	0	0	0	0	0	0
Indications	Normal end: 0 Abnormal	Command error: 1 Otherwise:	Program- ming error: 1	Erase error: 1 Otherwise:	_	_	Count exceeded: 1	Effective address error: 1
	end: 1	0	Otherwise: 0	0			Otherwise: 0	Otherwise: 0

#### Notes on status read mode

After exiting auto-program mode or auto-erase mode, status read mode must be executed without dropping the power supply.

Immediately after powering on, or once powering off, the return command is undefined.

#### 15.8.8 PROM Mode Transition Time

Commands cannot be accepted during the oscillation stabilization period or the PROM mode setup period. After the PROM mode setup time, a transition is made to memory read mode.

Table 15.20 Stipulated Transition Times to Command Wait State

Item	Symbol	Min	Max	Unit	Notes
Standby release (oscillation settling time)	t <sub>osc1</sub>	20	_	ms	
PROM mode setup time	t <sub>bmv</sub>	10	_	ms	
V <sub>cc</sub> hold time	t <sub>dwn</sub>	0	_	ms	

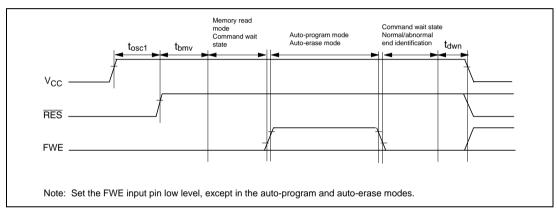


Figure 15.23 Oscillation Stabilization Time, Boot Program Transfer Time

# 15.8.9 Notes on Memory Programming

- When programming addresses which have previously been programmed, carry out autoerasing before auto-programming (figure 15.24).
- When performing programming using PROM mode on a chip that has been programmed/erased in an on-board programming mode, auto-erasing is recommended before carrying out auto-programming.
- Notes: 1. The flash memory is initially in the erased state when the device is shipped by Renesas. For other chips for which the erasure history is unknown, it is recommended that autoerasing be executed to check and supplement the initialization (erase) level.
  - 2. In the PROM mode, auto-programming to a 128-byte programming unit block should be performed only once.

Do not perform additional programming to a programmed 128-byte programming unit block.

To reprogram, perform auto-programming after auto-erasing.

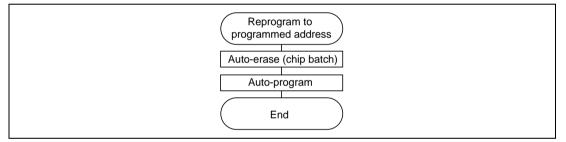


Figure 15.24 Reprogramming to Programmed Address

# 15.9 Notes on Flash Memory Programming/Erasing

The following describes notes when using the on-board programming mode, RAM emulation function, and PROM mode.

(1) Program/erase with the specified voltage and timing.

Applied voltages in excess of the rating can permanently damage the device.

Use a PROM writer that supports the Renesas Technology 128 kbytes flash memory on-board microcomputer device type.

Do not set the PROM writer at the HN28F101. If the PROM writer is set to the HN28F101 by mistake, a high level can be input to the FWE pin and the LSI can be destroyed.

(2) Notes on powering on/powering off (See figures 15.25 to 15.27.)

Input a high level to the FWE pin after verifying Vcc. Before turning off Vcc, set the FWE pin to a low level.

When powering on and powering off the Vcc power supply, fix the FWE pin a low level and set the flash memory to the hardware protection mode.

Be sure that the powering on and powering off timing is satisfied even when the power is turned off and back on in the event of a power interruption, etc. If this timing is not satisfied, microcomputer runaway, etc., may cause overprogramming or overerasing and the memory cells may not operate normally.

(3) Notes on FWE pin High/Low switching (See figures 15.25 to 15.27.)

Input FWE in the state microcomputer operation is verified. If the microcomputer does not satisfy the operation confirmation state, fix the FWE pin at a low level to set the protection mode.

To prevent erroneous programming/erasing of flash memory, note the following in FWE pin High/Low switching:

- Apply an input to the FWE pin after the Vcc voltage has stabilized within the rated voltage. If an input is applied to the FWE pin when the microcomputer Vcc voltage does not satisfy the rated voltage, flash memory may be erroneously programmed or erased because the microcomputer is in the unconfirmed state.
- Apply an input to the FWE pin when the oscillation has stabilized (after the oscillation stabilization time).

When turning on the Vcc power, apply an input to the FWE pin after holding the  $\overline{\text{RES}}$  pin at a low level during the oscillation stabilization time ( $t_{osc1}$ =20ms). Do not apply an input to the FWE pin when oscillation is stopped or unstable.

• In the boot mode, perform FWE pin High/Low switching during reset.

In transition to the boot mode, input FWE = High level and set  $MD_2$  to  $MD_0$  while the  $\overline{RES}$  input is low. At this time, the FWE and  $MD_2$  to  $MD_0$  inputs must satisfy the mode programming setup time  $(t_{MDS})$  relative to the reset clear timing. The mode programming setup time is necessary for  $\overline{RES}$  reset timing even in transition from the boot mode to another mode.

In reset during operation, the  $\overline{\text{RES}}$  pin must be held at a low level for at least 20 system clocks.

• In the user program mode, FWE = High/Low switching is possible regardless of the RES input.

FWE input switching is also possible during program execution on flash memory.

# Renesas

- Apply an input to FWE when the program is not running away. When applying an input to the FWE pin, the program execution state must be supervised using a watchdog timer, etc.
- Input low level to the FWE pin when the SWE, ESU, PSU, EV, PV, E, and P bits in FLMCR have been cleared.

Do not erroneously set the SWE, ESU, PSU, EV, PV, E, and P bits when FWE High/Low.

(4) Do not input a constant high level to the FWE pin.

To prevent erroneous programming/erasing in the event of program runaway, etc., input a high level to the FWE pin only when programming/erasing flash memory (including flash memory emulation by RAM). Avoid system configurations that constantly input a high level to the FWE pin. Handle program runaway, etc. by starting the watchdog timer so that flash memory is not overprogrammed/overerased even while a high level is input to the FWE pin.

(5) Program/erase the flash memory in accordance with the recommended algorithms. The recommended algorithms can program/erase the flash memory without applying voltage stress to the device or sacrificing the reliability of the program data. When setting the PSU and ESU bits in FLMCR, set the watchdog timer for program runaway, etc.

- (6) Do not set/clear the SWE bit while a program is executing on flash memory. Before performing flash memory program execution or data read, clear the SWE bit. If the SWE bit is set, the flash data can be reprogrammed, but flash memory cannot be accessed for purposes other than verify (verify during programming/erase). Similarly perform flash memory program execution and data read after clearing the SWE bit even when using the RAM emulation function with a high level input to the FWE pin. However, RAM area that overlaps flash memory space can be read/programmed whether the SWE bit is set or cleared.
- (7) Do not use an interrupt during flash memory programming or erasing. Since programming/erase operations (including emulation by RAM) have priority when a high level is input to the FWE pin, disable all interrupt requests, including NMI.
- (8) Do not perform additional programming. Reprogram flash memory after erasing.
  With on-board programming, program to 32-byte programming unit blocks one time only.
  Program to 128-byte programming unit blocks one time only even in the PROM mode. Erase all the programming unit blocks before reprogramming.
  Bus release must also be disabled.

- (9) Before programming, check that the chip is correctly mounted in the PROM programmer. Overcurrent damage to the device can result if the index marks on the PROM programmer socket, socket adapter, and chip are not correctly aligned.
- (10) Do not touch the socket adapter or chip during programming. Touching either of these can cause contact faults and write errors.

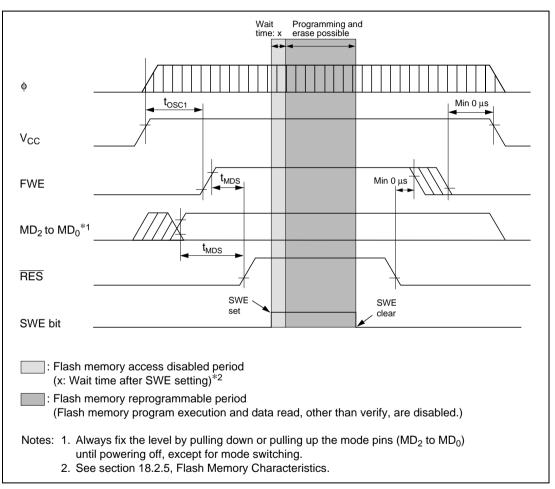


Figure 15.25 Powering On/Off Timing (Boot Mode)

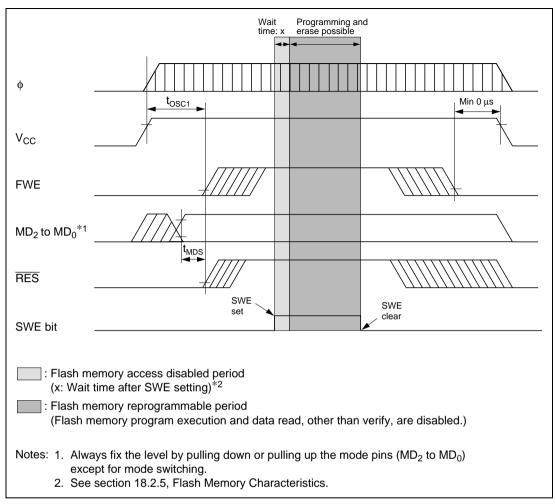
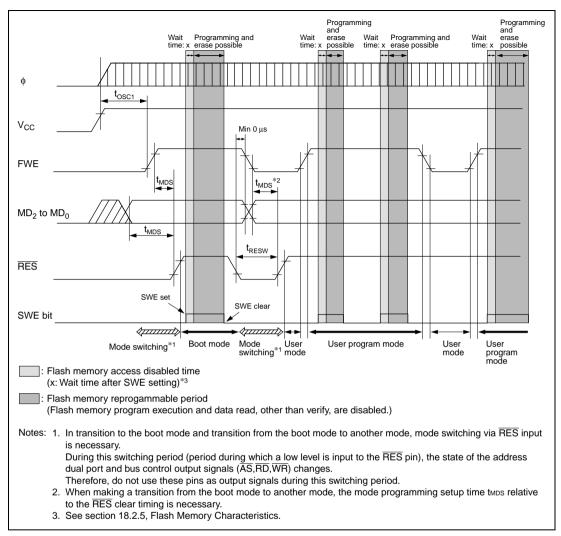


Figure 15.26 Powering On/Off Timing (User Program Mode)



# Figure 15.27 Mode Transition Timing (Example: Boot Mode → User Mode ↔ User Program Mode)

# 15.10 Mask ROM Overview

# 15.10.1 Block Diagram

Figure 15.28 shows a block diagram of the ROM.

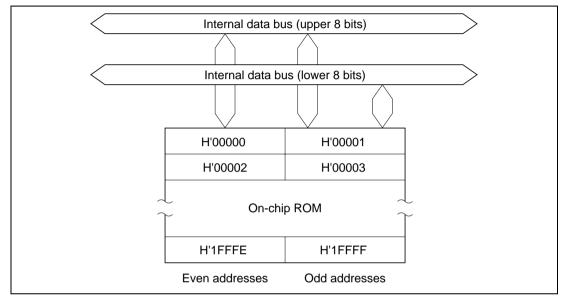


Figure 15.28 ROM Block Diagram (H8/3039)

# 15.11 Notes on Ordering Mask ROM Version Chip

When ordering the H8/3039 Group chips with a mask ROM, note the following.

- When ordering through an EPROM, use a 128-kbyte one.
- Fill all the unused addresses with H'FF as shown in figure15.29 to make the ROM data size 128 kbytes for all H8/3039 Group chips, which incorporate different sizes of ROM. This applies to ordering through an EPROM and through electrical data transfer.
- The flash memory versions only registers for flash memory control (FLMCR, EBR, RAMCR, and FLMSR) are not provided in the mask ROM versions. Reading the corresponding addresses in a mask ROM version will always return 1s, and writes to these addresses are disabled. This must be borne in mind when switching from the flash memory versions to a mask ROM version.

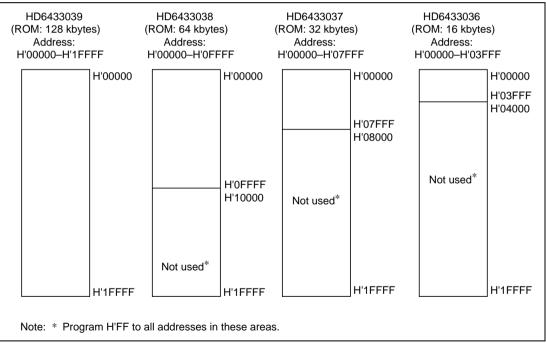


Figure 15.29 Mask ROM Addresses and Data

# Renesas

# Section 16 Clock Pulse Generator

# 16.1 Overview

This LSI has a built-in clock pulse generator (CPG) that generates the system clock ( $\phi$ ) and other internal clock signals ( $\phi/2$  to  $\phi/4096$ ). After duty adjustment, a frequency divider divides the clock frequency to generate the system clock ( $\phi$ ). The system clock is output at the  $\phi$  pin<sup>\*1</sup> and furnished as a master clock to prescalers that supply clock signals to the on-chip supporting modules. Frequency division ratios of 1/1, 1/2, 1/4, and 1/8 can be selected for the frequency divider by settings in a division control register (DIVCR). Power consumption in the chip is reduced in almost direct proportion to the frequency division ratio<sup>\*2</sup>.

- Notes: 1. Usage of the φ pin differs depending on the chip operating mode and the PSTOP bit setting in the module standby control register (MSTCR). For details, see section 17.7, System Clock Output Disabling Function.
  - 2. The division ratio of the frequency divider can be changed dynamically during operation. The clock output at the  $\phi$  pin also changes when the division ratio is changed. The frequency output at the  $\phi$  pin is shown below.

 $\phi = EXTAL \times n$ 

EXTAL: Frequency of crystal resonator or external clock signal n: Frequency division ratio (n = 1/1, 1/2, 1/4, or 1/8)

# Renesas

### 16.1.1 Block Diagram

Figure 16.1 shows a block diagram of the clock pulse generator.

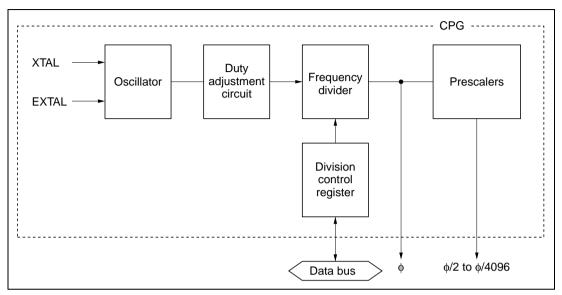


Figure 16.1 Block Diagram of Clock Pulse Generator

# 16.2 Oscillator Circuit

Clock pulses can be supplied by connecting a crystal resonator, or by input of an external clock signal.

### 16.2.1 Connecting a Crystal Resonator

# **Circuit Configuration**

A crystal resonator can be connected as in the example in figure 16.2. The damping resistance Rd should be selected according to table 16.1. An AT-cut parallel-resonance crystal should be used.

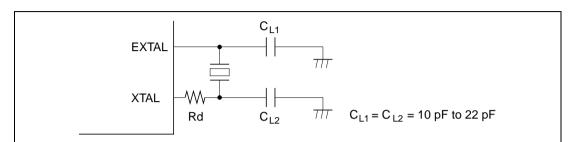


Figure 16.2 Connection of Crystal Resonator (Example)

<b>Table 16.1</b>	Damping	Resistance	Value	(Example)
-------------------	---------	------------	-------	-----------

Frequency (MHz)	2	4	8	10	12	16	18
Rd (Ω)	1 k	500	200	0	0	0	0

#### **Crystal Resonator**

Figure 16.3 shows an equivalent circuit of the crystal resonator. The crystal resonator should have the characteristics listed in table 16.2.

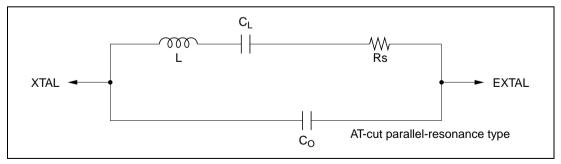


Figure 16.3 Crystal Resonator Equivalent Circuit

<b>Table 16.2</b>	<b>Crystal Resonator Parameters</b>
-------------------	-------------------------------------

Frequency (MHz)	2	4	8	10	12	16	18	
Rs max (Ω)	500	120	80	70	60	50	40	
Co max (pF)	7	7	7	7	7	7	7	

Use a crystal resonator with a frequency equal to the system clock frequency  $(\phi)$ .

#### Notes on Board Design

When a crystal resonator is connected, the following points should be noted:

Other signal lines should be routed away from the oscillator circuit to prevent induction from interfering with correct oscillation. See figure 16.4.

When the board is designed, the crystal resonator and its load capacitors should be placed as close as possible to the XTAL and EXTAL pins.

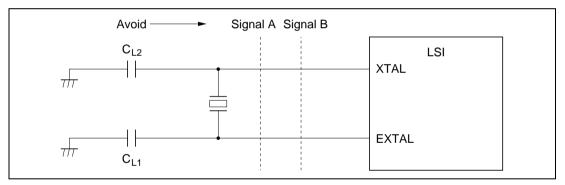


Figure 16.4 Example of Incorrect Board Design

#### 16.2.2 External Clock Input

#### **Circuit Configuration**

An external clock signal can be input as shown in the examples in figure 16.5. In example b, the clock should be held high in standby mode.

If the XTAL pin is left open, the stray capacitance should not exceed 10 pF.

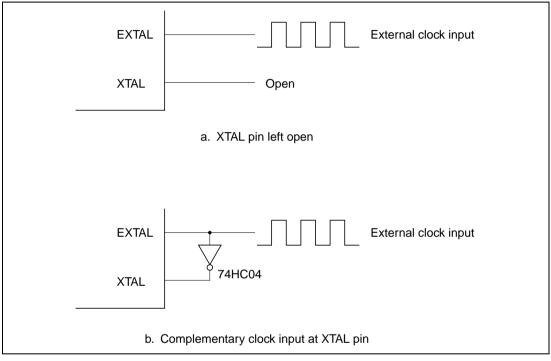


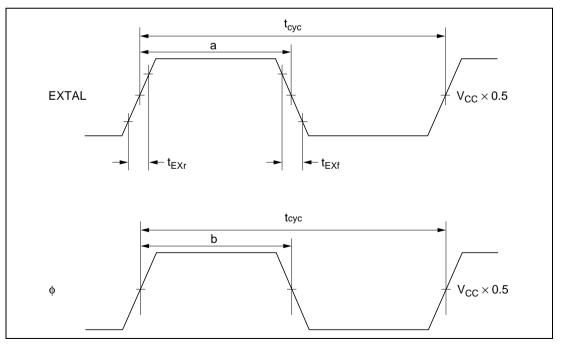
Figure 16.5 External Clock Input (Examples)

### **External Clock**

The external clock frequency should be equal to the system clock frequency ( $\phi$ ). Table 16.3 and figure 16.6 indicate the clock timing.

### Table 16.3 Clock Timing

			V <sub>cc</sub> = / to 5.5 V		V <sub>cc</sub> = V ±10%		
ltem	Symbol	Min	Max	Min	Max	Unit	Test Conditions
External clock rise time	$\mathbf{t}_{_{\mathrm{EXr}}}$	_	10	_	5	ns	Figure 16.6
External clock fall time	t <sub>exf</sub>	—	10	—	5	ns	_
External clock	_	30	70	30	70	%	$\phi \ge 5 \text{ MHz}$ Figure
input duty (a/t <sub>cyc</sub> )		40	60	40	60	%	φ < 5 MHz <sup>16.6</sup>
$\phi$ clock width duty (b/t <sub>cyc</sub> )	_	40	60	40	60	%	



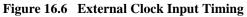
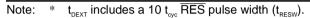


Table 16.4 and figure 16.7 show the timing for the external clock output stabilization delay time. The oscillator and duty correction circuit have the function of regulating the waveform of the external clock input to the EXTAL pin. When the specified clock signal is input to the EXTAL pin, internal clock signal output is confirmed after the elapse of the external clock output stabilization delay time ( $t_{DEXT}$ ). As clock signal output is not confirmed during the  $t_{DEXT}$  period, the reset signal should be driven low and the reset state maintained during this time.

### Table 16.4 External Clock Output Stabilization Delay Time

Conditions:  $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ 

Item	Symbol	Min	Мах	Unit	Notes
External clock output stabilization delay time	t <sub>dext</sub> *	500	—	μs	Figure 16.7



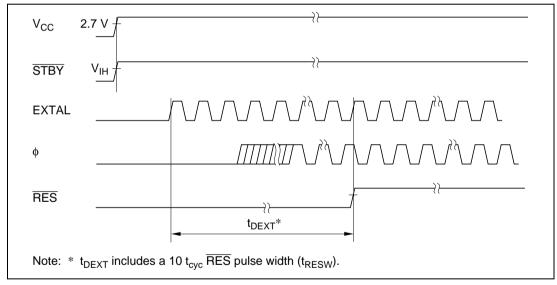


Figure 16.7 External Clock Output Stabilization Delay Time

# 16.3 Duty Adjustment Circuit

When the oscillator frequency is 5 MHz or higher, the duty adjustment circuit adjusts the duty cycle of the clock signal from the oscillator to generate the system clock ( $\phi$ ).

# 16.4 Prescalers

The prescalers divide the system clock ( $\phi$ ) to generate internal clocks ( $\phi$ /2 to  $\phi$ /4096).

# 16.5 Frequency Divider

The frequency divider divides the duty-adjusted clock signal to generate the system clock ( $\phi$ ). The frequency division ratio can be changed dynamically by modifying the value in DIVCR, as described below. Power consumption in the chip is reduced in almost direct proportion to the frequency division ratio. The system clock generated by the frequency divider can be output at the  $\phi$  pin.

# 16.5.1 Register Configuration

Table 16.5 summarizes the frequency division register.

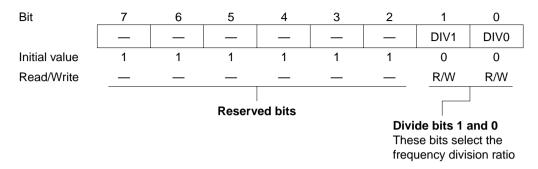
# Table 16.5Frequency Division Register

Address*	Name	Abbreviation	R/W	Initial Value
H'FF5D	Division control register	DIVCR	R/W	H'FC
Note: * The lower 16 bits of the address are shown				

Note: \* The lower 16 bits of the address are shown.

#### 16.5.2 Division Control Register (DIVCR)

DIVCR is an 8-bit readable/writable register that selects the division ratio of the frequency divider.



DIVCR is initialized to H'FC by a reset and in hardware standby mode. It is not initialized in software standby mode.

Bits 7 to 2—Reserved: These bits cannot be modified and are always read as 1.

**Bits 1 and 0—Divide (DIV1 and DIV0):** These bits select the frequency division ratio, as follows.

Bit 1 DIV1	Bit 0 DIV0	Frequency Division Ratio	
0	0	1/1	(Initial value)
0	1	1/2	
1	0	1/4	
1	1	1/8	

# 16.5.3 Usage Notes

The DIVCR setting changes the  $\phi$  frequency, so note the following points.

• Select a frequency division ratio that stays within the assured operation range specified for the clock cycle time  $t_{cyc}$  in the AC electrical characteristics. Note that  $\phi_{MIN} = 1$  MHz. Avoid settings that give system clock frequencies less than 1 MHz.

All on-chip module operations are based on  $\phi$ . Note that the timing of timer operations, serial communication, and other time-dependent processing differs before and after any change in the division ratio. The waiting time for exit from software standby mode also changes when the division ratio is changed. For details, see section 17.4.3, Selection of Oscillator Waiting Time after Exit from Software Standby Mode

# Section 17 Power-Down State

# 17.1 Overview

This LSI has a power-down state that greatly reduces power consumption by halting CPU functions, and a module standby function that reduces power consumption by selectively halting on-chip modules. The power-down state includes the following three modes:

- Sleep mode
- Software standby mode
- Hardware standby mode

The module standby function can halt on-chip supporting modules independently of the powerdown state. The modules that can be halted are the ITU, SCI0, SCI1, and A/D converter.

Table 17.1 indicates the methods of entering and exiting these power-down modes and the status of the CPU and on-chip supporting modules in each mode.

			<u>a</u>		ск		
	Exiting Methods	<ul> <li>Interrupt</li> <li>RES</li> <li>STBY</li> </ul>	• NMI • IRQ <sub>0</sub> to IRQ <sub>1</sub> • <u>RES</u>	• <u>STBY</u> • RES	STBY     RES     Clear MSTCR     bit to 0*3		
	l/O Ports	Held	Held	High impedance	I		· (MSTCR). mode. nodule, first
	<pre></pre>	øoutput	High output	High impedance	High impedance*1		control Register tware standby Fo restart the n
	RAM	Held	Held	Held*2	I		tandby C te to harc itialized. <sup>-</sup>
	Supporting Modules	Active	Halted and reset	Halted and reset	Active		<ol> <li>State in which the corresponding MSTCR bit was set to 1. For details see section 17.2.2, Module Standby Control Register (MSTCR).</li> <li>The RAME bit must be cleared to 0 in SYSCR before the transition from the program execution state to hardware standby mode.</li> <li>When a MSTCR bit is set to 1, the registers of the corresponding on-chip supporting module are initialized. To restart the module, first clear the MSTCR bit to 0, then set up the module registers again.</li> </ol>
State	AD	Active	Halted and reset	Halted and reset	Halted <sup>*1</sup> and reset		e section 1 the progra supportin
	SCI1	Active	Halted and reset	Halted and reset	Halted <sup>*1</sup> and reset		details see tion from ig on-chip in.
	SCI0	Active	Halted and reset	Halted and reset	Halted <sup>*1</sup> and reset		to 1. For- the trans rrespondii isters aga
	E	Active	Halted and reset	Halted and reset	Halted <sup>*1</sup> and reset		it was set CR before s of the co nodule reg
	CPU Registers	Held	Held	Undeter mined	1		g MSTCR I to 0 in SYS he registers set up the n
	СРU	Halted	Halted	Halted	Active	ar ol registe	espondin cleared set to 1, t 0, then s
	Clock	Active	Halted	Halted	Active	ol registe Idby bit Iby contre	the corre must be CR bit is 5 CR bit to
	Entering Conditions	SLEEP instruc- Active tion executed while SSBY = 0 in SYSCR	SLEEP instruc- tion executed while SSBY = 1 in SYSCR	Hardware Low input at standby <u>STBY</u> pin mode	Corresponding bit set to 1 in MSTCR	System control register Software standby bit Module standby control register	<ul> <li>State in which the corresponding MSTCR bit was set to 1. For de The RAME bit must be cleared to 0 in SYSCR before the transitic When a MSTCR bit is set to 1, the registers of the corresponding clear the MSTCR bit to 0, then set up the module registers again.</li> </ul>
	Mode	Sleep mode	Software standby mode	Hardware standby mode	Module standby function	Legend: SYSCR: SSBY: MSTCR:	Notes: 1 3. 2.

 Table 17.1
 Power-Down State and Module Standby Function

Section 17 Power-Down State

# 17.2 Register Configuration

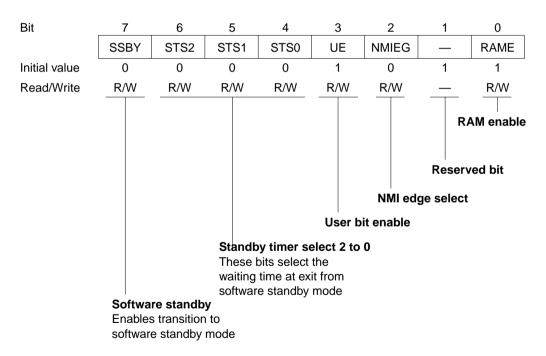
This LSI has a system control register (SYSCR) that controls the power-down state, and a module standby control register (MSTCR) that controls the module standby function. Table 17.2 summarizes this register.

 Table 17.2
 Register Configuration

Address*	Name	Abbreviation	R/W	Initial Value
H'FFF2	System control register	SYSCR	R/W	H'0B
H'FF5E	Module standby control register	MSTCR	R/W	H'40
Nata: * L	awar 40 bita af tha address			

Note: \* Lower 16 bits of the address.

#### 17.2.1 System Control Register (SYSCR)



SYSCR is an 8-bit readable/writable register. Bit 7 (SSBY) and bits 6 to 4 (STS2 to STS0) control the power-down state. For information on the other SYSCR bits, see section 3.3, System Control Register (SYSCR).

**Bit 7—Software Standby (SSBY):** Enables transition to software standby mode. When software standby mode is exited by an external interrupt, this bit remains set to 1 after the return to normal operation. To clear this bit, write 0.

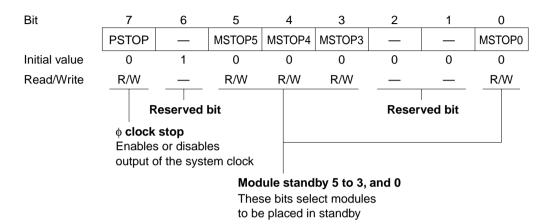
Bit 7		
SSBY	Description	
0	SLEEP instruction causes transition to sleep mode	(Initial value)
1	SLEEP instruction causes transition to software standby mode	

**Bits 6 to 4—Standby Timer Select (STS2 to STS0):** These bits select the length of time the CPU and on-chip supporting modules wait for the clock to settle when software standby mode is exited by an external interrupt. If the clock is generated by a crystal resonator, set these bits according to the clock frequency so that the waiting time (for the clock to stabilize) will be at least 7 ms. See table 17.3. If an external clock is used, any setting is permitted.

Bit 6 STS2	Bit 5 STS1	Bit 4 STS0	Description	
0	0	0	Waiting time = 8192 states	(Initial value)
		1	Waiting time = 16384 states	
	1	0	Waiting time = 32768 states	
		1	Waiting time = 65536 states	
1	0	0	Waiting time = 131072 states	
	0	1	Waiting time = 1024 states	
	1	—	Illegal setting	

### 17.2.2 Module Standby Control Register (MSTCR)

MSTCR is an 8-bit readable/writable register that controls output of the system clock ( $\phi$ ). It also controls the module standby function, which places individual on-chip supporting modules in the standby state. Module standby can be designated for the ITU, SCI0, SCI1, and A/D converter modules.



MSTCR is initialized to H'40 by a reset and in hardware standby mode. It is not initialized in software standby mode.

**Bit 7—φ Clock Stop (PSTOP):** Enables or disables output of the system clock (φ).

Bit 7 PSTOP	Description	
0	System clock output is enabled	(Initial value)
1	System clock output is disabled	

Bit 6—Reserved: This bit cannot be modified and is always read as 1.

Bit 5—Module Standby 5 (MSTOP5): Selects whether to place the ITU in standby.

Bit 5 MSTOP5	Description	
0	ITU operates normally	(Initial value)
1	ITU is in standby state	

Bit 4—Module Standby 4 (MSTOP4): Selects whether to place SCI0 in standby.

Bit 4 MSTOP4	Description	
0	SCI0 operates normally	(Initial value)
1	SCI0 is in standby state	

Bit 3—Module Standby 3 (MSTOP3): Selects whether to place SCI1 in standby.

Bit 3 MSTOP3	Description	
0	SCI1 operates normally	(Initial value)
1	SCI1 is in standby state	

Bits 2 to 1—Reserved: Bits 2 to 1 are reserved.

Bit 0—Module Standby 0 (MSTOP0): Selects whether to place the A/D converter in standby.

Bit 0 MSTOP0	Description	
0	A/D converter operates normally	(Initial value)
1	A/D converter is in standby state	

# 17.3 Sleep Mode

### 17.3.1 Transition to Sleep Mode

When the SSBY bit is cleared to 0 in the system control register (SYSCR), execution of the SLEEP instruction causes a transition from the program execution state to sleep mode. Immediately after executing the SLEEP instruction the CPU halts, but the contents of its internal registers are retained. The on-chip supporting modules do not halt in sleep mode. On-chip supporting modules which have been placed in standby by the module standby function, however, remain halted.

### 17.3.2 Exit from Sleep Mode

Sleep mode is exited by an interrupt, or by input at the  $\overline{\text{RES}}$  or  $\overline{\text{STBY}}$  pin.

**Exit by Interrupt:** An interrupt terminates sleep mode and causes a transition to the interrupt exception handling state. Sleep mode is not exited by an interrupt source in an on-chip supporting module if the interrupt is disabled in the on-chip supporting module. Sleep mode is not exited by an interrupt other then NMI if the interrupt is masked by interrupt priority settings (IPR) and the settings of the I and UI bits in CCR.

**Exit by RES** Input: Low input at the  $\overline{RES}$  pin exits from sleep mode to the reset state.

**Exit by STBY Input:** Low input at the STBY pin exits from sleep mode to hardware standby mode.

# 17.4 Software Standby Mode

# 17.4.1 Transition to Software Standby Mode

To enter software standby mode, execute the SLEEP instruction while the SSBY bit is set to 1 in SYSCR.

In software standby mode, current dissipation is reduced to an extremely low level because the CPU, clock, and on-chip supporting modules all halt. The on-chip supporting modules are reset and halted. As long as the specified voltage is supplied, however, CPU register contents and on-chip RAM data are retained. The settings of the I/O ports are also held.

# 17.4.2 Exit from Software Standby Mode

Software standby mode can be exited by input of an external interrupt at the NMI,  $\overline{IRQ}_0$ ,  $\overline{IRQ}_1$ , or by input at the  $\overline{RES}$  or  $\overline{STBY}$  pin.

**Exit by Interrupt:** When an NMI,  $IRQ_0$ , or  $IRQ_1$  interrupt request signal is received, the clock oscillator begins operating. After the oscillator settling time selected by bits STS2 to STS0 in SYSCR, stable clock signals are supplied to the entire chip, software standby mode ends, and interrupt exception handling begins. Software standby mode is not exited if the interrupt enable bits of interrupts  $IRQ_0$ , and  $IRQ_1$  are cleared to 0, or if these interrupts are masked in the CPU.

**Exit by RES Input:** When the RES input goes low, the clock oscillator starts and clock pulses are supplied immediately to the entire chip. The RES signal must be held low long enough for the clock oscillator to stabilize. When RES goes high, the CPU starts reset exception handling.

**Exit by STBY Input:** Low input at the STBY pin causes a transition to hardware standby mode.

#### 17.4.3 Selection of Oscillator Waiting Time after Exit from Software Standby Mode

Bits STS2 to STS0 in SYSCR, and its DIV1 and DIV0 in DIVCR should be set as follows.

#### **Crystal Resonator**

Set STS2 to STS0, and DIV1 and DIV0 so that the waiting time (for the clock to stabilize) is at least 7 ms. Table 17.3 indicates the waiting times that are selected by STS2 to STS0, and DIV1 and DIV0 settings at various system clock frequencies.

#### **External Clock**

Any value may be set.

#### Table 17.3 Clock Frequency and Waiting Time for Clock to Settle

DIV1	DIV0	STS2	STS1	STS0	Waiting Time	18 MHz	16 MHz	12 MHz	10 MHz	8 MHz	6 MHz	4 MHz	2 MHz	1 MHz	Unit
0	0	0	0	0	8192 states	0.46	0.51	0.65	0.8	1.0	1.3	2.0	4.1	8.2	ms
		0	0	1	16384 states	0.91	1.0	1.3	1.6	2.0	2.7	4.1	8.2	16.4	
		0	1	0	32768 states	1.8	2.0	2.7	3.3	4.1	5.5	8.2	16.4	32.8	
		0	1	1	65536 states	3.6	4.1	5.5	6.6	8.2	10.9	16.4	32.8	65.5	
		1	0	0	131072 states	7.3	8.2	10.9	13.1	16.4	21.8	32.8	65.5	131.1	
		1	0	1	1024 states	0.057	0.064	0.085	0.10	0.13	0.17	0.26	0.51	1.0	
		1	1	_	Illegal setting										
0	1	0	0	0	8192 states	0.91	1.02	1.4	1.6	2.0	2.7	4.1	8.2	16.4	ms
		0	0	1	16384 states	1.8	2.0	2.7	3.3	4.1	5.5	8.2	16.4	32.8	
		0	1	0	32768 states	3.6	4.1	5.5	6.6	8.2	10.9	16.4	32.8	65.5	
		0	1	1	65536 states	7.3	8.2	10.9	13.1	16.4	21.8	32.8	65.5	131.1	
		1	0	0	131072 states	14.6	16.4	21.8	26.2	32.8	43.7	65.5	131.1	262.1	
		1	0	1	1024 states	0.11	0.13	0.17	0.20	0.26	0.34	0.51	1.0	2.0	
		1	1	_	Illegal setting										
1	0	0	0	0	8192 states	1.8	2.0	2.7	3.3	4.1	5.5	8.2	16.4	32.8	ms
		0	0	1	16384 states	3.6	4.1	5.5	6.6	8.2	10.9	16.4	32.8	65.5	
		0	1	0	32768 states	7.3	8.2	10.9	13.1	16.4	21.8	32.8	65.5	131.1	
		0	1	1	65536 states	14.6	16.4	21.8	26.2	32.8	43.7	65.5	131.1	262.1	
		1	0	0	131072 states	29.1	32.8	43.7	52.4	65.5	87.4	131.1	262.1	524.3	
		1	0	1	1024 states	0.23	0.26	0.34	0.41	0.51	0.68	1.02	2.0	4.1	
		1	1	_	Illegal setting										
1	1	0	0	0	8192 states	3.6	4.1	5.5	6.6	8.2	10.9	16.4	32.8	65.5	ms
		0	0	1	16384 states	7.3	8.2	10.9	13.1	16.4	21.8	32.8	65.5	131.1	
		0	1	0	32768 states	14.6	16.4	21.8	26.2	32.8	43.7	65.5	131.1	262.1	
		0	1	1	65536 states	29.1	32.8	43.7	52.4	65.5	87.4	131.1	262.1	524.3	
		1	0	0	131072 states	58.3	65.5	87.4	104.9	131.1	174.8	262.1	524.3	1048.6	
		1	0	1	1024 states	0.46	0.51	0.68	0.82	1.0	1.4	2.0	4.1	8.2	
		1	1	—	Illegal setting										

: Recommended setting

# 17.4.4 Sample Application of Software Standby Mode

Figure 17.1 shows an example in which software standby mode is entered at the fall of NMI and exited at the rise of NMI.

With the NMI edge select bit (NMIEG) cleared to 0 in SYSCR (selecting the falling edge), an NMI interrupt occurs. Next the NMIEG bit is set to 1 (selecting the rising edge) and the SSBY bit is set to 1; then the SLEEP instruction is executed to enter software standby mode.

Software standby mode is exited at the next rising edge of the NMI signal.

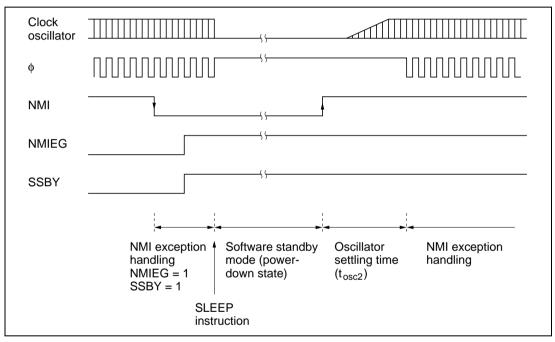


Figure 17.1 NMI Timing for Software Standby Mode (Example)

# 17.4.5 Usage Note

The I/O ports retain their existing states in software standby mode. If a port is in the high output state, its output current is not reduced.

### 17.5 Hardware Standby Mode

### 17.5.1 Transition to Hardware Standby Mode

Regardless of its current state, the chip enters hardware standby mode whenever the STBY pin goes low. Hardware standby mode reduces power consumption drastically by halting all functions of the CPU and on-chip supporting modules. All modules are reset except the on-chip RAM. As long as the specified voltage is supplied, on-chip RAM data is retained. I/O ports are placed in the high-impedance state.

Clear the RAME bit to 0 in SYSCR before  $\overline{\text{STBY}}$  goes low to retain on-chip RAM data.

The inputs at the mode pins  $(MD_2 \text{ to } MD_0)$  should not be changed during hardware standby mode.

### 17.5.2 Exit from Hardware Standby Mode

Hardware standby mode is exited by inputs at the  $\overline{\text{STBY}}$  and  $\overline{\text{RES}}$  pins. While  $\overline{\text{RES}}$  is low, when  $\overline{\text{STBY}}$  goes high, the clock oscillator starts running.  $\overline{\text{RES}}$  should be held low long enough for the clock oscillator to settle. When  $\overline{\text{RES}}$  goes high, reset exception handling begins, followed by a transition to the program execution state.

### 17.5.3 Timing for Hardware Standby Mode

Figure 17.2 shows the timing relationships for hardware standby mode. To enter hardware standby mode, first drive  $\overline{\text{RES}}$  low, then drive  $\overline{\text{STBY}}$  low. To exit hardware standby mode, first drive  $\overline{\text{STBY}}$  high, wait for the clock to settle, then bring  $\overline{\text{RES}}$  from low to high.

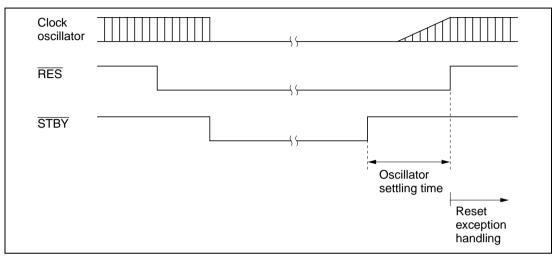


Figure 17.2 Hardware Standby Mode Timing

### 17.6 Module Standby Function

### 17.6.1 Module Standby Timing

The module standby function can halt several of the on-chip supporting modules (the ITU, SCI0, SCI1, and A/D converter) independently of the power-down state. This standby function is controlled by bits MSTOP5 to MSTOP3 and MSTOP0 in MSTCR. When one of these bits is set to 1, the corresponding on-chip supporting module is placed in standby and halts at the beginning of the next bus cycle after the MSTCR write cycle.

### 17.6.2 Read/Write in Module Standby

When an on-chip supporting module is in module standby, read/write access to its registers is disabled. Read access always results in H'FF data. Write access is ignored.

### 17.6.3 Usage Notes

When using the module standby function, note the following points.

**Cancellation of Interrupt Handling:** When an on-chip supporting module is placed in standby by the module standby function, its registers are initialized, including registers with interrupt request flags. Consequently, if an interrupt occurs just before the MSTOP bit is set to 1, the interrupt will not be recognized. The interrupt source will not be held pending.

**Pin States:** Pins used by an on-chip supporting module lose their module functions when the module is placed in module standby. What happens after that depends on the particular pin. For details, see section 7, I/O Ports. Pins that change from the input to the output state require special care. For example, if SCI1 is placed in module standby, the receive data pin loses its receive data function and becomes a generic I/O pin. If its data direction bit is set to 1, the pin becomes a data output pin, and its output may collide with external serial data. Data collisions should be prevented by clearing the data direction bit to 0 or taking other appropriate action.

**Register Resetting:** When an on-chip supporting module is halted by the module standby function, all its registers are initialized. To restart the module, after its MSTOP bit is cleared to 0, its registers must be set up again. It is not possible to write to the registers while the MSTOP bit is set to 1.

### Renesas

### 17.7 System Clock Output Disabling Function

Output of the system clock ( $\phi$ ) can be controlled by the PSTOP bit in MSTCR. When the PSTOP bit is set to 1, output of the system clock halts and the  $\phi$  pin is placed in the high-impedance state. Figure 17.3 shows the timing of the stopping and starting of system clock output. When the PSTOP bit is cleared to 0, output of the system clock is enabled. Table 17.4 indicates the state of the  $\phi$  pin in various operating states.

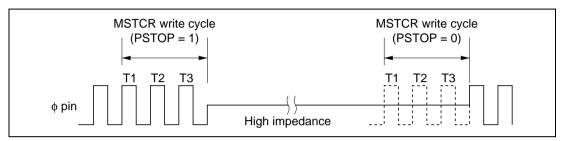


Figure 17.3 Starting and Stopping of System Clock Output

Operating State	PSTOP = 0	PSTOP = 1
Hardware standby	High impedance	High impedance
Software standby	Always high	High impedance
Sleep mode	System clock output	High impedance
Normal operation	System clock output	High impedance

## Section 18 Electrical Characteristics

### 18.1 Electrical Characteristics of Mask ROM Version

### 18.1.1 Absolute Maximum Ratings

Table 18.1 lists the absolute maximum ratings.

<b>Table 18.1</b>	Absolute	Maximum	Ratings
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Item	Symbol	Value	Unit
Power supply voltage	V <sub>cc</sub>	-0.3 to +7.0	V
Input voltage (except port 7)*	$V_{in}$	-0.3 to V <sub>cc</sub> +0.3	V
Input voltage (port 7)	$V_{in}$	–0.3 to AV <sub>cc</sub> +0.3	V
Analog power supply voltage	$AV_{cc}$	-0.3 to +7.0	V
Analog input voltage	V <sub>AN</sub>	–0.3 to AV <sub>cc</sub> +0.3	V
Operating temperature	T <sub>opr</sub>	Regular specifications: -20 to +75	°C
		Wide-range specifications: -40 to +85	°C
Storage temperature	T <sub>stg</sub>	-55 to +125	°C

Caution: Permanent damage to the chip may result if absolute maximum ratings are exceeded.

Note: \* 12 V must not be applied to any pin, as this will cause permanent damage to the chip.

### Renesas

### **18.1.2 DC Characteristics**

Table 18.2 lists the DC characteristics. Table 18.3 lists the permissible output currents.

### Table 18.2 DC Characteristics (1)

Conditions:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $AV_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = AV_{ss} = 0 \text{ V}^{*1}$ ,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt	Port A,	$V_{\tau}^{-}$	1.0		—	V	
trigger input voltages	P8 <sub>0</sub> to P8 <sub>1</sub> ,	$V_{T}^{+}$	_	_	$V_{cc}  imes 0.7$	V	_
Voltagoo	PB <sub>0</sub> to PB <sub>3</sub>	$V_{T}^{+} - V_{T}^{-}$	0.4		_	V	
Input high voltage	$\overline{\text{RES}}, \overline{\text{STBY}}, \\ \text{NMI}, \text{MD}_2, \\ \text{MD}_1, \text{MD}_0$	V <sub>IH</sub>	V <sub>cc</sub> -0.7	—	V <sub>cc</sub> +0.3	V	
	EXTAL	_	$V_{cc} \times 0.7$	_	V <sub>cc</sub> +0.3	V	_
	Port 7	_	2.0	_	AV <sub>cc</sub> +0.3	V	_
	Ports 1, 2, 3, 5, 6, 9, PB <sub>4</sub> , PB <sub>5</sub> , PB <sub>7</sub>	_	2.0	—	V <sub>cc</sub> +0.3	V	
Input low voltage	$\overline{\text{RES}}, \overline{\text{STBY}}, \\ \text{MD}_2, \text{MD}_1, \text{MD}_0$	V <sub>IL</sub>	-0.3	—	0.5	V	
	NMI, EXTAL, ports 1, 2, 3, 5, 6, 7, 9, PB <sub>4</sub> , PB <sub>5</sub> , PB <sub>7</sub>	_	-0.3	_	0.8	V	_
Output high	All output pins	V <sub>oh</sub>	V <sub>cc</sub> –0.5	_	_	V	I <sub>oH</sub> = -200 μA
voltage	(except RESO)		3.5	_	_	V	I <sub>он</sub> = —1 mA
Output low voltage	All output pins (except RESO)	V <sub>ol</sub>	—	—	0.4	V	I <sub>oL</sub> = 1.6 mA
	Ports 1, 2, 5, B		_	_	1.0	V	I <sub>oL</sub> = 10 mA
	RESO		_		0.4	V	I <sub>oL</sub> = 2.6 mA

Section 18	Electrical	Characteristics
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Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Input leakage current	$\label{eq:stby_norm} \overline{\frac{\text{STBY}}{\text{RES}}, \text{NMI},} \\ \overline{\text{RES}}, \text{MD}_2, \text{MD}_1, \\ \overline{\text{MD}}_0$	I <sub>in</sub>	_	_	1.0	μA	$V_{in} = 0.5$ to $V_{cc} - 0.5$ V
	Port 7	_	_		1.0	μA	$V_{in} = 0.5 \text{ to}$ AV <sub>cc</sub> -0.5 V
Three-state leakage	Ports 1, 2, 3, 5, 6, 8 to B	I <sub>tsi</sub>	_	_	1.0	μA	$V_{in} = 0.5 \text{ to}$ $V_{cc} - 0.5 \text{ V}$
current (off state)	RESO	_	_	_	10.0	μA	_
Input pull-up MOS current	Ports 2, 5	<b>—I</b> <sub>p</sub>	50	_	300	μA	$V_{in} = 0 V$
Input	NMI, RES	C <sub>in</sub>		—	50	pF	$V_{in} = 0 V,$
capacitance	All input pins except NMI and RES	_	_	_	20		f = 1 MHz, $T_a = 25^{\circ}C$
Current dissipation* <sup>2</sup>	Normal operation	I <sub>cc</sub>	_	50	70	mA	f = 18 MHz
	Sleep mode	-	_	35	50		f = 18 MHz
	Standby mode*3	_	_	0.01	5.0	μA	$T_a \le 50^{\circ}C$
			_	_	20.0		50°C < T <sub>a</sub>
Analog power	During A/D conversion	$AI_{cc}$	_	1.7	2.8	mA	
supply current	Idle	_	_	0.02	10.0	μA	
RAM standby	v voltage	$V_{RAM}$	2.0	_	_	V	

Notes: 1. If the A/D converter is not used, do not leave the AV<sub>cc</sub> and AV<sub>ss</sub> pins open. Connect AV<sub>cc</sub> to V<sub>cc</sub>, and connect AV<sub>ss</sub> to V<sub>ss</sub>.

2. Current dissipation values are for V<sub>IH</sub> min = V<sub>cc</sub> –0.5 V and V<sub>IL</sub> max = 0.5 V with all output pins unloaded and the on-chip pull-up transistors in the off state.

3. The values are for V<sub>RAM</sub>  $\leq$  V<sub>cc</sub> < 4.5 V, V<sub>IH</sub> min = V<sub>cc</sub>  $\times$  0.9, and V<sub>IL</sub> max = 0.3 V.

### Table 18.2 DC Characteristics (2)

Conditions:  $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}^{*1}$ ,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt	Port A,	V <sub>T</sub> <sup>-</sup>	$V_{cc}  imes 0.2$			V	
trigger input voltages	$P8_0$ to $P8_1$ , $PB_0$ to $PB_3$	$V_{T}^{+}$		_	$V_{cc}  imes 0.7$	V	_
voltages		$V_{_{T}}^{^{+}}-V_{_{T}}^{^{-}}$	$V_{cc}  imes 0.04$	—	_	V	_
Input high voltage	$\begin{tabular}{c} \hline $RES, $\overline{STBY},$\\ $NMI, $MD_2, $MD_1,$\\ $MD_0$ \end{tabular}$	V <sub>IH</sub>	$V_{cc} \times 0.9$	_	V <sub>cc</sub> +0.3	V	
	EXTAL		$V_{cc}  imes 0.7$	—	V <sub>cc</sub> +0.3	V	_
	Port 7	_	$V_{cc} \times 0.7$	_	AV <sub>cc</sub> +0.3	V	_
	Ports 1, 2, 3, 5, 6, 9, PB <sub>4</sub> , PB <sub>5</sub> , PB <sub>7</sub>	_	$V_{cc} \times 0.7$	_	V <sub>cc</sub> +0.3	V	_
Input low voltage	$\overline{\text{RES}}, \overline{\text{STBY}}, \\ \text{MD}_2, \text{MD}_1, \text{MD}_0$	V <sub>IL</sub>	-0.3	—	$V_{cc} \times 0.1$	V	
	NMI, EXTAL,	_	-0.3	_	$V_{cc}  imes 0.2$	V	V <sub>cc</sub> < 4.0 V
	ports 1, 2, 3, 5, 6, 7, 9, PB <sub>4</sub> , PB <sub>5</sub> , PB <sub>7</sub>				0.8	V	V <sub>cc</sub> = 4.0 V to 5.5 V
Output high	All output pins	V <sub>oh</sub>	V <sub>cc</sub> –0.5	_	_	V	I <sub>OH</sub> = -200 μA
voltage	(except RESO)		V <sub>cc</sub> –1.0	_	_	V	I <sub>он</sub> = -1 mA
Output low voltage	All output pins (except RESO)	$V_{\rm ol}$	_	—	0.4	V	I <sub>oL</sub> = 1.0 mA
	Ports 1, 2, 5, B		_	_	1.0	V	
	RESO	_	_	_	0.4	V	I <sub>oL</sub> = 1.6 mA
Input leakage current	$\label{eq:stby_norm} \begin{array}{c} \overline{\text{STBY}},  \text{NMI}, \\ \overline{\text{RES}},  \text{MD}_2,  \text{MD}_1, \\ \overline{\text{MD}}_0 \end{array}$	I <sub>in</sub>	_	_	1.0	μA	$V_{in} = 0.5 \text{ V to}$ $V_{cc} - 0.5 \text{ V}$
	Port 7	_		_	1.0	μA	$V_{in} = 0.5 V \text{ to}$ AV <sub>cc</sub> -0.5 V

ltem		Symbol	Min	Тур	Max	Unit	Test Conditions
Three-state leakage	Ports 1, 2, 3, 5, 6, 8, 9, A, B	I <sub>tsi</sub>	_	—	1.0	μA	$V_{in} = 0.5 V \text{ to}$ $V_{cc} - 0.5 V$
current (off state)	RESO	_	_	_	10.0	μA	_
Input pull-up MOS current	Ports 2, 5	−I <sub>p</sub>	10	_	300	μA	V <sub>cc</sub> = 2.7 V to 5.5 V, V <sub>in</sub> = 0 V
Input	NMI, RES	C <sub>in</sub>			50	pF	$V_{in} = 0 V,$
capacitance	All input pins except NMI and RES	_	_	_	20	_	f = 1 MHz , Ta = 25°C
Current dissipation* <sup>2</sup>	Normal operation	I <sub>CC</sub> * <sup>4</sup>	_	12 (3.0 V)	33.8 (5.5 V)	mA	f = 8 MHz
	Sleep mode	-	_	8 (3.0 V)	25.0 (5.5 V)	mA	f = 8 MHz
	Standby mode*3		_	0.01	5.0	μA	$T_a \le 50^{\circ}C$
			_	_	20.0	μA	50°C < T <sub>a</sub>
Analog	During A/D	Al <sub>cc</sub>	_	1.3	2.5	mA	$AV_{cc} = 3.0 V$
power supply	conversion		_	1.7	2.8	mA	$AV_{cc} = 5.0 V$
current	Idle	_	_	0.02	10.0	μA	
RAM standby	y voltage	$V_{RAM}$	2.0		_	V	

Notes: 1. If the A/D converter is not used, do not leave the  $AV_{cc}$  and  $AV_{ss}$  pins open.

Connect AV<sub>cc</sub> to V<sub>cc</sub>, and connect AV<sub>ss</sub> to V<sub>ss</sub>.

- 2. Current dissipation values are for  $V_{IH}$  min =  $V_{cc}$  –0.5 V and  $V_{IL}$  max = 0.5 V with all output pins unloaded and the on-chip pull-up MOS transistors in the off state.
- 3. The values are for V\_{\_{RAM}} \le V\_{\_{CC}} < 2.7 V, V\_{\_{IH}} min = V\_{\_{CC}} \times 0.9, and V\_{\_{IL}} max = 0.3 V.
- 4.  $I_{cc}$  depends on  $V_{cc}$  and f as follows:

 $I_{cc}$  max = 3.0 (mA) + 0.7 (mA/MHz · V) ×  $V_{cc}$  × f [normal mode]

 $I_{cc} max = 3.0 \text{ (mA)} + 0.5 \text{ (mA/MHz} \cdot \text{V}) \times \text{V}_{cc} \times \text{f [sleep mode]}$ 

### Renesas

### Table 18.2DC Characteristics (3)

Conditions:  $V_{cc} = 3.0 \text{ V}$  to 5.5 V,  $AV_{cc} = 3.0 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}^{*1}$ ,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

Item		Symbol	Min	Тур	Мах	Unit	Test Conditions
Schmitt	Port A,	V <sub>T</sub> <sup>-</sup>	$V_{cc}  imes 0.2$	_	_	V	
trigger input voltages	$P8_0$ to $P8_1$ , $PB_0$ to $PB_3$	$V_{T}^{+}$	_	_	$V_{cc}  imes 0.7$	V	_
voltages		$\overline{V_T^+ - V_T^-}$	$V_{cc}  imes 0.04$	_	_	V	_
Input high voltage	$\begin{tabular}{c} \hline $RES$, $\overline{STBY}$, $$NMI, $MD_2$, $MD_1$, $$MD_0$ \end{tabular}$	V <sub>IH</sub>	$V_{cc}  imes 0.9$	_	V <sub>cc</sub> +0.3	V	
	EXTAL		$V_{cc}  imes 0.7$	—	V <sub>cc</sub> +0.3	V	_
	Port 7	_	$V_{cc}  imes 0.7$	—	AV <sub>cc</sub> +0.3	V	_
	Ports 1, 2, 3, 5, 6, 9, PB <sub>4</sub> , PB <sub>5</sub> , PB <sub>7</sub>	_	$V_{cc} \times 0.7$	_	V <sub>cc</sub> +0.3	V	_
Input low voltage	$\overline{\text{RES}}, \overline{\text{STBY}}, \\ \text{MD}_2, \text{MD}_1, \text{MD}_0$	V <sub>IL</sub>	-0.3	—	$V_{cc} \times 0.1$	V	
	NMI, EXTAL,		-0.3		$V_{cc}  imes 0.2$	V	V <sub>cc</sub> < 4.0 V
	ports 1, 2, 3, 5, 6, 7, 9, PB₄, PB₅, PB <sub>7</sub>				0.8	V	$V_{cc} = 4.0 V$ to 5.5 V
Output high	All output pins	V <sub>oh</sub>	$V_{cc}$ –0.5	_	_	V	I <sub>oH</sub> = -200 μA
voltage	(except RESO)		V <sub>cc</sub> –1.0	_	_	V	I <sub>он</sub> = -1 mA
Output low voltage	All output pins (except RESO)	$V_{\rm ol}$	_	—	0.4	V	I <sub>oL</sub> = 1.0 mA
	Ports 1, 2, 5, B		_	_	1.0	V	
	RESO	_	_	—	0.4	V	I <sub>oL</sub> = 1.6 mA
Input leakage current	$\label{eq:stby_state} \frac{\overline{\text{STBY}},  \text{NMI}, }{\overline{\text{RES}},  \text{MD}_2,  \text{MD}_1, } \\ \frac{\text{MD}_0}{\overline{\text{MD}}_0}$	I <sub>in</sub>		_	1.0	μA	$V_{in} = 0.5 V to$ $V_{cc} -0.5 V$
	Port 7	_	_		1.0	μA	$V_{in} = 0.5 \text{ V to}$ AV <sub>cc</sub> -0.5 V

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Three-state leakage	Ports 1, 2, 3, 5, 6, 8 to B	I <sub>tsi</sub>	_	_	1.0	μA	$V_{in} = 0.5 V \text{ to}$ $V_{cc} -0.5 V$
current (off state)	RESO	_	_	—	10.0	μA	_
Input pull-up MOS current	Ports 2, 5	<sub>p</sub>	10	_	300	μA	$V_{cc} = 3.0 V \text{ to } 5.5 V,$ $V_{in} = 0 V$
Input	NMI, RES	C <sub>in</sub>	_	_	50	pF	$V_{in} = 0 V,$
capacitance	All input pins except NMI and RES	_	_	_	20		f = 1 MHz, T <sub>a</sub> = 25°C
Current dissipation* <sup>2</sup>	Normal operation	<b>I</b> <sub>cc</sub> * <sup>4</sup>	_	15 (3.0 V)	41.5 (5.5 V)	mA	f = 10 MHz
	Sleep mode	_	_	10 (3.0 V)	30.5 (5.5 V)	mA	f = 10 MHz
	Standby mode*3	-	_	0.01	5.0	μΑ	$T_a \le 50^{\circ}C$
			_	_	20.0	μΑ	50°C < T <sub>a</sub>
Analog	During A/D	Al <sub>cc</sub>	_	1.3	2.5	mA	$AV_{cc} = 3.0 V$
power supply	conversion	_	_	1.7	_	mA	$AV_{cc} = 5.0 V$
current	ldle	_	_	0.02	10.0	μΑ	
RAM standby	v voltage	$V_{RAM}$	2.0	_	_	V	

Notes: 1. If the A/D converter is not used, do not leave the  $AV_{cc}$  and  $AV_{ss}$  pins open.

Connect AV<sub>cc</sub> to V<sub>cc</sub>, and connect AV<sub>ss</sub> to V<sub>ss</sub>.

- 2. Current dissipation values are for V<sub>IH</sub> min = V<sub>cc</sub> –0.5 V and V<sub>IL</sub> max = 0.5 V with all output pins unloaded and the on-chip pull-up transistors in the off state.
- 3. The values are for V\_{\_{RAM}} \le V\_{\_{CC}} < 3.0 V, V\_{\_{IH}} min = V\_{\_{CC}} \times 0.9, and V\_{\_{IL}} max = 0.3 V.
- 4.  $I_{cc}$  depends on  $V_{cc}$  and f as follows:

 $I_{cc} max = 3.0 (mA) + 0.7 (mA/MHz \cdot V) \times V_{cc} \times f [normal mode]$  $I_{cc} max = 3.0 (mA) + 0.5 (mA/MHz \cdot V) \times V_{cc} \times f [sleep mode]$ 

### Table 18.3 Permissible Output Currents

Conditions:  $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

Item		Symbol	Min	Тур	Мах	Unit
Permissible output low current (per pin)	Ports 1, 2, 5 and B	I <sub>ol</sub>			10	mA
	Other output pins		_	_	2.0	mA
Permissible output low current (total)	Total of 27 pins including ports 1, 2, 5 and B	$\Sigma I_{OL}$	_	_	80	mA
	Total of 23 pins, including ports 8, 9, A and B			_	75* <sup>2</sup> /65* <sup>1</sup>	mA
	Total of all output pins, including the above			_	120	mA
Permissible output high current (per pin)	All output pins	I <sub>он</sub>	_	_	2.0	mA
Permissible output high current (total)	Total of all output pins	$\Sigma I_{_{OH}}$	_	_	40	mA

Notes: To protect chip reliability, do not exceed the output current values in table 18.3.

When driving a Darlington pair or LED, always insert a current-limiting resistor in the output line, as shown in figures 18.1 and 18.2.

1. The value is for conditions: V\_{cc} = 2.7 V to 5.5 V, AV $_{cc}$  = 2.7 V to 5.5 V

2. The value is for conditions: V<sub>cc</sub> = 4.5 V to 5.5 V, AV<sub>cc</sub> = 4.5 V to 5.5 V

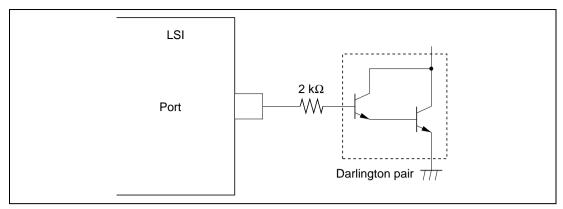


Figure 18.1 Darlington Pair Drive Circuit (Example)

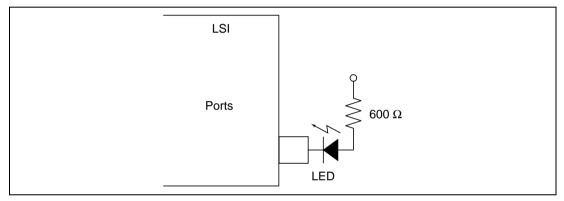


Figure 18.2 LED Drive Circuit (Example)

### **18.1.3** AC Characteristics

Bus timing parameters are listed in table 18.4. Control signal timing parameters are listed in table 18.5. Timing parameters of the on-chip supporting modules are listed in table 18.6.

### Table 18.4 Bus Timing

- Condition A:  $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 8 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)
- Condition B:  $V_{cc} = 3.0 \text{ V}$  to 5.5 V,  $AV_{cc} = 3.0 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 10 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

Condition C:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $AV_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 18 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

		Condition A		Condition B		Condition C		;	
		8	MHz	10	) MHz	18	MHz	_	Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Clock cycle time	t <sub>cyc</sub>	125	500	100	500	55.5	500	ns	Figure 18.7,
Clock low pulse width	t <sub>c∟</sub>	40	_	30	_	17	_		Figure 18.8
Clock high pulse width	t <sub>сн</sub>	40	_	30	_	17	_	_	
Clock rise time	t <sub>cr</sub>	_	20	_	15	_	10		
Clock fall time	t <sub>cr</sub>	—	20	—	15	_	10	_	
Address delay time	t <sub>AD</sub>	—	60	—	50	_	25	_	
Address hold time	t <sub>AH</sub>	25	_	20	_	10	—	_	
Address strobe delay time	t <sub>ASD</sub>	—	60	—	40	_	25	_	
Write strobe delay time	t <sub>wsp</sub>	—	60	—	50	_	25	_	
Strobe delay time	t <sub>sp</sub>	_	60	—	50	_	25	_	
Write data strobe pulse width 1	t <sub>wsw1</sub> *	85	_	60	_	32	—	_	
Write data strobe pulse width 2	t <sub>wsw2</sub> *	150	_	110	_	62	_	_	
Address setup time 1	t <sub>AS1</sub>	20	_	15	_	10	—	_	
Address setup time 2	t <sub>AS2</sub>	80	_	65	_	38	_	_	
Read data setup time	t <sub>RDS</sub>	50	_	35	_	15	_	_	
Read data hold time	t <sub>RDH</sub>	0		0		0		-	

		Con	dition A	Condition B		Condition C		;	
		8	MHz	10	MHz	18	8 MHz	_	Test
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
Write data delay time	t <sub>wdd</sub>	—	75	—	75	—	55	ns	Figure 18.7,
Write data setup time 1	t <sub>wDS1</sub>	60	_	40	_	10	—	_	Figure 18.8
Write data setup time 2	t <sub>wDS2</sub>	5	_	-10	_	-10	_		
Write data hold time	t <sub>wDH</sub>	25	_	20	_	20	—		
Read data access time 1	t <sub>ACC1</sub> *	_	120	_	100	_	50		
Read data access time 2	t <sub>ACC2</sub> *		240	_	200		105	_	
Read data access time 3	t <sub>ACC3</sub> *	—	70	—	50	—	20	_	
Read data access time 4	t <sub>ACC4</sub> *	_	180	_	150	_	80		
Precharge time	t <sub>PCH</sub> *	85	_	60	_	40	_		
Wait setup time	t <sub>wrs</sub>	40	_	40	_	25	—	_	Figure 18.9
Wait hold time	t <sub>wth</sub>	10	_	10	_	5			

Note: \* For Condition A, the following times depend on the clock cycle time as shown below.

$t_{ACC1} = 1.5 \times t_{Cyc} - 68$	(ns)
$t_{ACC2} = 2.5 \times t_{cyc} - 73$	(ns)
$t_{ACC3} = 1.0 \times t_{cyc} - 55$	(ns)
$t_{\rm ACC4} = 2.0 \times t_{\rm cyc} - 70$	(ns)

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$t_{wsw1} = 1.0 \times t_{cyc} - 40$	(ns)
$t_{wsw2} = 1.5 \times t_{cyc} - 38$	(ns)
$t_{_{PCH}} = 1.0 \times t_{_{cyc}} -40$	

For Condition B, the following times depend on the clock cycle time as shown below.

$t_{ACC1} = 1.5 \times t_{cyc} - 50$	(ns)	$t_{wsw1} = 1.0 \times t_{cyc} - 40$	(ns)
$t_{ACC2} = 2.5 \times t_{cyc} - 50$	(ns)	$t_{wsw2} = 1.5 \times t_{cyc} - 40$	(ns)
$t_{ACC3} = 1.0 \times t_{cyc} - 50$	(ns)	$t_{PCH} = 1.0 \times t_{cyc} - 40$	(ns)
$t_{ACC4} = 2.0 \times t_{cyc} - 50$	(ns)		

For Condition C, the following times depend on the clock cycle time as shown below.

$t_{ACC1} = 1.5 \times t_{cyc} - 34$	(ns)	$t_{wsw1} = 1.0 \times t_{cyc} -24$	(ns)
$t_{ACC2} = 2.5 \times t_{cyc} - 34$	(ns)	$t_{wsw2} = 1.5 \times t_{cyc} - 22$	(ns)
$t_{ACC3} = 1.0 \times t_{cyc} - 36$	(ns)	$t_{PCH} = 1.0 \times t_{cyc} - 21$	(ns)
$t_{ACC4} = 2.0 \times t_{cvc} - 31$	(ns)		

### Table 18.5 Control Signal Timing

- Condition A:  $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 8 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)
- Condition B:  $V_{cc} = 3.0 \text{ V}$  to 5.5 V,  $AV_{cc} = 3.0 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 10 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)
- Condition C:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $AV_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 18 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

		Condition A C		Con	Condition B		Condition C		
		8	MHz	10	) MHz	18	3 MHz	-	
Item	Symbol	Min	Max	Min	Max	Min	Max	Unit	Test Conditions
RES setup time	t <sub>ress</sub>	200	_	200	_	200	_	ns	Figure 18.10
RES pulse width	$\mathbf{t}_{\text{RESW}}$	10	—	10		10	_	tcyc	
Mode programming setup time (MD <sub>0</sub> , MD <sub>1</sub> , MD <sub>2</sub> )	t <sub>MDS</sub>	200	—	200	—	200	_	ns	
RESO output delay time	$\mathbf{t}_{\text{resd}}$	_	100	_	100	_	100	ns	Figure 18.11
RESO output pulse width	$\mathbf{t}_{\text{resow}}$	132	_	132	_	132	_	tcyc	_
NMI setup time (NMI, $\overline{IRQ}_0$ , $\overline{IRQ}_1$ , $\overline{IRQ}_4$ , $\overline{IRQ}_5$ )	t <sub>nmis</sub>	200	_	200	_	150	—	ns	Figure 18.12
NMI hold time (NMI, $\overline{IRQ}_0$ , $\overline{IRQ}_1$ , $\overline{IRQ}_4$ , $\overline{IRQ}_5$ )	t <sub>nmin</sub>	10	_	10	_	10	_	_	
Interrupt pulse width (NMI, IRQ,, IRQ <sub>o</sub> when exiting software standby mode)	t <sub>nmw</sub>	200		200	_	200	_	_	
Clock oscillator settling time at reset (crystal)	t <sub>osc1</sub>	20	_	20	_	20	_	ms	Figure 18.13
Clock oscillator settling time in software standby (crystal)	t <sub>osc2</sub>	8	_	8		7		ms	Figure 17.1

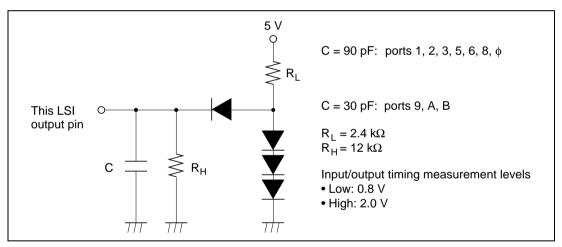
### Table 18.6 Timing of On-Chip Supporting Modules

- Condition A:  $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 8 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)
- Condition B:  $V_{cc} = 3.0 \text{ V}$  to 5.5 V,  $AV_{cc} = 3.0 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 10 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)
- Condition C:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $AV_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 18 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

				Con	dition A	Con	dition B	Con	dition C		Test Conditions
				8	MHz	10	) MHz	18	3 MHz	-	
ltem			Symbol	Min	Max	Min	Max	Min	Max	Unit	
ITU	Timer o time	output delay	$\mathbf{t}_{\text{tocd}}$	_	100	_	100	_	100	ns	Figure 18.15
	Timer i time	nput setup	$\mathbf{t}_{\mathrm{TICS}}$	50	—	50	—	50	_	-	
	Timer of setup ti	clock input me	$t_{\rm TCKS}$	50	—	50	_	50	_	_	Figure 18.16
	Timer clock	Single edge	$\mathbf{t}_{\mathrm{TCKWH}}$	1.5	_	1.5	_	1.5	_	t <sub>cyc</sub>	_
	pulse width	Both edges	t <sub>TCKWL</sub>	2.5	—	2.5	_	2.5	_	_	
SCI	Input clock	Asynchro- nous	t <sub>scyc</sub>	4	—	4	_	4	_	_	Figure 18.17
	cycle	Synchro- nous		6	_	6	_	6	_	_	
	Input cl	ock rise time	t <sub>scкr</sub>	_	1.5	_	1.5	_	1.5	-	
	Input cl	ock fall time	t <sub>sckf</sub>	_	1.5	—	1.5	—	1.5	-	
	Input cl width	ock pulse	t <sub>scкw</sub>	0.4	0.6	0.4	0.6	0.4	0.6	$\mathbf{t}_{_{\text{Scyc}}}$	

#### Section 18 Electrical Characteristics

			Condition A		Con	dition B	B Condition C			
			8	8 MHz		10 MHz		18 MHz		Test
Item		Symbol	Min	Max	Min	Max	Min	Max	Unit	Conditions
SCI	Transmit data delay time	$\mathbf{t}_{_{\mathrm{TXD}}}$	_	100	_	100	_	100	ns	Figure 18.18
	Receive data setup time (synchronous)	t <sub>RXS</sub>	100	—	100	_	100	_	_	
	Receive data hold time (synchronous clock input)	t <sub>RXH</sub>	100	_	100	_	100	_	-	
	Receive data hold time (synchronous clock output)	_	0	_	0	—	0	_	_	
Ports and	Output data delay time	t <sub>PWD</sub>	_	100	_	100	_	100	ns	Figure 18.14
TPC	Input data setup time	e t <sub>PRS</sub>	50	_	50	_	50	—	_	
	Input data hold time	t <sub>PRH</sub>	50	_	50	_	50	_	_	





#### 18.1.4 A/D Conversion Characteristics

Table 18.7 lists the A/D conversion characteristics.

#### Table 18.7 A/D Converter Characteristics

- Condition A:  $V_{cc} = 2.7 \text{ V to } 5.5 \text{ V}, \text{ AV}_{cc} = 2.7 \text{ V to } 5.5 \text{ V}, \text{ V}_{ss} = \text{AV}_{ss} = 0 \text{ V}, \phi = 2 \text{ MHz to}$ 8 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)
- Condition B:  $V_{cc} = 3.0 \text{ V}$  to 5.5 V,  $AV_{cc} = 3.0 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 10 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)
- Condition C:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $AV_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 18 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

	Condition A 8 MHz				Condition B			Condition C			
					10 MH	łz	_				
Item	Min	Тур	Мах	Min	Тур	Мах	Min	Тур	Max	Unit	
Resolution	10	10	10	10	10	10	10	10	10	bits	
Conversion time	_	_	16.8	_	_	13.4	_	_	7.5	μs	
Analog input capacitance	—		20	_	—	20	—	—	20	pF	
Permissible signal-	_	_	10* <sup>1</sup>	_	_	10* <sup>1</sup>	_	_	10* <sup>4</sup>	kΩ	
source impedance	_	_	5* <sup>2</sup>	_	_	5* <sup>3</sup>	_	_	<b>5</b> * <sup>5</sup>		
Nonlinearity error	_	_	±7.5	_	—	±7.5	_	_	±3.5	LSB	
Offset error	_	—	±7.5	_	_	±7.5	—	—	±3.5	LSB	
Full-scale error	_	—	±7.5	_	_	±7.5	—	—	±3.5	LSB	
Quantization error		—	±0.5		_	±0.5	_	_	±0.5	LSB	
Absolute accuracy	_	_	±8.0	_	—	±8.0	_	_	±4.0	LSB	

Notes: 1. The value is for 4.0 V  $\leq$  AV<sub>cc</sub>  $\leq$  5.5 V.

2. The value is for 2.7 V  $\leq$  AV  $_{cc}$  < 4.0 V.

3. The value is for 3.0 V  $\leq$  AV  $_{\rm cc}$  < 4.0 V.

- 4. The value is for  $\phi \leq 12$  MHz.
- 5. The value is for  $\phi$  > 12 MHz.

### **18.2** Electrical Characteristics of Flash Memory Version

### 18.2.1 Absolute Maximum Ratings

Table 18.8 lists the absolute maximum ratings.

### Table 18.8 Absolute Maximum Ratings

Item	Symbol	Value	Unit
Power supply voltage	V <sub>cc</sub>	-0.3 to +7.0	V
Input voltage (except port 7)*1	$V_{in}$	-0.3 to V <sub>cc</sub> +0.3	V
Input voltage (port 7)	$V_{in}$	–0.3 to AV <sub>cc</sub> +0.3	V
Analog power supply voltage	$AV_{cc}$	-0.3 to +7.0	V
Analog input voltage	V <sub>AN</sub>	–0.3 to AV <sub>cc</sub> +0.3	V
Operating temperature	T <sub>opr</sub>	Regular specifications: -20 to +75* <sup>2</sup>	°C
		Wide-range specifications: -40 to +85* <sup>2</sup>	°C
Storage temperature	T <sub>stg</sub>	-55 to +125	°C

Caution: Permanent damage to the chip may result if absolute maximum ratings are exceeded.

Notes: 1. 12 V must not be applied to any pin, as this will cause permanent damage to the chip.

2. The operating temperature range when programming/erasing flash memory is  $T_a = 0$  to +75°C (regular specifications) or  $T_a = 0$  to +85°C (wide-range specifications).

#### **18.2.2 DC** Characteristics

Table 18.9 lists the DC characteristics. Table 18.10 lists the permissible output currents.

### Table 18.9 DC Characteristics (1)

Conditions:  $V_{cc} = 5.0 V \pm 10\%$ ,  $AV_{cc} = 5.0 V \pm 10\%$ ,  $V_{ss} = AV_{ss} = 0 V^{*1}$ ,  $T_a = -20^{\circ}C$  to +75°C (regular specifications),  $T_a = -40^{\circ}C$  to +85°C (wide-range specifications) (Programming/Erasing Conditions:  $T_a = 0^{\circ}C$  to +75°C (regular specifications),  $T_a = 0^{\circ}C$  to +85°C (wide-range specifications))

ltem		Symbol	Min	Тур	Мах	Unit	Test Conditions
Schmitt	Port A,	V <sub>T</sub> <sup>-</sup>	1.0			V	
trigger input voltages	$P8_0$ to $P8_1$ , $PB_0$ to $PB_3$	V <sub>T</sub> <sup>+</sup>	_	_	$V_{cc}  imes 0.7$	V	_
vollagoo		$V_{T}^{+} - V_{T}^{-}$	0.4	_	_	V	_
Input high voltage	$\begin{tabular}{lllllllllllllllllllllllllllllllllll$	V <sub>IH</sub>	V <sub>cc</sub> -0.7	_	V <sub>cc</sub> +0.3	V	
	EXTAL		$V_{cc}  imes 0.7$		V <sub>cc</sub> +0.3	V	
	Port 7		2.0	_	$AV_{cc}$ +0.3	V	
	Ports 1, 2, 3, 5, 6, 9, PB <sub>4</sub> , PB <sub>5</sub> , PB <sub>7</sub>	_	2.0	—	V <sub>cc</sub> +0.3	V	
Input low voltage	$\begin{tabular}{lllllllllllllllllllllllllllllllllll$	V <sub>IL</sub>	-0.3	—	0.5	V	
	NMI, EXTAL, ports 1, 2, 3, 5, 6, 7, 9, PB₄, PB₅, PB <sub>7</sub>		-0.3	_	0.8	V	
	All output pins	V <sub>oh</sub>	V <sub>cc</sub> –0.5	_	_	V	I <sub>он</sub> = -200 μA
voltage			3.5	_	_	V	I <sub>он</sub> = -1 mA
Output low	All output pins	V <sub>ol</sub>		_	0.4	V	I <sub>oL</sub> = 1.6 mA
voltage	Ports 1, 2, 5, B		_		1.0	V	I <sub>oL</sub> = 10 mA

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Input leakage current	$\label{eq:stby_norm} \frac{\overline{\text{STBY}},  \text{NMI}, }{\overline{\text{RES}},  \text{MD}_2,  \text{MD}_1, } \\ \overline{\text{MD}}_0$	I <sub>in</sub>	_		1.0	μA	$V_{in} = 0.5 V$ to $V_{cc} -0.5 V$
	Port 7	_	_	_	1.0	μA	$V_{in} = 0.5 \text{ V to}$ AV <sub>cc</sub> -0.5 V
	FWE	_	_	_	10		
Three-state leakage current (off state)	Ports 1, 2, 3, 5, 6, 8, 9, A, B	I <sub>tsi</sub>	_	_	1.0	μA	$V_{in} = 0.5 V$ to $V_{cc} - 0.5 V$
Input pull-up current	Ports 2, 5	<sub>p</sub>	50	_	300	μA	V <sub>in</sub> = 0 V
Input	NMI, RES	C <sub>in</sub>	_	_	50	pF	$V_{in} = 0 V,$
capacitance	All input pins except NMI and RES	_	_	_	20		f = 1 MHz, $T_a = 25^{\circ}C$
Current dissipation	Normal operation	I <sub>cc</sub>	—	50	70	mA	f = 18 MHz
* <sup>2</sup> * <sup>4</sup>	Sleep mode	_	_	35	50	_	f = 18 MHz
	Standby mode*3	_	_	0.01	5.0	μA	$T_a \le 50^{\circ}C$
			_	_	20.0	_	50°C < T <sub>a</sub>
Analog power	During A/D conversion	Al <sub>cc</sub>	—	1.7	2.8	mA	
supply current	Idle	_	_	0.02	10.0	μA	
RAM standb	y voltage	$V_{RAM}$	2.0		_	V	

#### Section 18 Electrical Characteristics

Notes: 1. If the A/D converter is not used, do not leave the AV<sub>cc</sub> and AV<sub>ss</sub> pins open. Connect AV<sub>cc</sub> to V<sub>cc</sub>, and connect AV<sub>ss</sub> to V<sub>ss</sub>.

- 2. Current dissipation values are for V<sub>IH</sub> min = V<sub>cc</sub> –0.5 V and V<sub>IL</sub> max = 0.5 V with all output pins unloaded and the on-chip pull-up transistors in the off state.
- 3. The values are for V\_{\_{RAM}} \le V\_{\_{CC}} < 4.5 V, V  $_{_{IH}}$  min = V  $_{_{CC}} \times 0.9$ , and V  $_{_{IL}}$  max = 0.3 V.
- 4. Power supply current value when programming/erasing in flash memory (T<sub>a</sub> = 0°C to +75°C (regular specifications), T<sub>a</sub> = 0°C to +85°C (wide-range specifications)) is 20 mA (max) higher than the power supply current value in normal operation.

### Table 18.9 DC Characteristics (2)

Conditions:  $V_{cc} = 3.0 \text{ V}$  to 5.5 V,  $AV_{cc} = 3.0 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}^{*1}$ ,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications) (Programming/Erasing Conditions:  $V_{cc} = 3.0 \text{ V}$  to 3.6 V,  $T_a = 0^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = 0^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications))

ltem		Symbol	Min	Тур	Max	Unit	Test Conditions
Schmitt	Port A,	V <sub>T</sub>	$V_{cc}  imes 0.2$	_	_	V	
trigger input voltages	$P8_0$ to $P8_1$ , $PB_0$ to $PB_3$	V <sub>T</sub> <sup>+</sup>	_	_	$V_{cc} \times 0.7$	V	-
voltages		$V_T^+ - V_T^-$	$V_{cc}  imes 0.04$	_	_	V	-
Input high voltage	RES, STBY, NMI, MD <sub>2</sub> to MD <sub>0</sub> , FWE	V <sub>IH</sub>	$V_{cc} \times 0.9$	—	V <sub>cc</sub> +0.3	V	
	EXTAL	=	$V_{cc} \times 0.7$	_	V <sub>cc</sub> +0.3	V	-
	Port 7	=	$V_{cc} \times 0.7$	_	AV <sub>cc</sub> +0.3	V	-
	Ports 1 to 3, 5, 6, 9, PB <sub>4</sub> , PB <sub>5</sub> , PB <sub>7</sub>	,	$V_{cc} \times 0.7$		V <sub>cc</sub> +0.3	V	-
Input low voltage	$\frac{\text{RES, STBY,}}{\text{FWE, MD}_2 \text{ to}}$ $\text{MD}_0, \text{FWE}$	V <sub>IL</sub>	-0.3		$V_{cc} \times 0.1$	V	
	NMI, EXTAL, ports 1 to 3,	_	-0.3	_	$V_{cc} \times 0.2$	V	$V_{cc}$ < 4.0 V
	5 to 7, 9, PB <sub>4</sub> , PB <sub>5</sub> , PB <sub>7</sub>				0.8		V <sub>cc</sub> = 4.0 V to 5.5V
Output high	All output pins	V <sub>OH</sub>	V <sub>cc</sub> –0.5		_	V	I <sub>он</sub> = -200 μA
voltage			V <sub>cc</sub> -1.0	_	_	V	I <sub>он</sub> = —1 mA
Output low	All output pins	V <sub>ol</sub>	_	—	0.4	V	I <sub>oL</sub> = 1.0 mA
voltage	Ports 1, 2, 5, B	-	_	—	1.0	V	

Item		Symbol	Min	Тур	Max	Unit	Test Conditions
Input leakage current	$\frac{\overline{\text{STBY}}, \text{ NMI},}{\overline{\text{RES}}, \text{ MD}_2,}$ $\text{MD}_1, \text{ MD}_0$	<sub>in</sub>	_	_	1.0	μA	$V_{in} = 0.5 V to$ $V_{cc} -0.5 V$
	Port 7	-	—	_	1.0	μA	$V_{in} = 0.5 \text{ V to}$ AV <sub>cc</sub> -0.5 V
	FWE	-	_	—	10	μA	$V_{in} = 0.5 V to$ $V_{cc} - 0.5 V$
Three-state leakage current (off state)	Ports 1, 2, 3, 5, 6, 8 to B	<sub>TSI</sub>	_	_	1.0	μA	$V_{in} = 0.5 V$ to $V_{cc} - 0.5 V$
Input pull-up current	Ports 2 and 5	- <b>I</b> <sub>p</sub>	10	—	300	μA	$V_{cc} = 3.0 \text{ V to}$ 5.5 V, $V_{in} = 0 \text{ V}$
Input	NMI, RES	C <sub>in</sub>		_	50	pF	$V_{in} = 0 V,$
capacitance	All input pins except NMI and RES	-	_	—	20	pF	<sup>-</sup> f = 1 MHz, T <sub>a</sub> = 25°C
Current dissipation	Normal operation	Ι <sub>cc</sub> * <sup>4</sup>	—	15 (3.0 V)	41.5 (5.5 V)	mA	f = 10 MHz
* <sup>2</sup> * <sup>5</sup>	Sleep mode	_	_	10 (3.0 V)	30.5 (5.5 V)	mA	f = 10 MHz
	Standby mode*3	-	_	0.01	5.0	μΑ	$T_a \le 50^{\circ}C$
					20.0	_	50°C < T <sub>a</sub>
Analog	During A/D	Al <sub>cc</sub>		1.3	2.5	mA	$AV_{cc} = 3.0 V$
power supply	conversion	_	_	1.7	2.8		$AV_{cc} = 5.0 V$
current	Idel	-	_	0.02	10.0	μΑ	
RAM standby	v voltage	$V_{RAM}$	2.0	_	—	V	

#### Section 18 Electrical Characteristics

Notes: 1. If the A/D converter is not used, do not leave the  $AV_{cc}$  and  $AV_{ss}$  pins open. Connect  $AV_{cc}$  to  $V_{cc}$ , and connect  $AV_{ss}$  to  $V_{ss}$ .

- 2. Current dissipation values are for V<sub>IH</sub> min = VCC –0.5 V and V<sub>IL</sub> max = 0.5 V with all output pins unloaded and the on-chip pull-up transistors in the off state.
- 3. The values are for V\_{\_{RAM}} \le V\_{\_{CC}} < 3.0 V, V\_{\_{IH}} min = VCC  $\times$  0.9, and V\_{\_{IL}} max = 0.3 V.
- 4.  $I_{cc}$  depends on  $V_{cc}$  and f as follows:  $I_{cc}$  max = 3.0 (mA) + 0.7 (mA/MHz · V) ×  $V_{cc}$  × f [normal mode]  $I_{cc}$  max = 3.0 (mA) + 0.5 (mA/MHz · V) ×  $V_{cc}$  × f [sleep mode]
- 5. The current dissipation value when programming/erasing flash memory ( $T_a = 0^{\circ}C$  to +75°C (regular specifications),  $T_a = 0^{\circ}C$  to +85°C (wide-range specifications)) is 20 mA (max) higher than the current dissipation value in normal operation.

### Table 18.10 Permissible Output Currents

Conditions:  $V_{cc} = 3.0 \text{ V}$  to 5.5 V,  $AV_{cc} = 3.0 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

Item		Symbol	Min	Тур	Max	Unit
Permissible output low current (per pin)	Ports 1, 2, 5 and B	I <sub>ol</sub>	_	_	10	mA
	Other output pins				2.0	mA
Permissible output low current (total)	Total of 27 pins including ports 1, 2, 5 and B	Σl <sub>ol</sub>	_	_	80	mA
	Total of 23 pins, including ports 8, 9, A and B		—	_	75* <sup>2</sup> / 65* <sup>1</sup>	mA
	Total of all output pins, including the above		_	_	120	mA
Permissible output high current (per pin)	All output pins	I <sub>он</sub>	—	—	2.0	mA
Permissible output high current (total)	Total of all output pins	$\Sigma I_{_{OH}}$	_	_	40	mA

Notes: To protect chip reliability, do not exceed the output current values in table 18.10. When driving a Darlington pair or LED, always insert a current-limiting resistor in the output line, as shown in figures 18.4 and 18.5.

1. Conditions:  $V_{cc}$  = 3.0 V to 5.5 V,  $AV_{cc}$  = 3.0 V to 5.5 V

2. Conditions:  $V_{cc}$  = 4.5 V to 5.5 V,  $AV_{cc}$  = 4.5 V to 5.5 V

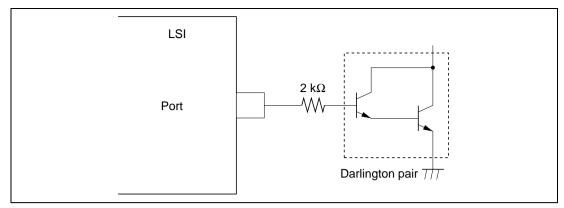


Figure 18.4 Darlington Pair Drive Circuit (Example)

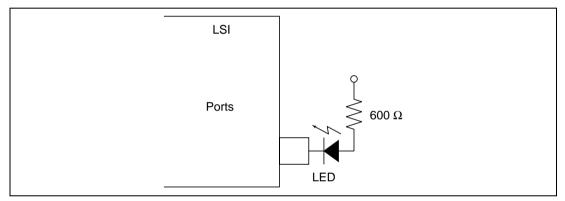


Figure 18.5 LED Drive Circuit (Example)

### **18.2.3** AC Characteristics

Bus timing parameters are listed in table 18.11. Control signal timing parameters are listed in table 18.12. Timing parameters of the on-chip supporting modules are listed in table 18.13.

### Table 18.11 Bus Timing

- Condition A:  $V_{cc} = 3.0 \text{ V}$  to 5.5 V,  $AV_{cc} = 3.0 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2$  to 10 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)
- Condition B:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $AV_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 18 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

		Condition A		Condition B			
		10	) MHz	18	MHz	-	Test
Item	Symbol	Min	Max	Min	Max	Unit	Conditions
Clock cycle time	t <sub>cyc</sub>	100	500	55.5	500	ns	Figure 18.7,
Clock low pulse width	t <sub>c∟</sub>	30	_	17	_	-	Figure 18.8
Clock high pulse width	t <sub>сн</sub>	30	_	17	_	-	
Clock rise time	t <sub>cr</sub>		15	_	10	-	
Clock fall time	t <sub>cf</sub>		15	_	10	-	
Address delay time	t <sub>AD</sub>		50	_	25	-	
Address hold time	t <sub>AH</sub>	20	_	10	—	-	
Address strobe delay time	t <sub>ASD</sub>		40	_	25	-	
Write strobe delay time	t <sub>wsp</sub>		50		25	-	
Strobe delay time	t <sub>sp</sub>		50	_	25	-	
Write data strobe pulse width 1	t <sub>wsw1</sub> *	60	_	32	_	-	
Write data strobe pulse width 2	t <sub>wsw2</sub> *	110	_	62	—	-	
Address setup time 1	t <sub>AS1</sub>	15	_	10	_	-	
Address setup time 2	t <sub>AS2</sub>	65	_	38	_	-	
Read data setup time	t <sub>RDS</sub>	35	_	15	—	-	
Read data hold time	t <sub>RDH</sub>	0	—	0	—		

		Condition A		Con	dition B			
		10	) MHz	18 MHz		-	Test	
Item	Symbol	Min	Max	Min	Max	Unit	Conditions	
Write data delay time	t <sub>wdd</sub>		75		55	ns	Figure 18.7,	
Write data setup time 1	t <sub>wDS1</sub>	40	_	10	—	-	Figure 18.8	
Write data setup time 2	t <sub>wDS2</sub>	-10	_	-10	—	-		
Write data hold time	t <sub>wdH</sub>	20	_	20	—	-		
Read data access time 1	t <sub>ACC1</sub> *		100		50	-		
Read data access time 2	t <sub>ACC2</sub> *	_	200		105	_		
Read data access time 3	t <sub>ACC3</sub> *		50		20	-		
Read data access time 4	t <sub>ACC4</sub> *	_	150		80	_		
Precharge time	t <sub>PCH</sub> *	60	_	40	—	_		
Wait setup time	t <sub>wrs</sub>	40	_	25	—	ns	Figure 18.9	
Wait hold time	t <sub>wth</sub>	10		5	_	_		

Note: \* For condition A, the following times depend on the clock cycle time as shown below.

 $\begin{array}{l} t_{_{ACC1}} = 1.5 \times t_{_{cyc}} -50 \quad (ns) \\ t_{_{ACC2}} = 2.5 \times t_{_{cyc}} -50 \quad (ns) \\ t_{_{ACC3}} = 1.0 \times t_{_{cyc}} -50 \quad (ns) \\ t_{_{ACC4}} = 2.0 \times t_{_{cyc}} -50 \quad (ns) \end{array}$ 

For condition B, the following times depend on the clock cycle time as shown below.

$t_{ACC1} = 1.5 \times t_{cyc} - 34$	(ns)
$t_{ACC2} = 2.5 \times t_{cyc} - 34$	(ns)
$t_{ACC3} = 1.0 \times t_{cyc} - 36$	(ns)
$t_{\rm ACC4} = 2.0 \times t_{\rm cyc} - 31$	(ns)

### Table 18.12 Control Signal Timing

- Condition A:  $V_{cc} = 3.0 \text{ V}$  to 5.5 V,  $AV_{cc} = 3.0 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ to } 10 \text{ MHz}$ ,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)
- Condition B:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $AV_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 18 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

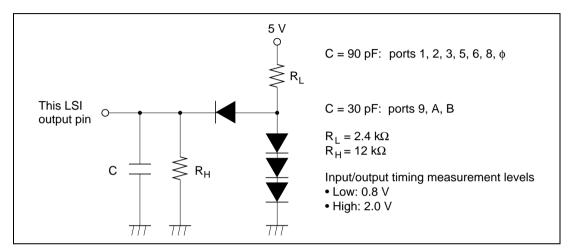
		Con	dition A	Con	dition B		
		10	) MHz	18	3 MHz	_	Test
Item	Symbol	Min	Max	Min	Max	Unit	Conditions
RES setup time	t <sub>ress</sub>	200		200	_	ns	Figure 18.10
RES pulse width	t <sub>resw</sub>	20	_	20	_	tcyc	_
Mode programming setup time	t <sub>MDS</sub>	200	_	200	_	ns	
NMI setup time (NMI, $\overline{IRQ}_0$ , $\overline{IRQ}_1$ , $\overline{IRQ}_4$ , $\overline{IRQ}_5$ )	t <sub>nmis</sub>	200	_	150	—	ns	Figure 18.12
NMI hold time (NMI, $\overline{IRQ}_0$ , $\overline{IRQ}_1$ , $\overline{IRQ}_4$ , $\overline{IRQ}_5$ )	t <sub>nmin</sub>	10	_	10	_		
Interrupt pulse width (NMI, $\overline{IRQ}_1$ , $\overline{IRQ}_0$ when exiting software standby mode)	t <sub>nmw</sub>	200	_	200	_	_	
Clock oscillator settling time at reset (crystal)	t <sub>osc1</sub>	20	_	20	—	ms	Figure 18.13
Clock oscillator settling time in software standby (crystal)	t <sub>osc2</sub>	8	_	7	_	ms	Figure 17.1

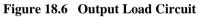
### Table 18.13 Timing of On-Chip Supporting Modules

- Condition A:  $V_{cc} = 3.0 \text{ V}$  to 5.5 V,  $AV_{cc} = 3.0 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 10 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)
- Condition B:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $AV_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 18 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

				Con	dition A	Condition B		3	
				10	MHz	18	MHz	_	Test
Item			Symbol	Min	Мах	Min	Мах	Unit	Conditions
ITU	Timer output	delay time	t <sub>TOCD</sub>	_	100	_	100	ns	Figure 18.15
	Timer input s	setup time	t <sub>TICS</sub>	50		50		_	
	Timer clock i time	nput setup	t <sub>TCKS</sub>	50	—	50	_		Figure 18.16
	Timer clock pulse width	Single edge	t <sub>тскwн</sub>	1.5	_	1.5		t <sub>cyc</sub>	_
		Both edges	$\mathbf{t}_{_{TCKWL}}$	2.5	—	2.5	—		
SCI	Input clock cycle	Asynchro- nous	t <sub>Scyc</sub>	4	_	4		_	Figure 18.17
		Synchro- nous	_	6		6	—	_	
	Input clock ri	se time	t <sub>scKr</sub>	_	1.5	_	1.5	_	
	Input clock fa	all time	t <sub>scкf</sub>	_	1.5	_	1.5	_	
	Input clock p	ulse width	t <sub>scкw</sub>	0.4	0.6	0.4	0.6	$\mathbf{t}_{_{\mathrm{Scyc}}}$	_
	Transmit dat	a delay time	t <sub>TXD</sub>	_	100	—	100	ns	Figure 18.18
	Receive data (synchronou		t <sub>RXS</sub>	100	_	100	—		
	Receive data (synchronous input)		t <sub>RXH</sub>	100		100	—	_	
	Receive data (synchronous output)		_	0	_	0	—	_	

			Condition A		Condition B				
			10	MHz	18	MHz	-	Test	
ltem		Symbol	Min	Max	Min	Max	Unit	Conditions	
Ports	Output data delay time	t <sub>PWD</sub>	_	100	_	100	ns	Figure 18.14	
and TPC	Input data setup time	t <sub>PRS</sub>	50	_	50		_		
	Input data hold time	t <sub>PRH</sub>	50		50		_		





### 18.2.4 A/D Conversion Characteristics

Table 18.14 lists the A/D conversion characteristics.

### Table 18.14 A/D Converter Characteristics

- Condition A:  $V_{cc} = 3.0 \text{ V}$  to 5.5 V,  $AV_{cc} = 3.0 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 10 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)
- Condition B:  $V_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $AV_{cc} = 5.0 \text{ V} \pm 10\%$ ,  $V_{ss} = AV_{ss} = 0 \text{ V}$ ,  $\phi = 2 \text{ MHz}$  to 18 MHz,  $T_a = -20^{\circ}\text{C}$  to  $+75^{\circ}\text{C}$  (regular specifications),  $T_a = -40^{\circ}\text{C}$  to  $+85^{\circ}\text{C}$  (wide-range specifications)

		_					
Item	Min	Тур	Max	Min	Тур	Max	Unit
Resolution	10	10	10	10	10	10	bits
Conversion time	_	_	13.4	_	_	7.5	μs
Analog input capacitance	_	—	20	_	—	20	pF
Permissible signal-	_	—	5* <sup>1</sup>	_	—	10* <sup>2</sup>	kΩ
source impedance				_	—	5* <sup>3</sup>	
Nonlinearity error	_	_	±7.5	_	_	±3.5	LSB
Offset error	_	_	±7.5	_		±3.5	LSB
Full-scale error	_	_	±7.5	_		±3.5	LSB
Quantization error	_	—	±0.5	—	—	±0.5	LSB
Absolute accuracy	_	_	±8.0	_	_	±4.0	LSB

Notes: 1. The value is for  $\phi = 10$  MHz.

- 2. The value is for  $\phi \leq 12$  MHz.
- 3. The value is for  $\phi$  > 12 MHz.

### 18.2.5 Flash Memory Characteristics

Table 18.15 shows the flash memory characteristics.

### Table 18.15 Flash Memory Characteristics (1)

# Conditions: $V_{cc} = 4.5 \text{ V}$ to 5.5 V, $AV_{cc} = 4.5 \text{ V}$ to 5.5 V, $V_{ss} = AV_{ss} = 0 \text{ V}$ $T_a = 0^{\circ}\text{C}$ to +75°C (program/erase operating temperature range: regular specifications), $T_a = 0^{\circ}\text{C}$ to +85°C (program/erase operating temperature range: wide-range specifications)

Item		Symbol	Min	Тур	Мах	Unit	Test condition
Programming	time* <sup>1</sup> * <sup>2</sup> * <sup>4</sup>	t <sub>P</sub>	_	10	200	ms/32 bytes	;
Erase time*1*3	3 *2	t <sub>e</sub>	—	100	300	ms/block	
Reprogrammir	ng count	$N_{\text{wec}}$	_	_	100	Times	
Programming	Wait time after SWE bit setting*1	х	10	—	—	μs	
	Wait time after PSU bit setting*1	у	50	_	_	μs	
	Wait time after P bit setting*1*4	z	150	_	500	μs	
	Wait time after P bit clear*1	α	10	_	_	μs	
	Wait time after PSU bit clear*1	β	10	—	_	μs	
	Wait time after PV bit setting*1	γ	4	_	—	μs	
	Wait time after H'FF dummy write*1	ε	2	_	—	μs	
	Wait time after PV bit clear*1	η	4	_	_	μs	
	Maximum programming count*1*4	Ν	_	_	403	Times	
Erase	Wait time after SWE bit setting*1	х	10	_	_	μs	
	Wait time after ESU bit setting*1	у	200	—	—	μs	
	Wait time after E bit setting*1*5	z	5	_	10	ms	
	Wait time after E bit clear*1	α	10	_	_	μs	
	Wait time after ESU bit clear*1	β	10	_	—	μs	
	Wait time after EV bit setting*1	γ	20	_	—	μs	
	Wait time after H'FF dummy write*1	ε	2	—	—	μs	
	Wait time after EV bit clear*1	η	5	—	—	μs	
	Maximum erase count*1*5	Ν	30	—	60	Times	

Notes: 1. Set the times according to the program/erase algorithms.

2. Programming time per 32 bytes (Shows the total time the flash memory control register (FLMCR) is set. It does not include the programming verification time.)

- 3. Block erase time (Shows the period the E bit in FLMCR is set. It does not include the erase verification time.)
- 4. To specify the maximum programming time (tp(max)) in the 32-byte programming flowchart, set the max value (403) for the maximum programming count (N). The wait time after P bit setting (z) should be changed as follows according to the programming counter value.

Programming counter value of 1 to 4:  $z = 150 \ \mu s$ Programming counter value of 5 to 403:  $z = 500 \ \mu s$ 

5. For the maximum erase time (t<sub>E</sub>(max)), the following relationship applies between the wait time after E bit setting (z) and the maximum erase count (N):

 $t_E(max)$  = Wait time after E bit setting (z) × maximum erase count (N)

To set the maximum erase time, the values of z and N should be set so as to satisfy the above formula.

Examples: When z = 5 [ms]: N = 60 times When z = 10 [ms]: N = 30 times

### Table 18.15 Flash Memory Characteristics (2)

Conditions:  $V_{cc} = 3.0 \text{ V}$  to 3.6 V,  $AV_{cc} = 3.0 \text{ V}$  to 3.6 V,  $V_{ss} = AV_{ss} = 0 \text{ V}$   $T_a = 0^{\circ}\text{C}$  to +75°C (Programming/erasing operating temperature range: regular specification)  $T_a = 0^{\circ}\text{C}$  to +85°C (Programming/erasing operating temperature range: wide-range specification)

Item		Symbol	Min	Тур	Max	Unit	Test condition
Programming time*1*2*4		t <sub>P</sub>	_	10	200	ms/32 bytes	6
Erase time*1*3*5		t <sub>e</sub>	_	100	300	ms/block	
Reprogramming count		$N_{\text{wec}}$	—	—	100	Times	
Programming	Wait time after SWE bit setting*1	х	10	_	_	μs	
	Wait time after PSU bit setting*1	У	50	_	_	μs	
	Wait time after P bit setting*1 *4	z	150	_	500	μs	
	Wait time after P bit clear*1	α	10	—	_	μs	
	Wait time after PSU bit clear*1	β	10	—	_	μs	
	Wait time after PV bit setting*1	γ	4	_	_	μs	
	Wait time after H'FF dummy write*1	ε	2	_	_	μs	
	Wait time after PV bit clear*1	η	4	—	_	μs	
	Maximum programming count*1.*4	N	_	_	403	Times	
Erase	Wait time after SWE bit setting*1	х	10	_	_	μs	
	Wait time after ESU bit setting*1	у	200	—	_	μs	
	Wait time after E bit setting*1*5	Z	5	_	10	ms	
	Wait time after E bit clear*1	α	10	_	_	μs	
	Wait time after ESU bit clear*1	β	10	_	_	μs	
	Wait time after EV bit setting*1	γ	20	_	_	μs	
	Wait time after H'FF dummy write*1	ε	2	_	_	μs	
	Wait time after EV bit clear*1	η	5	_	_	μs	
	Maximum erase count*1*5	Ν	30	_	60	Times	

Notes: 1. Make each time setting in accordance with the program/program-verify flowchart or erase/erase-verify flowchart.

- 2. Programming time per 32 bytes (Shows the total period for which the P-bit in the flash memory control register (FLMCR) is set. It does not include the programming verification time.)
- 3. Block erase time (Shows the total period for which the E-bit in FLMCR is set. It does not include the erase verification time.)

4. To specify the maximum programming time (t<sub>p</sub>(max)) in the 32-byte programming flowchart, set the maximum value (403) for the maximum programming count (N). The wait time after P bit setting (z) should be changed as follows according to the programming counter value.

Programming counter value of 1 to 4:  $z = 150 \ \mu s$ Programming counter value of 5 to 403:  $z = 500 \ \mu s$ 

 For the maximum erase time (t<sub>e</sub>(max)), the following relationship applies between the wait time after E bit setting (z) and the maximum erase count (N):

 $t_{E}(max)$  = Wait time after E bit setting (z) × maximum erase count (N)

To set the maximum erase time, the values of  $\boldsymbol{z}$  and  $\boldsymbol{N}$  should be set so as to satisfy the above formula.

Examples: When z = 5 [ms], N = 60 times When z = 10 [ms], N = 30 times

### **18.3** Operational Timing

This section shows timing diagrams.

### 18.3.1 Bus Timing

Bus timing is shown as follows:

- Basic bus cycle: two-state access Figure 18.7 shows the timing of the external two-state access cycle.
- Basic bus cycle: three-state access Figure 18.8 shows the timing of the external three-state access cycle.
- Basic bus cycle: three-state access with one wait state Figure 18.9 shows the timing of the external three-state access cycle with one wait state inserted.

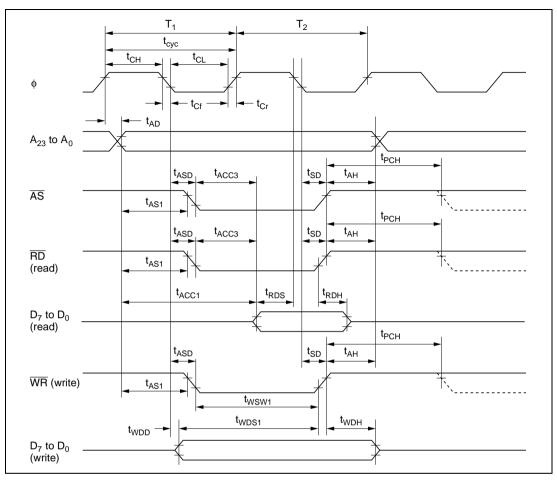


Figure 18.7 Basic Bus Cycle: Two-State Access

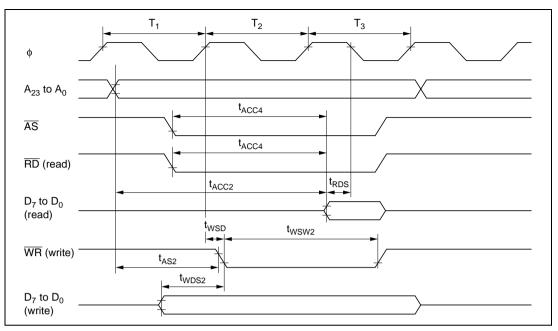


Figure 18.8 Basic Bus Cycle: Three-State Access

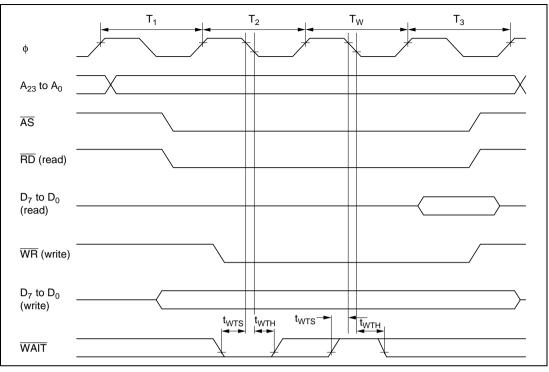


Figure 18.9 Basic Bus Cycle: Three-State Access with One Wait State

# 18.3.2 Control Signal Timing

Control signal timing is shown as follows:

- Reset input timing Figure 18.10 shows the reset input timing.
- Reset output timing

Figure 18.11 shows the reset output timing.

• Interrupt input timing

Figure 18.12 shows the interrupt input timing for NMI and  $\overline{IRQ}_5$ ,  $\overline{IRQ}_4$ ,  $\overline{IRQ}_1$ , and  $\overline{IRQ}_0$ .

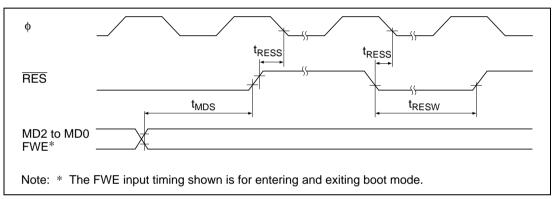


Figure 18.10 Reset Input Timing

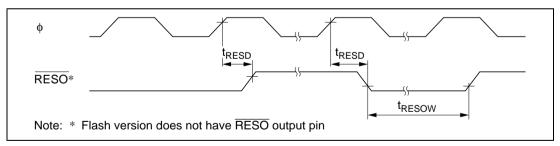


Figure 18.11 Reset Output Timing

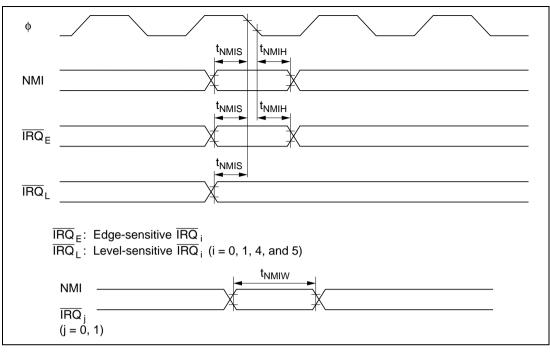


Figure 18.12 Interrupt Input Timing

#### 18.3.3 Clock Timing

Clock timing is shown below.

• Oscillator settling timing

Figure 18.13 shows the oscillator settling timing.

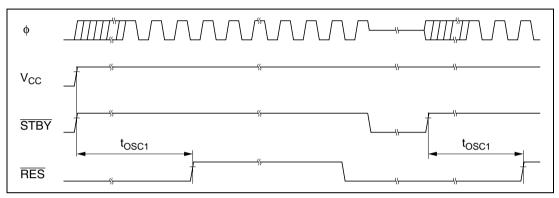


Figure 18.13 Oscillator Settling Timing

# 18.3.4 TPC and I/O Port Timing

TPC and I/O port timing is shown below.

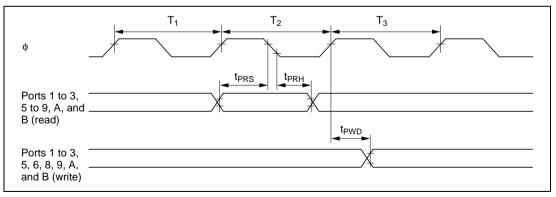
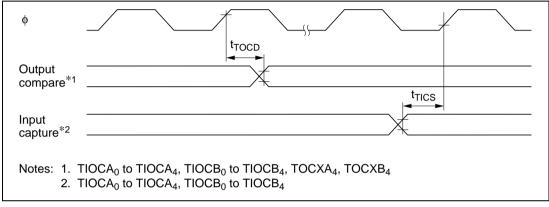


Figure 18.14 TPC and I/O Port Input/Output Timing

#### 18.3.5 ITU Timing

ITU timing is shown as follows:

- ITU input/output timing Figure 18.15 shows the ITU input/output timing.
- ITU external clock input timing Figure 18.16 shows the ITU external clock input timing.



#### Figure 18.15 ITU Input/Output Timing

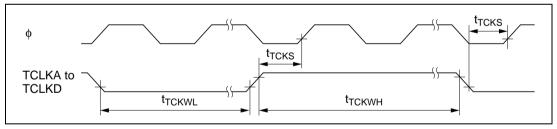


Figure 18.16 ITU External Clock Input Timing

# 18.3.6 SCI Input/Output Timing

SCI timing is shown as follows:

• SCI input clock timing

Figure 18.17 shows the SCI input clock timing.

• SCI input/output timing (synchronous mode)

Figure 18.18 shows the SCI input/output timing in synchronous mode.

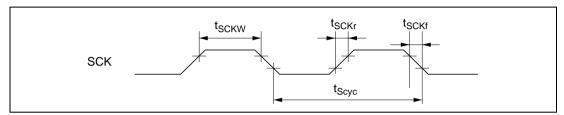


Figure 18.17 SCK Input Clock Timing

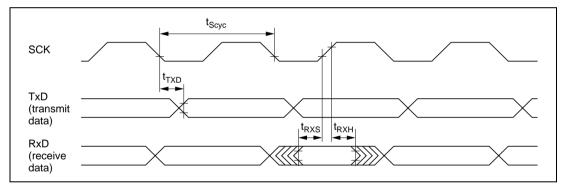


Figure 18.18 SCI Input/Output Timing in Synchronous Mode

# Appendix A Instruction Set

# A.1 Instruction List

# **Operand Notation**

Symbol	Description
Rd	General destination register*
Rs	General source register*
Rn	General register*
ERd	General destination register (address register or 32-bit register)
ERs	General source register (address register or 32-bit register)
ERn	General register (32-bit register)
(EAd)	Destination operand
(EAs)	Source operand
PC	Program counter
SP	Stack pointer
CCR	Condition code register
N	N (negative) flag in CCR
Z	Z (zero) flag in CCR
V	V (overflow) flag in CCR
С	C (carry) flag in CCR
disp	Displacement
$\rightarrow$	Transfer from the operand on the left to the operand on the right, or transition from the state on the left to the state on the right
+	Addition of the operands on both sides
_	Subtraction of the operand on the right from the operand on the left
×	Multiplication of the operands on both sides
÷	Division of the operand on the left by the operand on the right
^	Logical AND of the operands on both sides
•	Logical OR of the operands on both sides
$\oplus$	Exclusive logical OR of the operands on both sides
7	NOT (logical complement)
(), <>	Contents of operand
Note: *	General registers include 8-bit registers (R0H to R7H and R0L to R7L) and 16-bit registers (R0 to R7 and E0 to E7).

# **Condition Code Notation**

Symbol	Description
$\updownarrow$	Changed according to execution result
*	Undetermined (no guaranteed value)
0	Cleared to 0
1	Set to 1
_	Not affected by execution of the instruction
Δ	Varies depending on conditions, described in notes

#### Table A.1Instruction Set

# 1. Data transfer instructions

							-		le ai i (by		)							No. Stat	
	Operand Size		#xx	Rn	@ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	@ @ aa	mplied		Con					Normal	Advanced
		Operation	<b>#</b>	~	0	0	0	0	0	0	-	1	н	N ↑	Z ↑	V	C	_	-
MOV.B #xx:8, Rd	B	$\#xx:8 \rightarrow Rd8$	2	2								_	_	↓ ↓	↓ ↓	0	-	2	
MOV.B Rs, Rd	B B	$Rs8 \rightarrow Rd8$		2	2							_	_	↓ ↓	↓	-	-	4	
MOV.B @ERs, Rd MOV.B @(d:16, ERs), Rd	В	@ERs → Rd8 @(d:16, ERs) → Rd8			2	4						_	_	↓	↓ ↓	0	_	6	
MOV.B @(d:24, ERs), Rd	в	@(d:24, ERs) → Rd8				8						—	—	\$	€	0	—	10	)
MOV.B @ERs+, Rd	В	@ERs → RD8 ERs32+1 → ERs32					2					_	—	\$	\$	0	_	6	
MOV.B @aa:8, Rd	в	@aa:8 $\rightarrow$ Rd8						2				_	—	€	↕	0	_	4	
MOV.B @aa:16, Rd	в	@aa:16 $\rightarrow$ Rd8						4				_	—	€	↕	0	-	6	
MOV.B @aa:24, Rd	в	@aa:24 → Rd8						6				_	—	€	↕	0	_	8	
MOV.B Rs, @ERd	в	$Rs8 \rightarrow @ERd$			2							_	—	€	\$	0	-	4	
MOV.B Rs, @(d:16, ERd)	В	Rd8  ightarrow @(d:16, ERd)				4						_	—	\$	\$	0	_	6	
MOV.B Rs, @(d:24, ERd)	В	$Rd8 \to @(d:24,ERd)$				8						—	—	\$	\$	0	_	10	)
MOV.B Rs, @-ERd	В	$\begin{array}{l} ERd32-1 \rightarrow ERd32 \\ Rs8 \rightarrow @ERd \end{array}$					2					_	-	\$	\$	0	_	6	
MOV.B Rs, @aa:8	в	$Rs8 \rightarrow @aa:8$						2				_	—	\$	\$	0	_	4	
MOV.B Rs, @aa:16	в	$Rs8 \rightarrow @aa:16$						4				_	—	\$	\$	0	_	6	
MOV.B Rs, @aa:24	в	Rs8  ightarrow @aa:24						6				_	—	\$	\$	0	_	8	
MOV.W #xx:16, Rd	W	#xx:16 → Rd16	4									_	—	\$	↕	0	-	4	
MOV.W Rs, Rd	W	$Rs16 \rightarrow Rd16$		2								_	_	€	\$	0	_	2	
MOV.W @ERs, Rd	W	@ERs $\rightarrow$ Rd16			2							-	—	€	\$	0	-	4	
MOV.W @(d:16, ERs), Rd	w	@(d:16, ERs) → Rd16				4						_	-	\$	\$	0	_	6	
MOV.W @(d:24, ERs), Rd	w	@(d:24, ERs) → Rd16				8						_	-	\$	\$	0	-	10	)
MOV.W @ERs+, Rd	w	@ERs → Rd16 ERs32+2 → @ERd32					2					_	-	\$	\$	0	_	6	
MOV.W @aa:16, Rd	w	@aa:16 $\rightarrow$ Rd16						4				—	—	\$	\$	0	_	6	



									le ar by		)							No. Stat	of tes <sup>*1</sup>
	Operand Size				@ERn	d, ERn)	@-ERn/@ERn+	g	d, PC)	@ aa	Implied	(	Con	ditio	on C	Code	9	Normal	Advanced
Mnemonic	ð	Operation	XX#	Rn	8	@(d,	8	@ aa	@(d,	0	<u>n</u>	I	н	Ν	z	v	С	Noi	Adv
MOV.W @aa:24, Rd	W	$@aa:24 \rightarrow Rd16 \\$						6				—	—	$\updownarrow$	$\updownarrow$	0	_	8	3
MOV.W Rs, @ERd	w	$Rs16 \to @ERd$			2							—	—	$\updownarrow$	€	0	_	4	ŀ
MOV.W Rs, @(d:16, ERd)	W	$\text{Rs16} \rightarrow @(\text{d:16, ERd})$				4						—	_	\$	\$	0	_	6	5
MOV.W Rs, @(d:24, ERd)	W	$Rs16 \to @(d:24,ERd)$				8						—		€	\$	0		8	3
MOV.W Rs, @-ERd	W	$\begin{array}{l} ERd32-2 \rightarrow ERd32 \\ Rs16 \rightarrow @ERd \end{array}$					2					_		€	\$	0		6	5
MOV.W Rs, @aa:16	w	$\text{Rs16} \rightarrow @\text{aa:16}$						4				_	—	↕	€	0	_	6	;
MOV.W Rs, @aa:24	w	$\text{Rs16} \rightarrow @aa:24$						6				_	—	\$	\$	0	_	8	;
MOV.L #xx:32, Rd	L	$\#xx:\!32 \to Rd32$	6									_	—	$\updownarrow$	\$	0	_	6	;
MOV.L ERs, ERd	L	$ERs32 \to ERd32$		2								_	—	\$	\$	0	_	2	2
MOV.L @ERs, ERd	L	@ERs $\rightarrow$ ERd32			4							_	—	\$	\$	0	—	8	3
MOV.L @(d:16, ERs), ERd	L	@(d:16, ERs) → ERd32				6						_	_	\$	\$	0	_	1	0
MOV.L @(d:24, ERs), ERd	L	@(d:24, ERs) → ERd32				10						_	_	\$	\$	0	_	1	4
MOV.L @ERs+, ERd	L	@ERs → ERd32 ERs32+4 → ERs32					4					—	_	\$	\$	0	_	1	0
MOV.L @aa:16, ERd	L	@aa:16 $\rightarrow$ ERd32						6				_	—	\$	↕	0	-	1	0
MOV.L @aa:24, ERd	L	$@aa:24 \rightarrow ERd32$						8				—	—	\$	€	0	_	1:	2
MOV.L ERs, @ERd	L	$ERs32 \to @ERd$			4							_	—	\$	€	0	—	8	5
MOV.L ERs, @(d:16, ERd)	L	$ERs32 \rightarrow @(d:16, ERd)$				6						_	_	\$	\$	0	_	1	0
MOV.L ERs, @(d:24, ERd)	L	$ERs32 \to @(d:24, ERd)$				10						_	—	\$	\$	0	_	1	4
MOV.L ERs, @-ERd	L	$\begin{array}{c} ERd324 \rightarrow ERd32\\ ERs32 \rightarrow @ERd \end{array}$					4					_	_	€	\$	0	_	1	0
MOV.L ERs, @aa:16	L	$ERs32 \to @aa:16$						6				—	—	$\updownarrow$	\$	0	-	1	0
MOV.L ERs, @aa:24	L	$ERs32 \to @aa:24$						8				_	—	\$	\$	0	_	1:	2
POP.W Rn	W										2	—	—	\$	\$	0	_	6	5
POP.L ERn	L	$\begin{array}{l} @SP \to ERn32 \\ SP+4 \to SP \end{array}$									4	—	—	\$	\$	0	—	1	0

							-		le ai (by		)							No. Sta	of tes <sup>*1</sup>			
	Operand Size				@ERn	(d, ERn)	@-ERn/@ERn+	a	d, PC)	@ @ aa	lied		Con	ditio	on C	Code	Ð	Normal	Advanced			
Mnemonic	ope	Operation	XX#	R	8	0	0	@ aa	@(d,	0	ᇤ	Condition C L H N Z 2 1 1					С	Nor	Adv			
PUSH.W Rn	W	$SP-2 \rightarrow SP$ Rn16 $\rightarrow @SP$									2	_	—	\$	\$	0	_	6	5			
PUSH.L ERn	L	$SP-4 \rightarrow SP$ ERn32 $\rightarrow @SP$									4	_	$\begin{array}{c} \uparrow \\ \hline \\$					1	0			
MOVFPE @aa:16, Rd	В	Cannot be used in the H8/3039 Group						4							e H8/30	039						
MOVTPE Rs, @aa:16	В	Cannot be used in the H8/3039 Group						4				-	Group Cannot be used in the H8/3039 Group									

#### 2. Arithmetic instructions

							ng I Ler				)							No. Stat	
	Operand Size		×	-	@ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	@aa	Implied		Con			-		Normal	Advanced
Mnemonic	ō	Operation	XX#	å	0	0	0	0	0	0	Ē	Ι	н	N	z	V	С	ž	Ă
ADD.B #xx:8, Rd	В	$Rd8+#xx:8 \rightarrow Rd8$	2									—	\$	€	⊅	€	↕	2	2
ADD.B Rs, Rd	В	$Rd8+Rs8 \rightarrow Rd8$		2								—	$\updownarrow$	↕	$\updownarrow$	$\updownarrow$	$\updownarrow$	2	2
ADD.W #xx:16, Rd	w	$Rd16+#xx:16 \rightarrow Rd16$	4									—	(1)	€	$\updownarrow$	≎	$\Leftrightarrow$	4	Ļ
ADD.W Rs, Rd	w	$Rd16+Rs16 \rightarrow Rd16$		2								—	(1)	↕	$\updownarrow$	↕	$\updownarrow$	2	2
ADD.L #xx:32, ERd	L	ERd32+#xx:32 → ERd32	6									—	(2)	\$	$\Leftrightarrow$	€	≎	6	;
ADD.L ERs, ERd	L	ERd32+ERs32 → ERd32		2								—	(2)	\$	$\Leftrightarrow$	€	€	2	2
ADDX.B #xx:8, Rd	в	$Rd8+#xx:8 + C \rightarrow Rd8$	2									_	\$	\$	(3)	€	$\Leftrightarrow$	2	2
ADDX.B Rs, Rd	в	$Rd8+Rs8 + C \rightarrow Rd8$		2								_	\$	\$	(3)	\$	€	2	2
ADDS.L #1, ERd	L	$ERd32+1 \rightarrow ERd32$		2								—	—	—	_	_		2	2
ADDS.L #2, ERd	L	$ERd32+2 \rightarrow ERd32$		2								—	—	—	—	_	_	2	2
ADDS.L #4, ERd	L	$ERd32+4 \rightarrow ERd32$		2								_	—	_		_	_	2	2
INC.B Rd	в	$Rd8+1 \rightarrow Rd8$		2								—	—	\$	$\updownarrow$	€	-	2	2
INC.W #1, Rd	w	$Rd16+1 \rightarrow Rd16$		2								_	—	\$	\$	\$	_	2	2
INC.W #2, Rd	W	$Rd16+2 \rightarrow Rd16$		2								—	—	\$	$\updownarrow$	↕	—	2	2

					ddre													No.	of *1
			<u> </u>	Inst	ruct	tion	Ler	ngth	by	tes	)							State	∋s″'
	Operand Size				@ERn	d, ERn)	@-ERn/@ERn+	9	@(d, PC)	@aa	Implied		Con	ditio	on C	cod	e	Normal	Advanced
Mnemonic	ope	Operation	¥¥	R	8	@(d,	8	@ aa	0	0	ᇤ	Т	н	Ν	z	v	С	No.	Adv
INC.L #1, ERd	L	$ERd32+1 \rightarrow ERd32$		2								-	—	\$	\$	\$	_	2	
INC.L #2, ERd	L	$ERd32+2 \to ERd32$		2								—	—	€	\$	$\updownarrow$	—	2	
DAA Rd	В	Rd8 decimal adjust $\rightarrow$ Rd8		2								-	*	\$	\$	*	-	2	
SUB.B Rs, Rd	В	$Rd8-Rs8 \rightarrow Rd8$		2								-	\$	€	\$	\$	\$	2	
SUB.W #xx:16, Rd	W	$Rd16\#xx:16 \rightarrow Rd16$	4									—	(1)	$\updownarrow$	€	$\updownarrow$	\$	4	
SUB.W Rs, Rd	W	$Rd16\text{-}Rs16 \rightarrow Rd16$		2								-	(1)	€	≎	$\updownarrow$	\$	2	
SUB.L #xx:32, ERd	L	ERd32-#xx:32 $\rightarrow$ ERd32	6									-	(2)	€	\$	\$	\$	6	
SUB.L ERs, ERd	L	$\begin{array}{l} ERd32\text{-}ERs32\\ \rightarrow ERd32 \end{array}$		2								-	(2)	\$	\$	\$	\$	2	
SUBX.B #xx:8, Rd	в	$Rd8$ –#xx:8– $C \rightarrow Rd8$	2									-	\$	€	(3)	\$	\$	2	
SUBX.B Rs, Rd	в	Rd8–Rs8–C $\rightarrow$ Rd8		2								-	\$	€	(3)	\$	\$	2	
SUBS.L #1, ERd	L	ERd32–1 $\rightarrow$ ERd32		2								—	—	_	—	—	-	2	
SUBS.L #2, ERd	L	$ERd32-2 \rightarrow ERd32$		2								—	—	—	—	—	_	2	
SUBS.L #4, ERd	L	$ERd32-4 \rightarrow ERd32$		2								-	—	—	-	—	_	2	
DEC.B Rd	В	$Rd8-1 \rightarrow Rd8$		2								_	—	€	\$	$\updownarrow$	_	2	
DEC.W #1, Rd	W	$Rd16-1 \rightarrow Rd16$		2								-	—	€	\$	\$	_	2	
DEC.W #2, Rd	W	$Rd16-2 \rightarrow Rd16$		2								_	—	€	€	$\updownarrow$	_	2	
DEC.L #1, ERd	L	$ERd321\toERd32$		2								—	—	€	\$	$\updownarrow$	—	2	
DEC.L #2, ERd	L	$ERd32-\!\!\!\!-\!\!\!2 \to ERd32$		2								-	—	€	\$	$\updownarrow$	_	2	
DAS.Rd	В	Rd8 decimal adjust $\rightarrow$ Rd8		2								-	*	\$	\$	*	-	2	
MULXU. B Rs, Rd	В	$Rd8 \times Rs8 \rightarrow Rd16$ (unsigned multiplication)		2								-	—	—	-	—	-	14	ł
MULXU. W Rs, ERd	W	$Rd16 \times Rs16 \rightarrow ERd32$ (unsigned multiplication)		2								-	—	—	-	_	-	22	2
MULXS. B Rs, Rd	В	$Rd8 \times Rs8 \rightarrow Rd16$ (signed multiplication)		4								-	—	\$	\$	_	-	16	;
MULXS. W Rs, ERd	W	$Rd16 \times Rs16 \rightarrow ERd32$ (signed multiplication)		4								_	_	€	\$	_	_	24	ŀ
DIVXU. B Rs, Rd	В	$\begin{array}{l} Rd16 \div Rs8 \rightarrow Rd16 \\ (RdH: remainder, \\ RdL: quotient) \\ (unsigned division) \end{array}$		2										(6)	(7)			14	ļ



									le ai 1 (by		)							No. Stat	
	Operand Size		J		@ERn	@(d, ERn)	@-ERn/@ERn+	33	@(d, PC)	@aa	mplied	(	Con	ditio	on C	od	e	Normal	Advanced
Mnemonic	ď	Operation	XX#	R	0	0	0	@ aa	0	00	Ξ	I	н	N	z	v	С	No	Ρq
DIVXU. W Rs, ERd	W	ERd32 ÷ Rs16 →ERd32 (Ed: remainder, Rd: quotient) (unsigned division)		2										(6)	(7)	_	_	2	2
DIVXS. B Rs, Rd	В	$Rd16 \div Rs8 \rightarrow Rd16$ (RdH: remainder, RdL: quotient) (signed division)		4										(8)	(7)		-	1	6
DIVXS. W Rs, ERd	W	$\begin{array}{l} ERd32 \div Rs16 \to ERd32 \\ (Ed: remainder, \\ Rd: quotient) \\ (signed division) \end{array}$		4										(8)	(7)		_	2	4
CMP.B #xx:8, Rd	В	Rd8–#xx:8	2									—	\$	\$	\$	\$	\$	2	2
CMP.B Rs, Rd	В	Rd8–Rs8		2								—	\$	\$	↕	\$	\$	2	2
CMP.W #xx:16, Rd	W	Rd16–#xx:16	4									—	(1)	\$	\$	\$	\$	4	Ļ
CMP.W Rs, Rd	W	Rd16–Rs16		2								—	(1)	\$	\$	\$	\$	2	2
CMP.L #xx:32, ERd	L	ERd32–#xx:32	6									—	(2)	\$	↕	\$	\$	4	Ļ
CMP.L ERs, ERd	L	ERd32–ERs32		2								—	(2)	\$	$\updownarrow$	\$	€	2	2
NEG.B Rd	В	$0-Rd8 \rightarrow Rd8$		2								—	€	\$	$\updownarrow$	\$	\$	2	2
NEG.W Rd	W	$0-Rd16 \rightarrow Rd16$		2								—	\$	\$	$\updownarrow$	\$	\$	2	2
NEG.L ERd	L	$0ERd32 \to ERd32$		2								—	\$	\$	$\updownarrow$	\$	\$	2	2
EXTU.W Rd	w	$0 \rightarrow (\text{} \text{of Rd16})$		2								—	-	0	\$	0	-	2	2
EXTU.L ERd	L	$0 \rightarrow$ ( <bits 16="" 31="" to=""> of Rd32)</bits>		2								—	-	0	\$	0	-	2	2
EXTS.W Rd	W	( <bit 7=""> of Rd16) <math>\rightarrow</math> (<bits 15="" 8="" to=""> of Rd16)</bits></bit>		2									_	\$	\$	0	-	2	2
EXTS.L ERd	L	( <bit 15=""> of Rd32) <math>\rightarrow</math> (<bits 16="" 31="" to=""> of ERd32)</bits></bit>		2								_	_	\$	\$	0	_	2	2

#### 3. Logic instructions

					ddre ruct		•				)							No. Stat	of es <sup>*1</sup>
	Operand Size				@ERn	ERn)	@-ERn/@ERn+		PC)	@aa	Implied		Con	ditio	on C	Code	•	Normal	Advanced
Mnemonic	ð	Operation	XX#	Rn	0	@(d,	9	@aa	@(d,	0	Ξ	I	н	Ν	z	۷	С	Ŷ	Ρq
AND.B #xx:8, Rd	В	$Rd8 \land \#xx: 8 \rightarrow Rd8$	2									-	-	\$	\$	0	—	2	2
AND.B Rs, Rd	В	$Rd8 {\scriptscriptstyle \wedge} Rs8 \rightarrow Rd8$		2								-	—	↕	\$	0	—	2	2
AND.W #xx:16, Rd	W	$Rd16 {\scriptstyle \wedge} \#xx: 16 \rightarrow Rd16$	4									—	—	↕	€	0	—	4	ļ
AND.W Rs, Rd	W	$Rd16 {\wedge} Rs16 \rightarrow Rd16$		2								-	—	€	\$	0	—	2	2
AND.L #xx:32, ERd	L	$ERd32{\scriptstyle\wedge} \#xx:\!32 \rightarrow ERd32$	6									-	_	€	\$	0	-	6	i
AND.L ERs, ERd	L	$ERd32{\scriptstyle\wedge}ERs32\rightarrowERd32$		4								—	—	€	\$	0	—	4	ļ
OR.B #xx:8, Rd	в	$Rd8{\lor}\#xx{:}8\toRd8$	2									_	—	€	\$	0	—	2	
OR.B Rs, Rd	в	$Rd8{\scriptstyle\vee}Rs8\rightarrowRd8$		2								—	—	€	\$	0	—	2	2
OR.W #xx:16, Rd	W	$Rd16 \lor \#xx:16 \rightarrow Rd16$	4									—	—	€	\$	0	—	4	ļ
OR.W Rs, Rd	W	$Rd16 {\lor} Rs16 \rightarrow Rd16$		2								—	—	€	\$	0	—	2	2
OR.L #xx:32, ERd	L	$ERd32{\lor} \#xx:32 \to ERd32$	6									—	—	€	\$	0	—	6	;
OR.L ERs, ERd	L	$ERd32{\lor}ERs32\toERd32$		4								—	—	\$	\$	0	—	4	ļ
XOR.B #xx:8, Rd	в	$Rd8 \oplus \#xx: 8 \rightarrow Rd8$	2									_	—	↕	\$	0	—	2	2
XOR.B Rs, Rd	в	$Rd8{\oplus}Rs8\toRd8$		2								_	—	\$	\$	0	—	2	2
XOR.W #xx:16, Rd	w	Rd16⊕#xx:16 → Rd16	4									_	_	\$	\$	0	—	4	ļ
XOR.W Rs, Rd	W	$Rd16 \oplus Rs16 \rightarrow Rd16$		2								_	_	€	\$	0	_	2	2
XOR.L #xx:32, ERd	L	$ERd32 {\oplus} \#xx:32 \rightarrow ERd32$	6									_	-	\$	\$	0	-	6	i
XOR.L ERs, ERd	L	$ERd32{\oplus}ERs32 \to ERd32$		4								_	-	€	\$	0	-	4	ļ
NOT.B Rd	в	$\neg \text{ Rd8} \rightarrow \text{ Rd8}$		2								_	_	\$	\$	0	_	2	2
NOT.W Rd	w	¬ Rd16 → Rd16		2								_	-	\$	\$	0	-	2	2
NOT.L ERd	L	¬ Rd32 → Rd32		2								—	-	\$	\$	0	-	2	2

#### 4. Shift instructions

							ng I Ler				)							No. Stat	of es <sup>*1</sup>
	Operand Size				@ERn	d, ERn)	@-ERn/@ERn+	a	d, PC)	@ aa	Implied		Con	ditio	on C	od	9	Normal	Advanced
Mnemonic	ð	Operation	XX#	뙵	8	@(d,	ø	@aa	@(d,	00	E	I	н	Ν	z	v	С	Ŷ	Ρq
SHAL.B Rd	В			2								-	-	↕	\$	↕	\$	2	
SHAL.W Rd	w			2								-	-	\$	€	↕	↕	2	
SHAL.L ERd	L	C MSB LSB		2								-	-	\$	\$	↕	\$	2	
SHAR.B Rd	в			2								-	-	$\updownarrow$	\$	0	\$	2	
SHAR.W Rd	w			2								-	-	\$	\$	0	\$	2	1
SHAR.L ERd	L	MSB LSB C		2								_	-	↕	\$	0	\$	2	
SHLL.B Rd	в			2								_	—	\$	\$	0	\$	2	
SHLL.W Rd	w	↓ ↓ ↓ 0		2								_	-	↕	\$	0	\$	2	
SHLL.L ERd	L	C MSB LSB		2								_	_	\$	\$	0	\$	2	
SHLR.B Rd	в			2								—	-	\$	\$	0	\$	2	
SHLR.W Rd	w			2								_	-	↕	\$	0	\$	2	
SHLR.L ERd	L	MSB LSB C		2								—	-	\$	\$	0	\$	2	
ROTXL.B Rd	в			2								_	-	\$	\$	0	\$	2	
ROTXL.W Rd	w	╘┤╺╴		2								_	-	$\updownarrow$	\$	0	\$	2	
ROTXL.L ERd	L	C MSB 🗲 LSB		2								_	_	\$	\$	0	\$	2	
ROTXR.B Rd	в			2								_	—	\$	\$	0	\$	2	
ROTXR.W Rd	w	┝┶└─────┝しど		2								_	-	$\updownarrow$	\$	0	\$	2	
ROTXR.L ERd	L	MSB → LSB C		2								_	-	\$	\$	0	\$	2	
ROTL.B Rd	в			2								_	-	\$	\$	0	\$	2	!
ROTL.W Rd	w	╽║┩		2								_	-	\$	\$	0	\$	2	
ROTL.L ERd	L	C MSB 🔶 LSB		2								_	-	\$	\$	0	\$	2	
ROTR.B Rd	в			2								_	-	\$	\$	0	\$	2	
ROTR.W Rd	w	╞┶		2								_	-	\$	\$	0	\$	2	
ROTR.L ERd	L	MSB → LSB C		2								_	-	\$	\$	0	\$	2	

# 5. Bit manipulation instructions

							-		le ar (by		)							No. Stat	of es <sup>*1</sup>
Mnemonic	Operand Size	Operation	#xx	Rn	@ERn	@(d, ERn)	@-ERn/@ERn+	Qaa	@(d, PC)	@ @ aa	Implied	(	Con	ditio	on C	Code	e C	Normal	Advanced
BSET #xx:3, Rd	В	(#xx:3 of Rd8) ← 1		2								_	_	_	_	_	-	2	
BSET #xx:3, @ERd	в	(#xx:3 of @ERd) ← 1			4							_	_	_	_	_	_	8	
BSET #xx:3, @aa:8	в	(#xx:3 of @aa:8) ← 1						4				_	_	_	_	_	_	8	
BSET Rn, Rd	в	(Rn8 of Rd8) ← 1		2								_	_	_	_	_	_	2	
BSET Rn, @ERd	в	(Rn8 of @ERd) $\leftarrow$ 1			4							_	_	_	-	_	_	8	
BSET Rn, @aa:8	в	(Rn8 of @aa:8) ← 1						4				_	—	_	_	—	_	8	
BCLR #xx:3, Rd	в	(#xx:3 of Rd8) ← 0		2								_	—	_	_	—	_	2	
BCLR #xx:3, @ERd	в	(#xx:3 of @ERd) $\leftarrow$ 0			4							_	_	_	_	—	_	8	
BCLR #xx:3, @aa:8	в	(#xx:3 of @aa:8) ← 0						4				_	—	_	_	—	_	8	
BCLR Rn, Rd	в	(Rn8 of Rd8) ← 0		2								—	—	_	_	—	_	2	
BCLR Rn, @ERd	в	(Rn8 of @ERd) $\leftarrow 0$			4							_	—	_	_	—	_	8	
BCLR Rn, @aa:8	в	(Rn8 of @aa:8) ← 0						4				—	—	—	-	—	—	8	
BNOT #xx:3, Rd	В	(#xx:3 of Rd8) ← ¬ (#xx:3 of Rd8)		2								_		_	-	—	-	2	
BNOT #xx:3, @ERd	в	(#xx:3 of @ERd) ← ¬ (#xx:3 of @ERd)			4							_	—	-	-	—	-	8	
BNOT #xx:3, @aa:8	в	(#xx:3 of @aa:8) ← ¬ (#xx:3 of @aa:8)						4				_	—	-	-	—	-	8	
BNOT Rn, Rd	В	(Rn8 of Rd8) ← ¬ (Rn8 of Rd8)		2								_	—	-	-	—	-	2	
BNOT Rn, @ERd	В	(Rn8 of @ERd) ← ¬ (Rn8 of @ERd)			4							_	—	_	-	_	-	8	
BNOT Rn, @aa:8	В	(Rn8 of @aa:8) ← ¬ (Rn8 of @aa:8)						4				-	—	-	-	—	-	8	
BTST #xx:3, Rd	в	¬ (#xx:3 of Rd8) → Z		2								_	_	_	\$	—	_	2	
BTST #xx:3, @ERd	В	¬ (#xx:3 of @ERd) → Z			4							—	—	—	\$	—	-	6	
BTST #xx:3, @aa:8	В	¬ (#xx:3 of @aa:8) → Z						4				_	—	_	\$	—	-	6	
BTST Rn, Rd	В	¬ (Rn8 of @Rd8) → Z		2								—	—	—	\$	_	-	2	
BTST Rn, @ERd	В	¬ (Rn8 of @ERd) → Z			4							_	_	_	\$	_	_	6	
BTST Rn, @aa:8	В	¬ (Rn8 of @aa:8) → Z						4				_	_	_	\$	_	_	6	
BLD #xx:3, Rd	В	(#xx:3 of Rd8) $\rightarrow$ C		2								-	—	-	-	—	\$	2	

							ng I Ler				)							No. Sta	of tes <sup>*1</sup>
Mnemonic	Operand Size	Operation	XX#	Rn	@ERn	@(d, ERn)	@-ERn/@ERn+	@ aa	@(d, PC)	@ @ aa	Implied	1	Con	ditio	on C	code V	÷	Normal	Advanced
BLD #xx:3, @ERd	В	(#xx:3 of @ERd) $\rightarrow$ C		-	4						_	_	_	_	_	_	\$	- 6	;
BLD #xx:3, @aa:8	В	(#xx:3 of @aa:8) $\rightarrow$ C						4				_	—	_	_	_	\$	6	;
BILD #xx:3, Rd	В	¬ (#xx:3 of Rd8) → C		2								_	_	_	_	-	\$	2	2
BILD #xx:3, @ERd	В	¬ (#xx:3 of @ERd) → C			4							_	—	—	—	—	\$	6	;
BILD #xx:3, @aa:8	В	¬ (#xx:3 of @aa:8) → C						4				—	—	—	—	-	\$	6	;
BST #xx:3, Rd	в	$C \rightarrow (\#xx:3 \text{ of } Rd8)$		2								—	—	-	—	—	-	2	2
BST #xx:3, @ERd	В	$C \rightarrow (\#xx:3 \text{ of } @ERd24)$			4							_	-	-	—	-	-	8	}
BST #xx:3, @aa:8	В	$C \rightarrow$ (#xx:3 of @aa:8)						4				_	—	-	—	—	—	8	}
BIST #xx:3, Rd	В	$\neg \text{ C} \rightarrow (\text{\#xx:3 of Rd8})$		2								-	-	-	—	-	-	2	2
BIST #xx:3, @ERd	В	$\neg \text{ C} \rightarrow (\text{\#xx:3 of @ERd24})$			4							—	—	-	—	-	—	8	}
BIST #xx:3, @aa:8	В	$\neg \text{ C} \rightarrow (\text{\#xx:3 of @aa:8})$						4				—	—	—	—	—	—	8	}
BAND #xx:3, Rd	В	$C {\wedge} (\#xx:3 \text{ of } Rd8) \to C$		2								—	—	-	—	-	\$	2	2
BAND #xx:3, @ERd	В	$C \land (\#xx:3 \text{ of } @ERd24) \rightarrow C$			4							—	—	—	—	—	\$	6	;
BAND #xx:3, @aa:8	В	$C {\wedge} (\#xx:3 \text{ of } @aa:8) \rightarrow C$						4				—	—	—	—	-	\$	6	;
BIAND #xx:3, Rd	В	$C \land \neg (\#xx:3 \text{ of } Rd8) \to C$		2								—	—	—	—	—	\$	2	2
BIAND #xx:3, @ERd	В	$C \wedge \neg$ (#xx:3 of @ERd24) $\rightarrow$ C			4							—	—	—	—	—	\$	6	;
BIAND #xx:3, @aa:8	В	$C \wedge \neg$ (#xx:3 of @aa:8) $\rightarrow C$						4				—	—	—	—	—	\$	6	;
BOR #xx:3, Rd	В	$C {\scriptstyle \lor} (\#xx:3 \text{ of } Rd8) \rightarrow C$		2								—	—	—	—	—	\$	2	2
BOR #xx:3, @ERd	В	$C{\lor}(\texttt{\#xx:3 of @ERd24}) \to C$			4							—	—	—	—	—	\$	6	;
BOR #xx:3, @aa:8	В	$C{\scriptstyle\lor}(\#xx{:}3\text{ of }@aa{:}8)\rightarrow C$						4				_	-	-	—	-	\$	6	;
BIOR #xx:3, Rd	В	$C \lor \neg$ (#xx:3 of Rd8) $\rightarrow$ C		2								—	—	—	—	—	\$	2	2
BIOR #xx:3, @ERd	В	$C \lor \neg$ (#xx:3 of @ERd24) $\rightarrow$ C			4							_	—	_	—	—	\$	6	;
BIOR #xx:3, @aa:8	В	$C \lor \neg$ (#xx:3 of @aa:8) $\rightarrow C$						4				_	—	—	—	—	\$	6	;
BXOR #xx:3, Rd	В	$C \oplus (\texttt{\#xx:3 of Rd8}) \to C$		2								_	—	—	—	—	\$	2	2
BXOR #xx:3, @ERd	В	$C \oplus (\#xx:3 \text{ of } @ERd24) \rightarrow C$			4							_	—	_	—	_	\$	6	;
BXOR #xx:3, @aa:8	В	C ⊕ (#xx:3 of @aa:8) → C						4				_	—	—	—	—	$\updownarrow$	6	;
BIXOR #xx:3, Rd	В	$C \oplus \neg (\#xx:3 \text{ of } Rd8) \rightarrow C$		2								_	—	_	_	_	\$	2	2
BIXOR #xx:3, @ERd	В	$C \oplus \neg (\#xx:3 \text{ of } @ERd24) \rightarrow C$			4							_	—	-	—	—	\$	6	;
BIXOR #xx:3, @aa:8	В	$C \oplus \neg \ (\#xx:3 \text{ of } @aa:8) \! \rightarrow \! C$						4						-	_		\$	6	;

# 6. Branching instructions

					A Inst	ddre ruct						)							No. Stat	of es <sup>*1</sup>
Mnemonic	<b>Operand Size</b>	Operation		XX#	Rn	@ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	@ @ aa	Implied	(	Con H	ditio N	on C Z	Code V	e C	Normal	Advanced
BRA d:8 (BT d:8)	—	If condition	Always							2			-	—	-	-	—	-	4	
BRA d:16 (BT d:16)	—	is true then PC ←								4			_	—	—	_	—	-	6	
BRN d:8 (BF d:8)	—	PC+d else	Never							2			-	—	—	-	—	—	4	
BRN d:16 (BF d:16)	—	next;								4			_	_	_	_	_	_	6	
BHI d:8	—		$C \lor Z = 0$							2			-	—	—	-	—	-	4	
BHI d:16	—									4			_	—	_	_	_	_	6	
BLS d:8	—		$C \lor Z = 1$							2			_	—	—	_	—	—	4	
BLS d:16	—									4			-	—	—	-	—	_	6	
BCC d:8 (BHS d:8)	—		C = 0							2			—	—	—	—	—	—	4	
BCC d:16 (BHS d:16)	-									4			-	_	—	-	—	-	6	
BCS d:8 (BLO d:8)	_		C = 1							2			—	—	—	—	—	—	4	
BCS d:16 (BLO d:16)	-									4			_	—	—	_	—	_	6	
BNE d:8	_		Z = 0							2			—	_	—	—	—	_	4	
BNE d:16	-									4			—	—	—	—	—	_	6	
BEQ d:8	-		Z = 1							2			—	_	—	—	_	_	4	
BEQ d:16	_									4			—	—	—	—	—	—	6	
BVC d:8	_		V = 0							2			_	—	_	_	—	_	4	
BVC d:16	_									4			—	—	—	—	—	—	6	
BVS d:8	_		V = 1							2			_	—	—	_	—	_	4	
BVS d:16	_									4			—	—	—	—	—	—	6	
BPL d:8	_		N = 0							2			_	—	—	_	—	_	4	
BPL d:16	_									4			—	—	—	—	—	—	6	
BMI d:8	_		N = 1							2			—	—	—	_	—	_	4	
BMI d:16	_									4			—	—	—	—	—	—	6	
BGE d:8	—		$N \oplus V = 0$							2			_	—	_	_	—	_	4	
BGE d:16	_									4			—	—	—	—	—	_	6	
BLT d:8	_		N ⊕ V = 1							2			_	_	_	_	—	_	4	
BLT d:16	_									4			—	_	—	—	—	—	6	
BGT d:8	_		$Z \lor (N \oplus V)$							2			_	-	_	_	-	-	4	
BGT d:16	_		= 0							4			—	_	—	—	—	—	6	

										le ar (by		)							No. Stat	of tes <sup>*1</sup>
	Operand Size			×		@ERn	@(d, ERn)	@-ERn/@ERn+	Qaa	@(d, PC)	@ @ aa	Implied	0	Con	ditio	on C	Code	•	Normal	Advanced
Mnemonic	g	Operation		XX#	R	0	0	ø	0	0	0	<u></u>	Т	н	Ν	z	v	С	Ŷ	PA
BLE d:8	—	If condition	$Z \lor (N \oplus V)$							2			_	-	-	—	-	—	4	1
BLE d:16		is true then PC ← PC+d else next;	= 1							4									6	6
JMP @ERn	_	$PC \gets ERn$				2							_	-	-	_	-	_	4	1
JMP @aa:24		$PC \gets aa: 24$							4				_	_	_	_	_	_	6	6
JMP @@aa:8	_	PC ← @aa:8	3								2		_	_	_	_	_		8	10
BSR d:8		$PC \rightarrow @-SF$ $PC \leftarrow PC+d$								2			_	_	-	-	_	—	6	8
BSR d:16	_	$PC \rightarrow @-SF$ $PC \leftarrow PC+d$								4			_	_	-	—	_	—	8	10
JSR @ERn	_	$PC \rightarrow @-SF$ $PC \leftarrow @ERr$				2								—	-	—	—	—	6	8
JSR @aa:24	_	$PC \rightarrow @-SF$ $PC \leftarrow @aa:2$							4				_	_	_	_	_	—	8	10
JSR @@aa:8	_	$PC \rightarrow @-SF$ $PC \leftarrow @aa:8$									2			_	_		_	—	8	12
RTS	_	$PC \leftarrow @SP$	+									2	—	-	-		-	—	8	10

#### 7. System control instructions

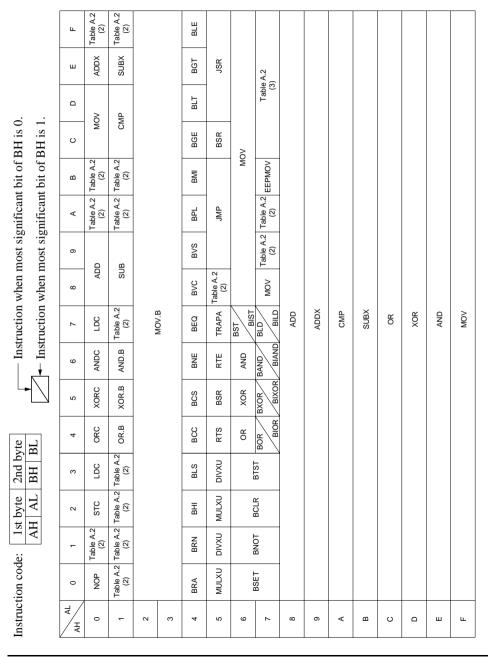
			Addressing Mode and Instruction Length (bytes)					)							No. Stat	of tes <sup>*1</sup>			
Mnemonic	Operand Size	Operation	xx#	Rn	@ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	@ @aa	Implied	- (	Con	ditio	on C	Code	e C	Normal	Advanced
TRAPA #x:2		$PC \rightarrow @-SP$ $CCR \rightarrow @-SP$ $ \rightarrow PC$									2	1	_	-	-	_	_	14	16
RTE	-	$CCR \leftarrow @SP+$ PC $\leftarrow @SP+$										\$	\$	\$	\$	\$	\$	1	0
SLEEP	—	Transition to power- down state										—	—	-	-	—	_	2	2
LDC #xx:8, CCR	В	#xx:8 → CCR	2									\$	\$	\$	\$	↕	\$	2	2
LDC Rs, CCR	В	$Rs8 \rightarrow CCR$		2								\$	↕	\$	\$	↕	\$	2	2
LDC @ERs, CCR	W	$@ERs \rightarrow CCR$			4							\$	\$	\$	\$	↕	\$	6	6
LDC @(d:16, ERs), CCR	W	@(d:16, ERs) → CCR				6						\$	\$	\$	\$	\$	\$	8	3
LDC @(d:24, ERs), CCR	w	@(d:24, ERs) → CCR				10						\$	\$	\$	\$	\$	\$	1	2
LDC @ERs+, CCR	W	@ERs → CCR ERs32+2 → ERs32					4					\$	\$	\$	\$	\$	\$	8	3
LDC @aa:16, CCR	W	@aa:16 → CCR						6				\$	↕	\$	\$	$\uparrow$	\$	8	3
LDC @aa:24, CCR	W	@aa:24 → CCR						8				\$	\$	\$	\$	€	\$	1	0
STC CCR, Rd	В	$CCR \rightarrow Rd8$		2								_	—	_	-	—	—	2	2
STC CCR, @ERd	W	$CCR \to @ERd$			4							—	—	_	—	—	—	6	6
STC CCR, @(d:16, ERd)	V	$CCR \rightarrow @(d:16, ERd)$				6						-	-	-	-	—	_	8	3
STC CCR, @(d:24, ERd)	w	$CCR \to @(d{:}24, ERd)$				10						_	—	-	_		_	1	2
STC CCR, @-ERd	W	$\begin{array}{c} ERd32-2 \rightarrow ERd32 \\ CCR \rightarrow @ERd \end{array}$					4					_	—	-	-	_	_	8	3
STC CCR, @aa:16	W	$CCR \rightarrow @aa:16$						6				_	_	_	_	—	—	ε	3
STC CCR, @aa:24	W	$CCR \rightarrow @aa:24$						8				_	—	-	-	—	—	1	0
ANDC #xx:8, CCR	В	CCR∧#xx:8 → CCR	2									\$	€	\$	\$	\$	\$	2	2
ORC #xx:8, CCR	В	$CCR \lor \#xx:8 \rightarrow CCR$	2									\$	$\uparrow$	\$	\$	\$	\$	2	2
XORC #xx:8, CCR	В	$CCR \oplus \#xx: 8 \rightarrow CCR$	2									\$	€	\$	\$	\$	\$	2	2
NOP	_	$PC \gets PC+2$									2	_	_	-	-	_	-	2	2

#### 8. Block transfer instructions

									le ar (by		)							No. Sta	of tes <sup>*1</sup>
	Operand Size				@ERn	@(d, ERn)	@-ERn/@ERn+	ŋ	d, PC)	@ aa	Implied	0	Con	ditio	on C	ode	•	Normal	Advanced
Mnemonic	ð	Operation	XX#	R	0	0	0	@ aa	@(d,	0	Ĕ	T	н	N	z	v	С	Ñ	Ad
EEPMOV. B	-	$\begin{array}{l} \text{if } R4L \neq 0 \text{ then} \\ \text{repeat} & @R5 \rightarrow @R6 \\ & R5+1 \rightarrow R5 \\ & R6+1 \rightarrow R6 \\ & R4L-1 \rightarrow R4L \\ \text{until} & R4L=0 \\ \text{else next} \end{array}$									4							8+4	4n <sup>*2</sup>
EEPMOV. W	_	$\begin{array}{l} \text{if } R4 \neq 0 \text{ then} \\ \text{repeat}  @R5 \rightarrow @R6 \\ R5+1 \rightarrow R5 \\ R6+1 \rightarrow R6 \\ R4-1 \rightarrow R4 \\ \text{until}  R4=0 \\ \text{else next} \end{array}$									4		_					8+4	4n <sup>*2</sup>

Notes: 1. The number of states is the number of states required for execution when the instruction and its operands are located in on-chip memory. For other cases see section A.3, Number of States Required for Execution.

- 2. n is the value set in register R4L or R4.
- (1) Set to 1 when a carry or borrow occurs at bit 11; otherwise cleared to 0.
- (2) Set to 1 when a carry or borrow occurs at bit 27; otherwise cleared to 0.
- (3) Retains its previous value when the result is zero; otherwise cleared to 0.
- (4) Set to 1 when the adjustment produces a carry; otherwise retains its previous value.
- (5) The number of states required for execution of an instruction that transfers data in synchronization with the E clock is variable.
- (6) Set to 1 when the divisor is negative; otherwise cleared to 0.
- (7) Set to 1 when the divisor is zero; otherwise cleared to 0.
- (8) Set to 1 when the quotient is negative; otherwise cleared to 0.



# \_ \_ \_

**Operation Code Maps** 

A.2

L	Table A.2 (3)		INC						EXTS		DEC		BLE		
ш													BGT		
۵	Table A.2 Table A.2 (3) (3)		INC						EXTS		DEC		вгт		
U	Table A.2 (3)	ADD		MOV						SUB		CMP	BGE		
ß		AI		W	SHAL	SHAR	ROTL	ROTR	NEG	ស		CI	BMI		
۲													BPL		
<u></u> б			ADDS		SHAL	SHAR	ROTL	ROTR	NEG		SUB		BVS		
8	SLEEP		AD		S.	Ъ	RC	RC	Z		S		BVC		
2			INC						ЕХТИ		DEC		BEQ		
و													BNE	AND	AND
2J			NC						ЕХТИ		DEC		BCS	XOR	XOR
4	LDC/STC												BCC	OR	OR
ю					SHLL	SHLR	ROTXL	ROTXR	NOT				BLS	SUB	SUB
5													BHI	СМР	CMP
-					SHLL	SHLR	ROTXL	ROTXR	NOT				BRN	ADD	ADD
0	NOM	INC	ADDS	DAA	Ϋ́	ST	RO	RO <sup>-</sup>	ž	DEC	SUBS	DAS	BRA	MOV	MOV
AH AL	10	OA	OB	OF	10	11	12	13	17	1A	1B	1Ε	58	62	7A

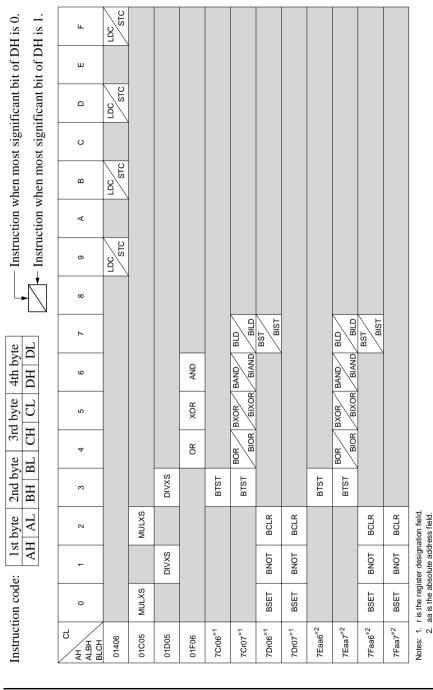
# Table A.2 Operation Code Map (2)

Appendix A Instruction Set

Rev.3.00 Mar. 26, 2007 Page 578 of 682 REJ09B0353-0300

1st byte2nd byteAHALBHBL

Instruction code:



Rev.3.00 Mar. 26, 2007 Page 579 of 682 REJ09B0353-0300

# A.3 Number of States Required for Execution

The tables in this section can be used to calculate the number of states required for instruction execution by the H8/300H CPU. Table A.3 indicates the number of states required per cycle according to the bus size. Table A.4 indicates the number of instruction fetch, data read/write, and other cycles occurring in each instruction. The number of states required for execution of an instruction can be calculated from these two tables as follows:

Number of states =  $I \times S_1 + J \times S_1 + K \times S_K + L \times S_L + M \times S_M + N \times S_N$ 

# Examples of Calculation of Number of States Required for Execution

**Examples:** Advanced mode, stack located in external address space, on-chip supporting modules accessed with 8-bit bus width, external devices accessed in three states with one wait state and 16-bit bus width.

#### BSET #0, @FFFFC7:8

From table A.3,  $S_1 = 4$  and  $S_L = 3$ From table A.4, I = L = 2 and J = K = M = N = 0Number of states  $= 2 \times 4 + 2 \times 3 = 14$ 

JSR @@30

From table A.3,  $S_1 = S_3 = S_k = 4$ From table A.4, I = J = K = 2 and L = M = N = 0Number of states  $= 2 \times 4 + 2 \times 4 + 2 \times 4 = 24$ 

				Acc	ess Cond	itions		
				-Chip		Externa	al Device	
			•	porting odule	8-Bit	Bus	16-B	it Bus
Cycle		On-Chip Memory	8-Bit Bus	16-Bit Bus	2-State Access		2-State Access	3-State Access
Instruction fetch	S,	2	6	3	4	6 + 2m	2	3 + m
Branch address read	S							
Stack operation	Sκ							
Byte data access	$S_{\scriptscriptstyle L}$		3		2	3 + m	_	
Word data access	S <sub>M</sub>		6		4	6 + 2m	_	
Internal operation	S <sub>N</sub>	1	1	1	1	1	1	1
Logondi								

#### Table A.3 Number of States per Cycle

Legend:

m: Number of wait states inserted in external device access

# Table A.4 Number of Cycles per Instruction

		Instruction Fetch	Addr. Read	-	Access	Word Data Access	Operation
Instruction	Mnemonic	I	J	к	L	Μ	Ν
ADD	ADD.B #xx:8, Rd	1					
	ADD.B Rs, Rd	1					
	ADD.W #xx:16, Rd	2					
	ADD.W Rs, Rd	1					
	ADD.L #xx:32, ERd	3					
	ADD.L ERs, ERd	1					
ADDS	ADDS #1/2/4, ERd	1					
ADDX	ADDX #xx:8, Rd	1					
	ADDX Rs, Rd	1					
AND	AND.B #xx:8, Rd	1					
	AND.B Rs, Rd	1					
	AND.W #xx:16, Rd	2					
	AND.W Rs, Rd	1					
	AND.L #xx:32, ERd	3					
	AND.L ERs, ERd	2					
ANDC	ANDC #xx:8, CCR	1					
BAND	BAND #xx:3, Rd	1					
	BAND #xx:3, @ERd	2			1		
	BAND #xx:3, @aa:8	2			1		
Bcc	BRA d:8 (BT d:8)	2					
	BRN d:8 (BF d:8)	2					
	BHI d:8	2					
	BLS d:8	2					
	BCC d:8 (BHS d:8)	2					
	BCS d:8 (BLO d:8)	2					
	BNE d:8	2					
	BEQ d:8	2					
	BVC d:8	2					
	BVS d:8	2					
	BPL d:8	2					
	BMI d:8	2					
	BGE d:8	2					
	BLT d:8	2					
	BGT d:8	2					
	BLE d:8	2					

1	••	Instruction Fetch	Addr. Read	•		Word Data Access	Internal Operation
Instruction	Mnemonic	1	J	К	L	М	N
Bcc	BRA d:16 (BT d:16)	2					2
	BRN d:16 (BF d:16)	2					2
	BHI d:16	2					2
	BLS d:16	2					2
	BCC d:16 (BHS d:16)	2					2
	BCS d:16 (BLO d:16)	2					2
	BNE d:16	2					2
	BEQ d:16	2					2
	BVC d:16	2					2
	BVS d:16	2					2
	BPL d:16	2					2
	BMI d:16	2					2
	BGE d:16	2					2
	BLT d:16	2					2
	BGT d:16	2					2
	BLE d:16	2					2
BCLR	BCLR #xx:3, Rd	1					
	BCLR #xx:3, @ERd	2			2		
	BCLR #xx:3, @aa:8	2			2		
	BCLR Rn, Rd	1					
	BCLR Rn, @ERd	2			2		
	BCLR Rn, @aa:8	2			2		
BIAND	BIAND #xx:3, Rd	1					
	BIAND #xx:3, @ERd	2			1		
	BIAND #xx:3, @aa:8	2			1		
BILD	BILD #xx:3, Rd	1					
	BILD #xx:3, @ERd	2			1		
	BILD #xx:3, @aa:8	2			1		
BIOR	BIOR #xx:8, Rd	1					
	BIOR #xx:8, @ERd	2			1		
	BIOR #xx:8, @aa:8	2			1		
BIST	BIST #xx:3, Rd	1					
	BIST #xx:3, @ERd	2			2		
	BIST #xx:3, @aa:8	2			2		
BIXOR	BIXOR #xx:3, Rd	1					
	BIXOR #xx:3, @ERd	2			1		
	BIXOR #xx:3, @aa:8	2			1		

#### Appendix A Instruction Set

Instruction	Mnemonic		Instruction Fetch I	Stack Operation K	-	Word Data Access M	Internal Operation N
BLD	BLD #xx:3, R	d	1				
	BLD #xx:3, @	ERd	2		1		
	BLD #xx:3, @	aa:8	2		1		
BNOT	BNOT #xx:3,	Rd	1				
	BNOT #xx:3,	@ERd	2		2		
	BNOT #xx:3,	@aa:8	2		2		
	BNOT Rn, Ro	ł	1				
	BNOT Rn, @	ERd	2		2		
	BNOT Rn, @	aa:8	2		2		
BOR	BOR #xx:3, R	d	1				
	BOR #xx:3, @	0ERd	2		1		
	BOR #xx:3, @	⊉aa:8	2		1		
BSET	BSET #xx:3,	Rd	1				
	BSET #xx:3,	@ERd	2		2		
	BSET #xx:3,	@aa:8	2		2		
	BSET Rn, Rd		1				
	BSET Rn, @I	ERd	2		2		
	BSET Rn, @a	aa:8	2		2		
BSR	BSR d:8	Normal	2	1			
		Advanced	2	2			
	BSR d:16	Normal	2	1			2
		Advanced	2	2			2
BST	BST #xx:3, R	d	1				
	BST #xx:3, @	ERd	2		2		
	BST #xx:3, @	aa:8	2		2		
BTST	BTST #xx:3, I	٦d	1				
	BTST #xx:3,	@ERd	2		1		
	BTST #xx:3,	@aa:8	2		1		
	BTST Rn, Rd		1				
	BTST Rn, @B	ERd	2		1		
	BTST Rn, @a	aa:8	2		1		
BXOR	BXOR #xx:3,	Rd	1				
	BXOR #xx:3,	@ERd	2		1		
	BXOR #xx:3,	@aa:8	2		1		

Instruction	Mnemonic		Instruction Fetch I		Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
CMP	CMP.B #xx:8, R	?d	1	-		-		
	CMP.B Rs, Rd		1					
	CMP.W #xx:16,	Rd	2					
	CMP.W Rs, Rd		1					
	CMP.L #xx:32,		3					
	CMP.L ERs, EF		1					
DAA	DAA Rd		1					
DAS	DAS Rd		1					
DEC	DEC.B Rd		1					
	DEC.W #1/2, R	d	1					
	DEC.L #1/2, ER	۲d	1					
DIVXS	DIVXS.B Rs, Ro	d	2					12
	DIVXS.W Rs, E	Rd	2					20
DIVXU	DIVXU.B Rs, R	d	1					12
	DIVXU.W Rs, E	Rd	1					20
EEPMOV	EEPMOV.B		2			2n +2*1		
	EEPMOV.W		2			2n +2*1		
EXTS	EXTS.W Rd		1					
	EXTS.L ERd		1					
EXTU	EXTU.W Rd		1					
	EXTU.L ERd		1					
INC	INC.B Rd		1					
	INC.W #1/2, Rd	I	1					
	INC.L #1/2, ER	d	1					
JMP	JMP @ERn		2					
	JMP @aa:24		2					2
	JMP @@aa:8	Normal	2	1				2
		Advanced	2	2				2
JSR	JSR @ERn	Normal	2		1			
		Advanced	2		2			
	JSR @aa:24	Normal	2		1			2
		Advanced	2		2			2
	JSR @@aa:8	Normal	2	1	1			
		Advanced	2	2	2			

#### Appendix A Instruction Set

Instruction	Mnemonic	Instruction Fetch I	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
LDC	LDC #xx:8, CCR	1				
	LDC Rs, CCR	1				
	LDC @ERs, CCR	2			1	
	LDC @(d:16, ERs), CCR	3			1	
	LDC @(d:24, ERs), CCR	5			1	
	LDC @ERs+, CCR	2			1	2
	LDC @aa:16, CCR	3			1	
	LDC @aa:24, CCR	4			1	
MOV	MOV.B #xx:8, Rd	1				
	MOV.B Rs, Rd	1				
	MOV.B @ERs, Rd	1		1		
	MOV.B @(d:16, ERs), Rd	2		1		
	MOV.B @(d:24, ERs), Rd	4		1		
	MOV.B @ERs+, Rd	1		1		2
	MOV.B @aa:8, Rd	1		1		
	MOV.B @aa:16, Rd	2		1		
	MOV.B @aa:24, Rd	3		1		
	MOV.B Rs, @ERd	1		1		
	MOV.B Rs, @(d:16, ERd)	2		1		
	MOV.B Rs, @(d:24, ERd)	4		1		
	MOV.B Rs, @-ERd	1		1		2
	MOV.B Rs, @aa:8	1		1		
	MOV.B Rs, @aa:16	2		1		
	MOV.B Rs, @aa:24	3		1		
	MOV.W #xx:16, Rd	2				
	MOV.W Rs, Rd	1				
	MOV.W @ERs, Rd	1			1	
	MOV.W @(d:16, ERs), Rd	2			1	
	MOV.W @(d:24, ERs), Rd	4			1	
	MOV.W @ERs+, Rd	1			1	2
	MOV.W @aa:16, Rd	2			1	
	MOV.W @aa:24, Rd	3			1	
	MOV.W Rs, @ERd	1			1	
	MOV.W Rs, @(d:16, ERd)	2			1	
	MOV.W Rs, @(d:24, ERd)	4			1	
	MOV.W Rs, @-ERd	1			1	2
	MOV.W Rs, @aa:16	2	 		1	

		Instruction Fetch	Addr. Read	Stack Operation	Byte Data Access	Word Data Access	Internal Operation
Instruction	Mnemonic	I	J	К	L	М	Ν
MOV	MOV.W Rs, @aa:24	3				1	
	MOV.L #xx:32, ERd	3					
	MOV.L ERs, ERd	1					
	MOV.L @ERs, ERd	2				2	
	MOV.L @(d:16, ERs), ERd	3				2	
	MOV.L @(d:24, ERs), ERd	5				2	
	MOV.L @ERs+, ERd	2				2	2
	MOV.L @aa:16, ERd	3				2	
	MOV.L @aa:24, ERd	4				2	
	MOV.L ERs, @ERd	2				2	
	MOV.L ERs, @(d:16, ERd)	3				2	
	MOV.L ERs, @(d:24, ERd)	5				2	
	MOV.L ERs, @-ERd	2				2	2
	MOV.L ERs, @aa:16	3				2	
	MOV.L ERs, @aa:24	4				2	
MOVFPE	MOVFPE @aa:16, Rd* <sup>2</sup>	2			1		
MOVTPE	MOVTPE Rs, @aa:16*2	2			1		
MULXS	MULXS.B Rs, Rd	2					12
	MULXS.W Rs, ERd	2					20
MULXU	MULXU.B Rs, Rd	1					12
	MULXU.W Rs, ERd	1					20
NEG	NEG.B Rd	1					
	NEG.W Rd	1					
	NEG.L ERd	1					
NOP	NOP	1					
NOT	NOT.B Rd	1					
	NOT.W Rd	1					
	NOT.L ERd	1					
OR	OR.B #xx:8, Rd	1					
	OR.B Rs, Rd	1					
	OR.W #xx:16, Rd	2					
	OR.W Rs, Rd	1					
	OR.L #xx:32, ERd	3					
	OR.L ERs, ERd	2					

#### Appendix A Instruction Set

Instruction	Mnemonic		Instruction Fetch I	Stack Operation K	Word Data Access M	Internal Operation N
ORC	ORC #xx:8, CC	R	1			
POP	POP.W Rn		1		1	2
	POP.L ERn		2		2	2
PUSH	PUSH.W Rn		1		1	2
	PUSH.L ERn		2		2	2
ROTL	ROTL.B Rd		1			
	ROTL.W Rd		1			
	ROTL.L ERd		1			
ROTR	ROTR.B Rd		1			
	ROTR.W Rd		1			
	ROTR.L ERd		1			
ROTXL	ROTXL.B Rd		1			
	ROTXL.W Rd		1			
	ROTXL.L ERd		1			
ROTXR	ROTXR.B Rd		1			
	ROTXR.W Rd		1			
	ROTXR.L ERd		1			
RTE	RTE		2	2		2
RTS	RTS	Normal	2	1		2
		Advanced	2	2		2
SHAL	SHAL.B Rd		1			
	SHAL.W Rd		1			
	SHAL.L ERd		1			
SHAR	SHAR.B Rd		1			
	SHAR.W Rd		1			
	SHAR.L ERd		1			
SHLL	SHLL.B Rd		1			
	SHLL.W Rd		1			
	SHLL.L ERd		1			
SHLR	SHLR.B Rd		1			
	SHLR.W Rd		1			
	SHLR.L ERd		1			
SLEEP	SLEEP		1			

Instruction	Mnemonic		Instruction Fetch I	Branch Addr. Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
STC	STC CCR, Rd		1					
	STC CCR, @E	Rd	2				1	
	STC CCR, @(	d:16, ERd)	3				1	
	STC CCR, @(	d:24, ERd)	5				1	
	STC CCR, @-	-ERd	2				1	2
	STC CCR, @a	a:16	3				1	
	STC CCR, @a	a:24	4				1	
SUB	SUB.B Rs, Rd		1					
	SUB.W #xx:16, Rd		2					
	SUB.W Rs, Rd		1					
	SUB.L #xx:32,	ERd	3					
	SUB.L ERs, El	Rd	1					
SUBS	SUBS #1/2/4,	ERd	1					
SUBX	SUBX #xx:8, R	Rd	1					
	SUBX Rs, Rd		1					
TRAPA	TRAPA #x:2	Normal	2	1	2			4
		Advanced	2	2	2			4
XOR	XOR.B #xx:8,	Rd	1					
	XOR.B Rs, Rd		1					
	XOR.W #xx:16	6, Rd	2					
	XOR.W Rs, Ro	b	1					
	XOR.L #xx:32,	ERd	3					
	XOR.L ERs, E	Rd	2					
XORC	XORC #xx:8, 0	CCR	1					

Notes: 1. n is the value set in register R4L or R4. The source and destination are accessed n+1 times each.

2. Not used with this LSI.

# Appendix B Internal I/O Register Field

# B.1 Addresses

Address	Register	Data Bus				Bit	Names				_
(low)	Name	Width	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name
H'1C											
H'1D	_										
H'1E	_										
H'1F	_										
H'20				_					_	—	
H'21	—		—	_	_	—	—	—	_	_	_
H'22	_		_	_	_	_	_	_	_	_	_
H'23	—		—	_	_	—	—	—	_	_	_
H'24	_		_	_	_	_	_	_	_	_	_
H'25	_		_	_	_	_	_	_	_	_	_
H'26	_		_	_	_	_	_	_	_	_	_
H'27	_		_	_	_	_	_	_	_	_	
H'28	—		—	—	—	—	—	—	—	—	
H'29	_		_	_	_	_	_	_	_	_	
H'2A	_		_	_	_	_	_	_	—	_	
H'2B	_		_	_	_	_	_	_	—	_	
H'2C	—		—	—	—	—	—	—	—	—	
H'2D	—		—	—	—	—	—	_	—	_	
H'2E	_		_	_	_	_	_	_	_	_	_
H'2F	_		_	—	—	_	_	_	_	—	
H'30	—		—	—	—	—	—	—	—	—	
H'31	—		—	—	—	—	—	—	—	—	
H'32	_		_	—	—	_	_	_	_	_	_
H'33	—		_	—	_	_	—	—	—	—	_
H'34				—	_	_	_		_	—	_
H'35	_		_	—	—	_	_	_	_	_	_
H'36	—		—	—	_	_	—	—	—	—	
H'37	—		—	—	—	—	—	—	—	—	
H'38	_		_	_	_	_	_	_	—	_	
H'39	_		_	_	_	_	_	_	—	_	
H'3A	—		-	—	_	_	_		—	_	
H'3B	_		_	_	_	_	_	_	_	_	
H'3C	_		_	_	_	_	_	_	_	_	
H'3D	_		_	_	_	_	_	_	_	_	
H'3E	_		_	_	_	_	_	_	_	_	
H'3F	_		_	_	—	_	—	—	—	_	

		Data	Bit Names									
Address (low)	Register Name	Bus Width	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name	
H'40	FLMCR	8	FWE	SWE	ESU	PSU	EV	PV	E	Р	Flash memory	
H'41	—		—		—	—		—	—	—		
H'42	EBR	8	EB7	EB6	EB5	EB4	EB3	EB2	EB1	EB0	_	
H'43	_		_	_	_	_	_	_	_	_		
H'44	_		_	_	_	_	_	_	_	_		
H'45	—		—		—	—		—	—	—		
H'46	_		_	_	_	_	_	_	_	_		
H'47	RAMCR	8	—		—	_	RAMS	RAM2	RAM1	—		
H'48	_		-	_	_	_	_	_	_	_	_	
H'49	_		_	_	_	_	_	_	_	_	—	
H'4A	_		_	_	_	_	_	_	_	_	_	
H'4B	_		_	_	_	_	_	_	_	_	_	
H'4C	_		_	_	_	_	_	_	_	_	_	
H'4D	FLMSR	8	FLER	_	_	_	_	_	_	_	_	
H'4E	_		_	_	_	_	_	_	_	_	_	
H'4F	_		_	_	_	_	_	_	_	_	_	
H'50	_		_		_	_	_	_	_	_		
H'51	_		_		_	_	_	_	_	_	_	
H'52	_		_		_	_	_	_	_	_	_	
H'53	_		_		_	_	_	_	_	_	_	
H'54	_		_	_	_	_	_	_	_	_	_	
H'55	_		_		_	_	_	_	_	_	_	
H'56	_		_	_	_	_	_	_	_	_	_	
H'57	_		_	_	_	_	_	_	_	_	_	
H'58	_		_		_	_	_	_	_	_	_	
H'59	_		_		_	_	_	_	_	_	_	
H'5A	_		_		_	_	_	_	_	_	_	
H'5B	_		_	_	_	_	_	_	_	_	—	
H'5C	_		_		_	_	_	_	_	_	—	
H'5D	DIVCR	8	_	_	_	_	_	_	DIV1	DIV0	System control	
H'5E	MSTCR	8	PSTOP	_	MSTOF	5 MSTOF	P4 MSTOP	3 —	_	MSTOP	0	
H'5F	_		_	_	_	_	_	_	_	_		

Appendix B	Internal I/O Register Field
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		Data										
Address (low)	Register Name	Bus Width	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name	
H'60	TSTR	8	_	_	_	STR4	STR3	STR2	STR1	STR0	ITU	
H'61	TSNC	8	—	—	—	SYNC4	SYNC3	SYNC2	SYNC1	SYNC0	(all channels)	
H'62	TMDR	8		MDF	FDIR	PWM4	PWM3	PWM2	PWM1	PWM0	_	
H'63	TFCR	8	_	_	CMD1	CMD0	BFB4	BFA4	BFB3	BFA3	_	
H'64	TCR0	8	_	CCLR1	CCLR0	CKEG1	CKEG0	TPSC2	TPSC1	TPSC0	ITU	
H'65	TIOR0	8	_	IOB2	IOB1	IOB0	_	IOA2	IOA1	IOA0	channel 0	
H'66	TIER0	8	_	_		_	_	OVIE	IMIEB	IMIEA	_	
H'67	TSR0	8	_	_	_	_	_	OVF	IMFB	IMFA	_	
H'68	TCNT0H	16									-	
H'69	TCNT0L	_									_	
H'6A	GRA0H	16									-	
H'6B	GRA0L	_									-	
H'6C	GRB0H	16									-	
H'6D	GRB0L	_									-	
H'6E	TCR1	8	_	CCLR1	CCLR0	CKEG1	CKEG0	TPSC2	TPSC1	TPSC0	ITU	
H'6F	TIOR1	8	_	IOB2	IOB1	IOB0	_	IOA2	IOA1	IOA0	channel 1	
H'70	TIER1	8	_	_	_	_	_	OVIE	IMIEB	IMIEA	-	
H'71	TSR1	8	_	_	_	_	_	OVF	IMFB	IMFA	_	
H'72	TCNT1H	16									-	
H'73	TCNT1L	_									-	
H'74	GRA1H	16									-	
H'75	GRA1L	_									-	
H'76	GRB1H	16									-	
H'77	GRB1L	_									-	
H'78	TCR2	8	—	CCLR1	CCLR0	CKEG1	CKEG0	TPSC2	TPSC1	TPSC0	ITU channel 2	
H'79	TIOR2	8	_	IOB2	IOB1	IOB0	_	IOA2	IOA1	IOA0	_	
H'7A	TIER2	8	_	_	_	_	_	OVIE	IMIEB	IMIEA	_	
H'7B	TSR2	8	_	_	_	_	_	OVF	IMFB	IMFA	_	
H'7C	TCNT2H	16									_	
H'7D	TCNT2L	-									-	
H'7E	GRA2H	16									-	
H'7F	GRA2L	-									_	
H'80	GRB2H	16									-	
H'81	GRB2L	_									-	

A	Register	Data Bus				Bit N	lames				_
Address (low)	Name	Bus Width	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name
H'82	TCR3	8	_	CCLR1	CCLR0	CKEG1	CKEG0	TPSC2	TPSC1	TPSC0	ITU channel 3
H'83	TIOR3	8	_	IOB2	IOB1	IOB0	_	IOA2	IOA1	IOA0	_
H'84	TIER3	8	_	_	_	_	_	OVIE	IMIEB	IMIEA	_
H'85	TSR3	8	_	_	_	_	_	OVF	IMFB	IMFA	_
H'86	TCNT3H	16									-
H'87	TCNT3L	_									_
H'88	GRA3H	16									-
H'89	GRA3L	_									_
H'8A	GRB3H	16									_
H'8B	GRB3L	_									_
H'8C	BRA3H	16									_
H'8D	BRA3L	_									_
H'8E	BRB3H	16									-
H'8F	BRB3L	_									_
H'90	TOER	8	_	_	EXB4	EXA4	EB3	EB4	EA4	EA3	ITU
H'91	TOCR	8	—	_	—	XTGD	_	_	OLS4	OLS3	(all channel)
H'92	TCR4	8	_	CCLR1	CCLR0	CKEG1	CKEG0	TPSC2	TPSC1	TPSC0	ITU channel 4
H'93	TIOR4	8	_	IOB2	IOB1	IOB0	_	IOA2	IOA1	IOA0	_
H'94	TIER4	8	—	_	—	_	_	OVIE	IMIEB	IMIEA	_
H'95	TSR4	8	—	_	—	_	_	OVF	IMFB	IMFA	_
H'96	TCNT4H	16									_
H'97	TCNT4L	_									_
H'98	GRA4H	16									-
H'99	GRA4L	_									_
H'9A	GRB4H	16									-
H'9B	GRB4L	_									_
H'9C	BRA4H	16									-
H'9D	BRA4L	_									_
H'9E	BRB4H	16									
H'9F	BRB4L	_									

# Appendix B Internal I/O Register Field

Address	Register	Data Bus	Bit Names								
(low)	Name	Width	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name
H'A0	TPMR	8	_	_	_	_	G3NOV	G2NOV	G1NOV	G0NOV	TPC
H'A1	TPCR	8	G3CMS1	G3CMS0	G2CMS1	G2CMS0	G1CMS1	G1CMS0	G0CMS1	G0CMS0	
H'A2	NDERB	8	NDER15	NDER14	NDER13	NDER12	NDER11	NDER10	NDER9	NDER8	
H'A3	NDERA	8	NDER7	NDER6	NDER5	NDER4	NDER3	NDER2	NDER1	NDER0	
H'A4	NDRB* <sup>1</sup>	8	NDR15	NDR14	NDR13	NDR12	NDR11	NDR10	NDR9	NDR8	
		8	NDR15	NDR14	NDR13	NDR12	_	_	—	—	
H'A5	NDRA* <sup>1</sup>	8	NDR7	NDR6	NDR5	NDR4	NDR3	NDR2	NDR1	NDR0	
		8	NDR7	NDR6	NDR5	NDR4		—	_	—	
H'A6	NDRB* <sup>1</sup>	8	_	_	_	_	_	_	_	_	
		8	_	_	_	_	NDR11	NDR10	NDR9	NDR8	
H'A7	NDRA* <sup>1</sup>	8	_	_	_	_	_	_	_	_	
		8	_	_	_	_	NDR3	NDR2	NDR1	NDR0	
H'A8	TCSR* <sup>2</sup>	8	OVF	WT/IT	TME	_	_	CKS2	CKS1	CKS0	WDT
H'A9	TCNT* <sup>2</sup>	8									
H'AA	_		_	_	_	_	_	_	_	_	
H'AB	RSTCSR*2	8	WRST	RSTOE	_	_	_	_	_	_	
H'AC	_		_	_	_	_	_	_	_	_	
H'AD	_		_	_	_	_	_	_	_	_	
H'AE	_		_	_	_	_	_	_	_	_	
H'AF	_		_	_	_	_	_	_	_	_	
H'B0	SMR	8	C/A	CHR	PE	O/Ē	STOP	MP	CKS1	CKS0	SCI0
H'B1	BRR	8									
H'B2	SCR	8	TIE	RIE	TE	RE	MPIE	TEIE	CKE1	CKE0	
H'B3	TDR	8									
H'B4	SSR	8	TDRE	RDRF	ORER	FER	PER	TEND	MPB	MPBT	
H'B5	RDR	8									
H'B6	SCMR	8	_	_	_	_	SDIR	SINV	_	SMIF	Smart card interface
H'B7											
H'B8	SMR	8	C/Ā	CHR	PE	O/Ē	STOP	MP	CKS1	CKS0	SCI1
H'B9	BRR	8	_	_	_	_	_	_	_	_	
H'BA	SCR	8	TIE	RIE	TE	RE	MPIE	TEIE	CKE1	CKE0	
H'BB	TDR	8	_	_	_	_	_	_	_	_	
H'BC	SSR	8	TDRE	RDRF	ORER	FER	PER	TEND	MPB	MPBT	
H'BD	RDR	8	_	_	_	_	_	_	_	_	
H'BE	_		_	_	_	_	_	_	_	_	
H'BF	_		_	_	_	_	_	_	_	_	

A	Deviator	Data	Bit Names									
Address (low)	Register Name	Bus Width	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name	
H'C0	P1DDR	8	P1,DDR	$P1_{6}DDR$	P1₅DDR	$P1_4DDR$	P1 <sub>3</sub> DDR	$P1_2DDR$	$P1_1DDR$	P1₀DDR	Port 1	
H'C1	P2DDR	8	P2,DDR	$P2_{6}DDR$	$P2_5DDR$	$P2_4DDR$	$P2_{3}DDR$	$P2_2DDR$	P2,DDR	P2₀DDR	Port 2	
H'C2	P1DR	8	P1,	P1 <sub>6</sub>	P1₅	P1 <sub>4</sub>	P1 <sub>3</sub>	P1 <sub>2</sub>	P1,	P1 <sub>0</sub>	Port 1	
H'C3	P2DR	8	P2,	P2 <sub>6</sub>	P2₅	P2 <sub>4</sub>	P2 <sub>3</sub>	P2 <sub>2</sub>	P2,	P2 <sub>0</sub>	Port 2	
H'C4	P3DDR	8	P3,DDR	P3₅DDR	P3₅DDR	P3₄DDR	P3₃DDR	$P3_2DDR$	P3,DDR	P3₀DDR	Port 3	
H'C5	_			—	—	—	—	—	—	—		
H'C6	P3DR	8	P3,	P3 <sub>6</sub>	P3₅	P3 <sub>4</sub>	P3 <sub>3</sub>	P3 <sub>2</sub>	P3,	P3 <sub>0</sub>	Port 3	
H'C7	_			—	—	—	—	—	—	—		
H'C8	P5DDR	8		—	—	—	$P5_{3}DDR$	$P5_2DDR$	P5,DDR	P5₀DDR	Port 5	
H'C9	P6DDR	8	_	_	P6₅DDR	$P6_4DDR$	P6₃DDR	_	_	P6₀DDR	Port 6	
H'CA	P5DR	8	_	_	_	_	P5 <sub>3</sub>	P5 <sub>2</sub>	P5,	P5 <sub>0</sub>	Port 5	
H'CB	P6DR	8	_	_	P6₅	P6 <sub>4</sub>	P6 <sub>3</sub>	_	_	P6 <sub>0</sub>	Port 6	
H'CC	_		_	_	_	_	_	_	_	_		
H'CD	P8DDR	8	_	_	_	_	_	_	P8,DDR	P8₀DDR	Port 8	
H'CE	P7DR	8	P7,	P7 <sub>6</sub>	P7₅	P7 <sub>4</sub>	P7 <sub>3</sub>	P7 <sub>2</sub>	P7 <sub>1</sub>	P7 <sub>0</sub>	Port 7	
H'CF	P8DR	8	_	_	_	_	_	_	P8,	P8 <sub>0</sub>	Port 8	
H'D0	P9DDR	8	_	_	P9₅DDR	P9₄DDR	P9₃DDR	P9 <sub>2</sub> DDR	P9₁DDR	P9₀DDR	Port 9	
H'D1	PADDR	8	PA, DDR	PA₀DDR	PA₅DDR	PA₄DDR	PA <sub>3</sub> DDR	PA <sub>2</sub> DDR	PA,DDR	PA₀DDR	Port A	
H'D2	P9DR	8	_	_	P9₅	P9 <sub>4</sub>	P9 <sub>3</sub>	P9 <sub>2</sub>	P9,	P9 <sub>0</sub>	Port 9	
H'D3	PADR	8	PA <sub>7</sub>	PA <sub>6</sub>	PA₅	$PA_4$	$PA_3$	PA <sub>2</sub>	PA <sub>1</sub>	PA₀	Port A	
H'D4	PBDDR	8	PB,DDR	_	PB₅DDR	PB₄DDR	PB <sub>3</sub> DDR	$PB_2DDR$	PB,DDR	PB₀DDR	Port B	
H'D5	_		_	_	_	_	_	_	_	_		
H'D6	PBDR	8	PB <sub>7</sub>	_	PB₅	$PB_4$	PB <sub>3</sub>	PB <sub>2</sub>	PB <sub>1</sub>	PB₀	Port B	
H'D7	_		_	_	_	_	_	_	_	_		
H'D8	P2PCR	8	P2,PCR	P2₅PCR	P2₅PCR	P2₄PCR	P2 <sub>3</sub> PCR	$P2_2PCR$	P2,PCR	P2₀PCR	Port 2	
H'D9	_		_	_	_	_	_	_	_	_		
H'DA	_		_	_	_	_	_	_	_	_		
H'DB	P5PCR	8	_	_	_	_	P5 <sub>3</sub> PCR	P5 <sub>2</sub> PCR	P5,PCR	P5₀PCR	Port 5	
H'DC	_		_	_	_	_	_	_	_	_		
H'DD	_		_	_	_	_	_	_	_	_	-	
H'DE	_		_	_	_	_	_	_	_	_	-	
H'DF	_		_	_	_	_	_	_	_	_	-	

#### Appendix B Internal I/O Register Field

A	Deviator	Data				Bit N	lames				_
Address (low)	Register Name	Bus Width	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name
H'E0	ADDRAH	8	AD9	AD8	AD7	AD6	AD5	AD4	AD3	AD2	A/D
H'E1	ADDRAL	8	AD1	AD0	_	_	_	_	_	_	-
H'E2	ADDRBH	8	AD9	AD8	AD7	AD6	AD5	AD4	AD3	AD2	-
H'E3	ADDRBL	8	AD1	AD0	_	_	_	_	_	_	-
H'E4	ADDRCH	8	AD9	AD8	AD7	AD6	AD5	AD4	AD3	AD2	-
H'E5	ADDRCL	8	AD1	AD0	_	_	_	_	_	_	-
H'E6	ADDRDH	8	AD9	AD8	AD7	AD6	AD5	AD4	AD3	AD2	-
H'E7	ADDRDL	8	AD1	AD0	_	_	_	_	_	_	-
H'E8	ADCSR	8	ADF	ADIE	ADST	SCAN	CKS	CH2	CH1	CH0	-
H'E9	ADCR	8	TRGE	_	_	_	_	_	_	_	-
H'EA	_		_	_	_	_	_	_	_	_	
H'EB	—		—	—	—		—	—	—		-
H'EC	—		—	—	—		—	—	—		Bus controller
H'ED	ASTCR	8	AST7	AST6	AST5	AST4	AST3	AST2	AST1	AST0	-
H'EE	WCR	8	_	_	_	_	WMS1	WMS0	WC1	WC0	-
H'EF	WCER	8	WCE7	WCE6	WCE5	WCE4	WCE3	WCE2	WCE1	WCE0	_
H'F0	_		_	_	_	_	_	_	_	_	
H'F1	MDCR	8	_	_	_	_	_	MDS2	MDS1	MDS0	System control
H'F2	SYSCR	8	SSBY	STS2	STS1	STS0	UE	NMIEG	_	RAME	-
H'F3	ADRCR	8	$A_{23}E$	$A_{22}E$	A <sub>21</sub> E	_	_	_	_	_	Bus controller
H'F4	ISCR	8	_	—	IRQ5SC	IRQ4SC	_	—	IRQ1SC	IRQ0SC	Interrupt controller
H'F5	IER	8	_	_	IRQ5E	IRQ4E	_	_	IRQ1E	IRQ0E	-
H'F6	ISR	8	_	_	IRQ5F	IRQ4F	_	_	IRQ1F	IRQ0F	-
H'F7	_		_	_	_	_	_	_	_	_	-
H'F8	IPRA	8	IPRA7	IPRA6	_	IPRA4	IPRA3	IPRA2	IPRA1	IPRA0	-
H'F9	IPRB	8	IPRB7	IPRB6	_	_	IPRB3	IPRB2	IPRB1	_	-
H'FA	_		_	_	_	_	_	_	_	_	
H'FB	_		_	_	_	_	_	_	_	_	-
H'FC											-
H'FD	_		_	_	_	_	_	_	_	_	_
H'FE	_		_	_	_	_	_	_	_	_	-
H'FF	—		_	_	_	_	_	_	_	_	-

Legend:

ITU: 16-bit integrated timer unit

TPC: Programmable timing pattern controller

WDT: Watchdog timer

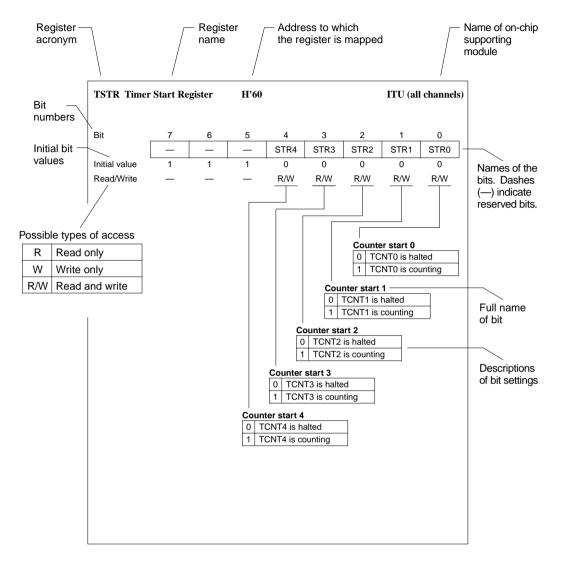
SCI: Serial communication interface

A/D: A/D converter

Notes: 1. The address depends on the output trigger setting.

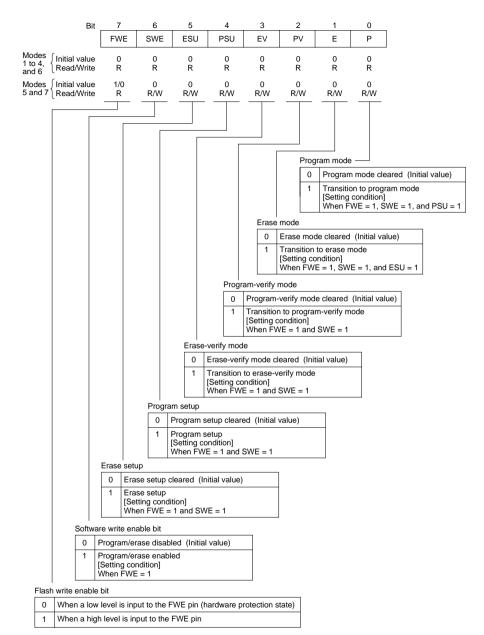
2. For write access to TCSR, TCNT, and RSTCSR, see section 10.2.4, Notes on Register Access.

# **B.2** Function



#### FLMCR—Flash Memory Control Register

H'40



Note: This register is used only in the flash memory versions.

Reading the corresponding address in a mask ROM version will always return 1s, and writes to this address are disabled. Fix the FWE pin low in mode 6.

EBR—Era	ase Block R	egister			H'42		Flash memory			
	Bit	7	6	5	4	3	2	1	0	1
		EB7	EB6	EB5	EB4	EB3	EB2	EB1	EB0	
Modes 1 to 4, and 6	Initial value Read/Write	0 R	0 R	0 R	0 R	0 R	0 R	0 R	0 R	_
Modes   5 to 7	Initial value Read/Write	0 R/W	0 R/W	0 R/W	0 R/W	0 R/W	0 R/W	0 R/W	0 R/W	

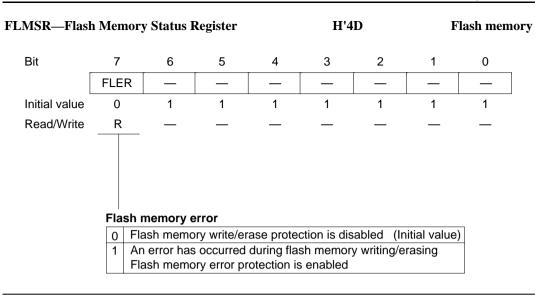
#### Block 7 to 0

0	Block EB7 to EB0 is not selected (Initial value)
1	Block EB7 to EB0 is selected

Note: When not erasing flash memory, EBR should be cleared to H'00. This register is used only in the flash memory versions. Reading the corresponding address in a mask ROM version will always return 1s, and writes to this address are disabled. 1s cannot be set in this register in mode 6.

RAMCR—RAM Cont	trol Re	gister				H'47		Flash Memory		
Bit	7	6	6	5	4	3	2		1	0
	—	-	-	—	—	RAMS	RAM	12 F	RAM1	—
Modes ∫ Initial value	1		1	1	1	0	0		0	1
1 to 4 CRead/Write	—	-	_	—	—	R	R	R F		_
Modes [ Initial value	1		1	1	1	0	0		0	1
5 to 7 Read/Write						R/W*	R/W	/*	R/W*	- -
		R	eserve	ed bits						
			C3C1 VC							
		RAM se	elect, R	AM2, R	AM1 ——					
		Bit 3	Bit 2	Bit 1	RAM	Area		PAM	Emulatio	on Status
		RAMS	RAM2	RAM1		Alea			Linulatio	in Status
		0	0/1	0/1	H'FFF000	to H'FFF3	FF	No er	nulation	
		1	0	0	H'000000	) to H'0003	F	Марр	ing RAM	
				1	H'000400	FF				
			1	0	H'000800	FF				
				1	H'000C00 to H'000FFF					

- Note: This register is used only in the flash memory versions. Reading the corresponding address in a mask ROM version will always return 1s, and writes to this address are disabled.
  - \* In mode 6 (single-chip normal mode), flash memory emulation by RAM is not supported; these bits can be modified, but must not be set to 1.



DIVCR—Divis	ion Contr	ol Registe	er		H'5I	)	System control			
Bit	7	6	5	7	3	2	1	0	_	
		_	—	_		—	DIV1	DIV0		
Initial value	1	1	1	1	1	1	0	0	1	
Read/Write	_	—	—	_	—	—	R/W	R/W		

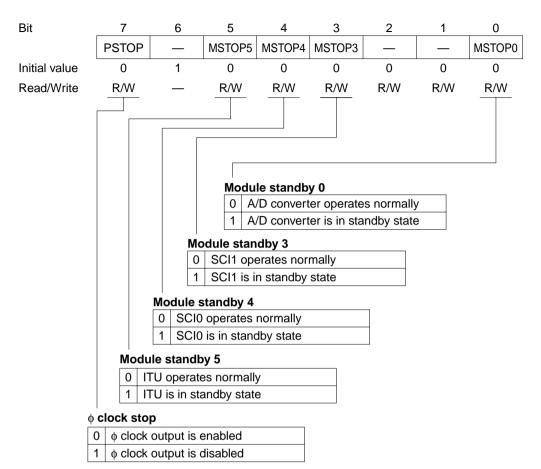
#### Divide bits 1 and 0

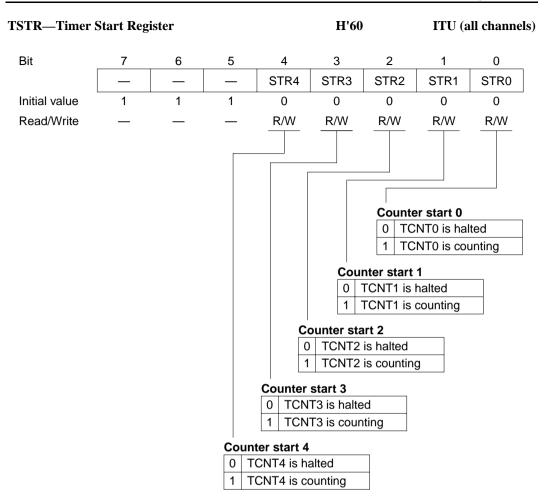
Bit 1	Bit 0	Frequency
DIV1	DIV0	Division Ratio
0	0	1/1initial value
	1	1/2
1	0	1/4
	1	1/8

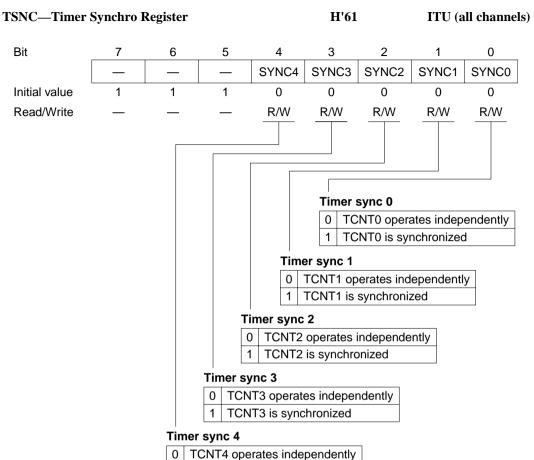
#### MSTCR—Module Standby Control Register



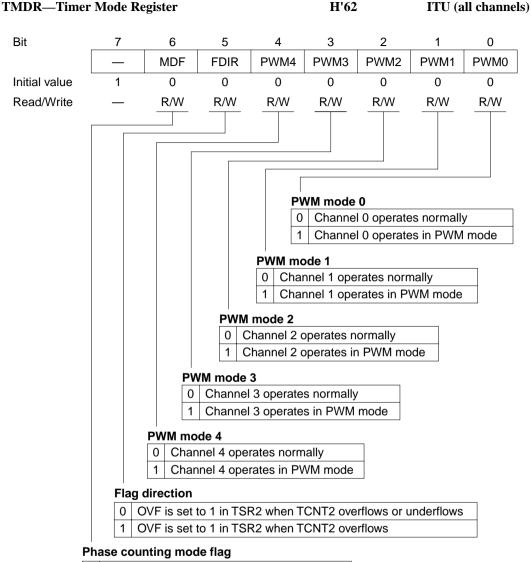
#### System control







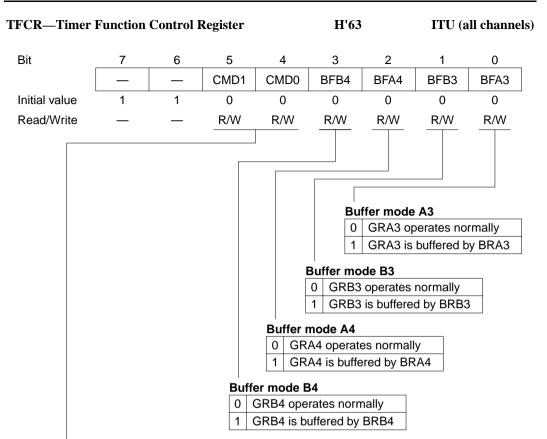
0	TCN14 operates independent
1	TCNT4 operates independent TCNT4 is synchronized



0 Channel 2 operates normally
-------------------------------

1 Channel 2 operates in phase counting mode

### Renesas



#### Combination mode 1 and 0

Bit 5	Bit 4	Operating Made of Channels 2 and 4
CMD1	CMD0	Operating Mode of Channels 3 and 4
0	0	Channels 3 and 4 operate normally
	1	
1	0	Channels 3 and 4 operate together in complementary PWM mode
	1	Channels 3 and 4 operate together in reset-synchronized PWM mode

TCR0—Timer Control Register 0

### H'64

ITU0

Bit	7	6		5	4	3	2	1	0				
	_	CCLF	אן c	CLR0	CKEG1	CKEG0	TPSC2	TPSC1	TPSC0				
Initial value	1	0		0	0	0	0	0	0				
Read/Write	_	R/W	V	R/W	R/W	R/W	R/W	R/W	R/W				
		Γ											
		Ti	mer p	rescale	r 2 to 0								
			Bit 2	Bit 1	Bit 0	TONI							
		TF	PSC2	TPSC1	TPSC0	ICN	Clock So	urce					
			0	0	0	Internal clo	ock: ø		7				
					1	Internal clo	ock:						
				1	0	Internal clock: φ/4							
					1	Internal clo	'						
			1	0	0	External cl	ock A: TCI	LKA input					
			1 External clock B: TCLKB										
				1	0	External cl							
					1	External cl	ock D: TC	LKD input					
	c	lock e	dge 1	and 0									
		Bit 4	Bit 3	Cou	untod Ed	non of Extor	nal Clask						
	C	KEG1	CKEG	i0 Cou	intea Eaé	ges of Exter	nal Clock						
		0	0	Risir	ng edges	counted							
			1			counted							
		1		Both	edges c	ounted							
	Coun	ter clea	ar 1 ai	nd 0									
	Bit 6						ear Source	<b>`</b>					
	CCLF	R1 CCL	R0					<del>,</del>					
	0	0	Т	CNT is	not clear	ed							
		1				oy GRA com							
	1	0	Т	CNT is	cleared b	oy GRB com	npare mate	ch or input	capture				

Synchronous clear: TCNT is cleared in synchronization with other synchronized timers

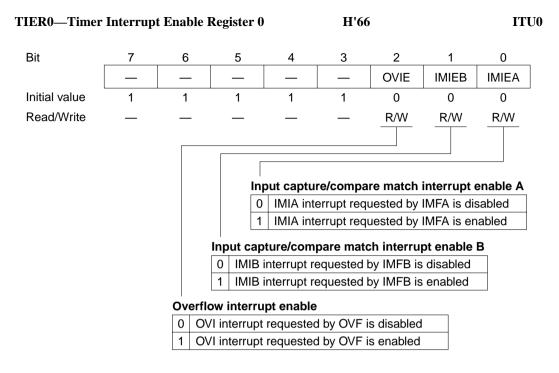
1

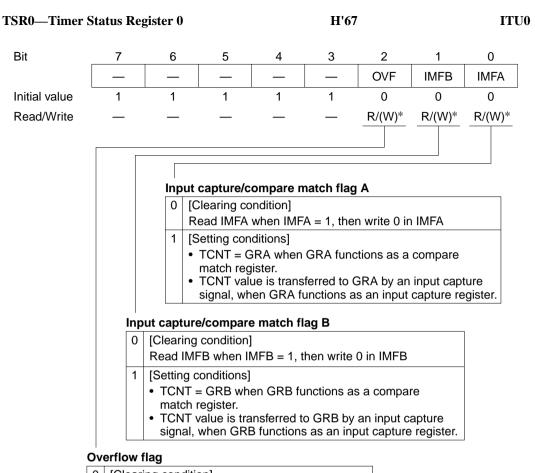
TIOR0—Tiu	ner I/(	) Cont	rol Reg	ister 0		H'65	;		П	CU0	
Bit	Bit 7		6	5	4	3	2	1	0	_	
		_	IOB2	IOB1	IOB0	—	IOA2	IOA1	IOA0		
Initial value	•	1	0	0	0	1	0	0	0	_	
Read/Write	;		R/W	R/W	R/W	_	R/W	R/W	R/W		
		ntrol A2	to A0								
	Bit 2	Bit 1	Bit 0							]	
	IOA2	IOA1	IOA0			GRA Fun	iction				
	0	0	0	GRA is an output No output at compare match							
	1 compare			compare re	egister	0 output at GRA compare match					
		4	0			1 output of		noro moto	h	1	

#### 1 output at GRA compare match 1 0 Output toggles at GRA compare match 1 GRA captures rising edge of input 0 GRA is an input 1 0 capture register GRA captures falling edge of input 1 GRA captures both edges of input 1 0 1

### I/O control B2 to B0

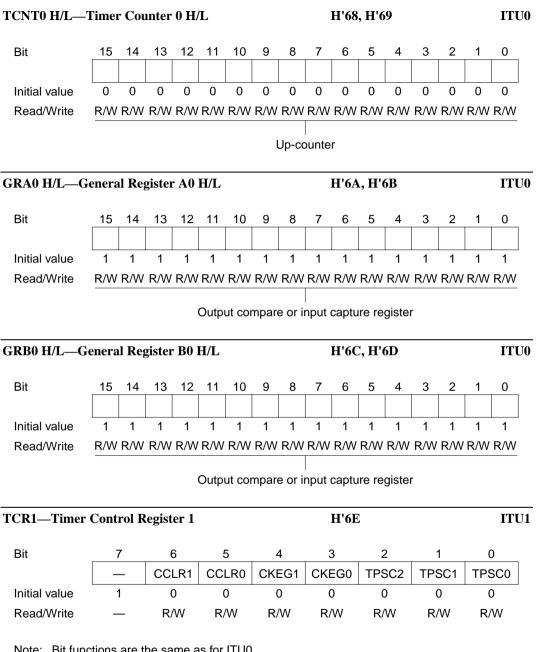
Bit 6	Bit 5	Bit 4	GRB Function								
IOB2	IOB1	IOB0		GRB Function							
0	0	0	GRB is an output	No output at compare match							
		1	compare register	0 output at GRB compare match							
	1	0		1 output at GRB compare match							
		1		Output toggles at GRB compare match							
1	0	0	GRB is an input	GRB captures rising edge of input							
		1	capture register	GRB captures falling edge of input							
	1	0		GRB captures both edges of input							
		1									





0	[Clearing condition] Read OVF when OVF = 1, then write 0 in OVF
1	[Setting condition] TCNT overflowed from H'FFFF to H'0000

Note: \* Only 0 can be written to clear the flag.



Note: Bit functions are the same as for ITU0.

### Renesas

IOR1—Time	r I/O Cont	rol Regis	ter 1		Н'6	F		ITU
Bit	7	6	5	4	3	2	1	0
	_	IOB2	IOB1	IOB0	—	IOA2	IOA1	IOA0
Initial value	1	0	0	0	1	0	0	0
Read/Write	—	R/W	R/W	R/W	—	R/W	R/W	R/W
Note: Bit fund	ctions are t	he same a	as for ITU0	).				
TIER1—Time	· Interrup	t Enable l	Register 1		H'7	0		ITU
Bit	7	6	5	4	3	2	1	0
	_	—	—	—	—	OVIE	IMIEB	IMIEA
Initial value	1	1	1	1	1	0	0	0
Read/Write	_	_	_	_	_	R/W	R/W	R/W
Bit	7	6	5	4	3	2	1	0
Bit		6	5	4	3	OVF	IMFB	IMFA
Initial value	1	1	1	1	1	0	0	0
Read/Write	—	—	—	—	—	R/(W)*	R/(W)*	R/(W)*
Notes: Bit fur * On	nctions are ly 0 can be							
CNT1 H/L—'	Timer Cou	inter 1 H/	L		H'7	2, H'73		ITU
Bit	15 14	13 12	11 10	98	7 6	5 4	3 2	1 0
Initial value	0 0	0 0	0 0	0 0	0 0	0 0	0 0	0 0
Read/Write	R/W R/W	R/W R/W	R/W R/W	R/W R/W	R/W R/V	V R/W R/W	R/W R/W	R/W R/W

GRA1 H/L—G	enera	al Re	giste	r A1	H/L					H'74	l, H'7	5				ITU1
Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Initial value	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
Read/Write	R/W	R/W	R/W	R/W	/ R/W	R/W	/ R/W	R/W	/ R/W	R/W	R/W	R/W	R/W	R/W	/ R/W	R/W
Note: Bit functions are the same as for ITU0.																
GRB1 H/L—G	enera	l Re	gister	r B1	H/L			H'76, H'77						ITU1		
Bit	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
Initial value	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1	1
Read/Write	R/W	R/W	R/W	R/W	/ R/W	R/W	/ R/W	R/W	/ R/W	R/W	R/W	R/W	R/W	R/W	/ R/W	R/W
Note: Bit fund	ctions	are t	he sa	ame a	as for	ITUC	).									
TCR2—Timer	Cont	rol R	Regist	ter 2						H'78	3					ITU2
Bit	7		6	6	5	5	4	ļ	З	3	2		1		C	)
	_	-	CCL	_R1	CCL	R0	CKE	G1	CKE	G0	TPS	C2	TPS	SC1	TPS	SC0
Initial value	1		C	)	0	)	C	)	C	)	0	)	C	)	C	)

Notes: 1. Bit functions are the same as for ITU0.

R/W

R/W

Read/Write

2. When channel 2 is used in phase counting mode, the counter clock source selection by bits CKEG1, CKEG0 and TPSC2 to TPSC0 is ignored.

R/W

R/W

R/W

R/W

R/W

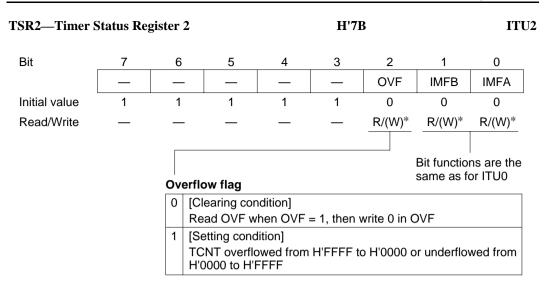
TIOR2—Timer	I/O Cont	rol Regist	ter 2		H'79						
Bit	7	6	5	4	3	2	1	0	-		
	_	IOB2	IOB1	IOB0	—	IOA2	IOA1	IOA0			
Initial value	1	0	0	0	1	0	0	0	-		
Read/Write	—	R/W	R/W	R/W	—	R/W	R/W	R/W			

Notes: 1. Bit functions are the same as for ITU0.

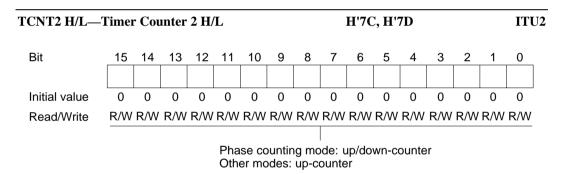
2. Channel 2 does not have a compare match toggle output function. If this setting is used, 1 output will be selected automatically.

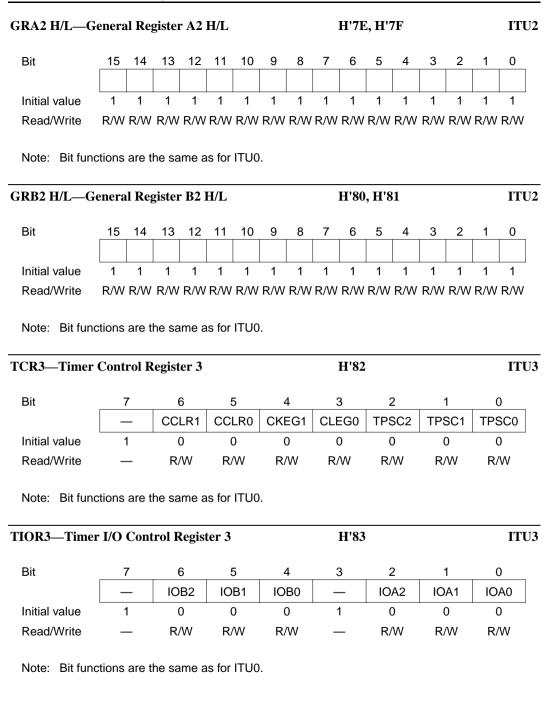
TIER2—Timer	Interrup	t Enable I	Register 2		<b>H'7</b> A	A		IT	U2
Bit	7	6	5	4	3	2	1	0	_
	—	—	—	_	_	OVIE	IMIEB	IMIEA	
Initial value	1	1	1	1	1	0	0	0	
Read/Write	—	—	—	—	—	R/W	R/W	R/W	

Note: Bit functions are the same as for ITU0.



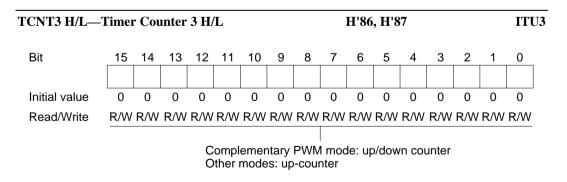
Note: \* Only 0 can be written to clear the flag.

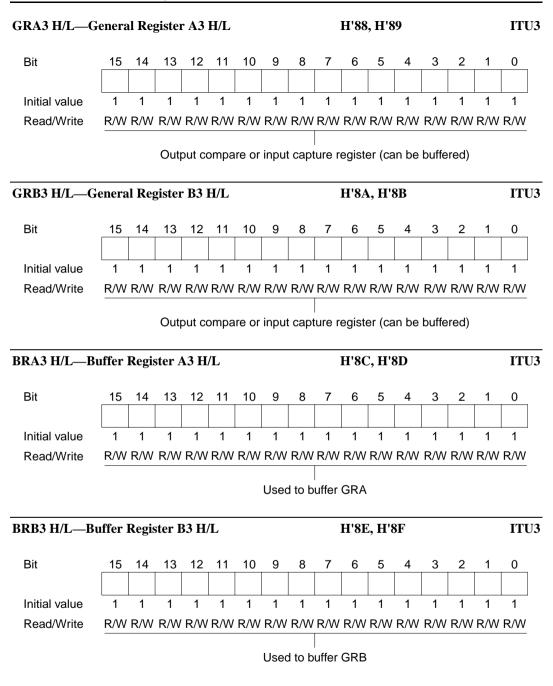


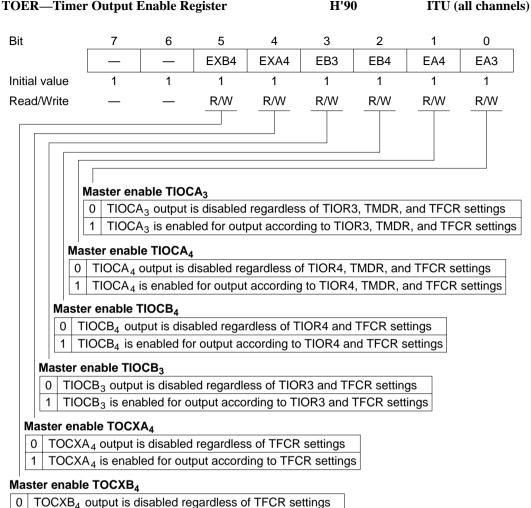


TIER3—Timer	Interrup	t Enable ]	Register 3	i	<b>H'8</b> 4	l I		ΙΤ
Bit	7	6	5	4	3	2	1	0
				_		OVIE	IMIEB	IMIEA
Initial value	1	1	1	1	1	0	0	0
Read/Write	—	_	—	—	—	R/W	R/W	R/W
Note: Bit func	tions are t	he same a	as for ITU0	).				
TSR3—Timer S	Status Reg	gister 3			H'85	5		ITU
Bit	7	6	5	4	3	2	1	0
	_		_		_	OVF	IMFB	IMFA
Initial value	1	1	1	1	1	0	0	0
Read/Write	—		—	—	_	R/(W)*	R/(W)*	R/(W)*
		Overflo	ow flag				Bit function same as for	
			earing cor	ndition]				
		Re	ad OVF w	hen OVF =	= 1, then v	vrite 0 in 0	DVF	
			etting cond	-				
			NT overflo		H'FFFF to	o H'0000 d	or underflow	wed from

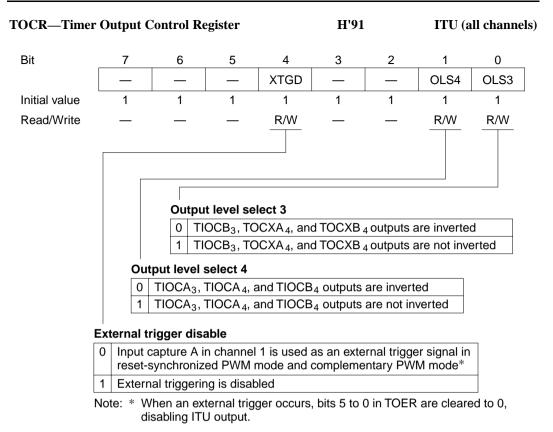
Note: \* Only 0 can be written to clear the flag.







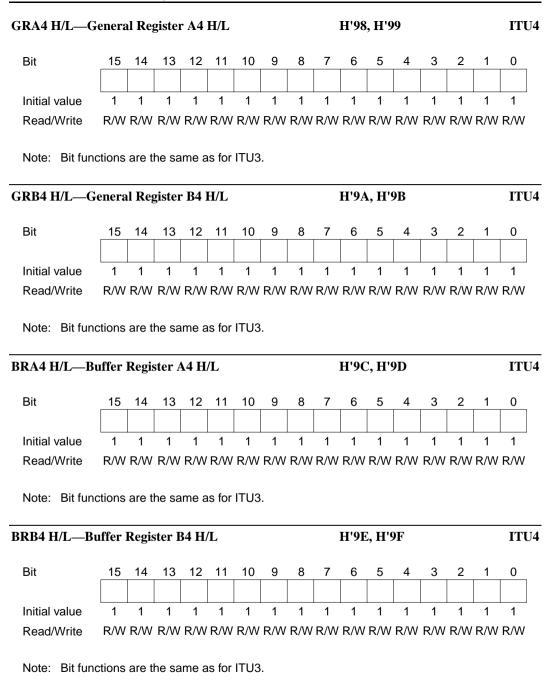
1 TOCXB₄ is enabled for output according to TFCR settings

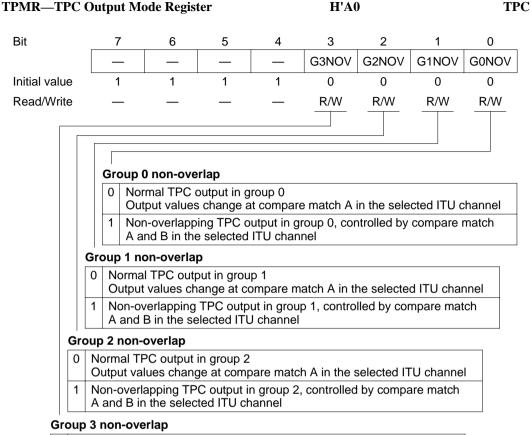


TCR4—Timer Control Register 4				ITU	J <b>4</b>				
Bit	7	6	5	4	3	2	1	0	
	—	CCLR1	CCLR0	CKEG1	CKEG0	TPSC2	TPSC1	TPSC0	
Initial value	1	0	0	0	0	0	0	0	
Read/Write	—	R/W	R/W	R/W	R/W	R/W	R/W	R/W	

Note: Bit functions are the same as for ITU0.

TIOR4—Timer I/O Control Register 4					H'93				
Bit	7	6	5	4	3	2	1	0	
		IOB2	IOB1	IOB0	_	IOA2	IOA1	IOA0	
Initial value	1	0	0	0	1	0	0	0	
Read/Write	—	R/W	R/W	R/W	—	R/W	R/W	R/W	
Note: Bit func	tions are t	he same a	as for ITUC	).					
TIER4—Timer	Interrup	t Enable 1	Register 4	ļ	H'9	4		ITU4	
Bit	7	6	5	4	3	2	1	0	
		_	_	_	_	OVIE	IMIEB	IMIEA	
Initial value	1	1	1	1	1	0	0	0	
Read/Write	—	—	—	—	—	R/W	R/W	R/W	
Note: Bit func					Н'9	5		ITU4	
Bit	7	6	5	4	3	2	1	0	
	_		_		_	OVF	IMFB	IMFA	
Initial value	1	1	1	1	1	0	0	0	
Read/Write	—	_	—	_	—	R/(W)*	R/(W)*	R/(W)*	
	0 can be	written to	clear the f						
TCNT4 H/L—1	Timer Co	unter 4 H	/L		Н'9	6, H'97		ITU4	
Bit	15 14	13 12	11 10	98	7 6	5 4	3 2	1 0	
Initial value	0 0	0 0	0 0	0 0	0 0	0 0	0 0	0 0	
Read/Write						V R/W R/W			
Note: Bit func									





0	Normal TPC output in group 3 Output values change at compare match A in the selected ITU channel
1	Non-overlapping TPC output in group 3, controlled by compare match A and B in the selected ITU channel

PCR—TPC Output Control Register					TPC			
Bit	7	6	5	4	3	2	1	0
	G3CMS	1 G3CMS0	G2CMS1	G2CMS0	G1CMS1	G1CMS0	G0CMS1	G0CMS0
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Group 0 c	ompare ma	tch select 1	and 0					
Bit 1 G0CMS1	Bit 0 G0CMS0	ITU Channel Selected as Output Trigger						

11			ITU Channel Selected as Output Trigger
	G0CMS1	G0CMS0	ITU Channel Selected as Output Trigger
	0	0	TPC output group 0 (TP <sub>3</sub> to TP <sub>0</sub> ) is triggered by compare match in ITU channel 0
		1	TPC output group 0 (TP <sub>3</sub> to TP <sub>0</sub> ) is triggered by compare match in ITU channel 1
	1	0	TPC output group 0 (TP <sub>3</sub> to TP <sub>0</sub> ) is triggered by compare match in ITU channel 2
		1	TPC output group 0 (TP <sub>3</sub> to TP <sub>0</sub> ) is triggered by compare match in ITU channel 3

### Group 1 compare match select 1 and 0

Bit 2					
G1CMS0	ITU Channel Selected as Output Trigger				
0	TPC output group 1 (TP <sub>7</sub> to TP <sub>4</sub> ) is triggered by compare match in ITU channel 0				
1	TPC output group 1 (TP <sub>7</sub> to TP <sub>4</sub> ) is triggered by compare match in ITU channel 1				
0	TPC output group 1 (TP <sub>7</sub> to TP <sub>4</sub> ) is triggered by compare match in ITU channel 2				
1	TPC output group 1 (TP <sub>7</sub> to TP <sub>4</sub> ) is triggered by compare match in ITU channel 3				
	G1CMS0 0 1				

#### Group 2 compare match select 1 and 0

Bit 4	ITU Channel Selected as Output Trigger					
G2CMS0	The original deletted as output migger					
0	TPC output group 2 (TP <sub>11</sub> to TP <sub>8</sub> ) is triggered by compare match in ITU channel 0					
1	TPC output group 2 (TP <sub>11</sub> to TP <sub>8</sub> ) is triggered by compare match in ITU channel 1					
0	TPC output group 2 (TP <sub>11</sub> to TP <sub>8</sub> ) is triggered by compare match in ITU channel 2					
1	TPC output group 2 (TP $_{11}$ to TP $_8$ ) is triggered by compare match in ITU channel 3					
	G2CMS0 0 1					

# Group 3 compare match select 1 and 0

Bit 7	Bit 6	ITH Channel Selected on Output Trigger					
G3CMS1	G3CMS0	ITU Channel Selected as Output Trigger					
0	0	TPC output group 3 (TP <sub>15</sub> to TP <sub>12</sub> )* is triggered by compare match in ITU channel 0					
	1	TPC output group 3 (TP <sub>15</sub> to TP <sub>12</sub> )* is triggered by compare match in ITU channel 1					
1	0	TPC output group 3 (TP <sub>15</sub> to TP <sub>12</sub> )* is triggered by compare match in ITU channel 2					
	1	TPC output group 3 (TP <sub>15</sub> to TP <sub>12</sub> )* is triggered by compare match in ITU channel 3					

Note: \* Since this LSI does not have a  $\rm TP_{14}$  pin, the  $\rm TP_{14}$  signal cannot be output off-chip.

NDERB—Next	Data Ena	ble Regist	ter B	B H'A2							
Bit	7	6	5	4	3	2	1	0			
	NDER15	NDER14	NDER13	NDER12	NDER11	NDER10	NDER9	NDER8			
Initial value	0	0	0	0	0	0	0	0			
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W			
	Next data enable 15 to 8 Bits 7 to 0										
	NDER1	5 to NDER	8		Descrip	otion					
		0		PC outputs TP <sub>15</sub> to TP <sub>8</sub> * are disabled DR15 to NDR8 are not transferred to PB <sub>7</sub> to PE							
		1		outputs TP 15 to NDR	<sub>0</sub> )						
	Note: *	Since this output off		s not have	a TP <sub>14</sub> pir	n, the TP <sub>14</sub>	signal car	nnot be			
NDERA—Next	t Data Ena	ble Regist	ter A		H'A	3		TPC			
Bit	7	6	5	4	3	2	1	0			
	NDER7	NDER6	NDER5	NDER4	NDER3	NDER2	NDER1	NDER0			
Initial value	0	0	0	0	0	0	0	0			
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W			
	Next da	ta enable	7 to 0								

Bits 7 to 0	Description					
NDER7 to NDER0	Description					
0	TPC outputs TP $_7$ to TP $_0$ are disabled (NDR7 to NDR0 are not transferred to PA $_7$ to PA $_0$ )					
1	TPC outputs TP <sub>7</sub> to TP <sub>0</sub> are enabled (NDR7 to NDR0 are transferred to PA <sub>7</sub> to PA <sub>0</sub> )					

#### NDRB-Next Data Register B

#### Same output trigger for TPC output groups 2 and 3 ٠ Address H'FFA4

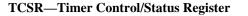
Bit	7	6	5	4	3	2	1	0			
	NDR15	NDR14	NDR13	NDR12	NDR11	NDR10	NDR9	NDR8			
Initial value	0	0	0	0	0	0	0	0			
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W			
		Next outpu TPC outpu		*		Next outp TPC outp					
Address H'FFA	46										
Bit	7	6	5	4	3	2	1	0			
	_	_	_	_			_	_			
Initial value	1	1	1	1	1	1	1	1			
Read/Write	—	—	—	—	—	—	—	—			
Different outpu Address H'FFA Bit		6	5	4	3	2	1	0			
	NDR15	NDR14	NDR13	NDR12	—		—	—			
Initial value	0	0	0	0	1	1	1	1			
Read/Write	R/W	R/W	R/W	R/W	—	—	—	—			
		Next outp									
Address H'FFA	46										
Bit	7	6	5	4	3	2	1	0			
	—	—	—	—	NDR11	NDR10	NDR9	NDR8			
Initial value	1	1	1	1	0	0	0	0			
Read/Write	—	—	—	—	R/W	R/W	R/W	R/W			
	— — — <u>R/W R/W R/W R/W</u> Next output data for TPC output group 2										

Since this LSI does not have a  $TP_{_{14}}$  pin, the  $TP_{_{14}}$  signal cannot be output off-chip. Note: \*

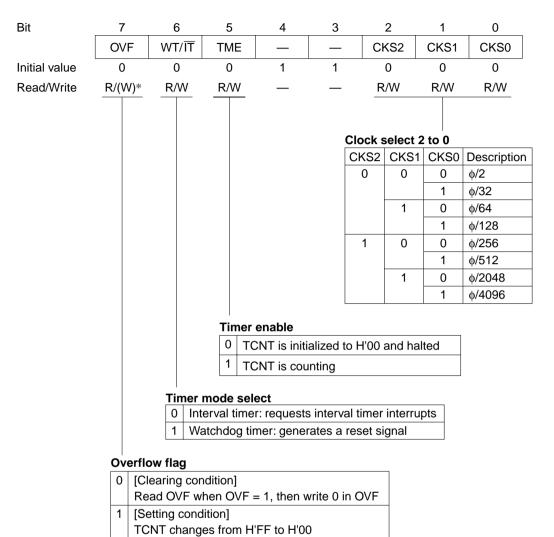
## RENESAS

H'A4/H'A6

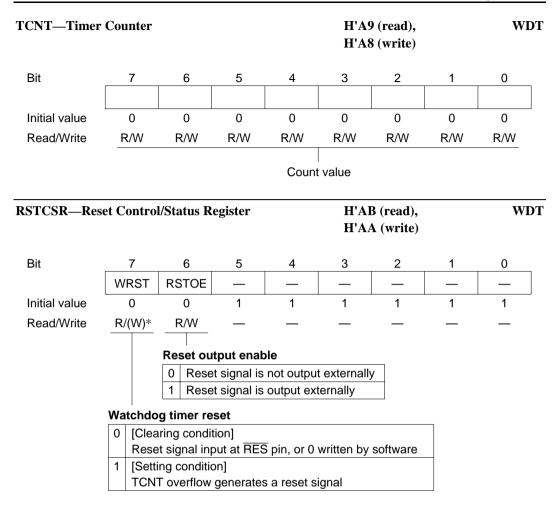
#### NDRA—Next Data Register A H'A5/H'A7 TPC • Same output trigger for TPC output groups 0 and 1 Address H'FFA5 Bit 7 6 5 4 3 2 1 0 NDR5 NDR7 NDR6 NDR4 NDR3 NDR2 NDR1 NDR0 Initial value 0 0 0 0 0 0 0 0 Read/Write R/W R/W R/W R/W R/W R/W R/W R/W Next output data for Next output data for TPC output group 1 TPC output group 0 Address H'FFA7 Bit 7 3 2 0 6 5 4 1 \_\_\_\_ 1 Initial value 1 1 1 1 1 1 1 Read/Write Different output triggers for TPC output groups 0 and 1 • Address H'FFA5 Bit 7 6 5 4 3 2 1 0 NDR7 NDR6 NDR5 NDR4 1 0 0 Initial value 0 0 1 1 1 Read/Write R/W R/W R/W R/W Next output data for TPC output group 1 Address H'FFA7 Bit 7 3 2 0 6 5 4 1 NDR3 NDR2 NDR1 NDR0 \_ \_\_\_\_ Initial value 1 1 1 1 0 0 0 0 Read/Write R/W R/W R/W R/W Next output data for TPC output group 0







Note: \* Only 0 can be written to clear the flag.

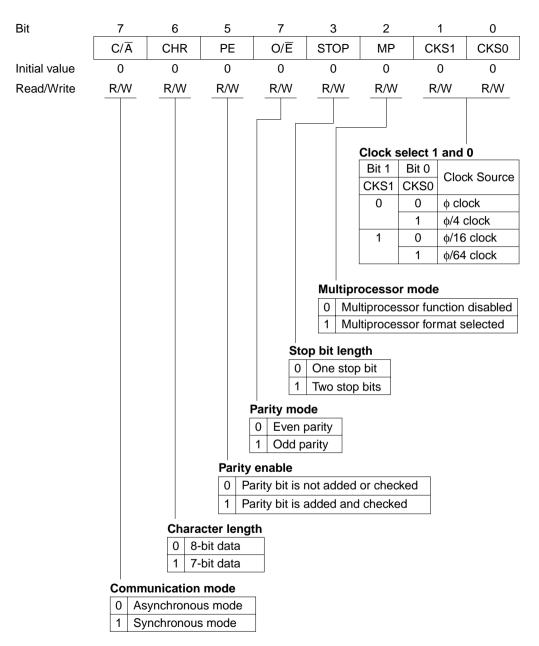


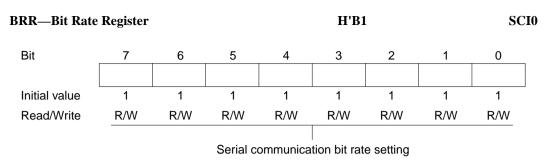
Note: \* Only 0 can be written in bit 7 to clear the flag.

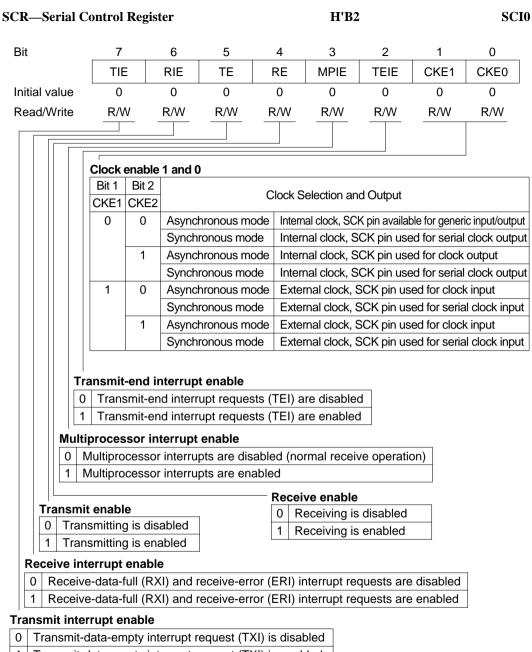
#### SMR—Serial Mode Register



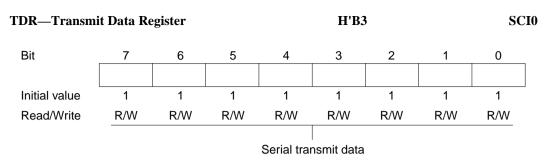
SCI0







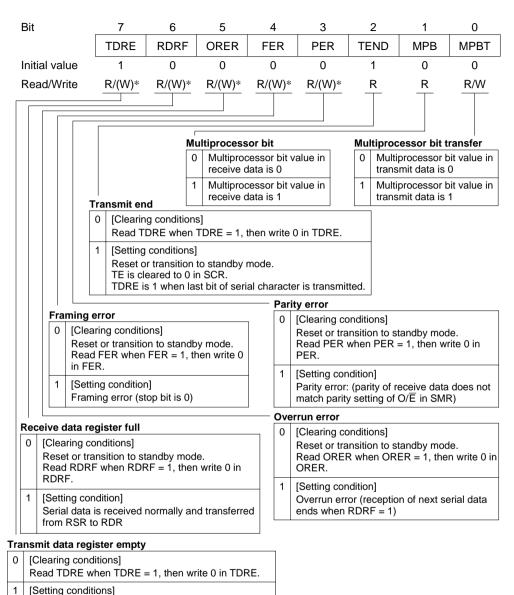
1 Transmit-data-empty interrupt request (TXI) is enabled



#### SSR—Serial Status Register



SCI0



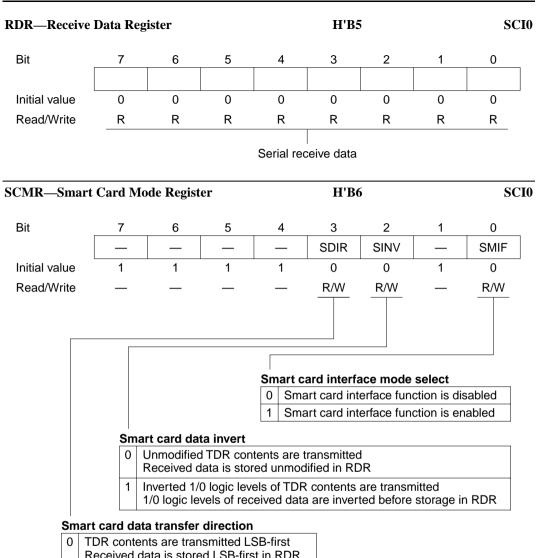
Note: \* Only 0 can be written to clear the flag.

Data is transferred from TDR to TSR, enabling new

Reset or transition to standby mode.

TE is 0 in SCR

data to be written in TDR.



1	TDR contents are transmitted MSB-first
	Received data is stored MSB-first in RDR

#### Renesas

MR—Serial M	lode Regi	ster			H'B8	3		SCI
Bit	7	6	5	7	3	2	1	0
	C/Ā	CHR	PE	O/Ē	STOP	MP	CKS1	CKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Note: Bit func	tions are t	he same a	as for SCI0	).				
RR—Bit Rate	Register				H'B9	)		SCI
Bit	7	6	5	4	3	2	1	0
Initial value	1	1	1	1	1	1	1	1
Deed/Muite	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
Read/Write Note: Bit func				).				
	tions are t	he same a		).	H'BA	A		SCI
Note: Bit func	tions are t	he same a		94	H'BA 3	2	1	SCI 0
Note: Bit func	tions are t	he same a gister	as for SCI0				1 CKE1	
Note: Bit func	tions are t	he same a gister 6	as for SCI0	4	3	2	1	0
Note: Bit func CR—Serial Co Bit	tions are t ontrol Reg 7 TIE	he same a gister 6 RIE	5 TE	4 RE	3 MPIE	2 TEIE	CKE1	0 CKE0
Note: Bit func CR—Serial Co Bit Initial value	tions are t ontrol Reg 7 TIE 0 R/W	he same a gister 6 RIE 0 R/W	5 TE 0 R/W	4 RE 0 R/W	3 MPIE 0	2 TEIE 0	CKE1	0 СКЕ0 0
Note: Bit func CR—Serial Co Bit Initial value Read/Write	tions are t ontrol Reg 7 TIE 0 R/W tions are t	he same a gister 6 RIE 0 R/W he same a	5 TE 0 R/W	4 RE 0 R/W	3 MPIE 0	2 TEIE 0 R/W	CKE1	0 СКЕ0 0
Note: Bit func CR—Serial Co Bit Initial value Read/Write Note: Bit func	tions are t ontrol Reg 7 TIE 0 R/W tions are t	he same a gister 6 RIE 0 R/W he same a	5 TE 0 R/W	4 RE 0 R/W	3 MPIE 0 R/W	2 TEIE 0 R/W	CKE1	0 CKE0 0 R/W
Note: Bit func <b>CR—Serial Co</b> Bit Initial value Read/Write Note: Bit func <b>TR—Transmi</b>	tions are t ontrol Reg 7 TIE 0 R/W tions are t	he same a gister 6 RIE 0 R/W he same a egister	5 TE 0 R/W as for SCI0	4 RE 0 R/W	3 MPIE 0 R/W H'BI	2 TEIE 0 R/W	CKE1 0 R/W	0 CKE0 0 R/W
Note: Bit func <b>CR—Serial Co</b> Bit Initial value Read/Write Note: Bit func <b>TR—Transmi</b>	tions are t ontrol Reg 7 TIE 0 R/W tions are t	he same a gister 6 RIE 0 R/W he same a egister	5 TE 0 R/W as for SCI0	4 RE 0 R/W	3 MPIE 0 R/W H'BI	2 TEIE 0 R/W	CKE1 0 R/W	0 CKE0 0 R/W

#### SSR—Serial Status Register

# H'BC

SCI1

Bit	7	6	5	4	3	2	1	0
	TDRE	RDRF	ORER	FER	PER	TEND	MPB	MPBT
Initial value	1	0	0	0	0	1	0	0
Read/Write	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R	R	R/W

Notes: Bit functions are the same as for SCI0.

\* Only 0 can be written to clear the flag.

RDR—Receive	H'BI	H'BD							
Bit	7	6	5	4	3	2	1	0	٦
Initial value	0	0	0	0	0	0	0	0	
Read/Write	R	R	R	R	R	R	R	R	

Note: Bit functions are the same as for SCI0.

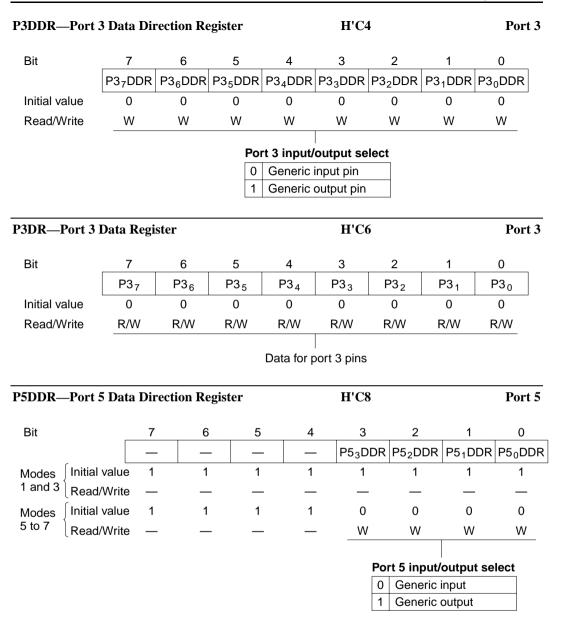
P1DDR-	-Port 1 Da	ta Directi	ion Regist	ter	H'C0				
Bit		7	6	5	4	3	2	1	0
		P17DDR	P1 <sub>6</sub> DDR	P1₅DDR	P1 <sub>4</sub> DDR	P1 <sub>3</sub> DDR	P1 <sub>2</sub> DDR	P11DDR	P10DDR
Modes	Initial value	e 1	1	1	1	1	1	1	1
1 and 3	Read/Write	e —	—		—	—	—	—	—
Modes	Initial value	e 0	0	0	0	0	0	0	0
5 to 7	Read/Write	e _ W	W	W	W	W	W	W	W

### Port 1 input/output select

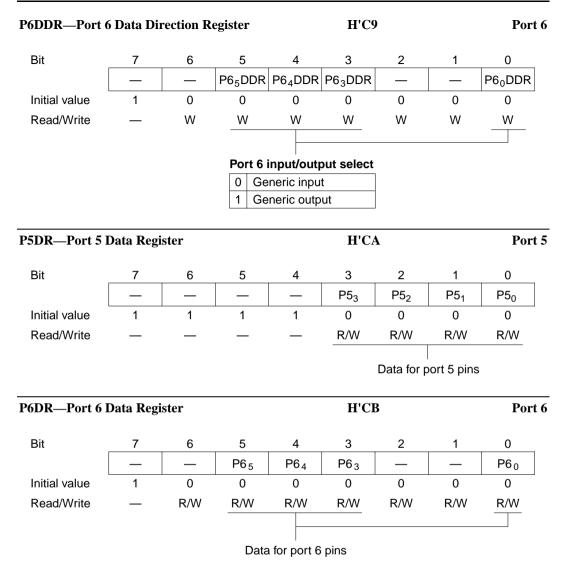
0	Generic input pin
1	Generic output pin

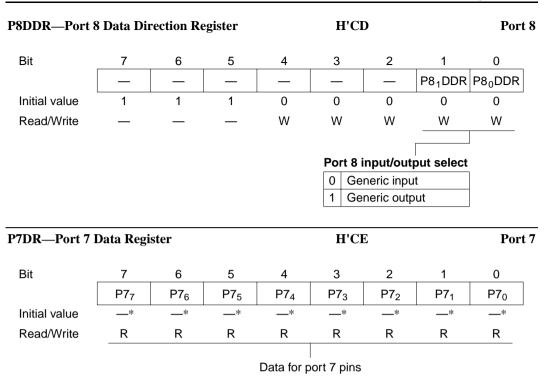
Bit       7       6       5       4       3       2       1       0 $P2_{7}DDR$ $P2_{6}DDR$ $P2_{6}DDR$ $P2_{4}DDR$ $P2_{3}DDR$ $P2_{2}DDR$ $P2_{1}DDR$ $P2_{0}DDR$ Modes       Initial value       1	P2DDR-	–Port 2 Da	ta Directi	on Regist	ter	H'C1					
Modes       Initial value       1 <th1< th="">       1       1</th1<>	Bit		7	6	5	4	3	2	1	0	
1 and 3       Read/Write       -			P27DDR	P2 <sub>6</sub> DDR	P25 DDR	P2 <sub>4</sub> DDR	P23DDR	P2 <sub>2</sub> DDR	P21 DDR	P20DDR	
Modes       Initial value       0	Modes	Initial valu	e 1	1	1	1	1	1	1	1	
5 to 7       Read/Write       W	1 and 3	Read/Writ	e —	_	_	—	—	—	—	—	
Or V W W W W W W W W W W W W W W W W W W		Initial valu	e 0	0	0	0	0	0	0	0	
$\begin{array}{c c c c c c c c c c c c c c c c c c c $	5 to 7	Read/Writ	e_W	W	W	W	W	W	W	W	
Bit       7       6       5       4       3       2       1       0         P17       P16       P15       P14       P13       P12       P11       P10         Initial value       0       0       0       0       0       0       0       0       0         Read/Write       R/W       R/W       R/W       R/W       R/W       R/W       R/W       R/W       R/W         Data for port 1 pins         P2DR—Port 2 Data Register       H'C3       Port 2         Bit       7       6       5       4       3       2       1       0         P27       P26       P25       P24       P23       P22       P21       P20         Initial value       0       0       0       0       0       0       0       0       0					0	Generic	input pin				
$\begin{array}{c ccccccccccccccccccccccccccccccccccc$	P1DR—I	Port 1 Data	a Register				H'C2			Port 1	
Initial value       0	Bit		7	6	5	4	3	2	1	0	
Read/Write       R/W       R/W		F	P1 <sub>7</sub> F	P1 <sub>6</sub> F	P1 <sub>5</sub>	P1 <sub>4</sub>	P1 <sub>3</sub>	P1 <sub>2</sub>	P1 <sub>1</sub>	P1 <sub>0</sub>	
Data for port 1 pins         Data for port 1 pins         P2DR—Port 2 Data Register       H'C3       Port 2         Bit       7       6       5       4       3       2       1       0         Bit       7       6       5       4       3       2       1       0         Distance       P27       P26       P25       P24       P23       P22       P21       P20         Initial value       0       0       0       0       0       0       0       0	Initial va	alue	0	0	0	0	0	0	0	0	
P2DR—Port 2 Data Register     H'C3     Port 2       Bit     7     6     5     4     3     2     1     0       P27     P26     P25     P24     P23     P22     P21     P20       Initial value     0     0     0     0     0     0     0	Read/W	/rite F	R/W F	R/W F	R/W F	R/W	R/W F	R/W I	R/W F	R/W	
Bit         7         6         5         4         3         2         1         0           P27         P26         P25         P24         P23         P22         P21         P20           Initial value         0         0         0         0         0         0         0         0					Da	ta for port	1 pins				
P27         P26         P25         P24         P23         P22         P21         P20           Initial value         0         0         0         0         0         0         0         0	P2DR—I	Port 2 Data	a Register				Н'С3			Port 2	
Initial value 0 0 0 0 0 0 0 0	Bit		7	6	5	4	3	2	1	0	
		F	P27	P2 <sub>6</sub> F	2 <sub>5</sub>   1	P2 <sub>4</sub>	P2 <sub>3</sub>	P2 <sub>2</sub>	P2 <sub>1</sub>	P2 <sub>0</sub>	
Read/Write R/W R/W R/W R/W R/W R/W R/W	Initial va	alue	0	0	0	0	0	0	0	0	
	Read/W	/rite F	R/W F	R/W F	R/W F	R/W	R/W F	R/W I	R/W F	R/W	

Data for port 2 pins



#### Renesas

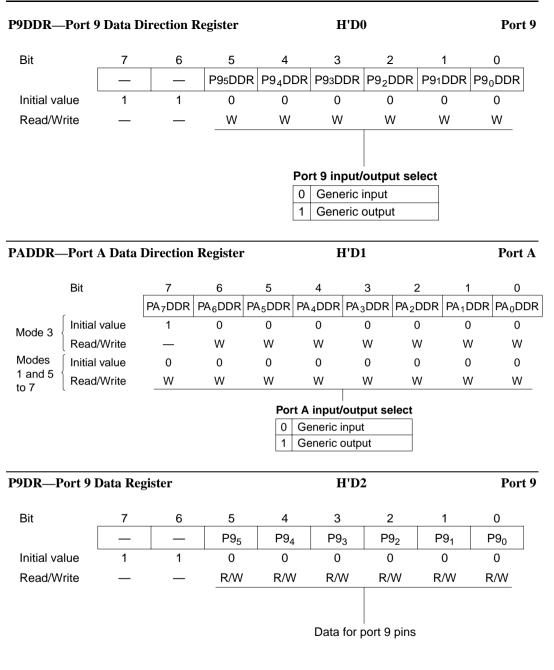




Note: \* Determined by pins  $P7_7$  to  $P7_0$ .

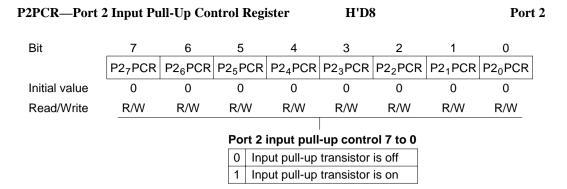
P8DR—Port 8	Data Regi	ster			H'CF				
Bit	7	6	5	4	3	2	1	0	_
		_	_	_	_	_	P8 1	P8 <sub>0</sub>	
Initial value	1	1	1	0	0	0	0	0	_
Read/Write	—	_	—	R/W	R/W	R/W	R/W	R/W	

Data for port 8 pins

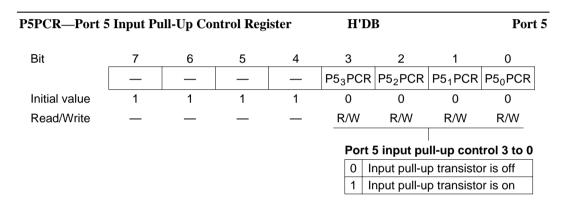


PADR—Port A	Data Reg	ister			H'D:	3		Port A			
Bit	7	6	5	4	3	2	1	0			
	PA <sub>7</sub>	PA <sub>6</sub>	PA <sub>5</sub>	PA <sub>4</sub>	PA <sub>3</sub>	PA <sub>2</sub>	PA <sub>1</sub>	PA <sub>0</sub>			
Initial value	0	0	0	0	0	0	0	0			
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W			
	Data for port A pins										
PBDDR—Port	B Data Di	rection <b>F</b>	Register		H'D4	l .		Port B			
Bit	7	6	5	4	3	2	1	0			
	PB7DDR	_	PB <sub>5</sub> DDR	PB <sub>4</sub> DDR	PB <sub>3</sub> DDR	PB <sub>2</sub> DDR	PB <sub>1</sub> DDR	PB <sub>0</sub> DDR			
Initial value	0	0	0	0	0	0	0	0			
Read/Write	W	W	W	W	W	W	W	W			
			Por	t B input/	/ output se	lect					
			0	Generic i							
			1	Generic	output						
PBDR—Port B	Data Reg	ister			H'De	6		Port B			
Bit	7	6	5	4	3	2	1	0			
	PB <sub>7</sub>	_	PB <sub>5</sub>	PB <sub>4</sub>	PB <sub>3</sub>	PB <sub>2</sub>	PB <sub>1</sub>	PB <sub>0</sub>			
Initial value	0	0	0	0	0	0	0	0			
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W			
	L				 / <b>5</b> · ·						

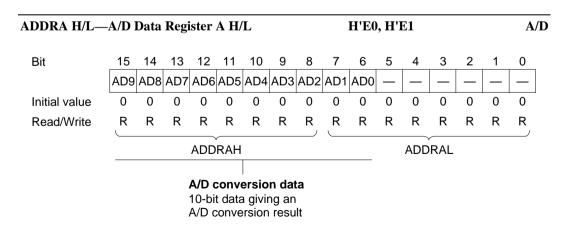
Data for port B pins

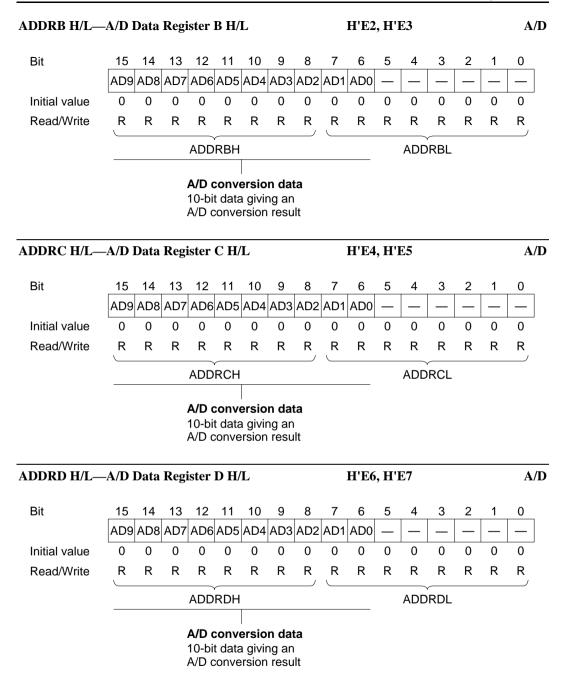


Note: Valid when the corresponding P2DDR bit is cleared to 0 (designating generic input).



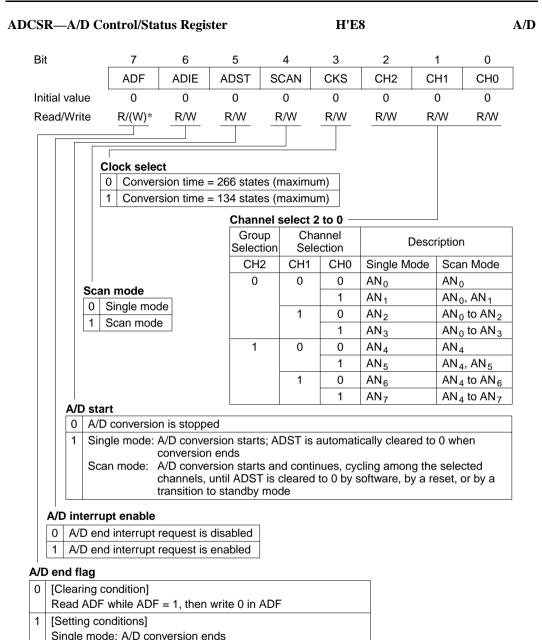
Note: Valid when the corresponding P5DDR bit is cleared to 0 (designating generic input).





#### Renesas

ADCR—A/D C		А	/D								
Bit	7	6	5	4	3	2	1	0	_		
	TRGE	_	—	_	_	_	—	_			
Initial value	0	1	1	1	1	1	1	1	_		
Read/Write	R/W	—	—	—	—	—	—	—			
	Trigger enable										
	0 A/D conversion cannot be externally triggered										
	1 A/[	) conversi	on starts a	at the fall c	of the exte	rnal trigge	r signal (Ā	DTRG)			



Scan mode: A/D conversion ends in all selected channels

Note: \* Only 0 can be written to clear flag.

Bit	7	6	5	4	3	2	1	0
	AST7	AST6	AST5	AST4	AST3	AST2	AST1	AST0
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W							

#### Area 7 to 0 access state control

Bits 7 to 0	Number of States in Access Cycle
AST7 to AST0	Number of States III Access Cycle
0	Areas 7 to 0 are two-state access areas
1	Areas 7 to 0 are three-state access areas

# WCR—Wait Control Register

Bit	7	6	5	4	3	2	1	0
	_		_		WMS1	WMS0	WC1	WC0
Initial value	1	1	1	1	0	0	1	1
Read/Write	—	—	—	—	R/W	R/W	R/W	R/W

Γ

#### Wait mode select 1 and 0

Bit 3	Bit 2							
WMS1	WMS0	Wait Mode						
0	0	Programmable wait mode						
	1	No wait states inserted by wait-state controller						
1	0	Pin wait mode 1						
	1	Pin auto-wait mode						

#### Wait count 1 and 0

H'EE

H'ED

**Bus controller** 

**Bus controller** 

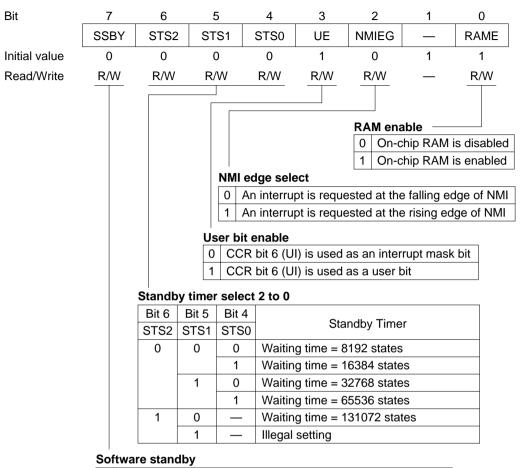
Bit 1	Bit 0	Number of Wait States						
WC1	WC0	Number of Walt States						
0	0	No wait states inserted by wait-state controller						
	1	1 state inserted						
1	0	2 states inserted						
	1	3 states inserted						

WCER—Wait Controller Enable Register					ł	I'EF		Bu	s controller
Bit	7	6	5	4	3		2	1	0
	WCE7	WCE6	WCE5	WCE4	WC	E3 \	WCE2	WCE1	WCE0
Initial value	1	1	1	1	1		1	1	1
Read/Write	R/W	R/W	R/W	R/W	R/\	N	R/W	R/W	R/W
			Wait state						
								ait mode 0)	
			1 Wait-s	state cont	trol is e	nabled			
MDCR—Mode	Control I	Dogistor			T	I'F1		Sve	tem control
WIDCK-WOUL	Control	xegistei			1	1 Г 1		5y5	
Bit	7	6	5	4	3		2	1	0
	_	—	—	_		-   1	MDS2	MDS1	MDS0
Initial value	1	1	0	0	0		*	*	*
Read/Write	—	_	—	—		-	R	R	R
				Madaaa		4- <b>0</b>			
				Mode se Bit 2	Bit 1	Bit 0			
				MD <sub>2</sub>	MD <sub>1</sub>	MD <sub>0</sub>	- Onei	ating mod	e
				WD2	0	0	_		
				0		1	Mode	1	
				0	1	0	—		
						1	1 Mode 3		
					0	0 —			
				1		1	Mode		_
					1	0	Mode		_
						1	Mode	1	

Note: \* Determined by the state of the mode pins (MD\_2 to MD\_0).







- 0 SLEEP instruction causes transition to sleep mode
- 1 SLEEP instruction causes transition to software standby mode

ADRCR	—Address C			H'F3	<b>Bus controller</b>				
	Bit	7 A <sub>23</sub> E	6 A <sub>22</sub> E	5 A <sub>21</sub> E	4	3	2	1	0
Modes	Initial value	1	1	1	1	1	1	1	0
1 and 5 to 7	Read/Write	—	—	—	—	—	—	—	R/W
Mada 0	Initial value	1	1	1	1	1	1	1	0
Mode 3	Read/Write	R/W	R/W	R/W	—	—	—	—	R/W
	A	ddress 23	to 21 ena	ble					
	(	Address	output						

1 I/O pins other than the above

#### ISCR—IRQ Sense Control Register

Bit	7	6	5	4	3	2	1	0
	_	—	IRQ5SC	IRQ4SC	_	—	IRQ1SC	IRQ0SC
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

#### IRQ5, IRQ4, IRQ1 and IRQ0 sense control

0	Interrupts are requested when $\overline{IRQ}_5, \overline{IRQ}_4, \overline{IRQ}_1, \text{ and } \overline{IRQ}_0$ inputs are low
1	Interrupts are requested by falling-edge input at $\overline{IRQ}_5, \overline{IRQ}_4, \overline{IRQ}_1$ and $\overline{IRQ}_0$

H'F4

Interrupt controller

IER—IRQ Enable Register						5	Interrupt controller		
Bit	7	6	5	4	3	2	1	0	_
			IRQ5E	IRQ4E	_	_	IRQ1E	IRQ0E	
Initial value	0	0	0	0	0	0	0	0	,
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W	
						1			
			IRO- II	RQ₄, IRQ₁	IRQ <sub>e</sub> en	 ahle			
			<b>•</b> ••		•		runts are d	lisabled	

IRQ<sub>5</sub>, IRQ<sub>4</sub>, IRQ<sub>1</sub> and IRQ<sub>0</sub> interrupts are disabled
 IRQ<sub>5</sub>, IRQ<sub>4</sub>, IRQ<sub>1</sub> and IRQ<sub>0</sub> interrupts are enabled

ISR—IRQ Status Register						H'F6		Interrupt controller		
Bit	7	6	5	4	3	2	1	0		
	_	_	IRQ5F	IRQ4F			IRQ1F	IRQ0F		
Initial value	0	0	0	0	0	0	0	0		
Read/Write	_	—	R/(W)*	R/(W)*	_	—	R/(W)*	R/(W)*		
	∣ IRQ <sub>5</sub> , IRQ₄	ı, IRQ₁ an	d IRQ <sub>0</sub> fla	gs						
	Bits 5, 4,	1 and 0	Setting and Clearing Conditions							
	IRQ									
	IRQ4F IRQ1F		Setting and Clearing Conditions							
	IRG									
	C	)	[Clearing	Clearing conditions]						
	• Read IRQnF when IRQnF = 1, then write 0 in IRQnF									
	• IRQnSC = 0, IRQn input is high, and interrupt exception handling is carried out.						xception			
			<ul> <li>IRQnSC = 1 and IRQn interrupt exception handling is carried out.</li> </ul>							
	1		[Setting conditions]							
			<ul> <li>IRQnSC = 0 and IRQn input is low.</li> <li>IRQnSC = 1 and IRQn input changes from high to low.</li> </ul>							

Note: n = 5, 4, 1 and 0

Note: \* Only 0 can be written to clear the flag.

PRA—Interru	pt Priorit	y Register	· A		H'F8	3	Interrup	ot controlle
Bit	7	6	5	4	3	2	1	0
	IPRA7	IPRA6	_	IPRA4	IPRA3	IPRA2	IPRA1	IPRA0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W
				-	el A7, A6, level 0 (lo level 1 (hi	w priority)		
	it7 Bi	•	it5 E			iit 2 PRA2	Bit 1 IPRA1	Bit 0 IPRA0
Interrupt IF source		RQ <sub>1</sub> —		RQ₄, V RQ₅		ſU hannel	ITU channel 1	ITU channel 2
PRB—Interru	pt Priorit	y Register	·B		H'F9	)	Interrup	ot controlle
Bit	7	6	5	4	3	2	1	0
	IPRB7	IPRB6	_	_	IPRB3	IPRB2	IPRB1	
			0	0	0	0	0	0
Initial value	0	0						D / / /
Initial value Read/Write	0 R/W	0 R/W	R/W	R/W		R/W	R/W	R/W
			R/W Priority	R/W	B6, B3 to	B1		R/W

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
	IPRB7	IPRB6	—	—	IPRB3	IPRB2	IPRB1	—
Interrupt source	ITU channel 3	ITU channel 4	-	-	SCI channel 0	SCI channel 1	A/D converter	—

# Appendix C I/O Block Diagrams

# C.1 Port 1 Block Diagram

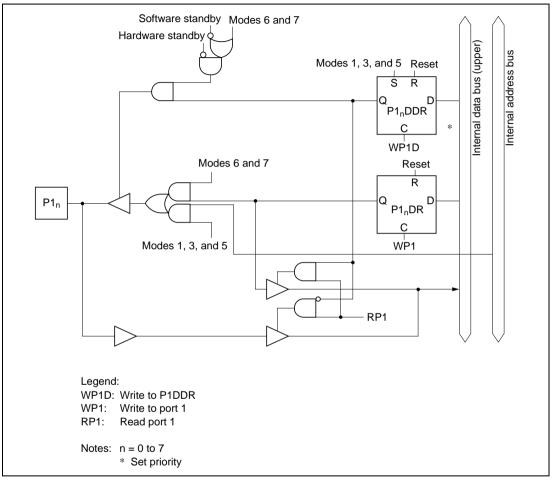


Figure C.1 Port 1 Block Diagram

# C.2 Port 2 Block Diagram

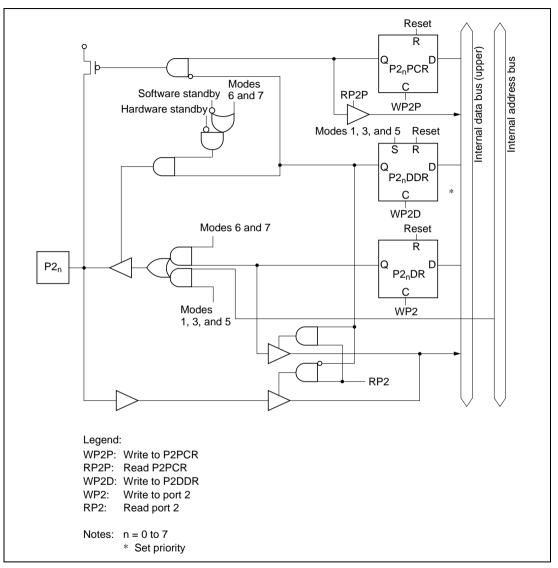


Figure C.2 Port 2 Block Diagram



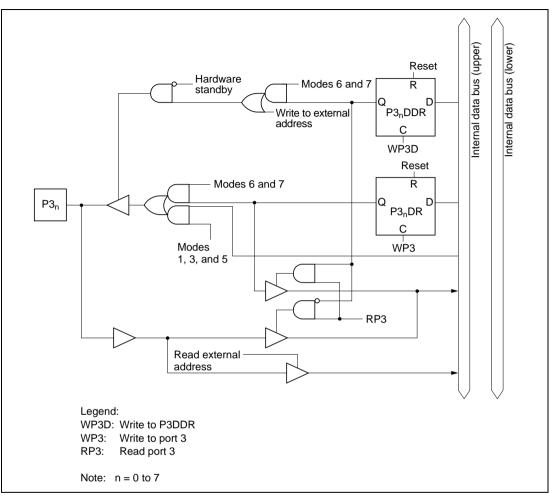


Figure C.3 Port 3 Block Diagram

# C.4 Port 5 Block Diagram

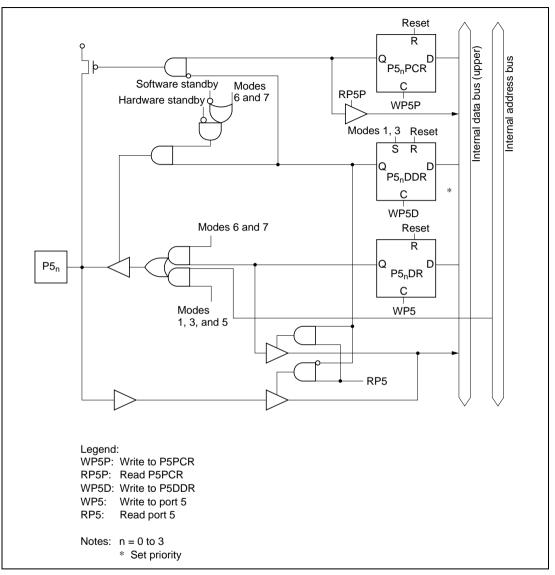


Figure C.4 Port 5 Block Diagram

# C.5 Port 6 Block Diagram

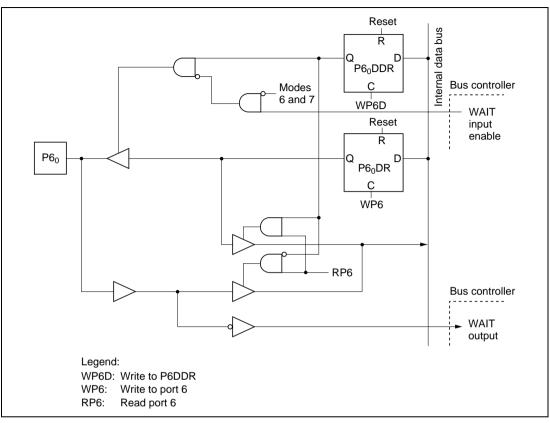


Figure C.5 (a) Port 6 Block Diagram (Pin P6<sub>0</sub>)

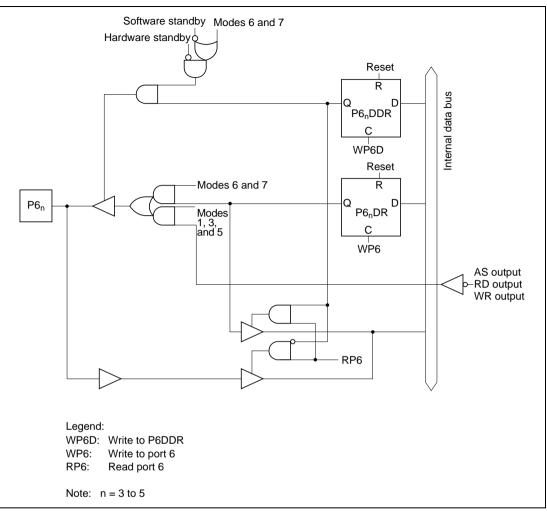


Figure C.5 (b) Port 6 Block Diagram (Pins P6, to P6,)

## C.6 Port 7 Block Diagram

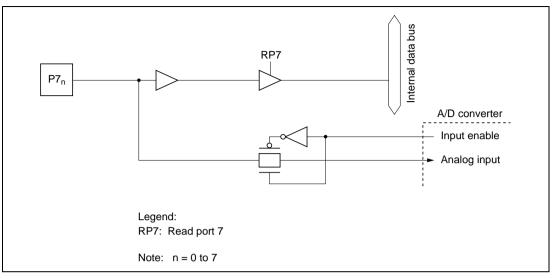


Figure C.6 Port 7 Block Diagram

## C.7 Port 8 Block Diagram

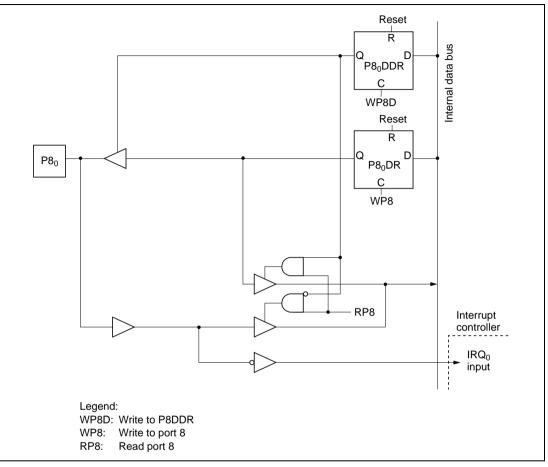


Figure C.7(a) Port 8 Block Diagram (Pin P8<sub>0</sub>)

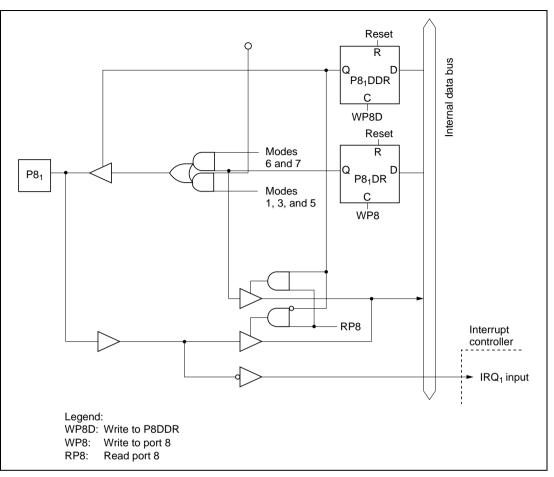


Figure C.7 (b) Port 8 Block Diagram (Pin P8<sub>1</sub>)

## C.8 Port 9 Block Diagram

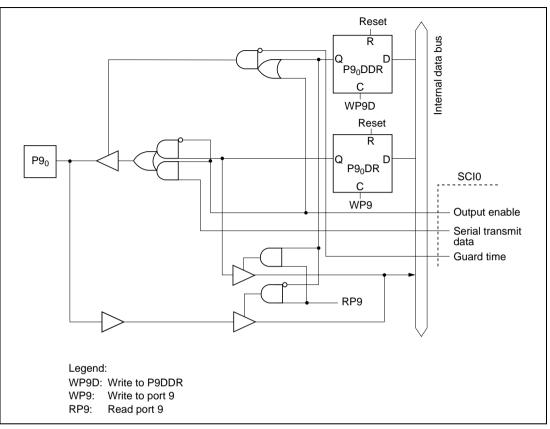


Figure C.8 (a) Port 9 Block Diagram (Pin P9<sub>0</sub>)

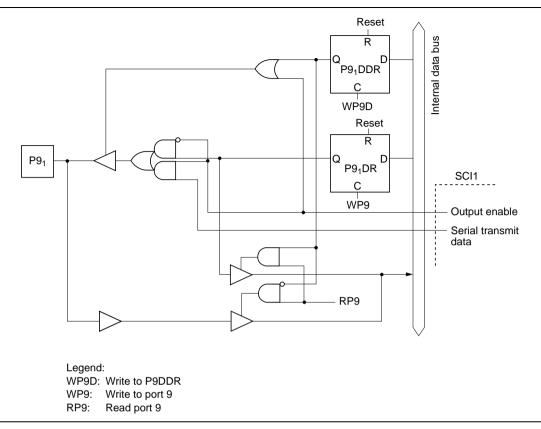


Figure C.8 (b) Port 9 Block Diagram (Pin P9<sub>1</sub>)

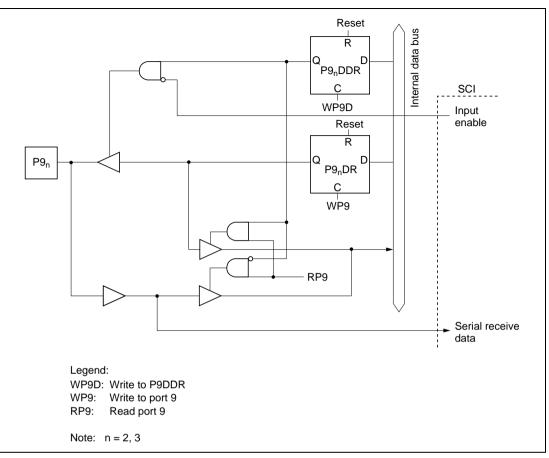


Figure C.8 (c) Port 9 Block Diagram (Pin P9<sub>2</sub>, P9<sub>3</sub>)

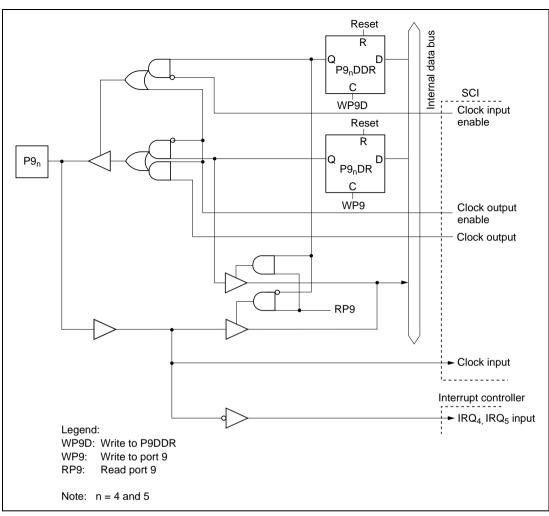


Figure C.8 (d) Port 9 Block Diagram (Pin P9<sub>4</sub>, P9<sub>5</sub>)

## C.9 Port A Block Diagram

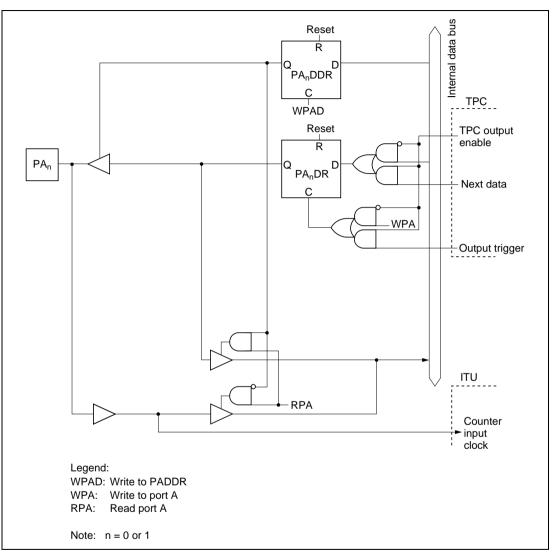


Figure C.9 (a) Port A Block Diagram (Pins PA<sub>0</sub>, PA<sub>1</sub>)

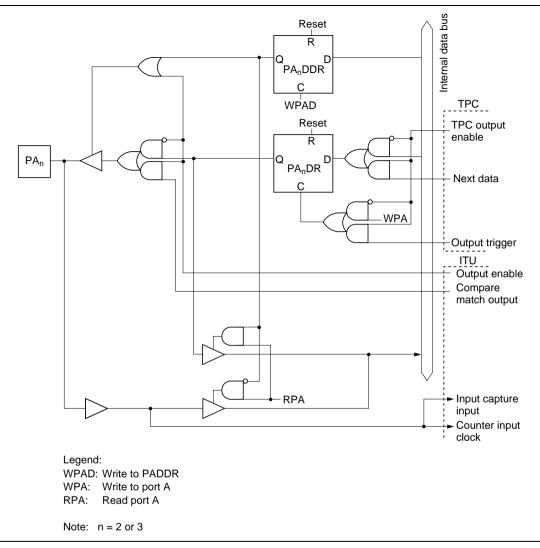


Figure C.9 (b) Port A Block Diagram (Pins PA<sub>2</sub>, PA<sub>3</sub>)

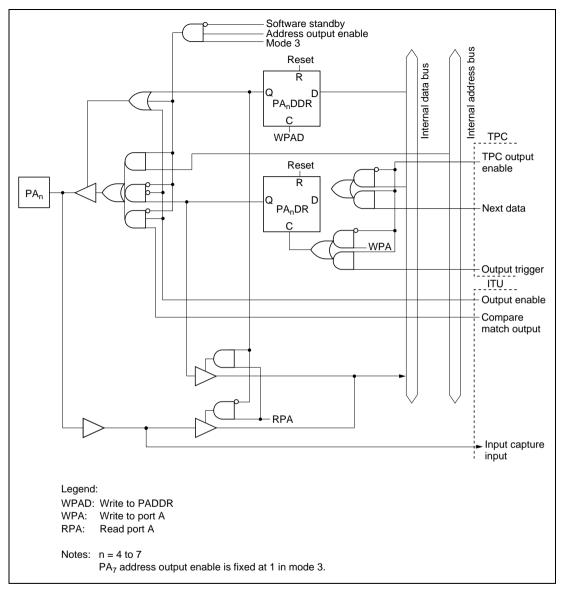


Figure C.9 (c) Port A Block Diagram (Pins PA<sub>4</sub> to PA<sub>7</sub>)

## C.10 Port B Block Diagram

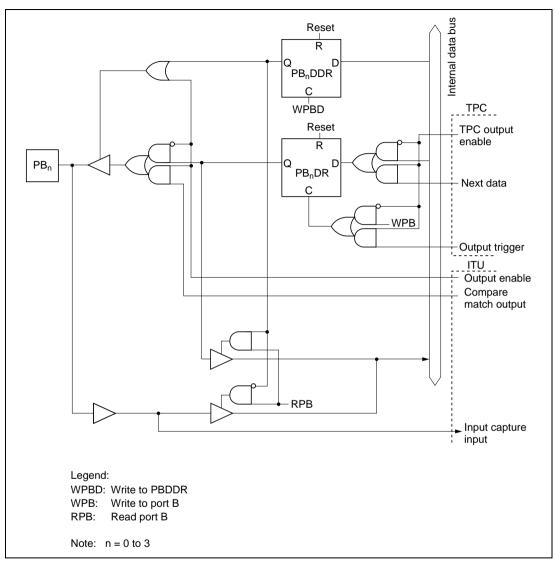


Figure C.10 (a) Port B Block Diagram (Pins PB<sub>0</sub> to PB<sub>3</sub>)

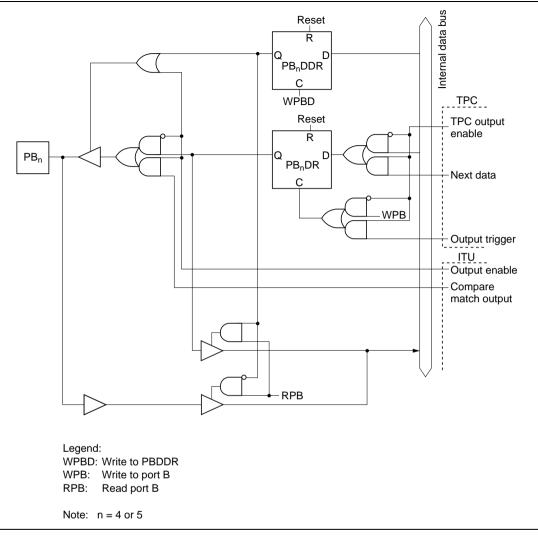


Figure C.10 (b) Port B Block Diagram (Pins PB<sub>4</sub>, PB<sub>5</sub>)

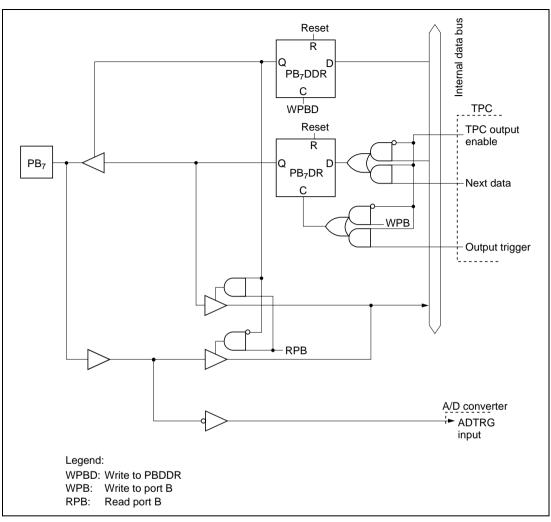


Figure C.10 (c) Port B Block Diagram (Pin PB<sub>7</sub>)

# Appendix D Pin States

# **D.1** Port States in Each Mode

## Table D.1Port States

Pin Name	Mode	Reset State	Hardware Standby Mode	Software Standby Mode	Program Execution State Sleep Mode
φ	—	Clock output	Т	Н	Clock output
RESO <sup>*1</sup>	_	T* <sup>2</sup>	Т	Т	RESO
P1, to P1,	1, 3	L	Т	Т	A <sub>7</sub> to A <sub>0</sub>
	5	Т	Т	keep	Input port (DDR = 0)
				Т	$A_7$ to $A_0$ (DDR = 1)
	6, 7	Т	Т	keep	I/O port
P2, to P2	1, 3	L	Т	Т	A <sub>15</sub> to A <sub>8</sub>
	5	Т	Т	keep	Input port (DDR = 0)
				Т	$A_{15}$ to $A_{8}$ (DDR = 1)
	6, 7	Т	Т	keep	I/O port
P3 <sub>7</sub> to P3 <sub>0</sub>	1, 3, 5	Т	Т	Т	D <sub>7</sub> to D <sub>0</sub>
	6, 7	Т	Т	keep	I/O port
P5 <sub>3</sub> to P5 <sub>0</sub>	1, 3	L	Т	Т	A <sub>19</sub> to A <sub>16</sub>
	5	Т	Т	keep	Input port (DDR = 0)
				Т	$A_{19}$ to $A_{16}$ (DDR = 1)
	6, 7	Т	Т	keep	I/O port
P6 <sub>0</sub>	1, 3, 5	Т	Т	keep	I/O port, WAIT
	6, 7	Т	Т	keep	I/O port
P6₅ to P6₃	1, 3, 5	Н	Т	Т	WR, RD, AS
	6, 7	Т	Т	keep	I/O port
P7 <sub>7</sub> to P7 <sub>0</sub>	1, 3, 5 to 7	Т	Т	Т	Input port
P8 <sub>0</sub>	1, 3, 5	Т	Т	keep	I/O port
	6, 7	Т	Т	keep	I/O port

Pin Name	e Mode	Reset State	Hardware Standby Mode	Software Standby Mode	Program Execution State Sleep Mode
P8,	1, 3, 5	Т	Т	T [DDR = 0]	Input port [DDR = 0]
				H [DDR = 1]	H [DDR = 1]
	6, 7	Т	Т	keep	I/O port
P9₅ to P9₀	1, 3, 5 to 7	Т	Т	keep	I/O port
PA <sub>3</sub> to PA	$A_0$ 1, 3, 5 to 7	Т	Т	keep	I/O port
PA <sub>6</sub> to PA	A <sub>4</sub> 3	Т	Т	[ADRCR = 0] T	(ADRCR = 0) A <sub>21</sub> to A <sub>23</sub>
				[ADRCR = 1] keep	(ADRCR = 1) I/O port
	1, 5, 6, 7	Т	Т	keep	I/O port
PA <sub>7</sub>	3	L	Т	Т	A <sub>20</sub>
	1, 5, 6, 7	Т	Т	keep	I/O port
PB <sub>7</sub> , PB <sub>5</sub> t PB <sub>0</sub>	to 1, 3, 5 to 7	Т	Т	keep	I/O port
Legend: H: L: T: keep: DDR: ADRCR:	High Low High-impedance Input pins are in Data direction re Address control	the high-ir gister bit	npedance state; out	put pins maintain	their previous state.

Notes: 1 Mask ROM version. Dedicated FWE input pin for the F-ZTAT version.

2 Low output only when WDT overflows causes a reset.

## D.2 Pin States at Reset

## **Reset in T1 State**

Figure D.1 is a timing diagram for the case in which  $\overline{\text{RES}}$  goes low during the T1 state of an external memory access cycle. As soon as  $\overline{\text{RES}}$  goes low, all ports are initialized to the input state.  $\overline{\text{AS}}$ ,  $\overline{\text{RD}}$ , and  $\overline{\text{WR}}$  go high, and the data bus goes to the high-impedance state. The address bus is initialized to the low output level 0.5 state after the low level of  $\overline{\text{RES}}$  is sampled. Sampling of  $\overline{\text{RES}}$  takes place at the fall of the system clock ( $\phi$ ).

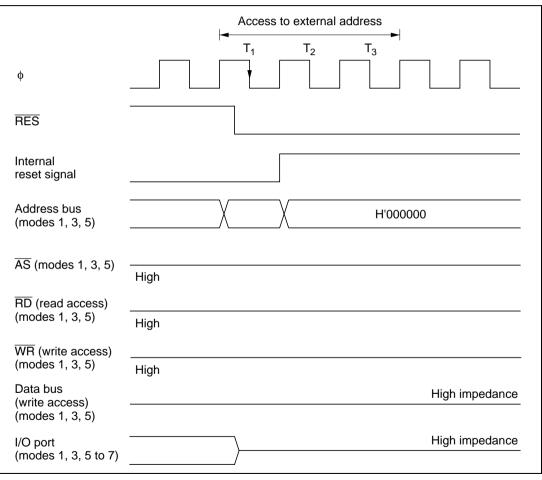


Figure D.1 Reset during Memory Access (Reset during T1 State)

### Reset in T2 State

Figure D.2 is a timing diagram for the case in which  $\overline{\text{RES}}$  goes low during the T2 state of an external memory access cycle. As soon as  $\overline{\text{RES}}$  goes low, all ports are initialized to the input state.  $\overline{\text{AS}}$ ,  $\overline{\text{RD}}$ , and  $\overline{\text{WR}}$  go high, and the data bus goes to the high-impedance state. The address bus is initialized to the low output level 0.5 state after the low level of  $\overline{\text{RES}}$  is sampled. The same timing applies when a reset occurs during a wait state (T<sub>w</sub>).

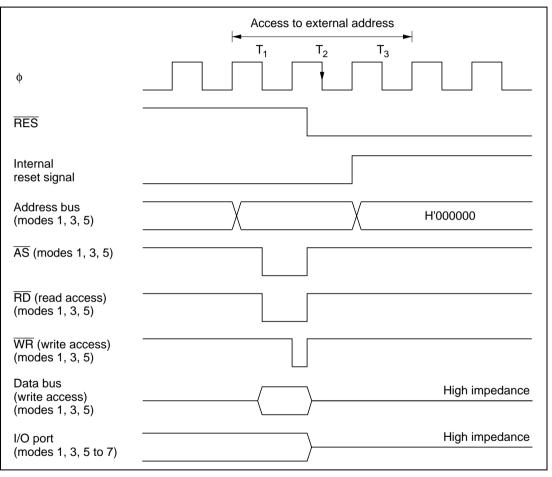


Figure D.2 Reset during Memory Access (Reset during T2 State)

## Reset in T3 State

Figure D.3 is a timing diagram for the case in which  $\overline{\text{RES}}$  goes low during the T<sub>3</sub> state of an external memory access cycle. As soon as  $\overline{\text{RES}}$  goes low, all ports are initialized to the input state.  $\overline{\text{AS}}$ ,  $\overline{\text{RD}}$ , and  $\overline{\text{WR}}$  go high, and the data bus goes to the high-impedance state. The address bus outputs are held during the T<sub>3</sub> state. The same timing applies when a reset occurs in the T<sub>2</sub> state of an access cycle to a two-state-access area.

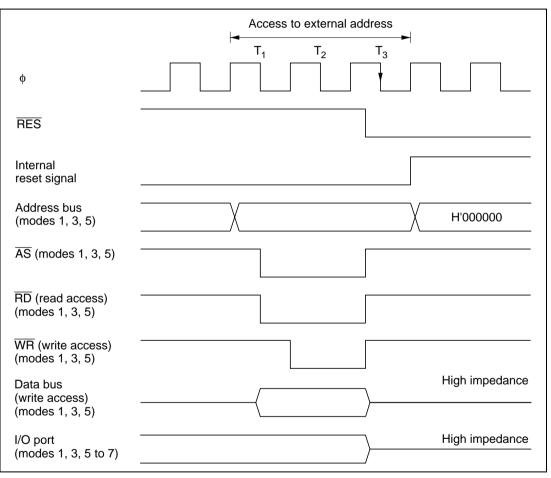
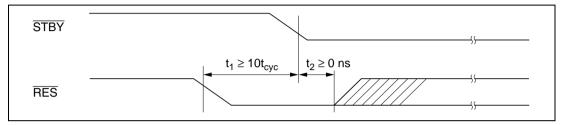


Figure D.3 Reset during Memory Access (Reset during T3 State)

# Appendix E Timing of Transition to and Recovery from Hardware Standby Mode

## Timing of Transition to Hardware Standby Mode

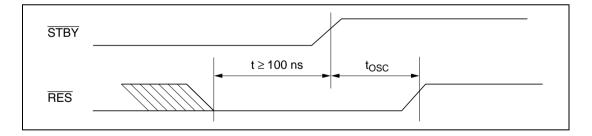
(1) To retain RAM contents with the RAME bit set to 1 in SYSCR, drive the RES signal low 10 system clock cycles before the STBY signal goes low, as shown below. RES must remain low until STBY goes low (minimum delay from STBY low to RES high: 0 ns).



(2) To retain RAM contents with the RAME bit cleared to 0 in SYSCR, RES does not have to be driven low as in (1).

### Timing of Recovery from Hardware Standby Mode

Drive the  $\overline{\text{RES}}$  signal low approximately 100 ns before  $\overline{\text{STBY}}$  goes high.



# Appendix F Product Lineup

## Table F.1 H8/3039 Group Product Lineup

Product Ty	уре		Part Number	Mark Code	Package (Package Code)
H8/3039	Flash memory version	5 V version	HD64F3039F	HD64F3039F	80-pin QFP (FP-80A)
			HD64F3039TE	HD64F3039TE	80-pin TQFP (TFP-80C)
		3 V version	HD64F3039VF	HD64F3039VF	80-pin QFP (FP-80A)
			HD64F3039VTE	HD64F3039VTE	80-pin TQFP (TFP-80C)
	Mask ROM version	5 V version	HD6433039F	HD6433039(***)F	80-pin QFP (FP-80A)
			HD6433039TE	HD6433039(***)TE	80-pin TQFP (TFP-80C)
		3 V version	HD6433039VF	HD6433039(***)VF	80-pin QFP (FP-80A)
			HD6433039VTE	HD6433039(***)VTE	80-pin TQFP (TFP-80C)
H8/3038	Mask ROM version	5 V version	HD6433038F	HD6433038(***)F	80-pin QFP (FP-80A)
			HD6433038TE	HD6433038(***)TE	80-pin TQFP (TFP-80C)
		3 V version	HD6433038VF	HD6433038(***)VF	80-pin QFP (FP-80A)
			HD6433038VTE	HD6433038(***)VTE	80-pin TQFP (TFP-80C)
H8/3037	Mask ROM version	5 V version	HD6433037F	HD6433037(***)F	80-pin QFP (FP-80A)
			HD6433037TE	HD6433037(***)TE	80-pin TQFP (TFP-80C)
		3 V version	HD6433037VF	HD6433037(***)VF	80-pin QFP (FP-80A)
			HD6433037VTE	HD6433037(***)VTE	80-pin TQFP (TFP-80C)
H8/3036	Mask ROM version	5 V version	HD6433036F	HD6433036(***)F	80-pin QFP (FP-80A)
			HD6433036TE	HD6433036(***)TE	80-pin TQFP (TFP-80C)
		3 V	HD6433036VF	HD6433036(***)VF	80-pin QFP (FP-80A)
		version	HD6433036VTE	HD6433036(***)VTE	80-pin TQFP (TFP-80C)

Note: (\*\*\*) in mask ROM versions is the ROM code.

# Appendix G Package Dimensions

The package dimension that is shown in the Renesas Semiconductor Package Data Book has priority.

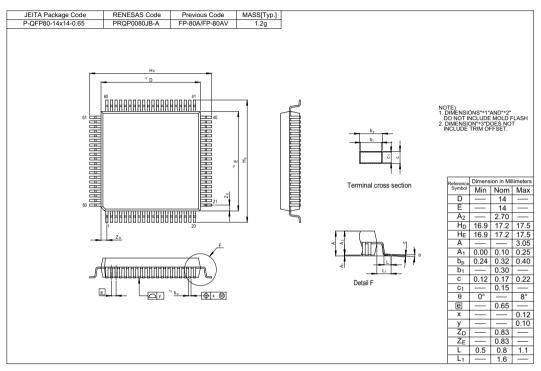


Figure G.1 Package Dimensions (FP-80A)

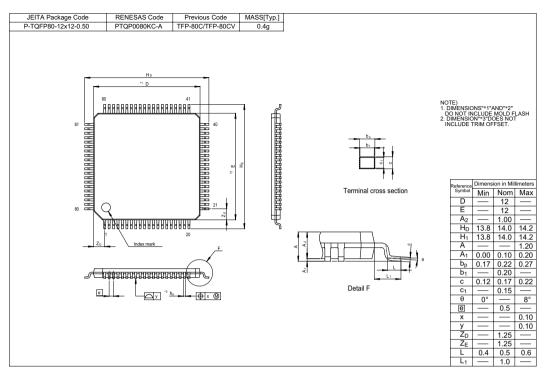


Figure G.2 Package Dimensions (TFP-80C)

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# H8/3039 Group, H8/3039F-ZTAT™ Hardware Manual



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